

# Modeling Self Triggered Control Using Hybrid System Framework

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## 1 Introduction

Cyber-physical systems(CPS) consist of a physical plants and a discrete(cyber) controllers. The interaction between the plant and controller is naturally modelled as a hybrid system.

This report looks at a control technique known as self-triggered control. Self-triggered controllers compute the next sampling period when the plant state is measured. Control is then applied for that interval. This technique is part of a broader category of control techniques known as event-triggered control.

First, the report looks at general event-triggered control model build on the hybrid system's framework of [1]. An event-triggered model relies on event-triggering-mechanisms(ETMs) that update sensor data and apply control input. This model is specialized to demonstrate application of self-triggered ETMs.

Next, two self-triggered ETMs from the literature are reviewed and compared.

## 2 An Event-Triggered CPS Model

The following assumptions on the CPS model are made.

1. The system is modeled as a differential equation
2. The control law  $\kappa$  is static
3. Plant and controller output operate under zero-order hold
4. Plant and controller output are triggered by synchronous ETMs.

### 2.1 Closed Loop CPS with ETMs in the Loop.

This section uses the work from [1] to develop a simplified event-triggered CPS model.

Consider a CPS with a with the plant state  $x_p \in \mathcal{X}_p \subset \mathbb{R}^{n_p}$ , and the plant output under zero-order-hold  $\hat{y} \in \mathcal{Y}$  and control under zero-order hold  $\hat{u} \in \mathcal{U}$ . By assumption  $\hat{y}$  and  $\hat{u}$  are updated at the same time.

This update occurs through an Event-Triggered Mechanism(ETM). An auxiliary state  $\chi \in X \subset \mathbb{R}^{n_\chi}$  is defined to hold variables related to a system's ETMs. ETMs are triggered using an event triggered function  $\gamma : \Xi \mapsto \{0, 1\}$ , where  $\Xi := \mathcal{Y} \times \mathcal{U} \times X$ . The input to  $\gamma$  is given by  $\xi := (\hat{y}, \hat{u}, \chi)$ . When  $\gamma(\xi) = 1$ ,  $\hat{y}$  and  $\hat{u}$  are updated. That is, the plant state is sampled and a new control is applied.

Consider the aggregate state  $z = (x_p, \hat{y}, \hat{u}, \chi) \in Z \subset \mathcal{X} \times \mathcal{Y} \times \mathcal{U} \times X$ . Then, the event triggered CPS with the above assumptions can be modeled with the following hybrid system  $\mathcal{H}_{et} := (F, G, C, D)$  defined as

$$\begin{aligned} \dot{z} = F(z) &= \begin{bmatrix} F_p(x_p, \hat{u}) \\ 0 \\ 0 \\ F_\chi(z) \end{bmatrix} & z^+ = G(z) &= \begin{bmatrix} x_p \\ H_p(x_p) \\ \kappa(\hat{y}) \\ G_\chi(z) \end{bmatrix} \\ C &= \{z \in Z : \gamma(\xi) = 0\} & D &= \{z \in Z : \gamma(\xi) = 1\} \end{aligned} \quad (1)$$

where  $F_p : \mathcal{X}_p \times \mathcal{U} \mapsto \mathcal{X}_p$  denotes the plant dynamics,  $F_\chi : Z \mapsto X$  denotes the dynamics of the auxiliary variables,  $H_p : \mathcal{X}_p \mapsto \mathcal{Y}$  denotes the sensor update function,  $\kappa : \mathcal{Y} \mapsto \mathcal{U}$  denotes a static control law, and  $G_\chi : Z \mapsto X$  denotes the auxiliary variable update function.

In order to define a hybrid model for a self-triggered control model the the auxiliary variable  $\chi$ , it's update function,  $G_\chi$ , and the ETM triggering function  $\gamma$  must be defined. In the next section the concept of self-triggered control is introduced and the identified variables and functions are defined.

### 3 Self-Triggered Control

Self-Triggered control is a aperiodic control method that predicts when next to sample and apply control. This contrasts periodic control where sampling period is constant. Regardless of the method, sampled data systems define a sequence of times  $\{t_i\}_0^\infty$  where the system is updated. The difference between periodic ETMs and self-triggered ETMs are are follows:

- **Periodic ETMs:**  $t_{i+1} - t_i = T_s \quad \forall i \in \mathbb{N}$
- **Self-Triggered ETMs:**  $t_{i+1} - t_i = \Gamma(\xi)$ ,

where the self-triggered sampler is defined by a mapping  $\Gamma : \Xi \mapsto \mathbb{R}_{\geq 0}$  that computes a sampling period from the current data available. A self-triggered control model therefore needs to define auxiliary variables to handle timers and their update functions.

Let  $\chi := (\tau, T_s^*, \chi_3, \dots, \chi_{n_\chi})^\top$  where  $\chi_3, \dots, \chi_{n_\chi}$  are free auxiliary variables. The auxiliary dynamics and the update function  $G_\chi$  defined as

$$G_\chi(z) := \begin{bmatrix} 0 \\ \Gamma(\xi) \\ \chi_3^+ \\ \dots \\ \chi_{n_\chi}^+ \end{bmatrix}, \quad F_\chi(z) := \begin{bmatrix} 1 \\ 0 \\ \dot{\chi}_3^+ \\ \dots \\ \dot{\chi}_{n_\chi}^+ \end{bmatrix} \quad (2)$$

Furthermore, the ETM triggering function can be defined as

$$\gamma(\xi) := \begin{cases} 0 & \text{if } \tau < T_s^* \\ 1 & \text{if } \tau \geq T_s^* \end{cases} \quad (3)$$

### 3.1 Model

A self triggered CPS can be modeled by a hybrid system  $\mathcal{H}_{et} = (F, G, C, D)$  as in (1) with the auxiliary variables defined in (8) and the ETM triggering function in (3). It remains up to the self-triggered control designer to design the control law  $\kappa$  and a function  $\Gamma$  such that the system operates under desired requirements such as stability, safety, and non-zeno behavior.

### 3.2 Zeno Behavior in Self-Triggered Methods

An important consideration is avoiding Zeno behavior in the sample period. Zeno behavior can be avoided if it can be guaranteed that there exists a minimum-inter-event time that satisfies the control requirements.

## 4 Self-Triggered Strategies

This section provides an overview and comparison of two self-triggering ETMs.

### 4.1 Tiberi, Johansson [3]

#### Assumptions

1. There exists a feedback law  $\kappa : \mathcal{X} \mapsto \mathcal{U}$  such that the origin is asymptotically stable.

The sampling sequence  $\{t_i\}_0^\infty$  is inductively defined as

$$\begin{aligned} t_{i+1} &= t_i + \Gamma_1(\hat{y}) \quad \text{for} \\ \Gamma_1(\hat{y}) &= \frac{1}{2L} \ln(1 + 2\delta / \|f(\hat{y}, \kappa(\hat{y}))\|) \end{aligned} \quad (4)$$

where  $\delta > 0$ , and  $L = L_{f,u} L_{\kappa,x}$  is the product of the Lipschitz constants  $L_{f,u}$  (Lipschitz constant of  $f$  with respect to  $u$ ) and  $L_{\kappa,x}$  (Lipschitz constant of  $\kappa$  with respect to  $x$ ). The triggering condition assumes that  $\|f(\hat{y}, \kappa(\hat{y}))\| > m$  for some  $m > 0$ .

This strategy results in system solutions that are globally ultimately uniformly bounded (GUUB). That is, for an arbitrarily large  $a$  and with  $\|x_o\| < a$  it holds that  $\|x_p\| \leq b > 0 \forall t \geq T$ .

### 4.2 Heemels, Johansson, Tabuada [2]

#### Assumptions

1.  $F_p$  represents linear dynamics.
2. There exists a feedback law  $\kappa : \mathcal{X} \mapsto \mathcal{U}$  such that the system is exponentially asymptotically stable (E.A.S) to the origin.
3. There is full state feedback. That is, at sample point  $t_i$ , the sampled state is  $\hat{x} := \hat{y} = x_p(t_i)$

Since the linear system is E.A.S there exists a Lyapunov function  $V(x)$  such that the following hold.

1.  $\alpha_1 \|x\|^2 \leq V(x) \leq \alpha_2 \|x\|^2$ ,

2.  $\dot{V}(x) \leq -\alpha_3 \|x\|^2$ ,
3.  $\|\frac{\partial V}{\partial x}(x)\| \leq \alpha_4 \|x\|$ .

Consequently, the ideal system solutions starting at  $x_o$  evolves as

$$V(x_p) \leq V(x_o)e^{\lambda t} \quad \forall t \in \text{dom } x_p, \quad \forall x_o \in \mathcal{X},$$

where  $\lambda \geq \alpha_3/(2\alpha_2)$ .

[2] shows that enforcing the strategy

$$h_c(\hat{x}, t) = V(x(t)) - V(\hat{x})e^{-\lambda} \leq 0 \quad \forall t \in [0, T_s], \quad (5)$$

keeps the system (E.A.S) to the origin. This condition states that the entire duration between samples must be exponentially decreasing as compared with the previous state  $\hat{x}$ .

Note that the sampling period  $T_s$  is not known before hand. Therefore, this self-triggered control method predicts solution over a finite horizon to estimate the sampling period  $T_s$  such that (5) holds. The range this finite horizon is  $[\tau_{min}, \tau_{max}]$ , where  $\tau_{min}$  and  $\tau_{max}$  are design parameters.

Next the range is discretized into  $N_{max}$  pieces, where

$$N_{max} = \frac{\tau_{max}}{\Delta},$$

and  $\Delta$  is the discretization step.

The self-triggered sampler is then given by:

$$\Gamma_2(\hat{x}) := \max\{\tau_{min}, n(\hat{x})\Delta\} \quad (6)$$

$$n(\hat{x}) := \arg \max\{h_c(\hat{x}, n\Delta) : n \in [0, N_{max}]\} \quad (7)$$

Implementation of this method has two steps that occur at each sampling period: Computing the function  $h_c$  and evaluate  $n(\hat{x})$  over the finite future horizon.

## 5 Comparison of Methods

This section provides a comparison between the self-triggered control methods given in Section 4. The comparison uses the hybrid self-triggered control model  $\mathcal{H}_{et} = (F, G, C, D)$  as in (1) with auxiliary variables defined as  $\chi = (\tau, T_s^*)$  with dynamics  $F_\chi$  and update function  $G_\chi$  defined as

$$G_\chi(z) := \begin{bmatrix} 0 \\ \Gamma(\xi) \end{bmatrix}, \quad F_\chi(z) := \begin{bmatrix} 1 \\ 0 \end{bmatrix},$$

and an ETM triggering condition given by (3).

### 5.1 Plant Model

A double integrator model is considered with a state vector  $x_p = (x_1, x_2)^\top \in \mathbb{R}^2$  and dynamics

$$\dot{x}_p = F_p(x_p) = (x_2, u)^\top. \quad (8)$$

Further, full state feedback is assumed. I.e,  $\hat{y} = H_p(x_p) = x_p$ . For this section we take  $\hat{x} := \hat{y}$ .

### 5.1.1 Continuous Time Stability

It can be shown that the plant dynamics  $F_p$  can be stabilized through the controller

$$u = \kappa(x) = -Kx_p = -[k_1 \ k_2] \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}. \quad (9)$$

Further, by the converse Lyapunov theorem, there exists a matrix  $P$  such that  $V(x) = x^\top Px$  is a Lyapunov certificate of stability.

## 5.2 Tiberi, Johansson [3]

Recall that the sampling function given by this paper is

$$\Gamma_1(\hat{y}) = \frac{1}{2L} \ln(1 + 2\delta / \|f(\hat{y}, \kappa(\hat{y}))\|),$$

where  $L = L_{f,u}L_{\kappa,x}$ , and  $\delta > 0$ . Using the controller in (9) we have that  $L_{f,u} = 1$  and  $L_{\kappa,x} = \max k_1, k_2$ . Further, the norm of the dynamics is  $\|f(\hat{y}, \kappa(\hat{y}))\| = \sqrt{x_2^2 + u^2}$ .

$k_1$	1
$k_2$	1
$L_{f,u}$	1
$L_{\kappa,x}$	1
$\delta$	0.5
$x(0)$	$(1, 0)^\top$
$\min \ f(\hat{y}, \kappa(\hat{y}))\ $	0.01

Table 1: Simulation Parameters

The norm  $\|f(\hat{y}, \kappa(\hat{y}))\|$  is artificially saturated above zero.

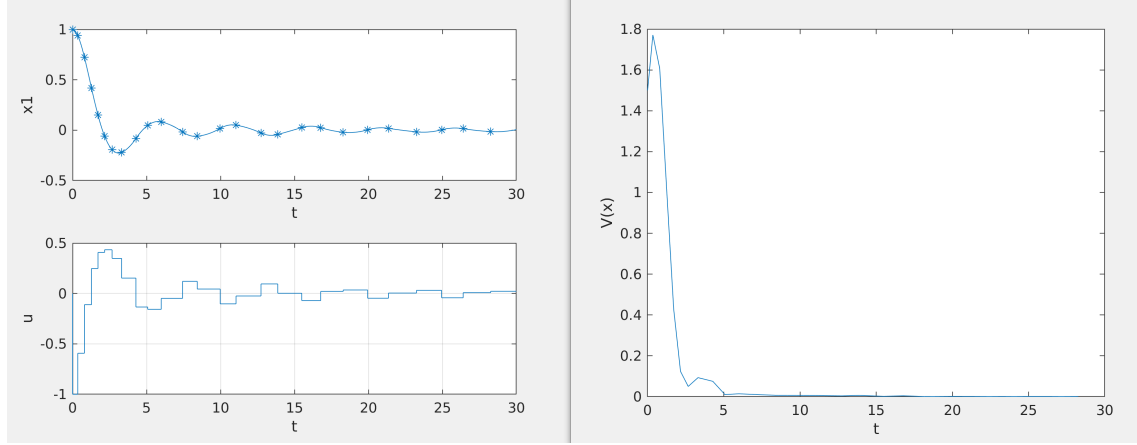


Figure 1: State Evolution, Control, and  $V(x)$

**Average Sample Time: 1.29s**  
**Average Sample Time( $t > 15$ ): 1.64s**

### 5.3 Heemels, Johansson, Tabuada [2]

Recall that the sampling function given by this paper is

$$\begin{aligned}\Gamma_2(\hat{x}) &:= \max\{\tau_{min}, n(\hat{x})\Delta\} \\ n(\hat{x}) &:= \arg \max\{h_c(\hat{x}, n\Delta) : n \in [0, N_{max}]\}\end{aligned}$$

A Lyapunov function for this system is found by applying the feedback law (9) to the system (8). Thus we have

$$F_p(x_p) = Ax_p - BKx_p = \hat{A}x_p,$$

where  $\hat{A} = A - BK$  is Hurwitz. Then, a Lyapunov function  $V(x_p) = x_p^\top Px_p$  is established by solving  $\hat{A}P + P\hat{A}^\top + I = 0$  for  $P$ . The following parameters are derived from this choice of  $V(x_p)$  and are used in simulations. Note,  $\lambda_{max}$  is the maximum eigenvalue of  $P$ .

$k_1$	1
$k_2$	1
$\lambda$	$1/(2\lambda_{max})$
$\Delta$	0.1
$\tau_{max}$	4
$\tau_{min}$	0.1
$x(0)$	$(1, 0)^\top$

Table 2: Simulation Parameters

The function  $h_c$  is calculated before every jump. This is done by seeing that the linear system (8) evolves between updates as

$$\dot{x}_p = Ax_p + BK\hat{x},$$

where  $\hat{x}$  is the last measurement taken. This solution is evaluated

The solution of this system can be written explicitly as

$$\begin{aligned}x(t) &= e^{At}x(0) - \int_0^t e^{-A\tau}BK\hat{x}d\tau \\ &= e^{At}x(0) - A^{-1}(e^{-At} - I)BK\hat{x}\end{aligned}$$

In the simulation  $A^{-1}$  is taken as approximation.

The implementation creates the matrices  $e^{A\Delta}$  and  $e^{-A\Delta}$  and uses these to propagate the states through discrete steps as in (6).

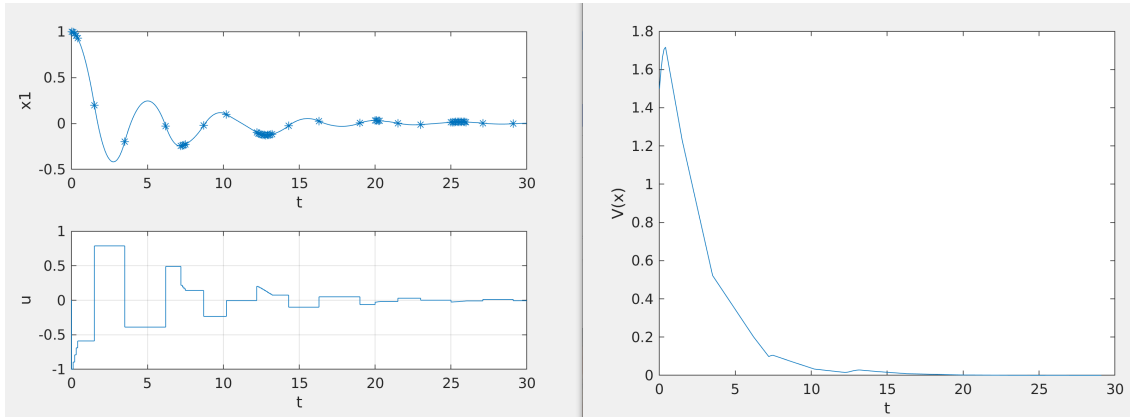


Figure 2: State Evolution, Control, and  $V(x)$

**Average Sample Time: 1.15s**

**Average Sample Time( $t > 15$ ): 1.01s**

## References

- [1] Jun Chai, Pedro Casau, and Ricardo G. Sanfelice. Analysis and design of event-triggered control algorithms using hybrid systems tools. In *2017 IEEE 56th Annual Conference on Decision and Control (CDC)*, pages 6057–6062, Melbourne, Australia, December 2017. IEEE.
- [2] W.P.M.H. Heemels, K.H. Johansson, and P. Tabuada. An introduction to event-triggered and self-triggered control. In *2012 IEEE 51st IEEE Conference on Decision and Control (CDC)*, Maui, HI, USA, December 2012. IEEE.
- [3] U. Tiberi and K.H. Johansson. A simple self-triggered sampler for perturbed nonlinear systems. *Nonlinear Analysis: Hybrid Systems*, 10, November 2013.