

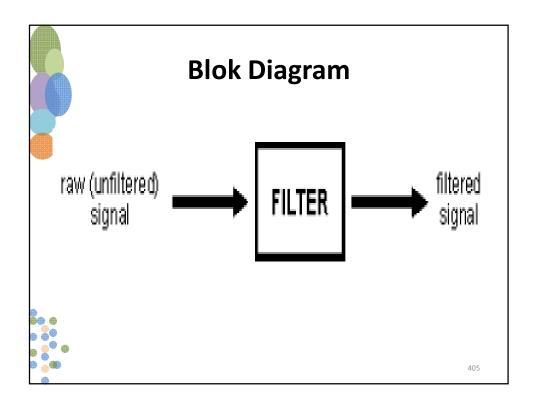
Bahan Ajar Sinyal dan Sistem Pascasarjana Terapan P E N S

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# Introduction to digital filters

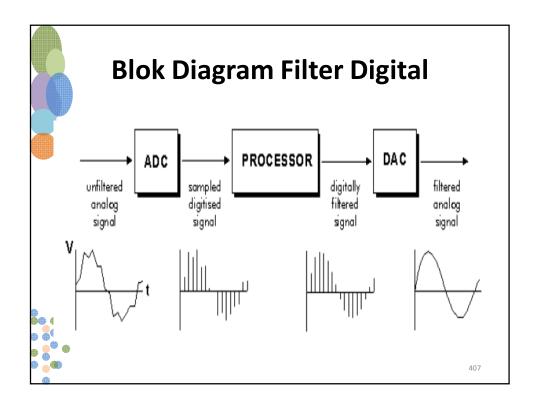
#### Analog and digital filters

In signal processing, the function of a *filter* is to remove unwanted parts of the signal, such as random noise, or to extract useful parts of the signal, such as the components lying within a certain frequency range.



# **Analog & Digital Filter**

- An **analog** filter uses analog electronic circuits made up from components such as resistors, capacitors and op amps to produce the required filtering effect. Such filter circuits are widely used in such applications as noise reduction, video signal enhancement, graphic equalisers in hi-fi systems, and many other areas.
- A digital filter uses a digital processor to perform numerical calculations on sampled values of the signal. The processor may be a general-purpose computer such as a PC, or a specialised DSP (Digital Signal Processor) chip.



# Advantages of using digital filters (1)

A digital filter is *programmable*, i.e. its operation is determined by a program stored in the processor's memory. This means the digital filter can easily be changed without affecting the circuitry (hardware). An analog filter can only be changed by redesigning the filter circuit.

- Digital filters are easily *designed*, *tested* and *implemented* on a general-purpose computer or workstation.
- The characteristics of analog filter circuits (particularly those containing active components) are subject to drift and are dependent on temperature. Digital filters do not suffer from these problems, and so are extremely *stable* with respect both to time and temperature.



- Unlike their analog counterparts, digital filters can handle low frequency signals accurately. As the speed of DSP technology continues to increase, digital filters are being applied to high frequency signals in the RF (radio frequency) domain, which in the past was the exclusive preserve of analog technology.
- Digital filters are very much more versatile in their ability to process signals in a variety of ways; this includes the ability of some types of digital filter to adapt to changes in the characteristics of the signal.
- Fast DSP processors can handle complex combinations of filters in parallel or cascade (series), making the hardware requirements relatively simple and compact in comparison with the equivalent analog circuitry.

#### 1. Unity gain filter:

$$y_n = x_n$$

Each output value  $y_n$  is exactly the same as the corresponding input value  $x_n$ :

$$y_0 = x_0$$

$$y_1 = x_1$$

$$y_2 = x_2$$

etc

This is a trivial case in which the filter has no effect on the signal.



#### 2. Simple gain filter:

$$y_n = Kx_n$$

where K = constant.

This simply applies a gain factor K to each input value.

K > 1 makes the filter an amplifier, while 0 < K < 1 makes it an attenuator. K < 0 corresponds to an inverting amplifier.



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# **Examples of simple digital filters**

#### 3. Pure delay filter:

$$y_n = x_{n-1}$$

The output value at time t = nh is simply the input at time t = (n-1)h, i.e. the signal is delayed by time h:

$$y_0 = x_{-1}$$

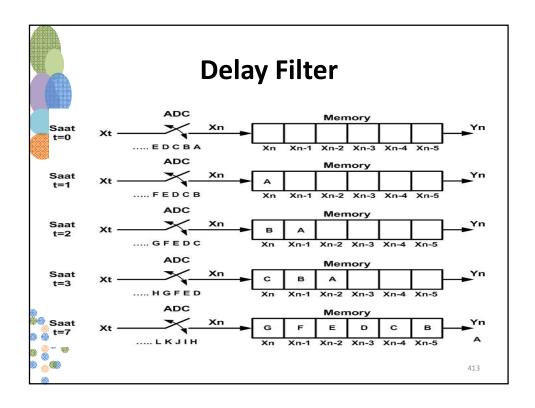
$$y_1 = x_0$$

$$y_2 = x_1$$

$$y_3 = x_2$$

eta

Note that as sampling is assumed to commence at t = 0, the input value  $x_{-1}$  at t = -h is undefined. It is usual to take this (and any other values of x prior to t = 0) as zero.



#### 4. Two-term difference filter:

$$y_n = x_n - x_{n-1}$$

The output value at t = nh is equal to the difference between the current input xn and the previous input xn-1:

$$y_0 = x_0 - x_{-1}$$

$$\mathbf{y}_1 = \mathbf{x}_1 - \mathbf{x}_0$$

$$y_2 = x_2 - x_1$$

$$y_3 = x_3 - x_2$$

... etc

i.e. the output is the *change* in the input over the most recent sampling interval *h*. The effect of this filter is similar to that of an analog differentiator circuit.



5. Two-term average filter:

$$y_n = \frac{x_n + x_{n-1}}{2}$$

The output is the average (arithmetic mean) of the current and previous input:

$$y_0 = \frac{x_0 + x_{-1}}{2}$$

$$y_1 = \frac{x_1 + x_0}{2}$$

$$y_2 = \frac{x_2 + x_1}{2}$$

This is a simple type of low pass filter it tends to smooth out high-frequency variations in a signal.

(We will look at more effective low pass filter designs later).







6. Three-term average filter:

$$y_n = \frac{x_n + x_{n-1} + x_{n-2}}{2}$$

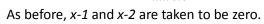
This is similar to the previous example, with the average being taken of the current and two previous inputs:

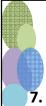
$$y_0 = \frac{x_0 + x_{-1} + x_{-2}}{2}$$

$$y_1 = \frac{x_1 + x_0 + x_{-1}}{2}$$

$$y_2 = \frac{x_2 + x_1 + x_0}{2}$$

.... etc







$$y_n = \frac{x_n - x_{n-2}}{2}$$

This is similar in its effect to example (4). The output is equal to half the change in the input signal over the previous two sampling intervals:

$$y_{0} = \frac{x_{0} + x_{-2}}{2}$$

$$y_{1} = \frac{x_{1} + x_{-1}}{2}$$

$$y_{2} = \frac{x_{2} + x_{0}}{2}$$

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The *order* of a digital filter is the number of *previous* inputs (stored in the processor's memory) used to calculate the current output.



## **Digital filter coefficients**

All of the digital filter examples given above can be written in the following general forms:

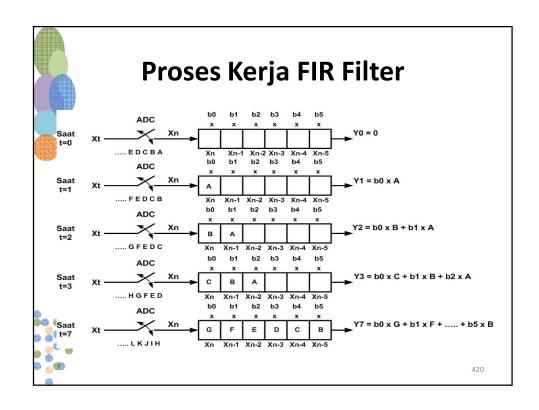
Zero order:  $y_n = a_0.x_n$ 

First order:  $y_n = a_0.x_n + a_1.x_{n-1}$ 

Second order:  $y_n = a_0.x_n + a_1.x_{n-1} + a_2.x_{n-2}$ 

Similar expressions can be developed for filters of any order.

The constants *a0*, *a1*, *a2*, ... appearing in these expressions are called the *filter coefficients*. It is the values of these coefficients that determine the characteristics of a particular filter.



# Program FIR

```
float buff[]={0,0,0}; // mengosongkan buffer memory

void FIR(int data)
{
    int orde = 2, i;
    float b[] = {0.5, 1, 0.5};
    float buff[10], temp,y;

    for(i=orde+1;i>=0;i--) // Menggeser isi buffer memory
        buff[i]=buff[i-1];
    buff[0]=data;

temp=0;
    for(i=0;i<orde+1;i++) // Perhitungan konvolusi output
        temp+=buff[i]*b[i];
    y=temp;
}
```

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# The transfer function of a digital filter

In the last section, we used two different ways of expressing the action of a digital filter: a form giving the output *yn* directly, and a "symmetrical" form with all the output terms on one side and all the input terms on the other.

- In this section, we introduce what is called the *transfer* function of a digital filter. This is obtained from the symmetrical form of the filter expression, and it allows us to describe a filter by means of a convenient, compact expression. We can also use the transfer function of a filter to work out its frequency response.
- First of all, we must introduce the delay operator, denoted by the symbol z<sup>-1</sup>.



- When applied to a sequence of digital values, this operator gives the *previous* value in the sequence. It therefore in effect introduces a delay of one sampling interval.
- Applying the operator  $z^{-1}$  to an input value (say  $x_n$ ) gives the previous input  $(x_{n-1})$ :

$$z^{-1} x_n = x_{n-1}$$

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Contoh:

$$x_0 = 5$$

$$x_1 = -2$$

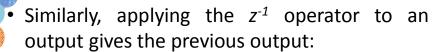
$$x_2 = 10$$

$$z^{-1}.x_1 = x_0 = 5$$

$$z^{-1}.x_2 = x_1 = -2$$







$$z^{-1}.y_n = y_{n-1}$$

• Applying the delay operator  $z^{-1}$  twice produces a delay of two sampling intervals:

$$z^{-1}.(z^{-1}.x_n) = z^{-1}.x_{n-1} = x_{n-2}$$

• We adopt the (fairly logical) convention :

$$z^{-1}$$
.  $z^{-1} = z^{-2}$ 

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## The transfer function of a digital filter

Aplikasi z<sup>-1</sup> dalam IIR Filter design :

$$a_{0} \cdot y_{n} + a_{1} \cdot y_{n-1} + a_{2} \cdot y_{n-2} = b_{0} \cdot x_{n} + b_{1} \cdot x_{n-1} + b_{2} \cdot x_{n-2}$$

$$a_{0} \cdot y_{n} + a_{1} \cdot y_{n} \cdot z^{-1} + a_{2} \cdot y_{n} \cdot z^{-2} = b_{0} \cdot x_{n} + b_{1} \cdot x_{n} \cdot z^{-1} + b_{2} \cdot x_{n} \cdot z^{-2}$$

$$(a_{0} + a_{1} \cdot z^{-1} + a_{2} \cdot z^{-2}) \cdot y_{n} = (b_{0} + b_{1} \cdot z^{-1} + b_{2} \cdot z^{-2}) \cdot x_{n}$$

$$H(z) = \frac{y_{n}}{x_{n}} = \frac{b_{0} + b_{1} \cdot z^{-1} + b_{2} \cdot z^{-2}}{a_{0} + a_{1} \cdot z^{-1} + a_{2} \cdot z^{-2}}$$

# The transfer function of a digital filter

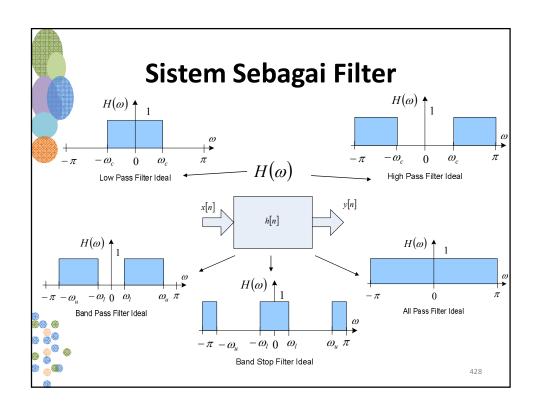
#### Aplikasi z<sup>-1</sup> dalam FIR Filter design :

$$a_{0} \cdot y_{n} = b_{0} \cdot x_{n} + b_{1} \cdot x_{n-1} + b_{2} \cdot x_{n-2}$$

$$a_{0} \cdot y_{n} = b_{0} \cdot x_{n} + b_{1} \cdot x_{n} \cdot z^{-1} + b_{2} \cdot x_{n} \cdot z^{-2}$$

$$a_{0} \cdot y_{n} = (b_{0} + b_{1} \cdot z^{-1} + b_{2} \cdot z^{-2}) \cdot x_{n}$$

$$H(z) = \frac{y_{n}}{x_{n}} = \frac{b_{0} + b_{1} \cdot z^{-1} + b_{2} \cdot z^{-2}}{a_{0}}$$





# Filter digital sebagai implementasi LCCDE

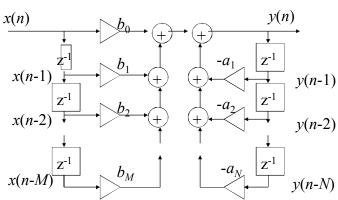
Persamaan perbedaan (Difference Equation)

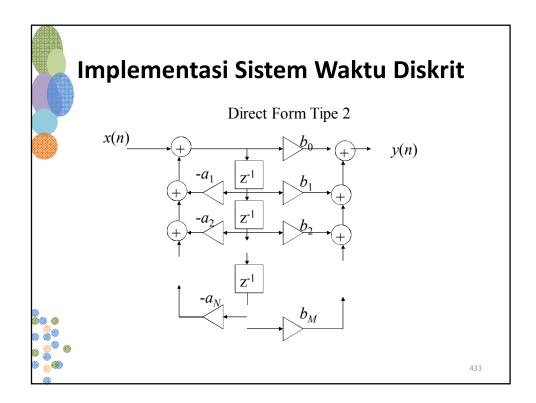
$$y(n) = -\sum_{k=1}^{N} a_k y(n-k) + \sum_{k=0}^{M} b_k x(n-k)$$

$$H(z) = \frac{\sum_{k=0}^{M} b_k z^{-k}}{1 + \sum_{k=1}^{N} a_k z^{-k}} \qquad H(\omega) = \frac{\sum_{k=0}^{M} b_k e^{-j\omega k}}{1 + \sum_{k=1}^{N} a_k e^{-j\omega k}}$$

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# **Respon Frekuensi Filter Digital**

Sebuah sistem filter digital secara umum digambarkan pada diagram blok sebagai berikut:

$$x(n) = A \cdot \sin(\omega n)$$

$$x(n) = e^{j\omega n}$$
Filter Digital
$$y(n) = B \cdot \sin(\omega n + \varphi)$$

$$y(n) = a \cdot e^{j(\omega n + \varphi)}$$



Sistem menerima masukan berupa sinyal dengan frekuensi tertentu  $\rightarrow x(n) = e^{j\omega n}$ 

# **Respon Frekuensi Filter Digital**

Jika konvolusi sbb:  $y(n) = \sum_{k=-\infty}^{\infty} h(k) \cdot x(n-k) = > domainwaktu$ 

Mendapatkan masukan:  $x(n) = e^{j\omega n}$ 

Maka: 
$$y(n) = \sum_{k=-\infty}^{\infty} h(k) \cdot e^{j\omega(n-k)}$$

$$y(n) = e^{j\omega n} \cdot \sum_{k=-\infty}^{\infty} h(k) \cdot e^{-j\omega k}$$

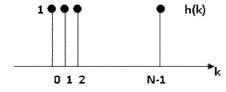
$$y(n) = e^{j\omega n} \cdot H(e^{j\omega})$$

Sinyal Harga amplitudo yang bergantung Semula pada frekuensi sinyal input ω

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### Contoh

Respon frekuensi LPF digital



Jika diketahui respon impulse h(k), maka carilah respon frekuensinya!



• Jawaban: 
$$H(e^{j\omega}) = \sum_{k=-\infty}^{\infty} h(k) \cdot e^{-j\omega k}$$

$$H(e^{j\omega}) = \sum_{k=0}^{N-1} 1 \cdot e^{-j\omega k}$$

$$H(e^{j\omega}) = \sum_{k=0}^{N-1} e^{-j\omega k}$$

$$H(e^{j\omega}) = \sum_{k=0}^{N-1} e^{-[j\omega]k}$$

Jika 
$$\sum_{k=0}^{N} a^k = \frac{1-a^{N+1}}{1-a}$$
, maka:

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#### Contoh

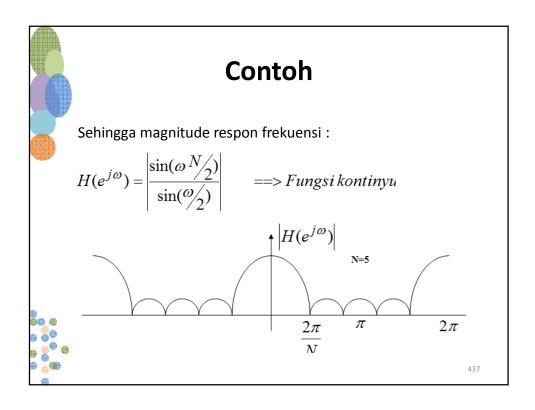
$$H(e^{j\omega}) = \sum_{k=0}^{N} [e^{-j\omega}]k = \frac{1 - e^{-j\omega N}}{1 - e^{-j\omega}}$$

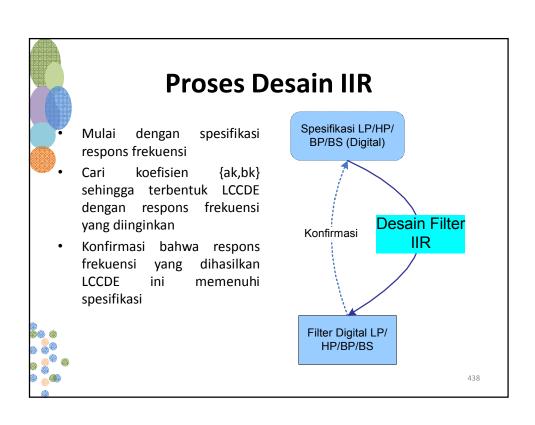
$$H(e^{j\omega}) = \frac{e^{-j\omega^{N}/2} [e^{j\omega^{N}/2} - e^{-j\omega^{N}/2}]}{e^{-j\omega^{N}/2} [e^{j\omega^{N}/2} - e^{-j\omega^{N}/2}]}$$

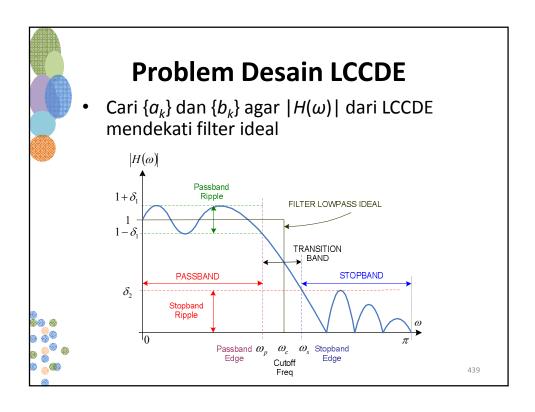
$$H(e^{j\omega}) = e^{-j(N-1)\frac{\omega}{2}} \cdot \frac{\left[e^{j\omega N/2} - e^{-j\omega N/2}\right]/2j}{\left[e^{j\omega/2} - e^{-j\omega/2}\right]/2j}$$

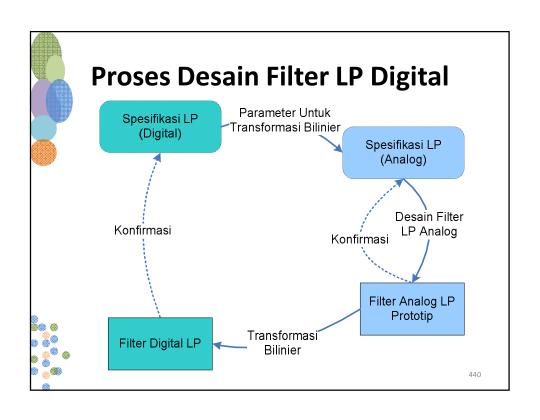
Jika 
$$\sin \alpha = \frac{e^{j\alpha} - e^{-j\alpha}}{2j}$$
, maka:

$$H(e^{j\omega}) = e^{-j(N-1)\omega/2} \cdot \frac{\sin(\omega^{N/2})}{\sin(\omega/2)}$$





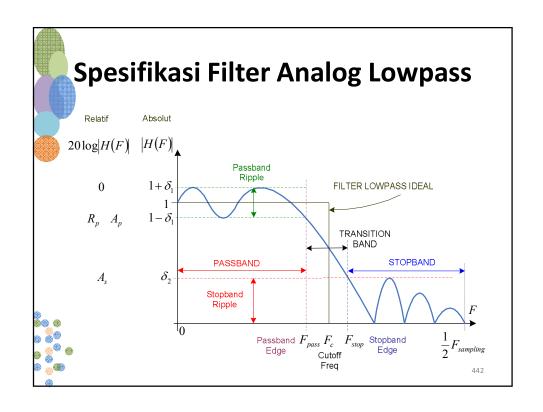


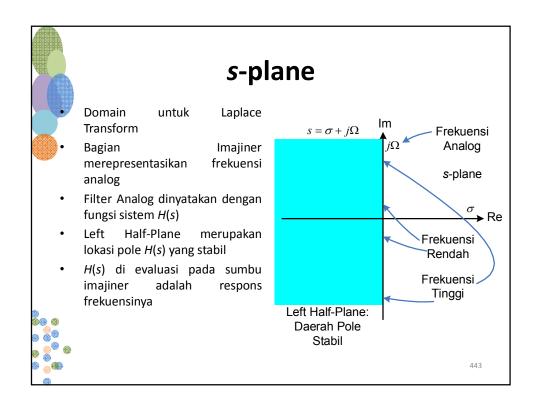


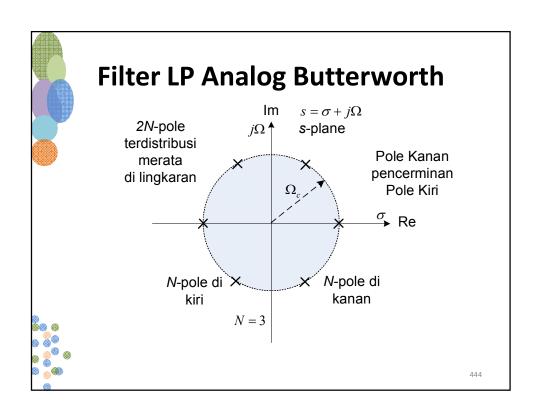


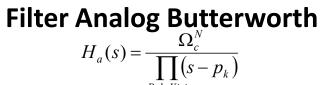


Contoh: LPF ini memiliki pass band edge dan stopband edge masing-masing  $F_{pass} = 800$  Hz dan  $F_{stop} = 1200$  Hz. Ripple pada passband dan stopband masing masing adalah  $\delta_1$  dan  $\delta_2$ , dan diinginkan  $R_p \le 1$  dB serta  $A_s > 50$  dB.





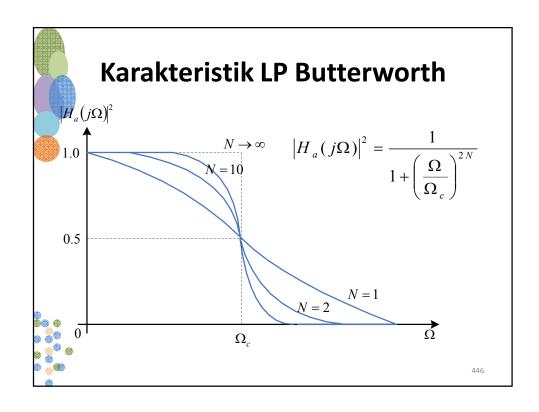




$$H_a(s) = \frac{\Omega_c^{N}}{\prod_{Pole\,Kiri} (s - p_k)}$$

$$\left|H_a(j\Omega)\right|^2 = H_a(j\Omega)H_a^*(j\Omega) = H_a(j\Omega)H_a(-j\Omega) = H_a(s)H_a(-s)\Big|_{s=j\Omega}$$

$$\begin{aligned} H_{a}(s)H_{a}(-s) &= \left| H_{a}(j\Omega) \right|^{2} \Big|_{\Omega = \frac{s}{j}} \qquad \left| H_{a}(j\Omega) \right|^{2} &= \frac{\Omega_{c}^{2N}}{\Omega^{2N} + \Omega_{c}^{2N}} \\ \left| H_{a}(j\Omega) \right|^{2} &= \frac{1}{1 + \left(\frac{\Omega}{\Omega_{c}}\right)^{2N}} \end{aligned}$$

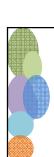




### **Soal Lowpass Filter Analog**



Contoh: LPF Butterworth ini memiliki pass band edge dan stopband edge masing-masing  $F_{pass} = 800$ Hz dan  $F_{stop} = 1200$  Hz. Ripple pada passband dan stopband masing masing adalah  $\delta_1$  dan  $\delta_2$ , dan diinginkan  $R_p \le 1$  dB serta  $A_s > 50$  dB. Cari N



#### **Jawaban**

$$R_p = -10\log_{10} \left| H_a \left( j\Omega_p \right) \right|^2$$

$$R_{p} = -10\log_{10}\left|H_{a}\left(j\Omega_{p}\right)^{2} \qquad R_{p} = -10\log_{10}\left[\frac{1}{1 + \left(\frac{\Omega_{p}}{\Omega_{c}}\right)^{2N}}\right]$$

$$A_s = -10\log_{10}|H_a(j\Omega_s)|^2$$

$$N = \frac{\log_{10} \left[ \left( 10^{R_p/10} - 1 \right) \left( 10^{A_s/10} - 1 \right) \right]}{2\log_{10} \left( \Omega_p / \Omega_s \right)}$$

$$N = \frac{\log_{10} \left[ \left( 10^{R_p/10} - 1 \right) \left( 10^{A_s/10} - 1 \right) \right]}{2 \log_{10} \left( \Omega_p / \Omega_s \right)} \qquad A_s = -10 \log_{10} \left( \frac{1}{1 + \left( \frac{\Omega_s}{\Omega_c} \right)^{2N}} \right)$$

$$\Omega_c = \frac{\Omega_p}{\sqrt[2N]{10^{\frac{R_p}{10}} - 1}}$$

$$\Omega_{c} = \frac{\Omega_{p}}{\sqrt[2N]{10^{\frac{R_{p}}{10}} - 1}} \qquad \Omega_{c} = \frac{\Omega_{s}}{\sqrt[2N]{10^{\frac{A_{s}}{10}} - 1}}$$

