Digital Communications and Laboratory Fourth Homework

Faccin Dario, Santi Giovanni

MATLAB code

```
clc; close all; clear global; clearvars;
  load('Useful.mat', 'qc');
  load('Input_symbols.mat');
  SNR_vect = 1.4:0.05:1.8;
  sigma_a = 2;
                                           % Input variance
  M = 4;
                                           % Constellation cardinality
  gm = conj(qc(end:-1:1));
                                           % Matched filter: complex conjugate of qc
  h = conv(qc,gm);
                                           % Impulse response
  t0_bar = find(h == max(h));
                                           % Timing phase: peak of h
  h = h(h>max(h)/100);
_{15} h = h(3:end-2);
  h_T = downsample(h,4);
  r_gm = xcorr(gm,gm);
                                           % Matched filter autocorrelation
  Pbit_DFE_code = zeros(length(SNR_vect),1);
  N2 = floor(length(h_T)/2);
  N1 = N2;
  M1 = 5;
  D = 4;
  M2 = N2 + M1 - 1 - D;
  parfor i=1:length(SNR_vect)
          snr_db = SNR_vect(i);
          snr_lin = 10^(snr_db/10);
          % Single carrier channel simulation
          [r_c, sigma_w, qc] = channel_sim(symbols_ak, snr_db, sigma_a);
30
          % Additive complex-Gaussian noise
          w = wgn(length(r_c),1, 10*log10(sigma_w), 'complex');
          r_c = r_c + w;
          r_c_prime = filter(gm,1,r_c);
          r_c_prime = r_c_prime(t0_bar:end);
35
          % Signal at the input of the DFE
          x_aa = downsample(r_c_prime,4);
          rw_tilde = sigma_w/4 .* downsample(r_gm, 4);
          % DFE
          [c_opt, Jmin] = Adaptive_DFE(h_T, rw_tilde, sigma_a, M1, M2, D);
          psi = conv(c_opt, h_T);
          b = - psi(end - M2 + 1:end);
          y_hat = x_aa/max(psi);
          % Equalized signal
45
          detected = equalization_DFE(y_hat, c_opt, b, M1, M2, D);
          % Decoder as LLR
          llr = zeros(2*length(detected),1);
          Jmin_lin = 10^(Jmin/10);
          noise_var = (Jmin_lin-sigma_a*abs(1-max(psi))^2)/(abs(max(psi))^2);
          llr(1:2:end) = -2*real(detected)/(noise_var/2);
          11r(2:2:end) = -2*imag(detected)/(noise_var/2);
          % Deinterleave and decode bits
          llr_deinter = deinterleaver(llr);
          decoded = LDPC_decoder(llr_deinter).';
```

```
Pbit_DFE_code(i) = length(find(x(1:length(decoded))~=decoded))/length(decoded);
end
toc
```

```
clc; close all; clear global; clearvars;
  load('Useful.mat', 'qc');
  load('Input_symbols.mat');
  symbols = bitmap(x).';
  SNR vect = 0:14;
  sigma_a = 2;
                                           % Input variance
  M = 4;
                                           % Constellation cardinality
  gm = conj(qc(end:-1:1));
                                           % Matched filter: complex conjugate of qc
_{10} h = conv(qc,gm);
                                           % Impulse response
  t0_bar = find(h == max(h));
                                           % Timing phase: peak of h
  h = h(h>max(h)/100);
  h = h(3:end-2);
  h_T = downsample(h,4);
15 r_gm = xcorr(gm,gm);
                                           % Matched filter autocorrelation
  Pbit_DFE_uncode = zeros(length(SNR_vect),1);
  N2 = floor(length(h_T)/2);
  N1 = N2;
  M1 = 5;
_{20}|_{D} = 4;
  M2 = N2 + M1 - 1 - D;
  tic
  parfor i=1:length(SNR_vect)
          snr_db = SNR_vect(i);
          snr_lin = 10^(snr_db/10);
25
          % Single carrier channel simulation
          [r_c, sigma_w, qc] = channel_sim(symbols, snr_db, sigma_a);
          % Additive complex-Gaussian noise
          w = wgn(length(r_c),1, 10*log10(sigma_w), 'complex');
          r_c = r_c + w;
30
          r_c_prime = filter(gm,1,r_c);
          r_c_prime = r_c_prime(t0_bar:end);
          % Signal at the input of the DFE
          x_aa = downsample(r_c_prime,4);
          rw_tilde = sigma_w/4 .* downsample(r_gm, 4);
35
          % DFE
          [c_opt, Jmin] = Adaptive_DFE(h_T, rw_tilde, sigma_a, M1, M2, D);
          psi = conv(c_opt, h_T);
          b = - psi(end - M2 + 1:end);
          y_hat = x_aa/max(psi);
40
          % Equalized signal
          detected = equalization_DFE(y_hat, c_opt, b, M1, M2, D);
          hard_d = zeros(length(detected),1);
          for k=1:length(hard_d)
                   hard_d(k) = QPSK_detector(detected(k));
45
          end
          hard_bits = ibmap(hard_d);
          Pbit_DFE_uncode(i) = length(find(x(1:length(hard_bits))~=hard_bits))/length(hard_bits);
  end
50 toc
```

```
r_c_prime = r_c;
detected = r_c_prime;
llr = zeros(2*length(detected),1);
llr(1:2:end) = -2*real(detected)/(sigma_w/2);
llr(2:2:end) = -2*imag(detected)/(sigma_w/2);
llr_deinter = deinterleaver(llr);
decoded = LDPC_decoder(llr_deinter).';
Pbit_AWGN_code(i) = length(find(x(1:length(decoded))~=decoded))/length(decoded);
end
toc
```

```
clc; close all; clear global; clearvars;
  load('Input_symbols.mat');
  symbols = bitmap(x).';
  SNR_vect = 0:14;
  sigma_a = 2;
  Pbit_AWGN_uncode = zeros(length(SNR_vect),1);
10 parfor i=1:length(SNR_vect)
          snr_db = SNR_vect(i);
snr_lin = 10^(snr_db/10);
           sigma_w = sigma_a / snr_lin;
           w = wgn(length(symbols),1, 10*log10(sigma_w), 'complex');
          r_c = symbols + w;
           r_c_prime = r_c;
           detected = r_c_prime;
           hard_d = zeros(length(detected),1);
           for u=1:length(detected)
20
                   hard_d(u) = QPSK_detector(detected(u));
           end
          hard_bits = ibmap(hard_d);
           Pbit_AWGN_uncode(i) = length(find(x(1:length(hard_bits))~=hard_bits))/length(hard_bits);
  end
25 toc
```

```
clc; close all; clear global; clearvars;
  load('Useful.mat', 'qc');
  load('Input_symbols.mat');
  SNR_vect = 0.8:0.05:1.4;
  sigma_a = 2;
                 % Input variance
  Pbit_OFDM_code = zeros(length(SNR_vect),1);
  M = 512;
                           % Sub-channels
  Npx = 18;
                           % Prefix length
  tic
  parfor i=1:length(SNR_vect)
          snr_db = SNR_vect(i);
          snr_lin = 10^(snr_db/10);
15
          [r_c, sigma_w, g_srrc, g, t0] = channel_OFDM(symbols_ak, snr_db, sigma_a, Npx);
          G = fft(g,512).
          a_matrix = reshape(r_c(1:end-mod(length(r_c),M+Npx)), M+Npx, []);
          rn = a_matrix(Npx+1:end,:);
          x_k = fft(rn);
20
          K_i = 1./G;
          y_matrix = x_k.*K_i;
          sigma_i = 0.5*sigma_w*M*abs(K_i).^2;
          llr_real = -2*real(y_matrix).*sigma_i.^(-1);
          llr_imag = -2*imag(y_matrix).*sigma_i.^(-1);
25
          llr_real_ar = reshape(llr_real, [], 1);
          llr_imag_ar = reshape(llr_imag, [], 1);
          llr = zeros(numel(llr_real) + numel(llr_imag), 1);
          llr(1:2:end) = llr_real_ar;
          11r(2:2:end) = 11r_imag_ar;
30
          llr = deinterleaver(llr);
          dec_bits = LDPC_decoder(llr).';
          nerr = length(find(x(1:length(dec_bits))~=dec_bits));
```

```
Pbit_OFDM_code(i) = nerr/length(dec_bits);
end
toc
```

```
clc; close all; clear global; clearvars;
  load('Input_symbols.mat');
  symbols = bitmap(x).';
  SNR_vect = 0:14;
  sigma_a = 2;
                  % Input variance
  Pbit_OFDM_uncode = zeros(length(SNR_vect),1);
  M = 512;
                           % Sub-channels
10 Npx = 18;
                           % Prefix length
  tic
  parfor k=1:length(SNR_vect)
          snr_db = SNR_vect(k);
          snr_lin = 10^(snr_db/10);
          [r_c, sigma_w, g_srrc, g, t0] = channel_OFDM(symbols, snr_db, sigma_a, Npx);
15
          G = fft(g,512).';
          a_matrix = reshape(r_c(1:end-mod(length(r_c),M+Npx)), M+Npx, []);
          rn = a_matrix(Npx+1:end,:);
          x_k = fft(rn);
          K_i = 1./G;
20
          y_matrix = x_k.*K_i;
          y = reshape(y_matrix,1,[]);
          hard_d = zeros(length(y),1);
          for i=1:length(hard_d)
                  hard_d(i) = QPSK_detector(y(i));
25
          end
          hard_bits = ibmap(hard_d);
          nerr = length(find(x(1:length(hard_bits))~=hard_bits));
          Pbit_OFDM_uncode(k) = nerr/length(hard_bits);
  end
30
  toc
```

```
clc; close all; clear global; clearvars;
  set(0, 'defaultTextInterpreter', 'latex')
                                                  % latex format
  %% Uncoded OFDM + DFE + AWGN
  SNR_vect_uncoded = 0:14;
  load('Pbit_DFE_uncoded.mat');
  load('Pbit_OFDM_uncoded.mat');
  load('Pbit_AWGN_uncoded.mat');
  figure();
10 semilogy(SNR_vect_uncoded, Pbit_DFE_uncode, 'g');
  hold on; grid on;
  semilogy(SNR_vect_uncoded, Pbit_OFDM_uncode, 'b');
semilogy(SNR_vect_uncoded, Pbit_AWGN_uncode, 'k');
  ylim([10<sup>-5</sup> 10<sup>-1</sup>]); xlim([4 14]);
15 xlabel('SNR'); ylabel('$P_{bit}$');
  legend('Uncoded DFE', 'Uncoded OFDM', 'Uncoded AWGN');
  set(legend, 'Interpreter', 'latex');
  %% Coded OFDM + DFE + AWGN
20 load('Pbit_DFE_coded.mat');
  load('Pbit_OFDM_coded.mat');
  load('Pbit_AWGN_coded.mat');
  figure();
25 semilogy (1.4:0.05:1.8, Pbit_DFE_code, 'g');
  hold on; grid on;
  semilogy(0.8:0.05:1.4, Pbit_OFDM_code, 'b');
  semilogy(0.35:0.05:1, Pbit_AWGN_code, 'k');
  ylim([10^-5 10^-1]); xlim([0 2]);
30 xlabel('SNR'); ylabel('$P_{bit}$');
  legend('Coded DFE', 'Coded OFDM', 'Coded AWGN');
  set(legend, 'Interpreter', 'latex');
```

```
function [outsym] = QPSK_detector(insym)
  if (real(insym)>0)
      if (imag(insym)>0)
          outsym = 1+1i;
      else
           outsym = 1-1i;
      end
  else
      if (imag(insym)>0)
          outsym = -1+1i;
      else
          outsym = -1-1i;
      end
15
  end
  end
```

```
function [c_opt, Jmin] = Adaptive_DFE(h, r_w, sigma_a, M1, M2, D)
  N1 = floor(length(h)/2);
  N2 = N1;
  padding = 60;
  hpad = padarray(h, padding);
  % Padding the noise correlation
  r_w_pad = padarray(r_w, padding);
    = zeros(M1 ,1);
  for i = 0 : M1-1
          p(i + 1) = sigma_a * conj(hpad(N1 + padding + 1 + D - i));
  end
  R = zeros(M1);
  for row = 0:(M1-1)
          for col = 0:(M1-1)
20
                   fsum = (hpad((padding + 1):(N1 + N2 + padding + 1))).' ...
                           * conj(hpad((padding + 1 - (row - col)):( N1 + N2 + ...
                           padding + 1 - (row - col))));
                   if M2 == 0
25
                           ssum=0;
                   else
                           ssum = (hpad((N1+padding+1+1+D-col):(N1+padding+1+M2+D-col))).' * ...
                                    conj((hpad((N1+padding+1+1+D-row):(N1+padding+1+M2+D-row))));
30
                   end
                   R(row + 1, col + 1) = sigma_a * (fsum - ssum) + r_w_pad(padding + 1 ...
                           + row - col + (floor(length(r_w) / 2 )));
           end
35 end
  c_{opt} = R \setminus p;
  temp2 = zeros(M1, 1);
40
  for 1 = 0:M1-1
          temp2(1 + 1) = c_opt(1 + 1) * hpad(N1 + padding + 1 + D - 1);
45 Jmin = 10*log10(sigma_a * (1 - sum(temp2)));
```

```
function [decisions] = equalization_DFE(x, c, b, M1, M2, D)
%EQUALIZATION for DFE

y = zeros(length(x) + D , 1); % output of ff filter
detected = zeros(length(x) + D, 1); % output of td
```

```
for k = 0:length(x) - 1 + D
      if (k < M1 - 1)
          xconv = [flipud(x(1:k+1)); zeros(M1 - k - 1, 1)];
      elseif k > length(x)-1 & k & k < length(x) - 1 + M1
10
          xconv = [zeros(k-length(x)+1, 1); flipud(x(end - M1 + 1 + k - length(x) + 1:end))];
      elseif k \ge length(x) - 1 + M1 \% just in case D is greater than M1
          xconv = zeros(M1, 1);
          xconv = flipud(x(k-M1+1 + 1:k + 1));
      end
      if (k \le M2)
          a_old = [flipud(detected(1:k)); zeros(M2 - k, 1)];
           a_old = flipud(detected(k-M2+1:k));
20
      y(k+1) = c.'*xconv;
      detected(k+1) = y(k+1) + b.**a_old;
25
  end
  % scatterplot(y)
  decisions = detected(D + 1:end);
```

```
function [pn] = PNSeq(L)
  r = log2(L+1);
  pn = zeros(L,1);
  % Initial conditions (set to one, arbitrary)
  % Must not be ALL zeros
  pn(1:r) = ones(1,r).;
10 for l=r+1:L
      switch r
          case 1
              pn(1) = pn(1-1);
           case 2
              pn(1) = xor(pn(1-1), pn(1-2));
           case 3
              pn(1) = xor(pn(1-2), pn(1-3));
           case 4
              pn(1) = xor(pn(1-3), pn(1-4));
           case 5
              pn(1) = xor(pn(1-3), pn(1-5));
           case 6
              pn(1) = xor(pn(1-5), pn(1-6));
           case 7
25
              pn(1) = xor(pn(1-6), pn(1-7));
           case 8
              pn(1) = xor(xor(pn(1-2),pn(1-3)),xor(pn(1-4),pn(1-8)));
           case 9
              pn(1) = xor(pn(1-5), pn(1-9));
30
           case 10
              pn(1) = xor(pn(1-7), pn(1-10));
           case 15
              pn(1) = xor(pn(1-14), pn(1-15));
           case 20
              pn(1) = xor(pn(1-17), pn(1-20));
35
      end
  end
  end
```

```
function [coded] = LDPC_encoder(input, h, N)

packets_num = N;
packet_length = 32400;
coded = zeros(length(input)*2,1);
for i=0:packets_num-1
```

```
uncoded = input(i*packet_length+1:i*packet_length+packet_length);
    coded(2 * i * packet_length + 1:2 * i * packet_length + 2 * packet_length) = step(h, uncoded);
end

coded = reshape(coded,length(input)*2,1);
end
```

```
function [output] = bitmap(input)
  % Check if the input array has even length
  L = length(input);
  output = zeros(1,L);
  for idx = 1:2:L-1
     if (isequal(input(idx:idx+1), [0; 0] ))
         output(idx) = -1-1i;
10
     elseif (isequal(input(idx:idx+1), [1; 0] ))
         output(idx) = 1-1i;
     elseif (isequal(input(idx:idx+1), [0; 1] ))
         output(idx) = -1+1i;
      elseif (isequal(input(idx:idx+1), [1; 1] ))
         output(idx) = +1+1i;
  end
  output = output(1:2:end);
  end
```

```
function [output] = ibmap(input)
      L = length(input);
       output = zeros(2*L,1);
       \mbox{\ensuremath{\%}} Map each couple of values to the corresponding symbol
       \% The real part gives the bit
           for k = 1:2:length(output)-1
           symbol = input((k+1)/2);
                    if (real(symbol) == 1)
              b2k = 1;
           else
10
              b2k = 0;
                    end
                    if (imag(symbol) == 1)
               b2k1 = 1;
15
           else
                b2k1 = 0;
                    output(k) = b2k;
                    output(k+1) = b2k1;
20
           end
  end
```

```
function [output, sigma_w, qc] = channel_sim(x, snr, sigma_a)
```

```
T = 1;
  Tc = T/4;
_{5} Q = T/Tc;
  alpha = 0.67;
  beta = 0.7424;
10 snr_db = snr;
  snr_lin = 10^(snr_db/10);
  qc_num = [0 0 0 0 0 beta];
  qc_denom = [1 -alpha];
15 qc = impz(qc_num, qc_denom);
  qc = [0; 0; 0; 0; qc(qc \ge max(qc)*10^(-2))];
  E_qc = sum(qc.^2);
  sigma_w = sigma_a * E_qc / snr_lin;
  a_prime = upsample(x,Q);
  s_c = filter(qc_num, qc_denom, a_prime);
  % to be added later
  % noise = wgn(length(s_c),1,sigma_w,'complex');
  r_c = s_c;
  output = r_c;
```

```
function[output, sigma_w, g_srrc, tot_ds, t0, tot] = channel_OFDM(input, snr, sigma_a, Npx)
  snr_db = snr;
  snr_lin = 10^(snr_db/10);
  M = 512;
  a_pad = [input; ones(M - mod(length(input), M), 1) * (1+1i)];
  a_matrix = reshape(a_pad, M, []); % Should mantain columnwise order
10 A_matrix = ifft(a_matrix);
  A_{matrix} = [A_{matrix}(M-Npx+1:M, :); A_{matrix}];
  r = reshape(A_matrix, [], 1);
_{15} Q = 4;
  in_upsampled = upsample(r, Q);
  % Square-root raised cosine
  ro = 0.0625;
20 span = 30;
  sps = 4;
  g_srrc = rcosdesign(ro, span, sps, 'sqrt');
  in_after_srrc = filter(g_srrc,1,in_upsampled);
  alpha = 0.67;
  beta = 0.7424;
  qc_num = [0 0 0 0 0 beta];
  qc_denom = [1 -alpha];
30 qc = impz(qc_num, qc_denom);
  qc = [0; 0; 0; 0; qc(qc \ge max(qc)*10^(-2))];
  in_after_qc = filter(qc,1,in_after_srrc);
  all_ch = conv(g_srrc, conv(g_srrc,qc));
_{35} E_tot = sum(conv(g_srrc,qc).^2);
  sigma_w = sigma_a/M * E_tot / snr_lin;
  in_after_qc = in_after_qc + wgn(length(in_after_qc),1,10*log10(sigma_w),'complex');
  tot = all_ch(abs(all_ch)>=(max(abs(all_ch))*1e-2));
40 tot_ds = downsample(tot, 4);
  in_after_srrc = filter(g_srrc, 1, in_after_qc);
```

```
t0 = find(in_after_srrc==max(in_after_srrc));

in_after_srrc = in_after_srrc(t0:end);
in_after_srrc = downsample(in_after_srrc,4);

output = in_after_srrc;
end
```

```
clc; close all; clear global; clearvars;

% Initial bits
L = 2^20-1;
x = [PNSeq(L); PNSeq(L)];
% Matlab LDPC Encoder
H = comm.LDPCEncoder();

sstep = 32400;
numbits = floor(length(x) / sstep) * sstep;
x = x(1:numbits + 54);
N = floor(length(x) / sstep);
encoded = LDPC_encoder(x,H,N);

interleaved = interleaver(encoded);
symbols_ak = bitmap(interleaved.').';
save('Input_symbols.mat', 'symbols_ak', 'x', 'H', 'sstep');
```

```
function [deinterleaved_bits] = deinterleaver(bits)
  % This function receives a sequence of bits and unscrambles it
  deinterleaved_bits = zeros(1,length(bits));
_{5}| % The deinterleaver is just an interleaver with rows and cols switched
  rows = 42;
  columns = 42;
  % We work with a rowsxcolumns matrix
10 for matrix = 0:(length(bits)/(rows*columns) - 1)
      curr_matrix = matrix * rows * columns;
      for col = 0:(columns-1)
          deinterleaved_bits(curr_matrix + col * rows + 1 : curr_matrix + col * rows + rows) = ...
              bits(curr_matrix + col + 1 : columns : curr_matrix + col + columns * rows);
      end
15
  end
  end
```