

Lab-4

1.a)

ROC, Causality, and Stability :

Here $N(z)=1$ and

$$H(z)=N(z)/D(z)$$

- N – number (a positive integer) of unique ROC possible for this $H(z)$
- ROC – $N \times 2$ matrix (of non-negative real numbers) with each row $[r1, r2]$ indicating an ROC of the form $r1 < |z| < r2$
- C – length N binary vector (1 if the corresponding system is causal, else 0)
- S – length N binary vector (1 if the corresponding system is stable, else 0)

Code for Function:

```
*function [N, ROC, C, S] = roc_cs(p2)
    N = length(p2)+1;
    ROC = zeros(N,2);
    C = zeros(N,1);
    S = zeros(N,1);
    C(N) = 1;
    for i = 1:N
        if i ~= N
            ROC(i,2) = abs(p2(i));
        end
        if i ~= 1
            ROC(i,1) = abs(p2(i-1));
        end
    end
    ROC(N,2) = Inf;
    v = 0;
    x=0;
    for j = 1:length(p2)
        if abs(p2(j)) == 1
            v = 1;
```

```

    end
end

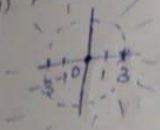
for i = 1:N-1
    if v ~= 1 && p2(i) < 1 && p2(i+1) > 1
        S(i+1) = 1;
        x = 1;
        break
    end
    if v~=1 && p2(i)>1
        S(i) = 1;
        x=1;
        break
    end
end
end
if v ~=1 && x ~= 1
    S(N) = 1;
end
end
1.b)

```

Lab-4
Z-Transform

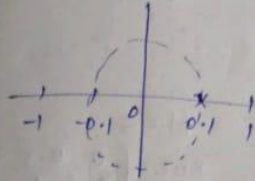
7.1) ROC, causality and Stability

b) (i) $P=3$



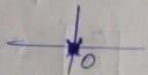
$$N=2 \quad ROC = \begin{bmatrix} 0 & 3 \\ 3 & \text{Inf} \end{bmatrix} \quad C = \begin{bmatrix} 0 \\ 1 \end{bmatrix} \quad S = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

(ii) $P=0.1$



$$N=2 \quad ROC = \begin{bmatrix} 0 & 0.1 \\ 0.1 & \text{Inf} \end{bmatrix} \quad C = \begin{bmatrix} 0 \\ 1 \end{bmatrix} \quad S = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

(iii) $P=0$



$$N=1 \quad ROC = \begin{bmatrix} 0 & \text{Inf} \end{bmatrix} \quad C = \begin{bmatrix} 1 \end{bmatrix} \quad S = \begin{bmatrix} 1 \end{bmatrix}$$

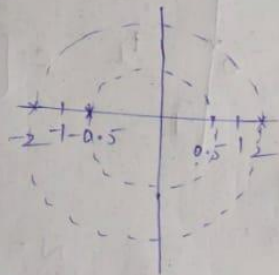
(iv) $P=[0, 0.5]$



$N=2$

$$ROC = \begin{bmatrix} 0 & 0.5 \\ 0.5 & \text{Inf} \end{bmatrix} \quad C = \begin{bmatrix} 0 \\ 1 \end{bmatrix} \quad S = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

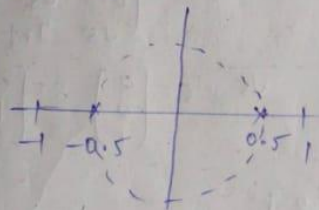
(v) $P = [2, -0.5]$
 $= [-0.5, 2]$



$N=3$

$$ROC = \begin{bmatrix} 0 & 0.5 \\ 0.5 & 2 \\ 2 & \text{Inf} \end{bmatrix} \quad S = \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix} \quad C = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$

(vi) $P=[0.5, 0.5]$



$N=2$

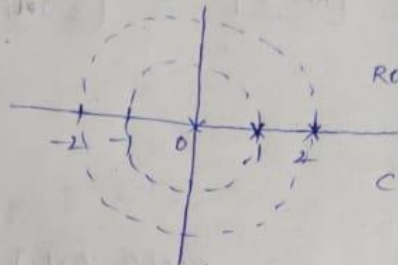
$$ROC = \begin{bmatrix} 0 & 0.5 \\ 0.5 & \text{Inf} \end{bmatrix} \quad C = \begin{bmatrix} 0 \\ 1 \end{bmatrix} \quad S = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

(vii) $P = [2, 2, 2]$
 $= [2]$



$N=2$
 $RDC = \begin{bmatrix} 0 & 2 \\ 2 & \text{Inf} \end{bmatrix}$ $C = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$ $S = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$

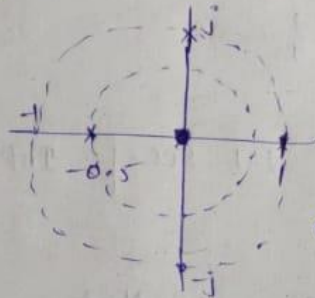
(viii) $P = [0, 1, 2]$



$N=3$
 $RDC = \begin{bmatrix} 0 & 1 & 2 \\ 1 & 2 & \text{Inf} \\ 2 & \text{Inf} & \text{Inf} \end{bmatrix}$
 $C = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$ $S = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}$

(ix) $P = [-0.5, j]$

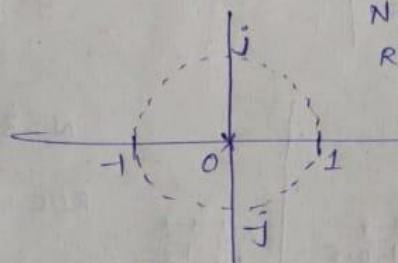
$|j|=1$



$N=3$
 $RDC = \begin{bmatrix} 0 & 0.5 & 1 \\ 0.5 & 1 & \text{Inf} \\ 1 & \text{Inf} & \text{Inf} \end{bmatrix}$
 $C = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$ $S = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}$

(x) $P = [0, j, -j]$

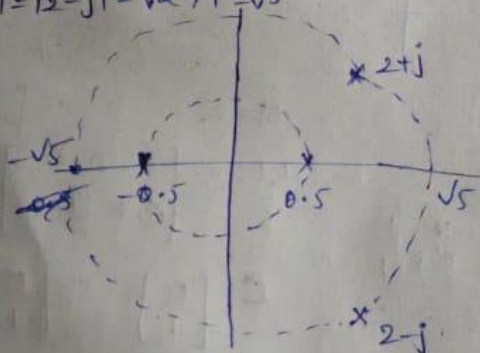
$|j|=|-j|=1$



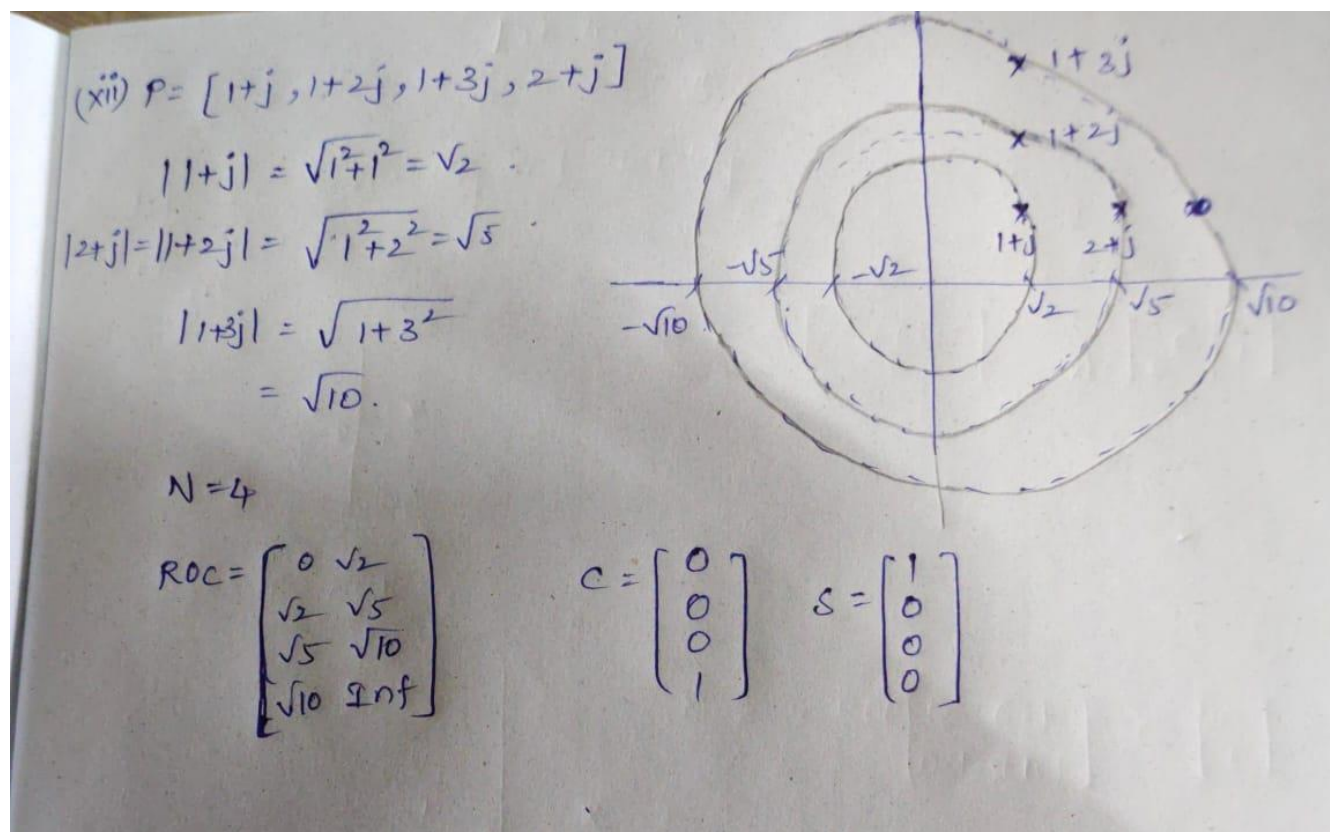
$N=2$
 $RDC = \begin{bmatrix} 0 & 1 \\ 1 & \text{Inf} \end{bmatrix}$
 $C = \begin{bmatrix} 0 \\ 1 \end{bmatrix}$ $S = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$

(xi) $P = [0.5, -0.5, 2+j, 2-j]$

$|2+j|=|2-j|=\sqrt{2^2+1}=\sqrt{5}$



$N=3$
 $RDC = \begin{bmatrix} 0 & 0.5 & \sqrt{5} \\ 0.5 & \sqrt{5} & \text{Inf} \\ \sqrt{5} & \text{Inf} & \text{Inf} \end{bmatrix}$
 $C = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$ $S = \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix}$



Verification with MATLAB:

When $p=3$

2

0 3

3 Inf

0

1

1

0

When $p=0.1$

2
0 0.1000
0.1000 Inf
0
1
0
1

When $p=0$

1
0 Inf
1
1

When $p=[0,0.5]$

2
0 0.5000
0.5000 Inf
0
1
0
1

When $p=[2,-0.5]$

3
0 0.5000
0.5000 2.0000
2.0000 Inf
0
0
1
0
1
0

When $p=[-0.5,0.5]$

2		
	0	0.5000
0.5000		Inf
0		
1		
0		
1		

When $p=[2,2,2]$

2		
0	2	
2		Inf
0		
1		
1		
0		

When $p=[0,1,2]$

$\frac{1}{3}$		
0	1	
1		2
2		Inf
0		
0		
1		
0		
0		
0		

When $p=[-0.5,j]$

3

0	0.5000
0.5000	1.0000
1.0000	Inf

0

0

1

0

0

0

When $p=[0,-j,j]$

2

0	1
1	Inf

0

1

0

0

When $p=[0.5,-0.5,2+j,2-j]$

3

0	0.5000
0.5000	2.2361
2.2361	Inf

0

0

1

0

1

0

When $p=[1+j,1+2j,1+3j,2+j]$

4

0	1.4142
1.4142	2.2361
2.2361	3.1623
3.1623	Inf

0

0

0

1

1

0

0

0

Code for script:

```
p = [1+1i,1+2i,1+3i,2+1i];  
p1 = unique(abs(p));
```

```
p2 = p1;
```

```
for i = 1 : length(p1)  
    if p1(i) == 0  
        k = 1;  
        p2 = p1(p1 ~= 0);  
        break  
    end  
end  
[N, ROC, C, S] = roc_cs(p2);  
disp(N);  
disp(ROC);  
disp(C);  
disp(S);
```

****Here we are getting the same output in the case of the calculation part and in the case of MATLAB part.**

2.

Given a LTI system

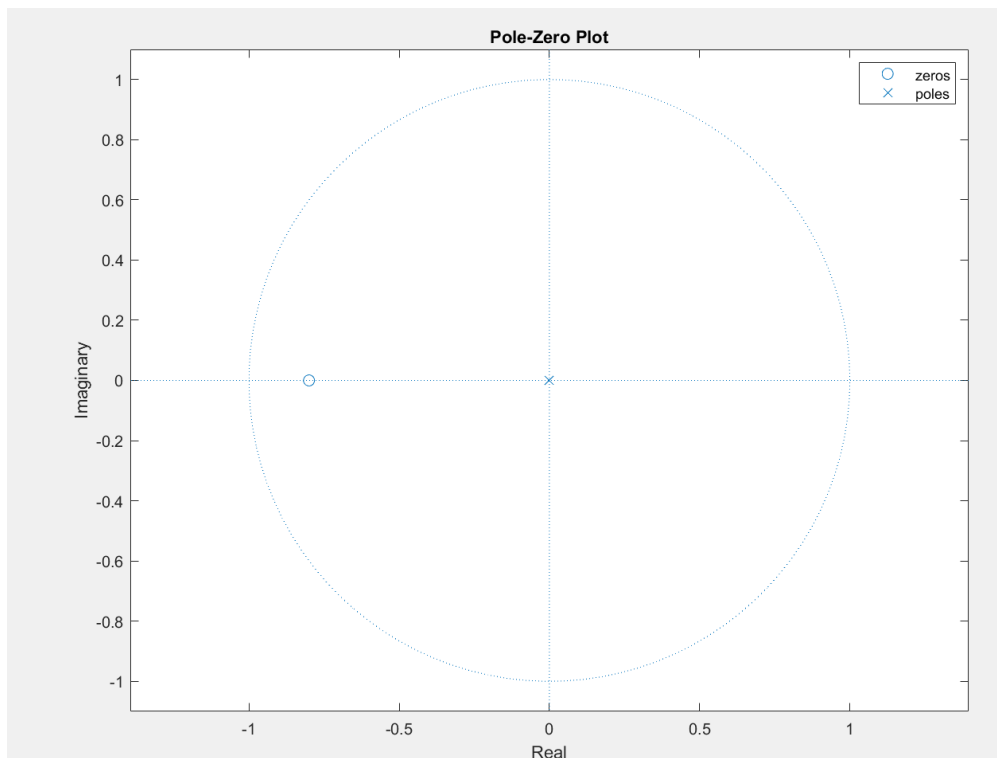
$$H(z) = z/(z+p) \quad \text{where } p \text{ belongs to } (-1,1)$$

a) $p=0.8$

Code in MATLAB:

```
p=0.8;  
d=[1,p];  
figure;  
zplane(d,1)  
title('Pole-Zero Plot');  
xlabel('Real');  
ylabel('Imaginary');  
legend('zeros','poles');
```

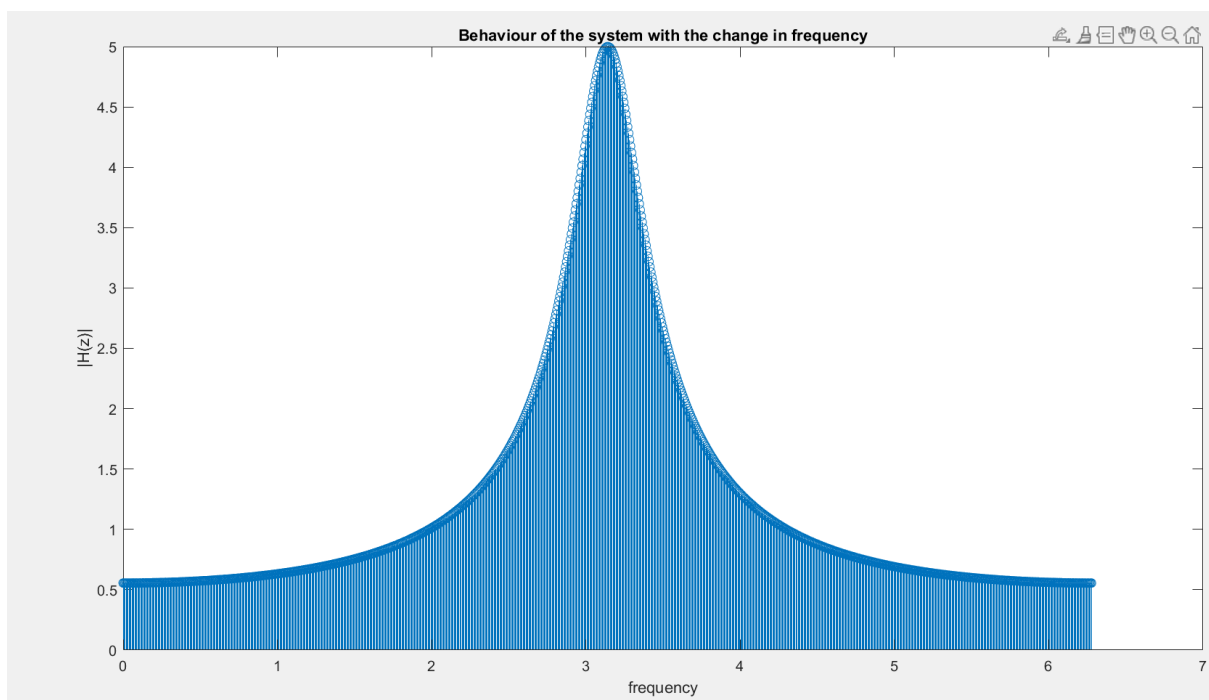
Plot:



b)

Code :

```
n=1001;
p=0.8;
d=[1,p];
[y,x]=freqz(1,d,n,"whole");
figure;
stem(x,abs(y));
title('Behaviour of the system with the change in frequency')
xlabel('frequency')
ylabel('|H(z)|')
```



c)

Code :

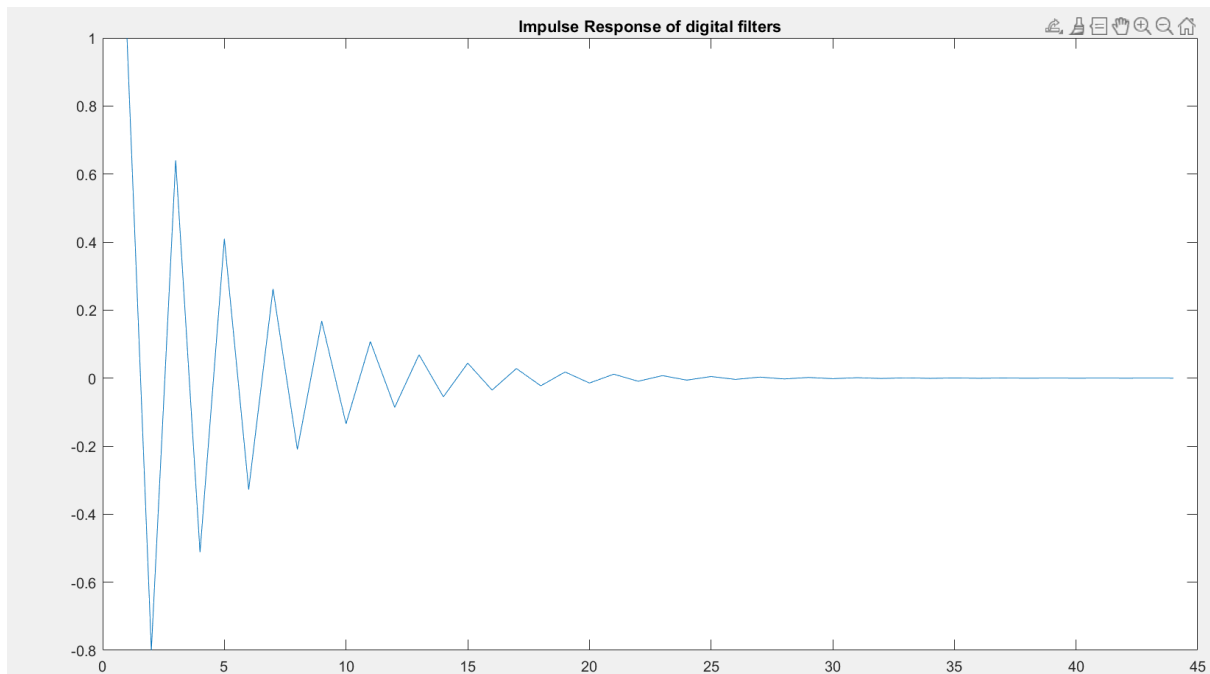
```
n=1001;
p=0.8;
d=[1,p];
[y,x]=freqz(1,d,n,"whole");
figure;
stem(x,abs(y));
```

```

title('Behaviour of the system with the change in frequency')
xlabel('frequency')
ylabel('|H(z)|')

```

Plot:



From the Plot we can say that it is an IIR filter.

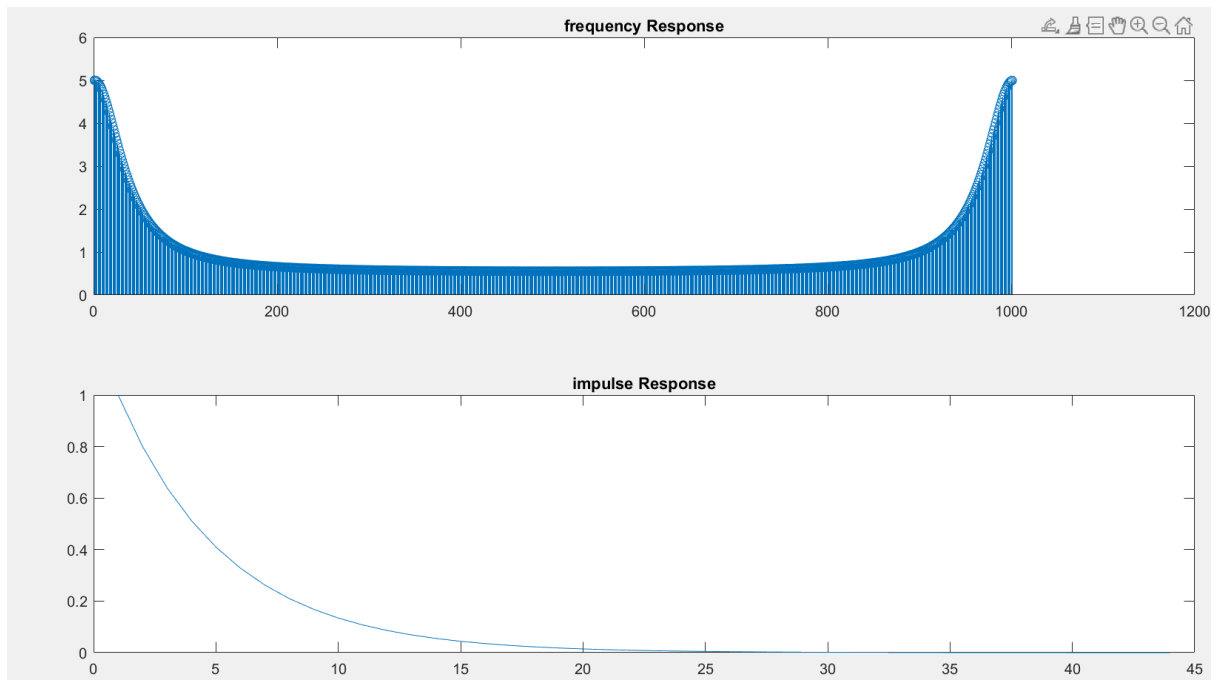
(from the definition of FIR and IIR filters

We know that the IIR is a type of filter that generates impulse response of infinite duration for a dynamic system. And the FIR filter provides an impulse response of a finite period.) here in the plot it is generating for the infinite duration .

In this case, the system is a first-order system, so there is only one possible impulse response

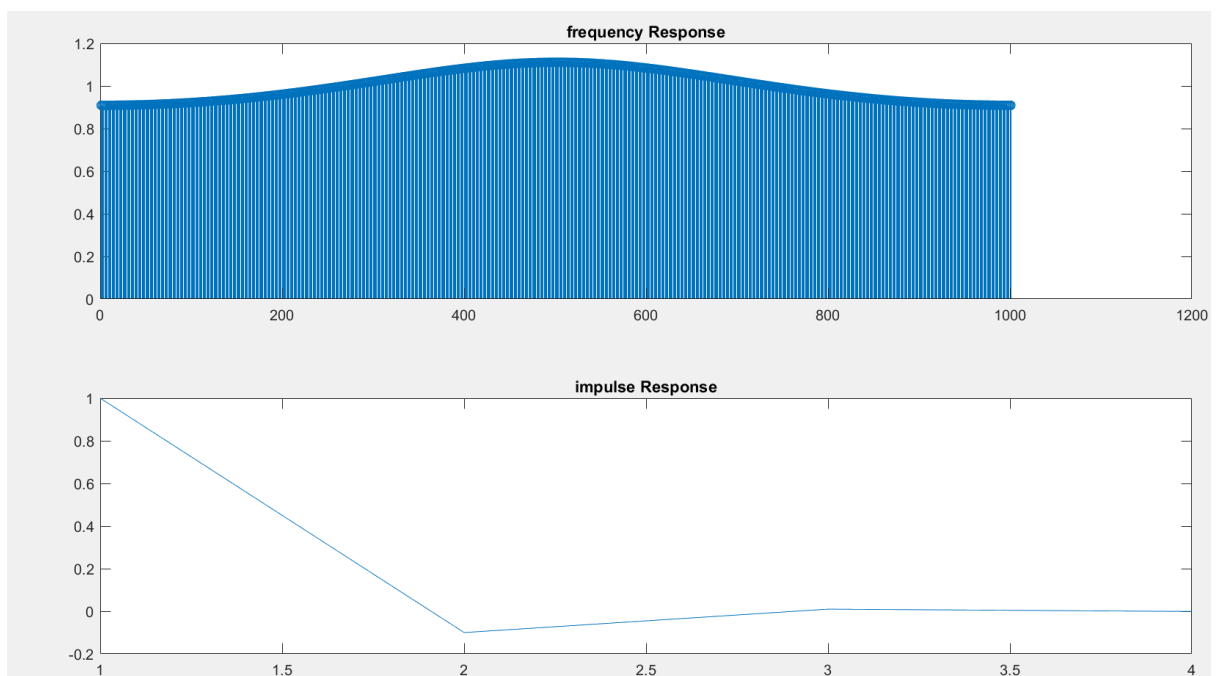
d)

$p = -0.8$



When you repeat the analysis for $p = -0.8$ in the magnitude response plot (Part d), you may see a response that resembles a high-pass or band-reject filter, depending on the exact value of 'p'. The response may not have a peak, but it should be stable.

P=0.1



For $p = 0.1$, the impulse response is likely to decay smoothly without significant oscillations, indicating a well-behaved, stable system.

Code:

```
n=1001;
p=0.1;
d=[1,p];
y1=freqz(1,d,n,"whole");
y2=impz(1,d);

figure
subplot(2,1,1)
stem(y1)
title('frequency Response')

subplot(2,1,2)
plot(y2)
title('impulse Response')
```

e)

Here the system function is

$$H(z)=(z-p^{-1})/(z-p) \text{ and}$$

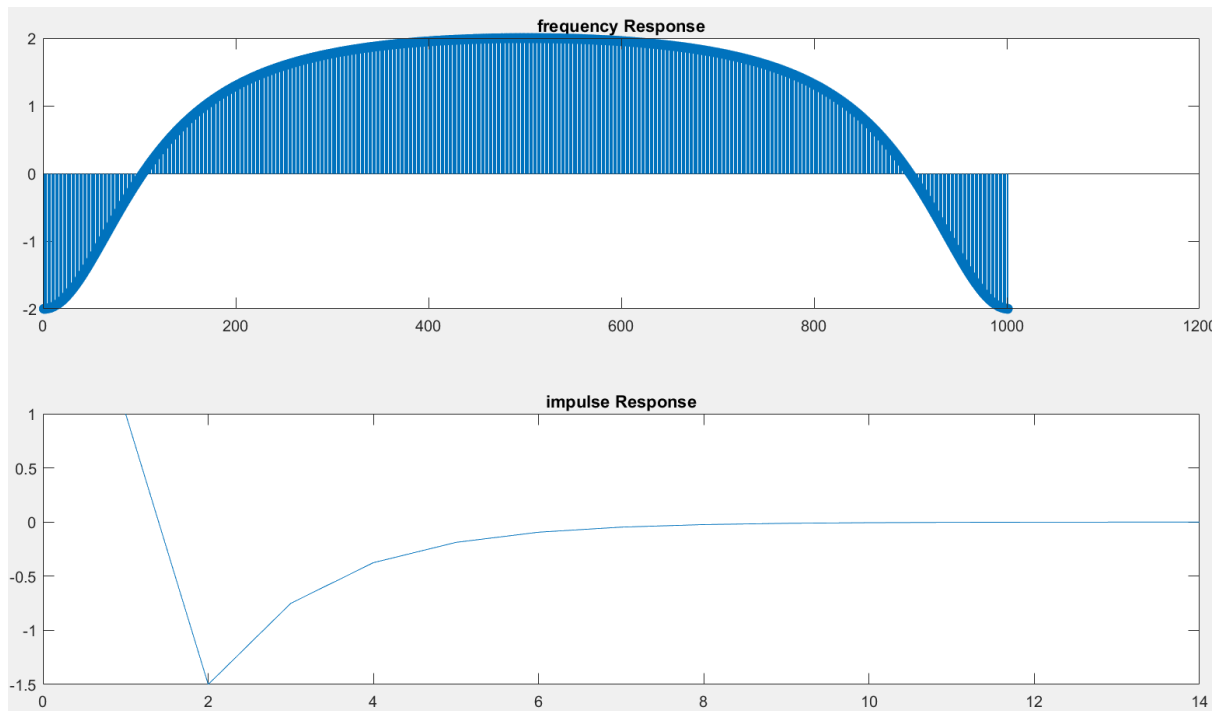
$$P=0.5$$

Code:

```
p=0.5;
N=1001;
n=[1,-1/p];
d=[1,-p];
y1=freqz(n,d,N,"whole");
y2=impz(n,d);
figure
subplot(2,1,1)
stem(y1,"filled")
title('frequency Response')

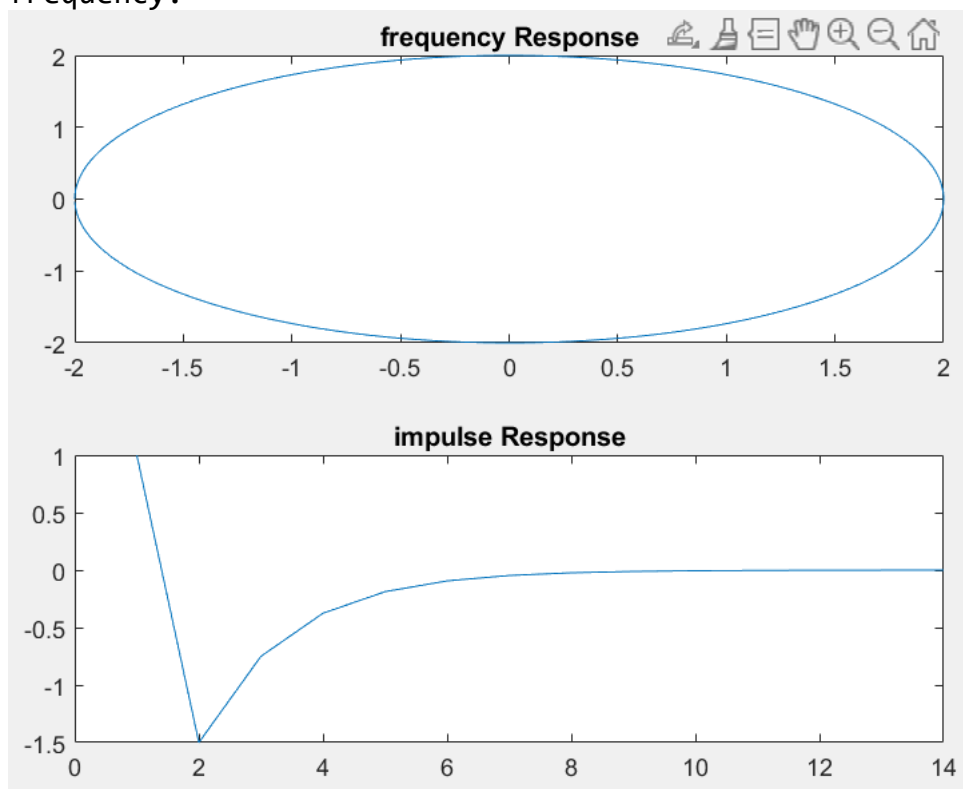
subplot(2,1,2)
plot(y2)
title('impulse Response')
```

Plot:



(Or)

The graphs are between magnitude of frequency response and frequency.



Here it is representing the IIR filter as per the plot.

3) LTI system with complex poles

Here the transfer function is

$H(z) = \frac{z^2 - (2\cos\theta)z + 1}{z^2 - (2r\cos\theta)z + r^2}$ where r belongs to $(0,1)$ and θ belongs to $[0, \pi]$

a)

Here

The Numerator is $z^2 - (2\cos\theta)z + 1$

and the denominator is $z^2 - (2r\cos\theta)z + r^2$

to solve the equation for the poles we need to equate the denominator to the zero. and for finding the zeros we need to equate the numerator to zero.

By solving we can observe the zeros and poles as

For the numerator, we can use the quadratic formula:

$$z_{\text{zero}} = \frac{2\cos(\theta) \pm \sqrt{4\cos^2(\theta) - 4}}{2} = \cos(\theta) \pm j\sqrt{1 - \cos^2(\theta)}$$

For the denominator, we can also use the quadratic formula:

$$z_{\text{pole}} = \frac{2r\cos(\theta) \pm \sqrt{4r^2\cos^2(\theta) - 4r^2}}{2} = r\cos(\theta) \pm jr\sin(\theta)$$

Code:

```
r_values = [0.25, 0.5, 0.75, 0.99];  
t_values = [pi/4, pi/2, 2*pi/3, 3*pi/4];
```

```
for i = 1:length(r_values)  
    for j = 1:length(t_values)  
        r = r_values(i);  
        t = t_values(j);  
  
        n = [1, -2*cos(t), 1];  
        d = [1, -2*r*cos(t), r*r];  
  
        figure;  
        subplot(2, 1, 1);
```



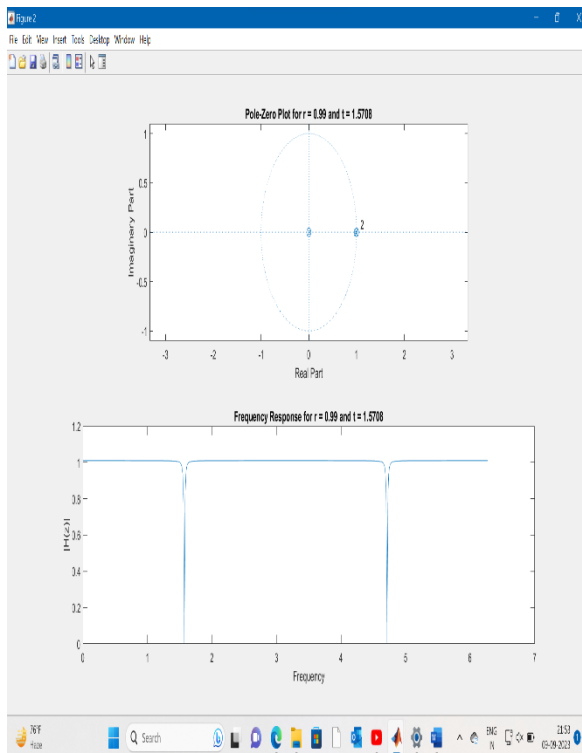
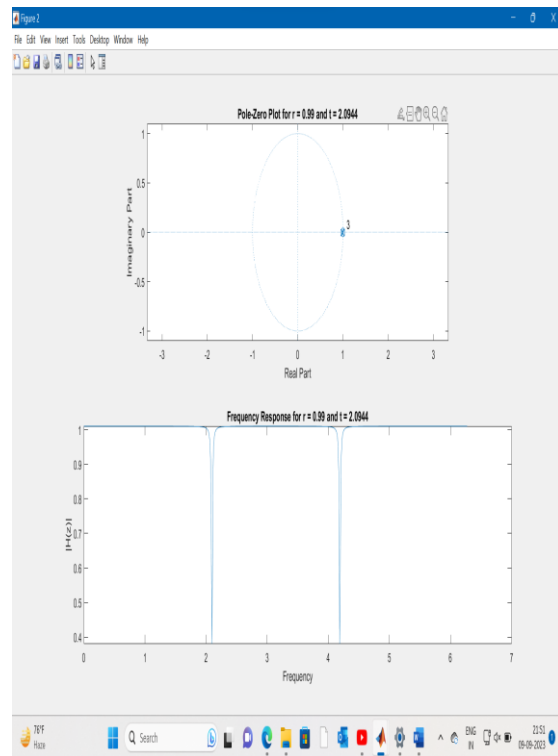
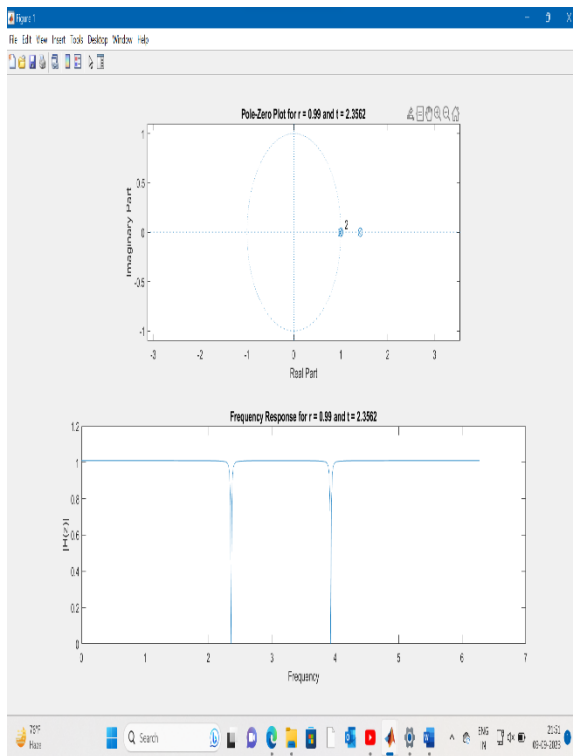
```

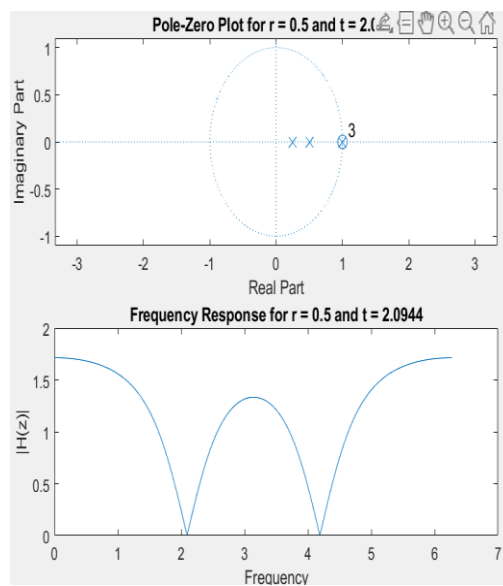
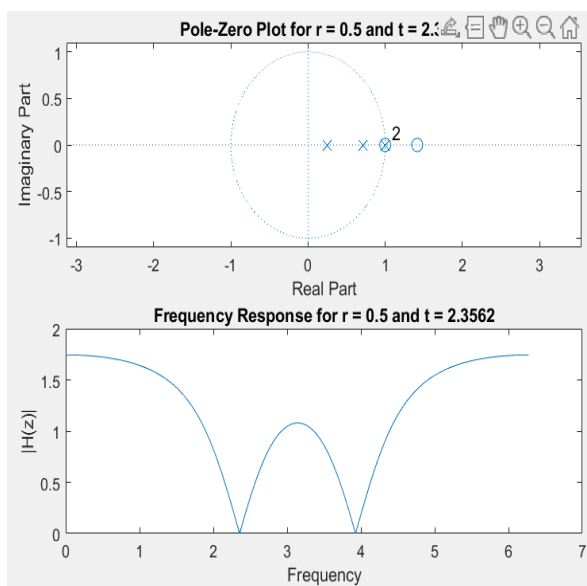
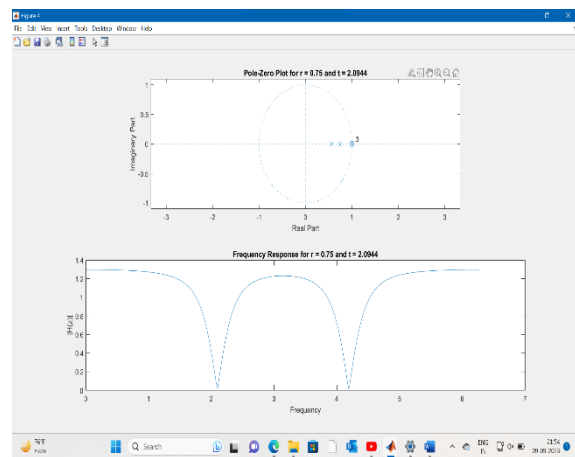
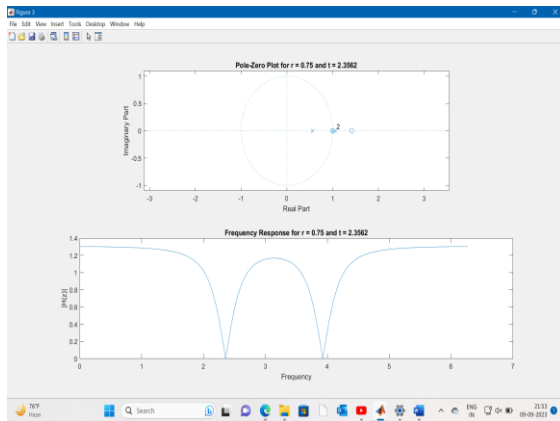
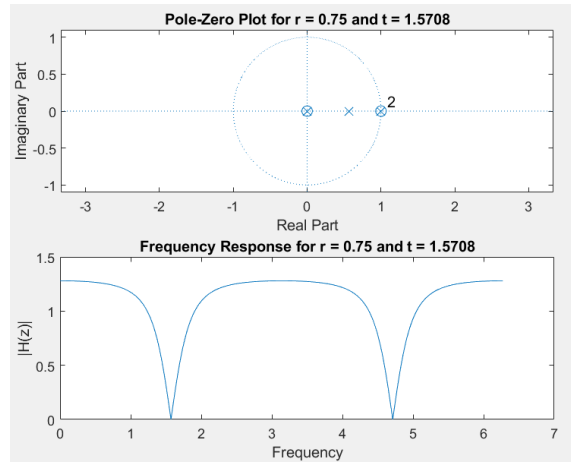
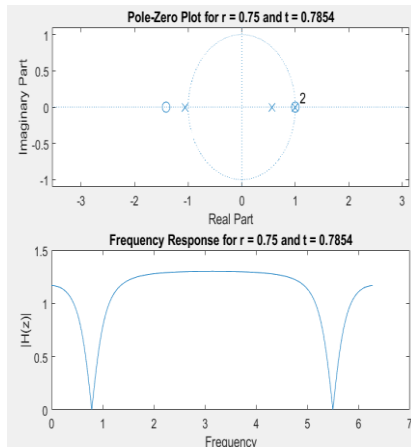
        zplaneplot(n, d);
        title(['Pole-Zero Plot for r = ', num2str(r), ' and t = ',
num2str(t)]);

        [y, x] = freqz(n, d, 'whole');
        subplot(2, 1, 2);
        plot(x, abs(y));
        title(['Frequency Response for r = ', num2str(r), ' and t =
', num2str(t)]);
        xlabel('Frequency');
        ylabel('|H(z)|');
    end
end

```

Plot:





3.b)

The system can be both causal and stable if all poles of the transfer function are located inside the unit circle ($|z| < 1$) in the complex plane. To find the ROC, we need to know the values of z for which the $H(z)$ converges.

Here the denominator of $H(z)$ is a quadratic polynomial, the ROC will be a ring-shaped region in the complex plane.

The ROC will include the unit circle ($|z| < 1$) and exclude the area outside the unit circle.

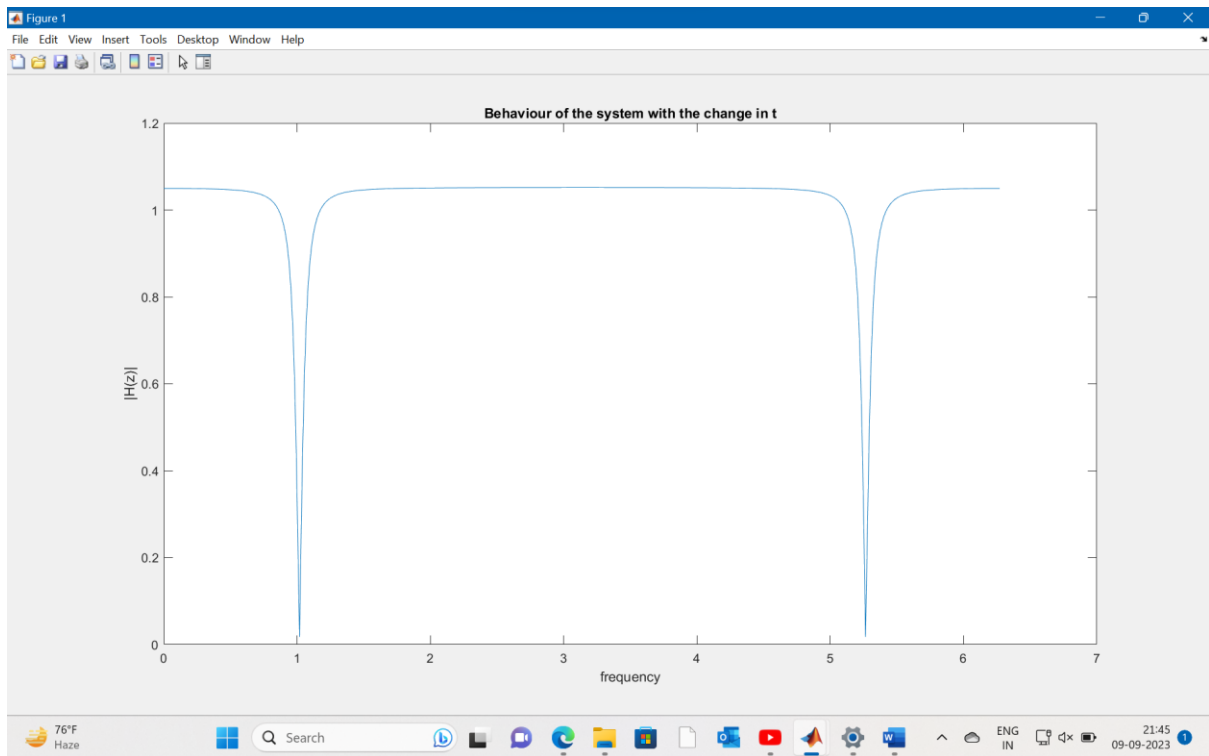
3.c)

Code:

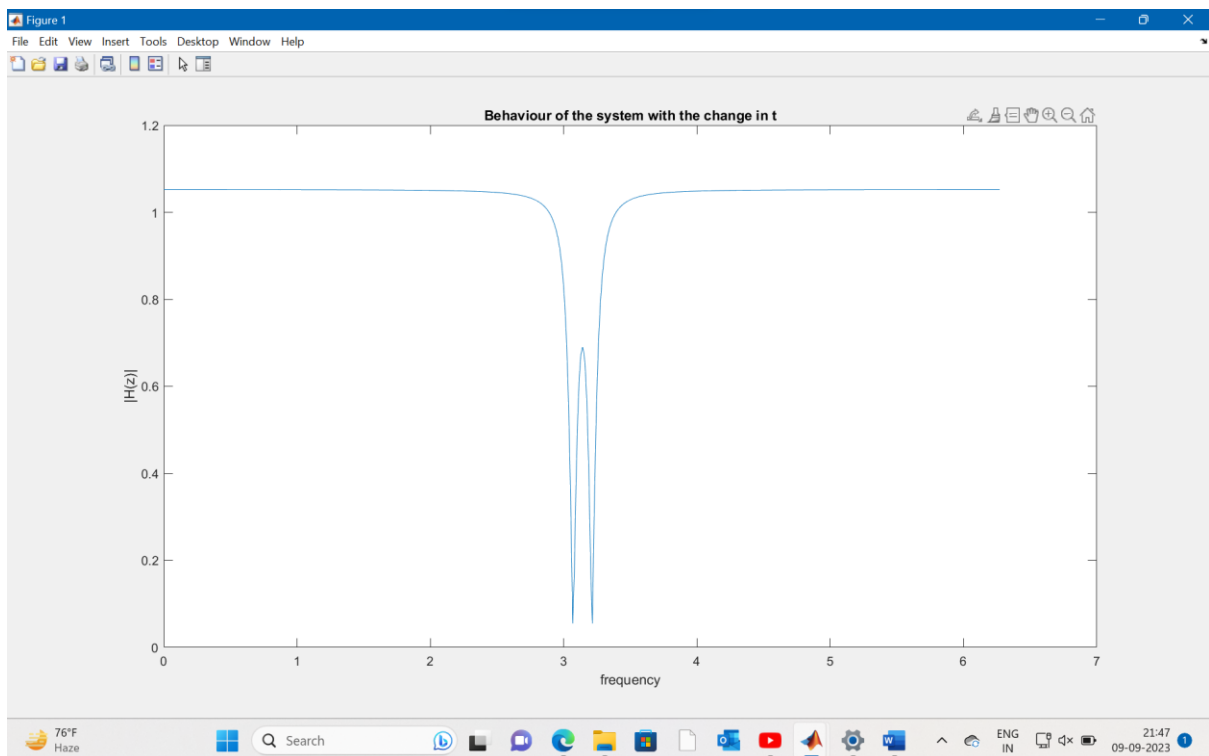
```
r=0.95;
t=60;
n=[1, -2*cos(t), 1];
d=[1, -2*r*cos(t), r*r];
[y,x]=freqz(n,d, 'whole');

figure;
plot(x,abs(y));
title('Behaviour of the system with the change in t')
xlabel('frequency')
ylabel('|H(z)|')
```

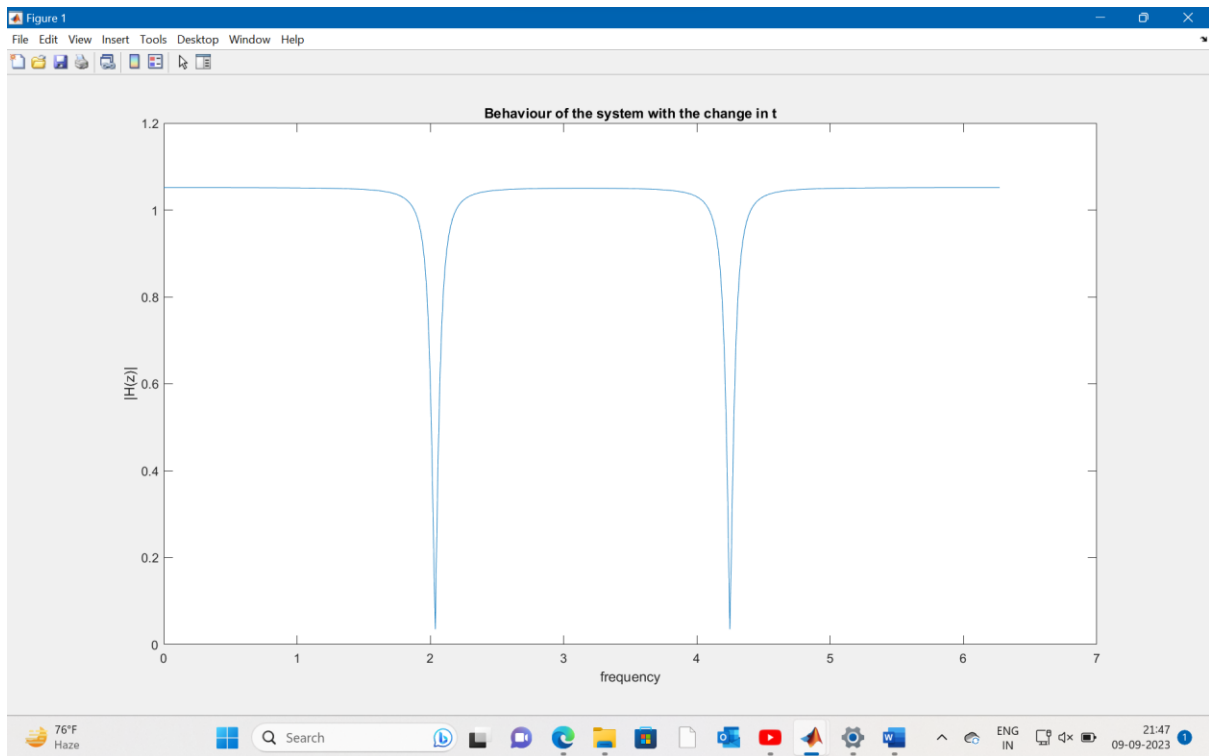
When $\theta = \pi/4$



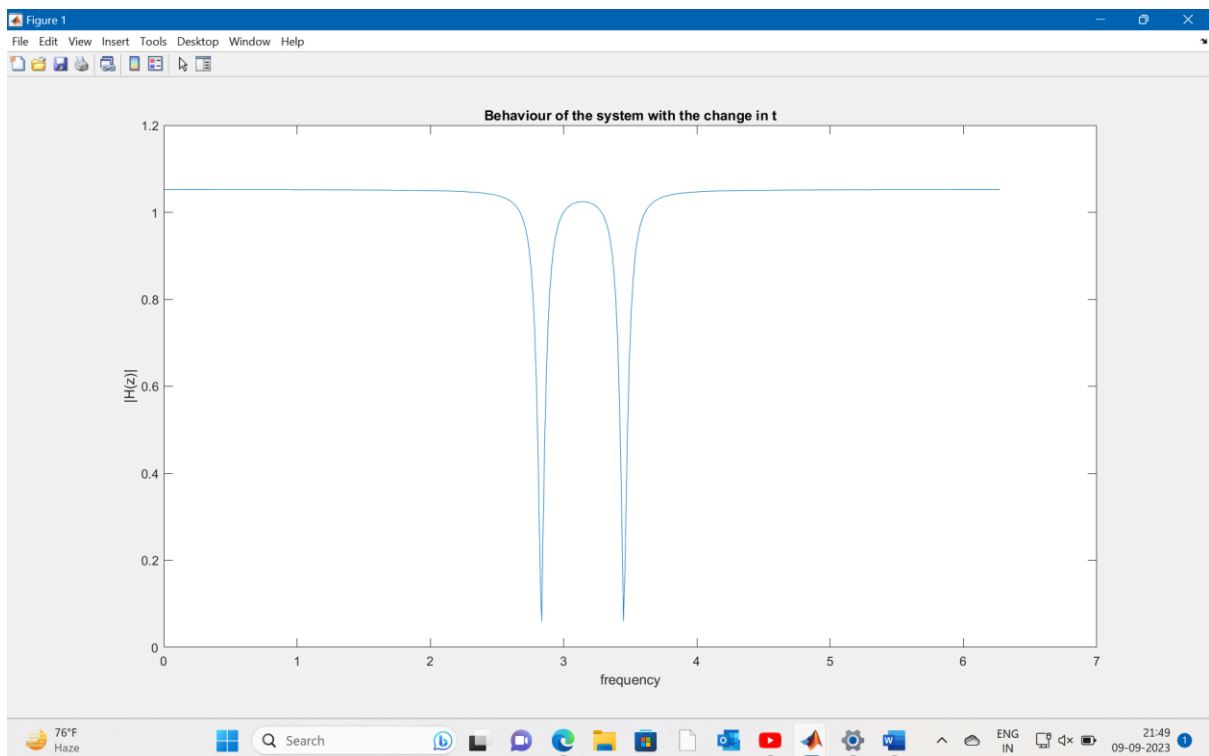
When $\theta=3*\pi/4$



When $\theta=\pi/2$



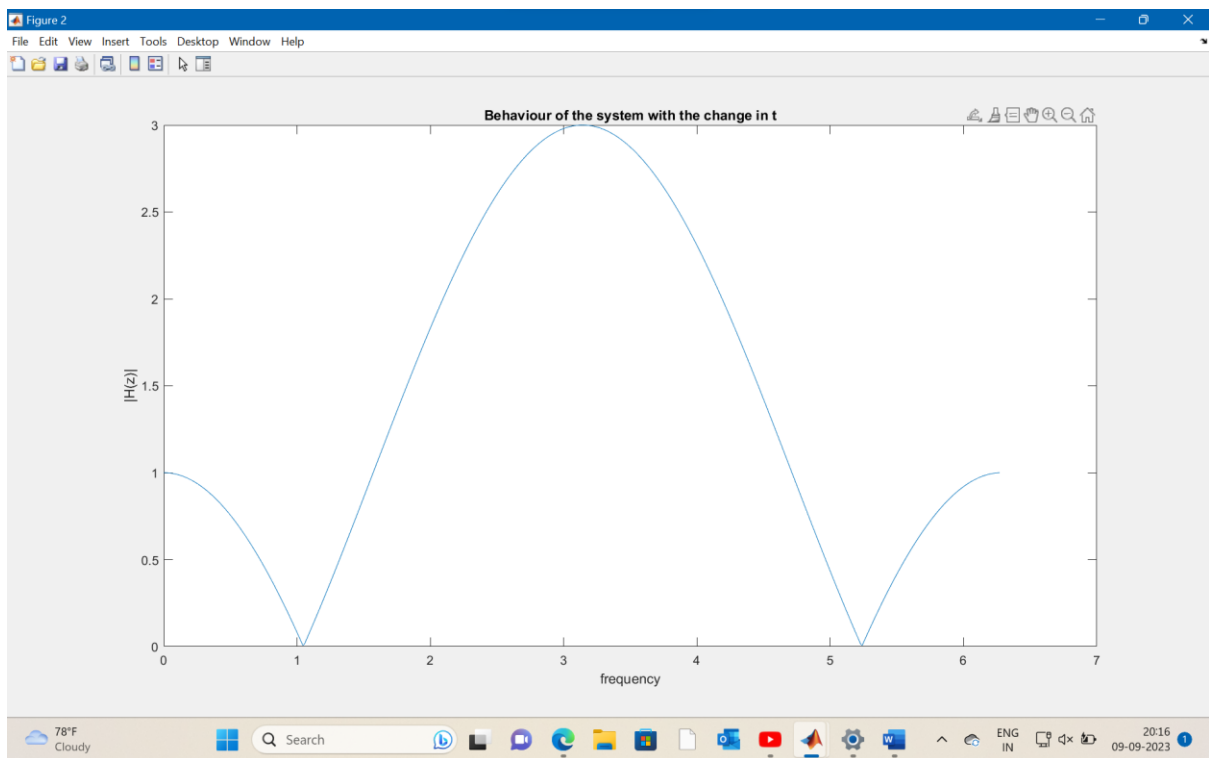
When $\theta = \pi/3$



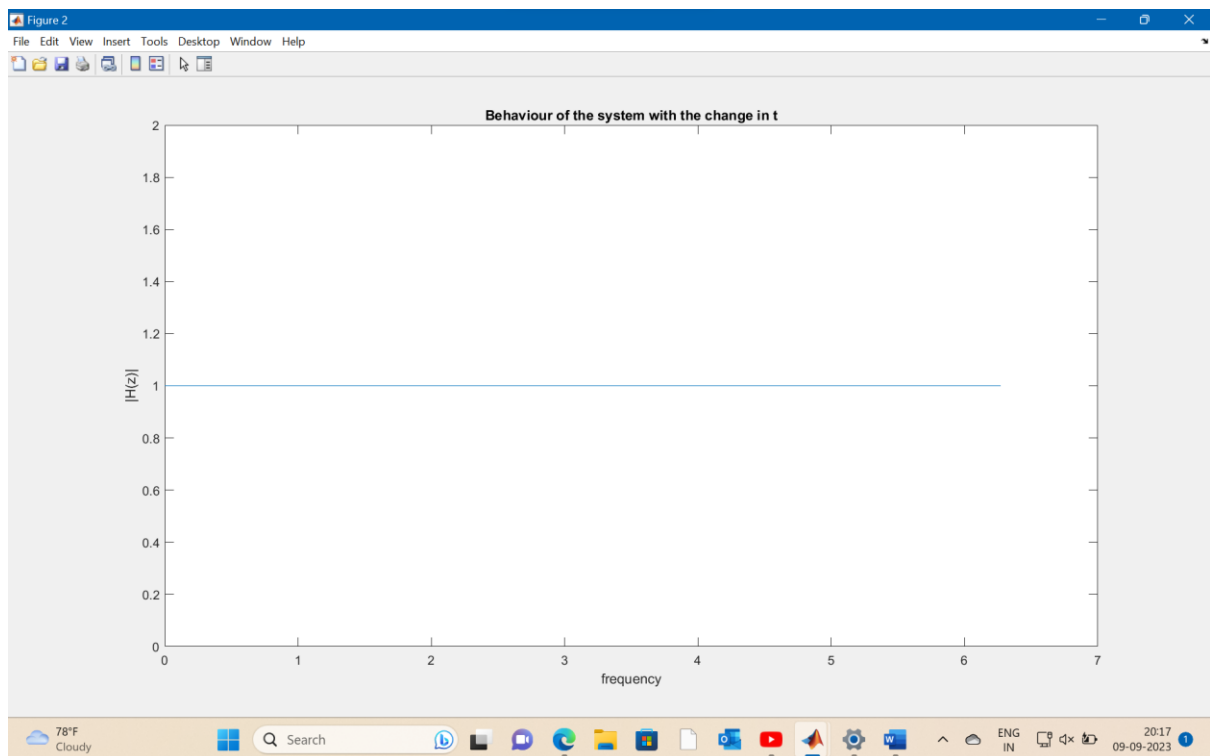
3.d)

here $\theta=60^\circ$

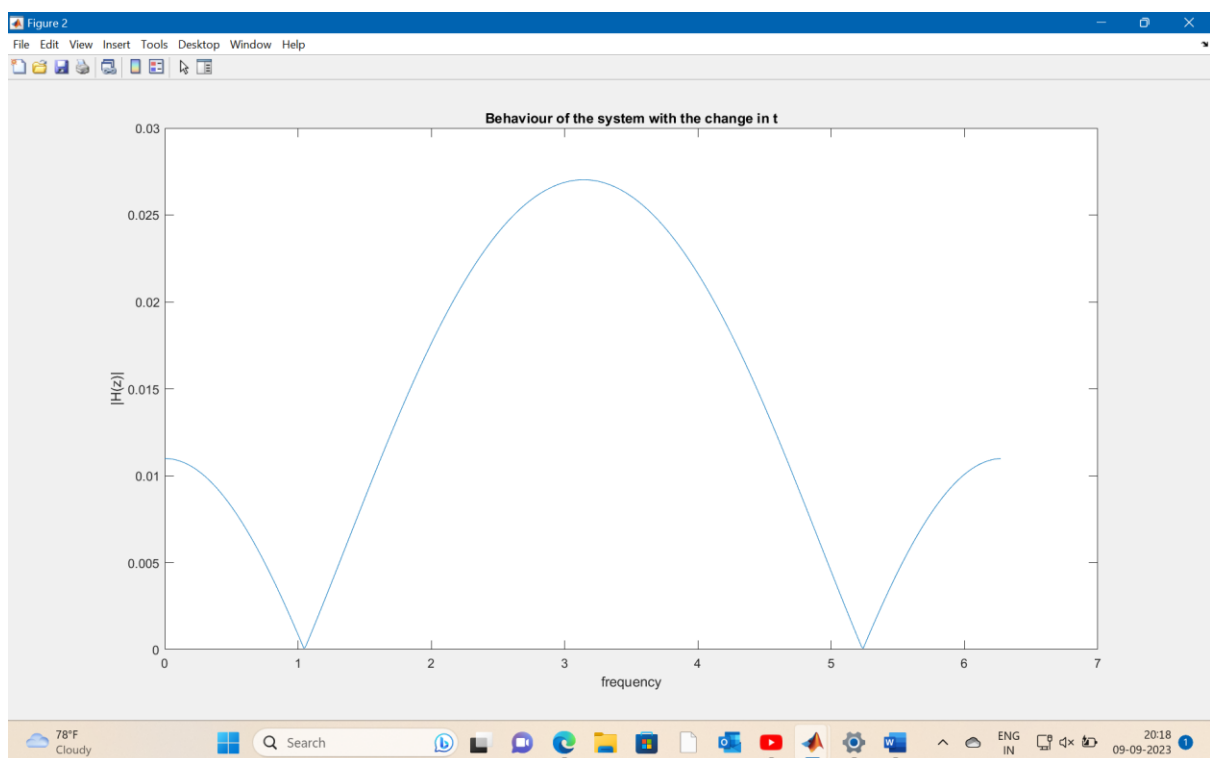
when $r=0$



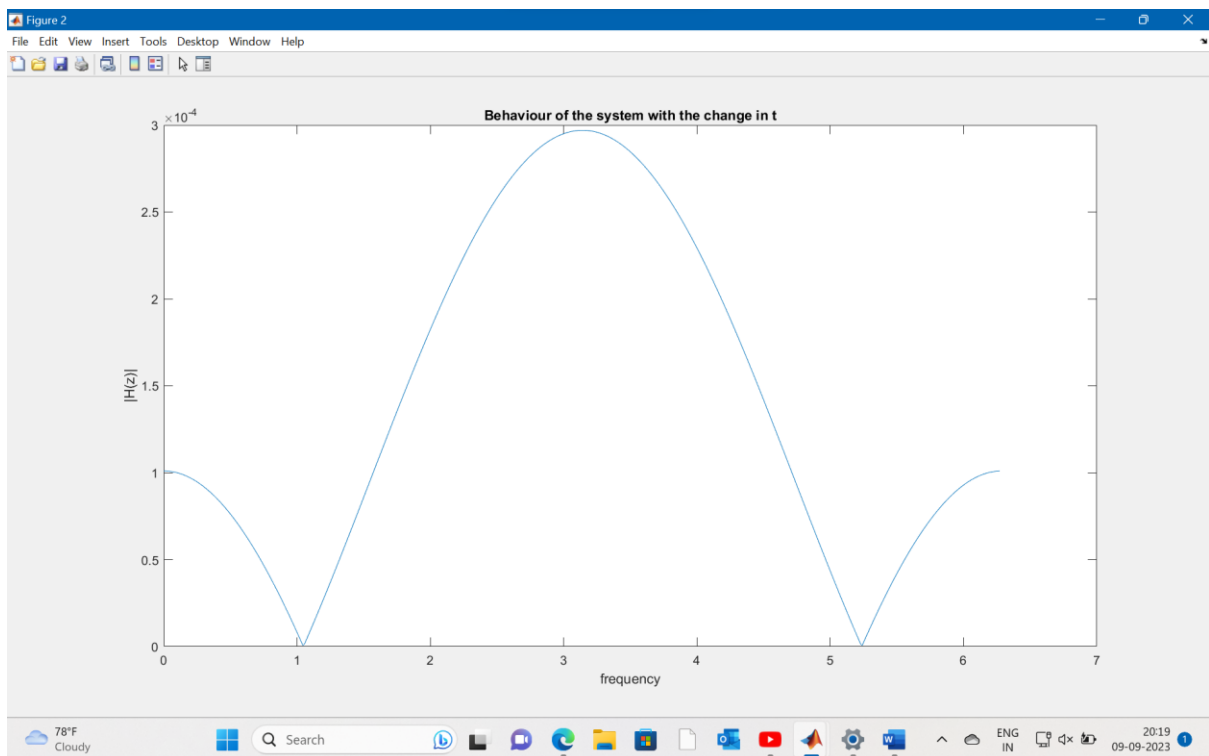
When $r=1$



When $r=10$



When $r=100$

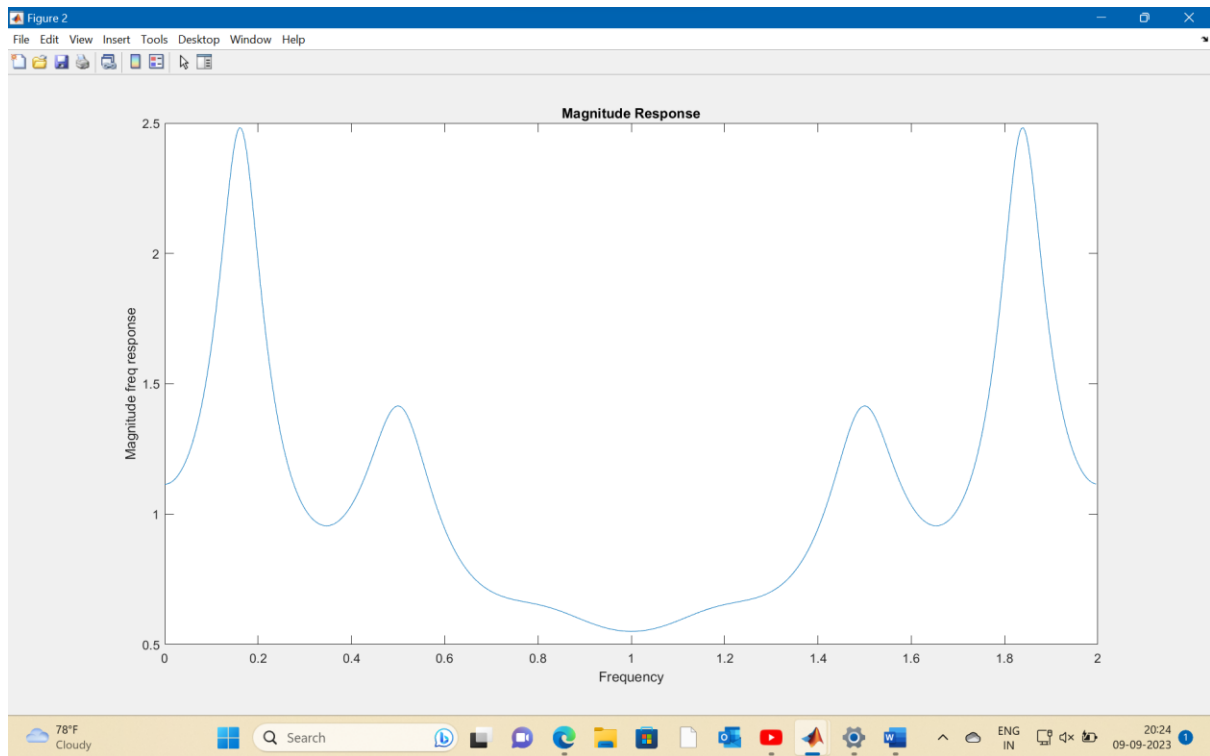


4. LTI system with multiple poles:

Here the transfer function is given as

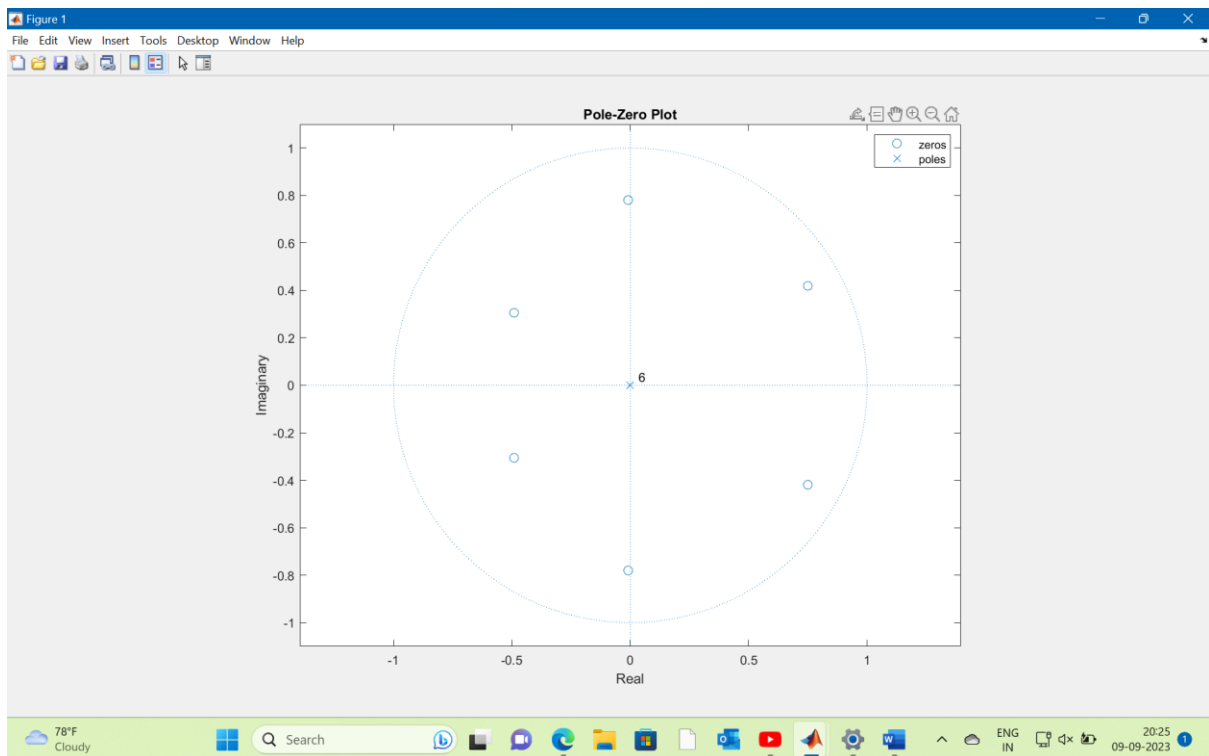
$$H(z) = 1 / (1 - 0.5z^{-1} + 0.2z^{-2} - 0.1z^{-3} + 0.007z^{-4} + 0.14z^{-5} + 0.15z^{-6})$$

a)



Here from the observation of plot we can say that the poles are located at the peaks of the curves . so here the peaks are symmetrical with respect to the frequency =1 .

4)b.



Here from the both plots we can say that the poles are conjugate pairs here there are 6 poles i.e 3 conjugate pairs of poles. Here the poles with highest peak are located near to the origin and the next second highest peak are next near to the origin.