



# Transformasi Z dan Implementasinya pada sistem

Bahan Ajar Sinyal dan Sistem  
Pascasarjana Terapan

P E N S

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## The Z-Transform

- In mathematics and signal processing, the **Z-transform** converts a discrete-time signal, which is a sequence of real or complex number, into a complex frequency domain representation.
- Convolution of discrete-time signals simply becomes multiplication of their z-transforms.

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## The Z-Transform

$$\begin{aligned}
 \mathcal{Z}\{x_1(n) * x_2(n)\} &= \mathcal{Z}\left\{\sum_{l=-\infty}^{\infty} x_1(l)x_2(n-l)\right\} \\
 &= \sum_{n=-\infty}^{\infty} \left[\sum_{l=-\infty}^{\infty} x_1(l)x_2(n-l)\right] z^{-n} \\
 &= \sum_{l=-\infty}^{\infty} x_1(l) \left[\sum_{n=-\infty}^{\infty} x_2(n-l)z^{-n}\right] \\
 &= \left[\sum_{l=-\infty}^{\infty} x_1(l)z^{-l}\right] \left[\sum_{n=-\infty}^{\infty} x_2(n)z^{-n}\right] \\
 &= X_1(z)X_2(z)
 \end{aligned}$$

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## The Z-Transform

Convolution

Multiplication

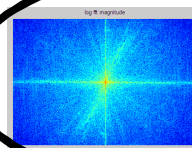


\*

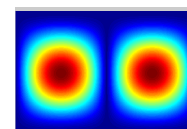


FFT

FFT

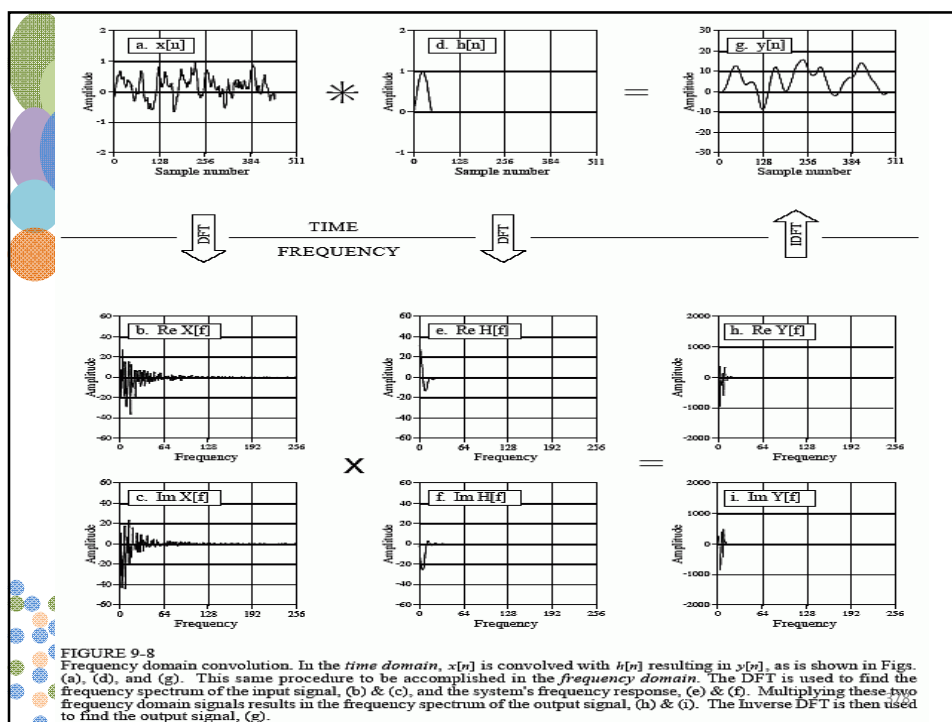


x



Frequency

Slide: Hoiem<sup>377</sup>



## The Z-Transform

- Analog filters are designed using the Laplace transform, Recursive digital filters are developed with a technique called the z-transform.
- The Laplace transform deals with differential equations, the s-domain, and the s-plane. Correspondingly, the z-transform deals with difference equations, the z-domain, and the z-plane.
- The s-plane is arranged in a rectangular coordinate system, while the z-plane uses a polar format.

## The Z-Transform

- Recursive digital filters are often designed by starting with one of the classic analog filters, such as the Butterworth, Chebyshev, or elliptic. A series of mathematical conversions are then used to obtain the desired digital filter. The z-transform provides the framework for this mathematics.

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## From Laplace Transform to Z-Transform

$$X(s) = \int_{t=-\infty}^{\infty} x(t) e^{-st} dt \quad \longrightarrow \quad X(\sigma, \omega) = \int_{t=-\infty}^{\infty} x(t) e^{-\sigma t} e^{-j\omega t} dt$$

Continuous

Exponential Forms :

$$X(\sigma, \omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-\sigma n} e^{-j\omega n}$$

Discrete

$$y[n] = e^{-\sigma n} \quad \text{or} \quad y[n] = r^{-n}$$



$$r^{-n} = [e^{\ln(r)}]^{-n} = e^{-n \ln(r)} = e^{-\sigma n}$$

$$\text{where: } \sigma = \ln(r)$$

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## From Laplace Transform to Z-Transform

a. Decreasing

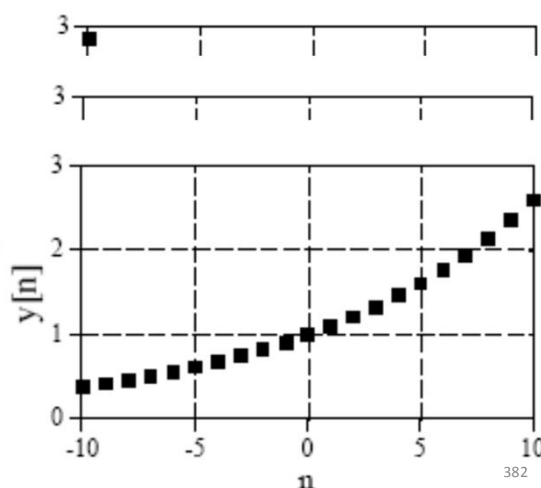
b. Constant

c. Increasing

$$y[n] = e^{-\sigma n}, \quad \sigma = -0.095$$

or

$$y[n] = r^{-n}, \quad r = 0.9$$



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## From Laplace Transform to Z-Transform

$$X(\sigma, \omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-\sigma n} e^{-j\omega n}$$



$$X(r, \omega) = \sum_{n=-\infty}^{\infty} x[n] r^{-n} e^{-j\omega n}$$

$$z = r e^{j\omega}$$

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## The Z-Transform

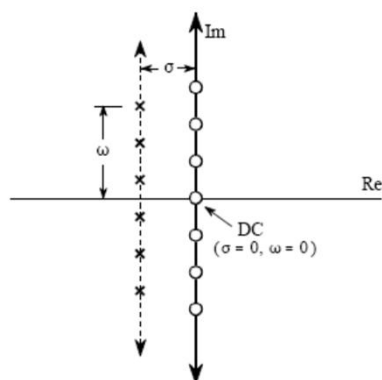
$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

$$z = r e^{j\omega}$$

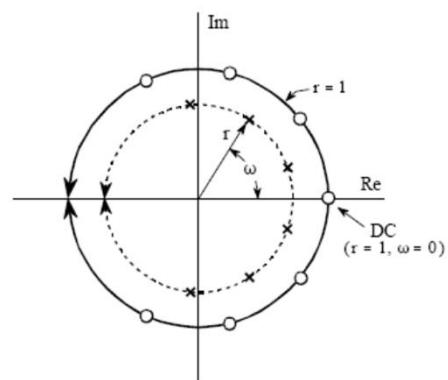
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## The Relationship between the s-plane and the z-plane

s - Plane



z - Plane



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## The Analysis of Recursive Systems

A recursive filter is described by a difference equation:

$$y[n] = a_0x[n] + a_1x[n-1] + a_2x[n-2] + \dots + b_1y[n-1] + b_2y[n-2] + b_3y[n-3] + \dots$$

The "a" and "b" terms are the recursion coefficients.

The system's transfer function,  $H[z] = Y[z] / X[z]$

$$H[z] = \frac{a_0 + a_1z^{-1} + a_2z^{-2} + a_3z^{-3} + \dots}{1 - b_1z^{-1} - b_2z^{-2} - b_3z^{-3} - \dots}$$

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## The recursion coefficients of a Digital Filter

$$\begin{array}{ll} a_0 = 0.389 & \\ a_1 = -1.558 & b_1 = 2.161 \\ a_2 = 2.338 & b_2 = -2.033 \\ a_3 = -1.558 & b_3 = 0.878 \\ a_4 = 0.389 & b_4 = -0.161 \end{array}$$

The system's transfer function

$$H[z] = \frac{0.389 - 1.558z^{-1} + 2.338z^{-2} - 1.558z^{-3} + 0.389z^{-4}}{1 - 2.161z^{-1} + 2.033z^{-2} - 0.878z^{-3} + 0.161z^{-4}}$$

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## Transfer function in pole-zero form

Just as with the s-domain, an important feature of the z-domain is that the transfer function can be expressed as **poles** and **zeros**. This provides the second general form of the z-domain:

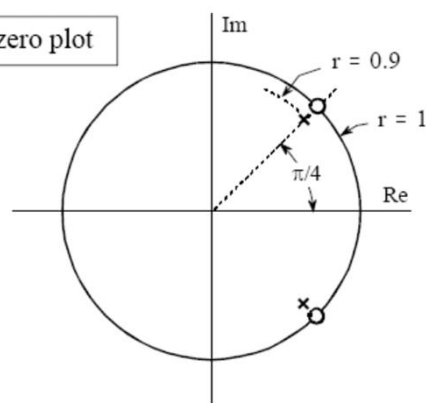
$$H[z] = \frac{(z - z_1)(z - z_2)(z - z_3)\dots}{(z - p_1)(z - p_2)(z - p_3)\dots}$$

Each of the poles ( $p$ ) and zeros ( $z$ ) is a complex number.

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## A Notch filter Design

a. Pole-zero plot



(1) specify the pole-zero placement in the z-plane

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## A Notch filter Design

In polar form:

$$z_1 = 1.00 e^{j(\pi/4)}$$

$$z_2 = 1.00 e^{j(-\pi/4)}$$

$$p_1 = 0.90 e^{j(\pi/4)}$$

$$p_2 = 0.90 e^{j(-\pi/4)}$$

In rectangular form:

$$z_1 = 0.7071 + j 0.7071$$

$$z_2 = 0.7071 - j 0.7071$$

$$p_1 = 0.6364 + j 0.6364$$

$$p_2 = 0.6364 - j 0.6364$$

**(2) Write down the transfer function in pole – zero form**

$$H(z) = \frac{[z - (0.7071 + j 0.7071)] [z - (0.7071 - j 0.7071)]}{[z - (0.6364 + j 0.6364)] [z - (0.6364 - j 0.6364)]}$$

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## A Notch filter Design

**(3) Rearrange the transfer function into the form of Z Polynomial**

$$H(z) = \frac{z^2 - 0.7071z + j0.7071z - 0.7071z + 0.7071^2 - j0.7071^2 - j0.7071z + j0.7071^2 - j^2 0.7071^2}{z^2 - 0.6364z + j0.6364z - 0.6364z + 0.6364^2 - j0.6364^2 - j0.6364z + j0.6364^2 - j^2 0.6364^2}$$

$$H[z] = \frac{1.000 - 1.414z + 1.000z^2}{0.810 - 1.273z + 1.000z^2}$$

$$H[z] = \frac{1.000 - 1.414z^{-1} + 1.000z^{-2}}{1.000 - 1.273z^{-1} + 0.810z^{-2}}$$

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## A Notch filter Design

(4) Identify the recursion coefficients needed to implement the filter.

$$a_0 = 1.000$$

$$a_1 = -1.414$$

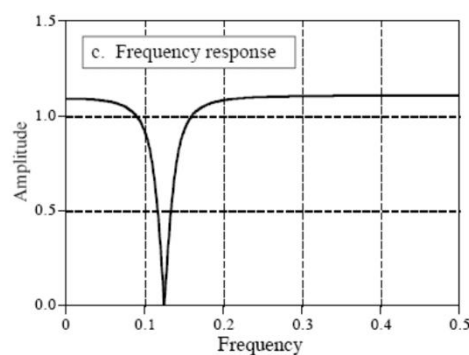
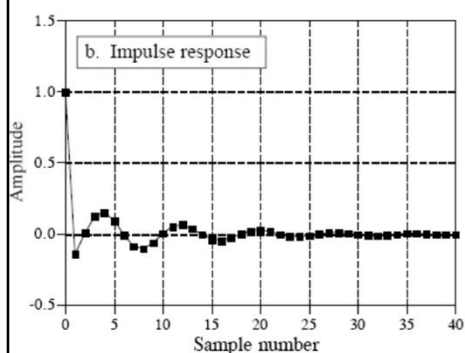
$$a_2 = 1.000$$

$$b_1 = 1.273$$

$$b_2 = -0.810$$

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## The resulting impulse and frequency response of a Notch Filter.



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## How do we find the frequency response?

### ✓ Mathematical

[ Find  $H(z)$  at  $r=1$ , replace each  $z$  with  $e^{j\omega} \rightarrow H(\omega)$  ]

### ✓ Computational (programming)

[ Define  $x$  equally spaced frequencies between  $\omega = 0$  and  $\omega = \pi$  ]

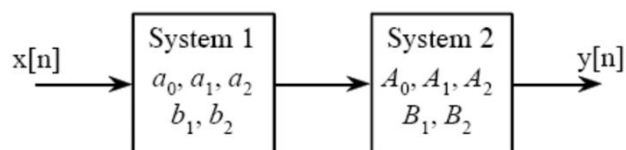
### ✓ Find the frequency response from the recursion coefficients that are actually used to implement the filter.

[ Find the impulse response of the filter by passing an impulse through the system, then take the FFT of the impulse response to find the system's frequency response. ]

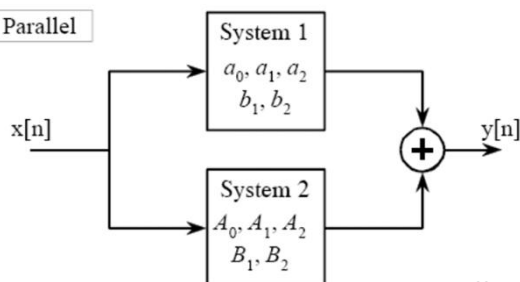
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## Cascade and Parallel Stages

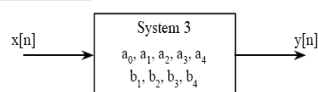
a. Cascade



b. Parallel



c. Replacement



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## Cascade and Parallel Stages

$$H[z] = \frac{a_0 + a_1 z^{-1} + a_2 z^{-2}}{1 - b_1 z^{-1} - b_2 z^{-2}} \times \frac{A_0 + A_1 z^{-1} + A_2 z^{-2}}{1 - B_1 z^{-1} - B_2 z^{-2}}$$

$$H[z] = \frac{a_0 A_0 + (a_0 A_1 + a_1 A_0) z^{-1} + (a_0 A_2 + a_1 A_1 + a_2 A_0) z^{-2} + (a_1 A_2 + a_2 A_1) z^{-3} + (a_2 A_2) z^{-4}}{1 - (b_1 + B_1) z^{-1} - (b_2 + B_2 - b_1 B_1) z^{-2} - (-b_1 B_2 - b_2 B_1) z^{-3} - (-b_2 B_2) z^{-4}}$$

$$a_0 = a_0 A_0$$

$$a_1 = a_0 A_1 + a_1 A_0$$

$$a_2 = a_0 A_2 + a_1 A_1 + a_2 A_0$$

$$a_3 = a_1 A_2 + a_2 A_1$$

$$a_4 = a_2 A_2$$

$$b_1 = b_1 + B_1$$

$$b_2 = b_2 + B_2 - b_1 B_1$$

$$b_3 = -b_1 B_2 - b_2 B_1$$

$$b_4 = -b_2 B_2$$

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In specific cases, it is possible to derive simpler equations directly relating the pole-zero positions to the recursion coefficients. For example, a system containing two poles and two zeros, called as **biquad**, has the following relations:

$$a_0 = 1$$

$$a_1 = -2r_0 \cos(\omega_0)$$

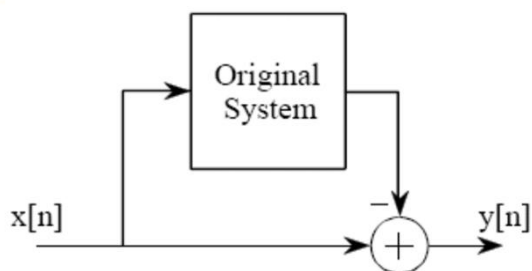
$$a_2 = r_0^2$$

$$b_1 = 2r_p \cos(\omega_p)$$

$$b_2 = -r_p^2$$

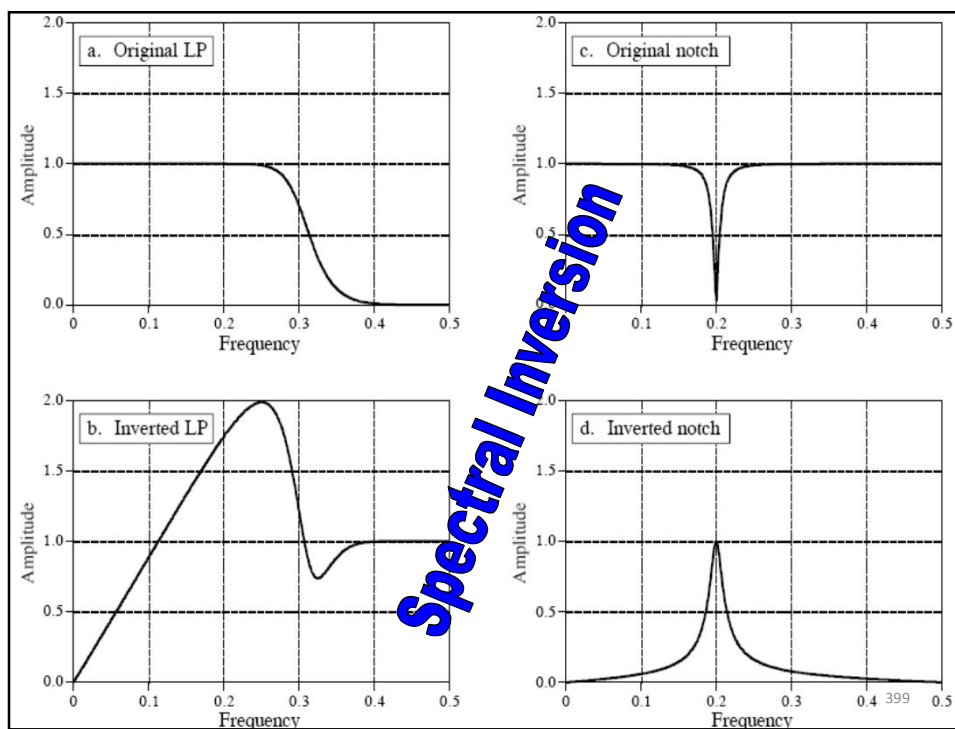
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# Spectral Inversion



$$\begin{aligned} a_0 &= 1 - a_0 \\ a_1 &= -a_1 - b_1 \\ a_2 &= -a_2 - b_2 \\ a_3 &= -a_3 - b_3 \\ &\vdots \end{aligned}$$

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## The Hilbert transformer.

Any system that has the frequency response:  
Magnitude = 1 and phase = 90 degrees, for all frequencies.

Hilbert transformers can be analog or discrete (hardware or software)

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## Gain Changes

Since the gain must be specified at a frequency in the *passband*, the procedure depends on the type of filter being used. Low-pass filters have their gain measured at a frequency of *zero*, while high-pass filters use a frequency of 0.5, the maximum frequency allowable.

$$y[n] = a_0x[n] + a_1x[n-1] + a_2x[n-2] + \dots + b_1y[n-1] + b_2y[n-2] + b_3y[n-3] + \dots$$

$$G = a_0 + a_1 + a_2 + a_3 + \dots + b_1G + b_2G + b_3G + b_4G \dots$$

$$G = \frac{a_0 + a_1 + a_2 + a_3 \dots}{1 - (b_1 + b_2 + b_3 \dots)}$$

To make a filter have a gain of *one* at DC, calculate the existing gain by using this relation, and then divide all the "a" coefficients by G.

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## Gain Changes

The gain at a frequency of 0.5 is found in a similar way: we force the input and output signals to operate at this frequency, and see how the system responds. At a frequency of 0.5, the samples in the input signal alternate between -1 and 1. That is, successive samples are: 1, -1, 1, -1, 1, -1, 1, etc. The corresponding output signal also alternates in sign, with an amplitude equal to the gain of the system:  $G$ ,  $-G$ ,  $G$ ,  $-G$ ,  $G$ ,  $-G$ , etc. Plugging these signals into the recursion equation:

$$G = a_0 - a_1 + a_2 - a_3 + \dots - b_1 G + b_2 G - b_3 G + b_4 G \dots$$

$$G = \frac{a_0 - a_1 + a_2 - a_3 + a_4 \dots}{1 - (-b_1 + b_2 - b_3 + b_4 \dots)}$$