

MAE 144 Homework 1

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Problem 1

Link to GitHub Page:

<https://github.com/davidlim4318>

Problem 2

2a

hw1_prob2a

```
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
x =
  RR_poly with properties:
poly:   -0.6710   -2.2869   10.1755   24.1641
roots:  -5.0009   -2.0030    3.5955
      n: 3
y =
  RR_poly with properties:
poly:    0.6710    2.2869  -21.5818  -62.0409   42.6886   81.5318
roots:   -6.0000   -3.0000   -1.0000    1.2681    5.3235
      n: 5
eps = 1.0000e-10
residual = 1.1369e-13
```

2b

The controller $D(s)$ is improper. The order of the denominator $x(s)$ is less than the order of the numerator $y(s)$.

hw1_prob2b

```
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
```


If my code is correct, k needs to be $= 5$ in order for the controller $D(s)$ to be proper (i.e. semi-causal, relying only on present and past inputs). This might be sufficient for a controller operating in continuous time (i.e. an analog controller). Increasing k further should eventually make $D(s)$ strictly proper (i.e. strictly causal, only relying on past inputs).

Problem 3

Symbolic:

```
syms s z1 p1
as = [1 p1 0];
bs = [1 z1];
Ds = RR_tf(bs,as);
h = 1;
Gz = DHL_C2D_matched(Ds,h)
```

Gz =

RR_tf with properties:

num: $((\text{abs}(z1) \cdot \text{abs}(\exp(-p1) - 1)) / (\text{abs}(p1) \cdot \text{abs}(\exp(-z1) - 1))), -(\exp(-z1) \cdot \text{abs}(z1) \cdot \text{abs}(\exp(-p1) - 1)) / (\text{abs}(p1) \cdot \text{abs}(\exp(-z1) - 1))$
den: $[1, -\exp(-p1) - 1, \exp(-p1)]$

Discrete-time transfer function with $h=1$

$m=1, n=2, n_r=n-m=1$, strictly proper, $K=(\text{abs}(z1) \cdot \text{abs}(\exp(-p1) - 1)) / (\text{abs}(p1) \cdot \text{abs}(\exp(-z1) - 1))$
 $z:\exp(-z1)$
 $p:[1, \exp(-p1)]$

By hand:

```
I = imread('hw1_prob3_byhand.jpeg');
imshow(I);
```

Problem 3: solving by hand, symbolically

$$D(s) = \frac{s + z_1}{s(s + p_1)}, \quad h = 1, \quad \bar{\omega} = 0$$

Step 1: $z = e^{sh}$

$$\text{poles: } s = 0 \Rightarrow z = e^{0 \cdot 1} = 1$$

$$s = -p_1 \Rightarrow z = e^{-p_1}$$

$$\text{zeros: } s = -z_1 \Rightarrow z = e^{-z_1}$$

Step 2:

$$\text{Add zero at } s = \infty \Rightarrow z = -1$$

$$G(z) = K \cdot \frac{(z - e^{-z_1}) (z + 1)}{(z - 1)(z - e^{-p_1})}$$

include (arrow pointing to $(z + 1)$)

*immediately remove it ($\rightarrow z = \infty$)
make it strictly causal* (arrow pointing to $(z + 1)$)

Step 3:

$$\text{DCgain} \left(\frac{s + z_1}{s(s + p_1)} \right) \Big|_{s=i\bar{\omega}=0} = \text{DCgain} \left(K \cdot \frac{(z - e^{-z_1})}{(z - 1)(z - e^{-p_1})} \right) \Big|_{z=e^{i\bar{\omega}h}=1}$$

$$\lim_{s \rightarrow 0} \frac{s + z_1}{s(s + p_1)} = \lim_{z \rightarrow 1} K \cdot \frac{(z - e^{-z_1})}{(\cancel{z - 1})(z - e^{-p_1})}$$

*in the limit,
0's should cancel,
right?*

$$\frac{z_1}{p_1} = K \frac{1 - e^{-z_1}}{1 - e^{-p_1}}$$

$$K = \frac{z_1 \cdot (1 - e^{-p_1})}{p_1 \cdot (1 - e^{-z_1})}$$

$$G(z) = \frac{z_1 \cdot (1 - e^{-p_1})}{p_1 \cdot (1 - e^{-z_1})} \frac{(z - e^{-z_1})}{(z - 1)(z - e^{-p_1})}$$

Numerical:

```
z1 = 1;
p1 = 10;
as = [1 p1 0];
bs = [1 z1];
Ds = RR_tf(bs,as);
```

```
eps = 1.0000e-10
eps = 1.0000e-10
```

```
h = 1;
Gz = DHL_C2D_matched(Ds,h)
```

```
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
```

```

eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
eps = 1.0000e-10
Gz =
  RR_tf with properties:
num:    0.1582   -0.0582
den:    1.0000   -1.0000    0.0000
Discrete-time transfer function with h=    1
      m=1, n=2, n_r=n-m=1, strictly proper, K=    0.1582
      z:    0.3679
      p:    0.0000    1.0000

```

Built-in function:

```

Ds = tf(bs,as);
Gz = zpkc(c2d(Ds,h,'matched'))

```

```

Gz =

      0.15819 (z-0.3679)
      -----
      (z-1) (z-4.54e-05)

```

```

Sample time: 1 seconds
Discrete-time zero/pole/gain model.
Model Properties

```

The result $D(z)$ generated using my function has nearly the same poles, zeros, and gain as the result generated using MATLAB's built-in function. My code is better because it works with symbolic variables and can give a symbolic result, which is pretty cool.

(My code can also handle $\omega_{\text{bar}} = 0$ by effectively taking the limit and canceling out the problematic poles in the calculation of the gain, arriving to the result as MATLAB.)