

$$\begin{aligned}
 d(t) &= \underline{2A_c} \cdot m(t) \cdot \underbrace{\cos(\omega_c t)}_y \cdot \underbrace{\cos(\omega_c t + \theta(t))}_x \\
 &= A_c m(t) \cdot [\cos(2\omega_c t + \theta(t)) + \cos(\theta(t))] \\
 &= \cancel{A_c \cdot m(t) \cdot \cos(2\omega_c t + \theta(t))} + A_c m(t) \cdot \cos(\theta(t))
 \end{aligned}$$

$2\cos x \cos y = \cos(x+y) + \cos(x-y)$

### Transmitted Signals in SSB

$$x_c(t) = \frac{1}{2} A_c m(t) \cdot \cos(2\pi f_c t) \quad \text{+} \quad \frac{1}{2} A_c \cdot \hat{m}(t) \cdot \sin(2\pi f_c t)$$

LSB signal

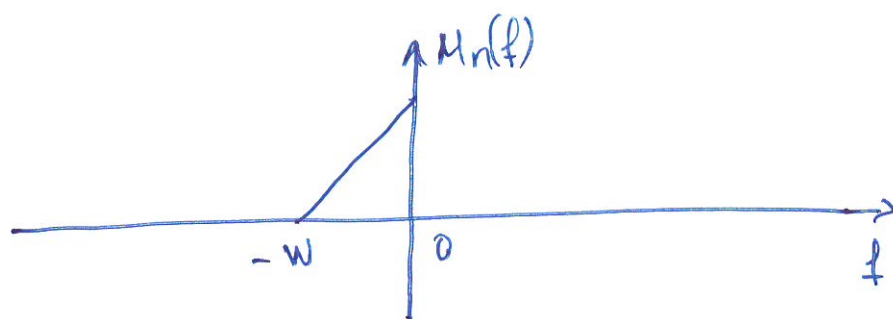
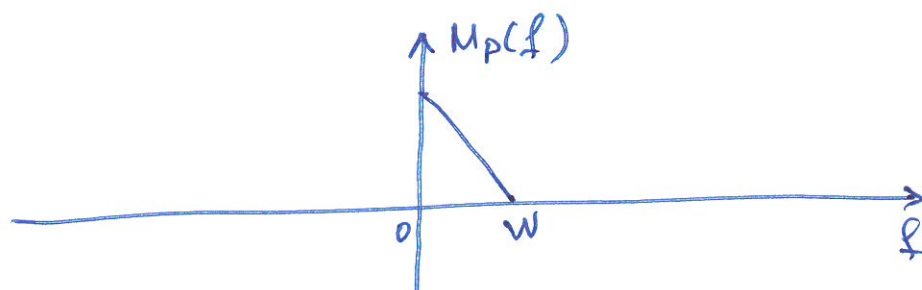
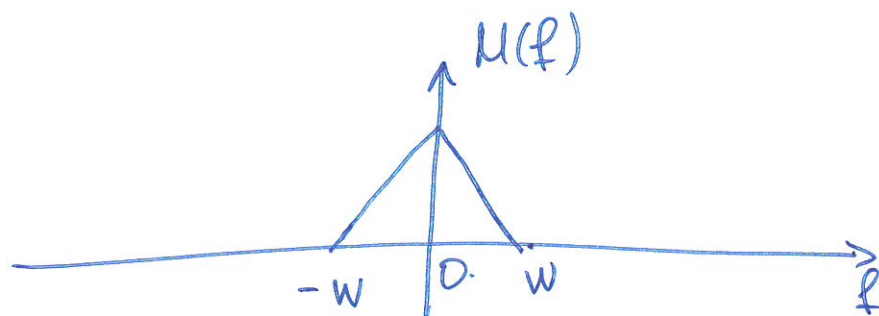
$$x_c(t) = \frac{1}{2} A_c m(t) \cos(2\pi f_c t) \quad \text{-} \quad \frac{1}{2} A_c \cdot \hat{m}(t) \cdot \sin(2\pi f_c t)$$

USB signal

B)

$$U_p(f) = \frac{1}{2} \mathcal{F} \{ m(t) + j \hat{m}(t) \}$$

$$U_n(f) = \frac{1}{2} \mathcal{F} \{ m(t) - j \hat{m}(t) \}$$



By definition :

$$\bar{X}_c(f) = \frac{1}{2} A_c M_p(f - f_c) + \frac{1}{2} A_c M_n(f + f_c)$$

$\Downarrow$

$$X_c(t) = \frac{1}{4} A_c [\underline{m(t)} + j\underline{\hat{m}(t)}] \cdot e^{j2\pi f_c t} + \frac{1}{4} A_c [\underline{m(t)} - j\underline{\hat{m}(t)}] e^{-j2\pi f_c t}$$

$$= \frac{1}{4} A_c \underline{m(t)} [e^{j2\pi f_c t} + e^{-j2\pi f_c t}] + j \frac{1}{4} A_c \underline{\hat{m}(t)} [e^{j2\pi f_c t} - e^{-j2\pi f_c t}]$$

$$= \frac{1}{2} A_c \cdot m(t) \cdot \cos(2\pi f_c t) - \frac{1}{2} A_c \cdot \hat{m}(t) \cdot \sin(2\pi f_c t)$$

USB

# Demodulation

$$x_c(t) = x_r(t)$$

3

$$A) d(t) = \left[ \frac{1}{2} A_c m(t) \cos(2\pi f_c t) \pm \frac{1}{2} A_c \hat{m}(t) \cdot \sin(2\pi f_c t) \right] \cdot$$

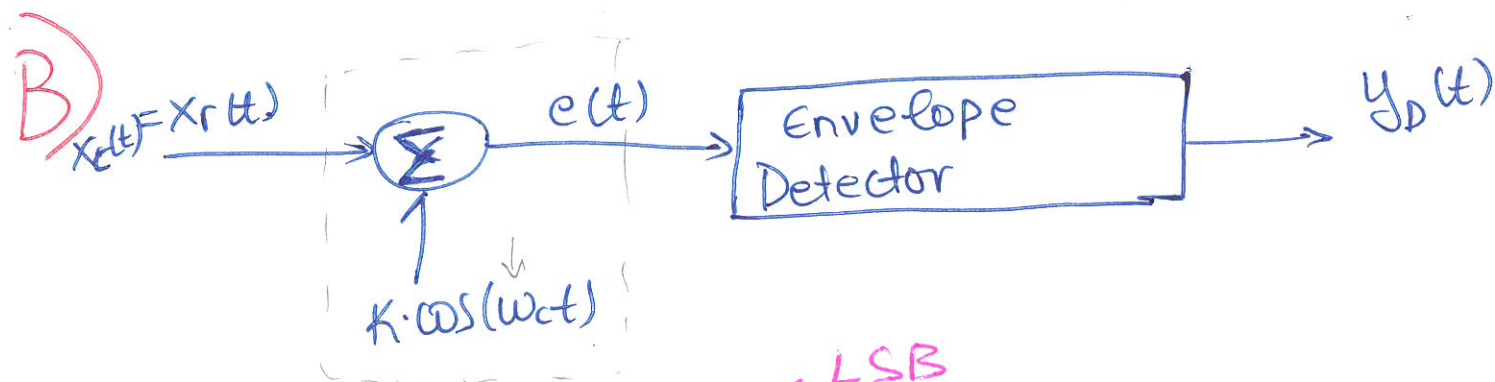
$$2 \cdot 2 \cdot 4 \cos(2\pi f_c t + \theta(t))$$

demodulation carrier

$$= A_c m(t) \cdot \cos(\theta(t)) + A_c \cdot m(t) \cdot \cos[4\pi f_c t + \theta(t)]$$

$$\mp A_c \hat{m}(t) \cdot \sin(\theta(t)) \pm A_c \hat{m}(t) \cdot \sin[4\pi f_c t + \theta(t)]$$

$$y_D(t) = \underbrace{A_c m(t) \cdot \cos(\theta(t))}_{\text{message signal}} \mp \underbrace{A_c \hat{m}(t) \cdot \sin(\theta(t))}_{\text{distortion}} \quad \text{Output signal}$$

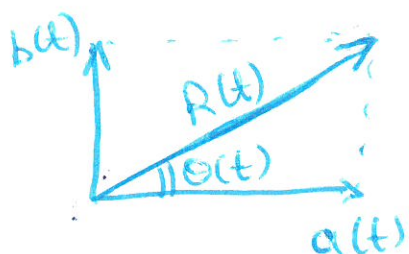


$$x_r(t) = \frac{1}{2} A_c m(t) \cos(2\pi f_c t) \pm \frac{1}{2} A_c \hat{m}(t) \cdot \sin(2\pi f_c t)$$

LSB (under the sine term)  
USB (under the cosine term)

$$e(t) = \underbrace{\left( \frac{1}{2} A_c m(t) + k \right) \cos(2\pi f_c t)}_{a(t)} \pm \underbrace{\frac{1}{2} A_c \hat{m}(t) \cdot \sin(2\pi f_c t)}_{b(t)}$$





$$a(t) = R(t) \cdot \cos(\theta(t))$$

$$b(t) = R(t) \sin(\theta(t))$$

$$e(t) = a(t) \cdot \cos(2\pi f_c t) \pm b(t) \sin(2\pi f_c t)$$

$$e(t) = R(t) [\cos(\theta(t)) \cdot \cos(2\pi f_c t) \pm \sin(\theta(t)) \cdot \sin(2\pi f_c t)]$$

$$e(t) = R(t) \cdot \cos[2\pi f_c t + \theta(t)]$$

$$R(t) = \sqrt{a^2(t) + b^2(t)}$$

$$y_D(t) = \sqrt{\left(\frac{1}{2} A_c m(t) + K\right)^2 + \left(\frac{1}{2} A_c \hat{m}(t)\right)^2}$$

$$\text{If } \frac{1}{2} A_c m(t) + K \gg \frac{1}{2} A_c \hat{m}(t).$$

$$\Rightarrow y_D(t) \approx \frac{1}{2} A_c m(t) + K$$

Example:

$$m(t) = \cos(2\pi f_1 t) - 0.4 \cdot \cos(4\pi f_1 t) + 0.9 \cos(6\pi f_1 t)$$

$$\hat{m}(t) = \sin(2\pi f_1 t) - 0.4 \cdot \sin(4\pi f_1 t) + 0.9 \sin(6\pi f_1 t)$$

$$x_c(t) = \frac{1}{2} A_c m(t) \cos(2\pi f_c t) \pm \frac{1}{2} A_c \hat{m}(t) \sin(2\pi f_c t)$$

LSB (pointing to the plus sign)  
USB (pointing to the minus sign)

$$x_c(t) = R(t) \cdot \cos(2\pi f_c t + \theta(t))$$

$$(L < \frac{1}{2} A_c m(t)) \quad \textcircled{S}$$

B) 
$$R(t) = \frac{A_c}{2} \sqrt{m^2(t) + \hat{m}^2(t)}$$

$$\tan \theta(t) = \frac{b(t)}{a(t)}$$

$$\theta(t) = \tan^{-1} \left( \frac{\frac{1}{2} A_c \hat{m}(t)}{\frac{1}{2} A_c m(t)} \right)$$

Phase:  $2\pi f_c t + \theta(t)$

$$\frac{d}{dt} [2\pi f_c t + \theta(t)] = 2\pi f_c \pm \frac{d}{dt} \left[ \tan^{-1} \left( \frac{\hat{m}(t)}{m(t)} \right) \right]$$

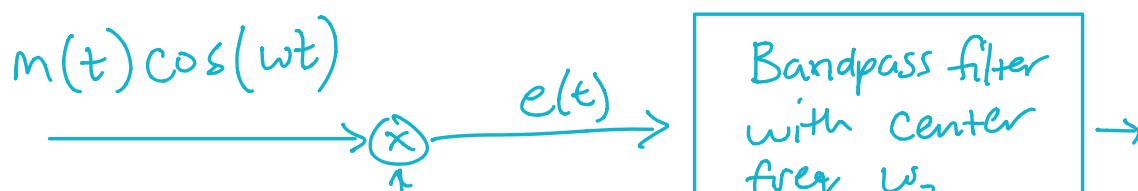
## Two kinds of Modulation



TDMA - Time-delay Modulation

## Translation of frequency

OFDMA & FDMA use this method<sup>\*</sup>  
(will learn in Digital systems)



$$\begin{array}{c}
 \text{local oscillator} \\
 2 \cos[(\omega_1 + \omega_2)t]
 \end{array}
 \quad
 \begin{array}{c}
 \boxed{\phantom{m(t) \cos(\omega_2 t)}} \\
 x_c(t) = \\
 m(t) \cos(\omega_2 t)
 \end{array}$$

amplitude modulation: [...]

pulse position: keep initial position, send info in different positions at same pulse  
which time position

pulse width: info captured in width of signal

## Pulse Modulation

Pulse Ampl. Mod. (P.A.M.)

Width (P.W.M.)

Position (P.P.M.)

↓

$$h(t) = \Pi\left(\frac{t - \frac{1}{2}\tau}{\tau}\right)$$

↓

$$m_c(t) = m_f(t) * h(t)$$

$$= m(nT_s) \cdot \Pi\left(\frac{(t - nT_s) + \frac{1}{2}\tau}{\tau}\right)$$

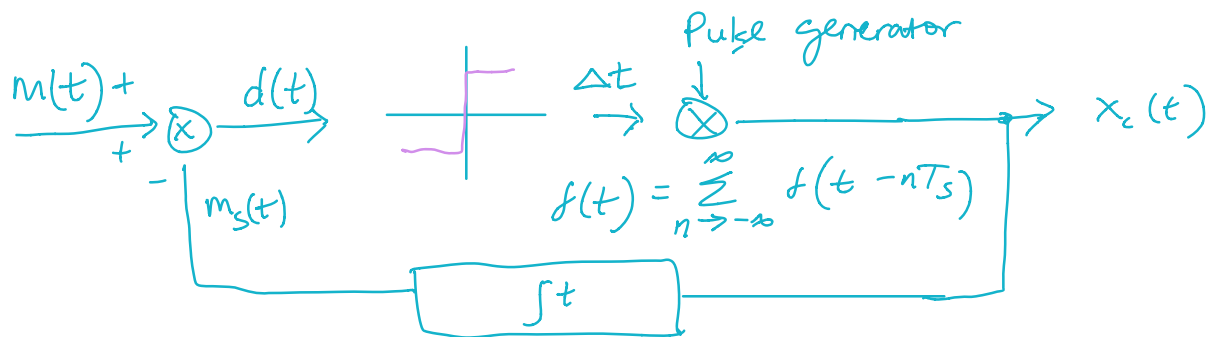
## Digital Pulse Modulation

✓  
Delta Mod

↓  
Pulse Code Mod

Compare delta 1 samples that we took and see how the deviate

limiter sets the boundaries of signal



$$d(t) = m(t) - m_s(t)$$

$$x_c(t) = \Delta(nT_s) \cdot \delta(t - nT_s)$$

$$m_s(t) = \Delta(nT_s) \cdot \int^t f(x - nT_s) dx$$



## Pulse Code Modulation

(Encoding = mapping amplitude to a specific code)



## Time-Division Multiplexing

| message signal

(1 user transmitting)

timeslot

Cont. Wave Mod

Pulse Mod

|  
Linear

|  
Angular

# Angle Mod. Techniques

↙  
PM

↘  
FM

general signal:

$$x_c(t) = A_c \cos [2\pi f_c t + \phi(t)]$$

Instant. phase |  $\theta(t) = 2\pi f_c t + \phi(t)$  — phase deviation

Instant. freq |  $f(t) \triangleq \frac{1}{2\pi} \frac{d\theta(t)}{dt}$

$$= f_c + \frac{1}{2\pi} \left( \frac{d\phi(t)}{dt} \right) \text{ — freq. deviation}$$

Phase mod |  $\phi(t) = k_p \cdot m(t)$