MAE 144 Homework 1

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Problem 1

Link to GitHub Page:

https://github.com/davidlim4318

Problem 2

2a

hw1_prob2a

```
RR_poly with properties:
                  -2.2869
                            10.1755
                                      24.1641
        -0.6710
poly:
        -5.0009
                  -2.0030
                             3.5955
roots:
  n: 3
 RR_poly with properties:
        0.6710
                  2.2869
                           -21.5818 -62.0409
                                                42.6886
                                                          81.5318
                                     1.2681
roots:
        -6.0000
                  -3.0000
                           -1.0000
                                                 5.3235
  n: 5
residual = 1.1369e-13
```

2b

The controller D(s) is improper. The order of the denominator x(s) is less than the order of the numerator y(s).

hw1_prob2b

```
k = 5
x =
  RR_poly with properties:
       1.0e+08 *
poly:
   0.0000
             0.0000
                      -0.0491 \quad -0.1903
                                           0.5827
                                                    1.5346
roots: 1.0e+03 *
                                           2.1584
  -2.2745 -0.0050
                      -0.0020
                                 0.0031
  n: 5
 RR poly with properties:
poly: 1.0e+08 *
   0.0492
             0.1921
                      -1.3847
                                -4.3786
                                           2.2250
        -6.0000 -3.0000
                           -1.0000
roots:
                                     1.1646
                                                4.9263
  n: 5
 RR_poly with properties:
poly: 1.0e+09 *
                       0.0000
                                                                                  2.7935
   0.0000
             0.0000
                                 0.0002
                                           0.0031
                                                    0.0339
                                                              0.2377
                                                                        1.0446
                                                                                            4.2883
roots: -20.0302 - 0.0219i -20.0302 + 0.0219i -19.9885 - 0.0354i -19.9885 + 0.0354i -19.9628 + 0.0000i -6
  n: 11
residual = 1.0832e-05
```

If my code is correct, k needs to be = 5 in order for the controller D(s) to be proper (i.e. semi-causal, relying only on present and past inputs). This might be sufficient for a controller operating in continuous time (i.e. an analog controller). Increasing k further should eventually make D(s) strictly proper (i.e. strictly causal, only relying on past inputs).

Problem 3

Symbolic:

```
syms s z1 p1
        as = [1 p1 0];
        bs = [1 z1];
        Ds = RR_t(bs, as);
        h = 1;
        Gz = DHL_C2D_matched(Ds,h)
        Gz =
                 RR tf with properties:
        num: [(abs(z1)*abs(exp(-p1) - 1))/(abs(p1)*abs(exp(-z1) - 1)), -(exp(-z1)*abs(z1)*abs(exp(-p1) - 1))/(abs(p1)*abs(exp(-p1) - 1))]
        den:[1, -exp(-p1) - 1, exp(-p1)]
        Discrete-time transfer function with h=
                  \  \  \, \text{m=1, n=2, n\_r=n-m=1, strictly proper, K=(abs(z1)*abs(exp(-p1) \ -\ 1))/(abs(p1)*abs(exp(-z1) \ -\ 1))/(abs(exp(-z1) \ -\ 1))/(abs(
                  z:exp(-z1)
                  p:[1, exp(-p1)]
By hand:
        I = imread('hw1_prob3_byhand.jpeg');
        imshow(I);
```

```
Problem 3: Solving by hand, symbolically D(s) = \frac{s + z_1}{s(s + p_1)} , h = 1 , \overline{\omega} = 0
Step 1: z = e^{sh}
poles: s = 0 \Rightarrow z = e^{p_1}
zeros: s = -z_1 \Rightarrow z = e^{-z_1}
Step 2:
Add zero at s = \infty \Rightarrow z = -1
(z - e^{-z_1}) (z + 1)
(z - 1)(z - e^{-p_1})
Step 3:
DC gain <math>\left(\frac{s + z_1}{s(s + p_1)}\right) \Big|_{s = i\overline{\omega} = 0} = DC gain \left(K \cdot \frac{(z - e^{-z_1})}{(z - 1)(z - e^{-p_1})}\right) \Big|_{z = e^{i\overline{\omega}h} = 1}
\lim_{s \to 0} \frac{s + z_1}{s(s + p_1)} = \lim_{z \to 1} K \cdot \frac{(z - e^{-z_1})}{(z - 1)(z - e^{-p_1})} \qquad \text{in the limit,}
0:s should cancel,
K = \frac{z_1 \cdot (1 - e^{-p_1})}{p_1 \cdot (1 - e^{-z_1})} \frac{(z - e^{-z_1})}{(z - 1)(z - e^{-p_1})}
K = \frac{z_1 \cdot (1 - e^{-p_1})}{p_1 \cdot (1 - e^{-z_1})} \frac{(z - e^{-z_1})}{(z - 1)(z - e^{-p_1})}
```

Numerical:

0.3679

0.0000

1.0000

z:

p:

Built-in function:

Model Properties

The result D(z) generated using my function has nearly the same poles, zeros, and gain as the result generated using MATLAB's built-in function. My code is better because it works with symbolic variables and can give a symbolic result, which is pretty cool.

(My code can also handle omega_bar = 0 by effectively taking the limit and canceling out the probelmatic poles in the calculation of the gain, arriving to the result as MATLAB.)