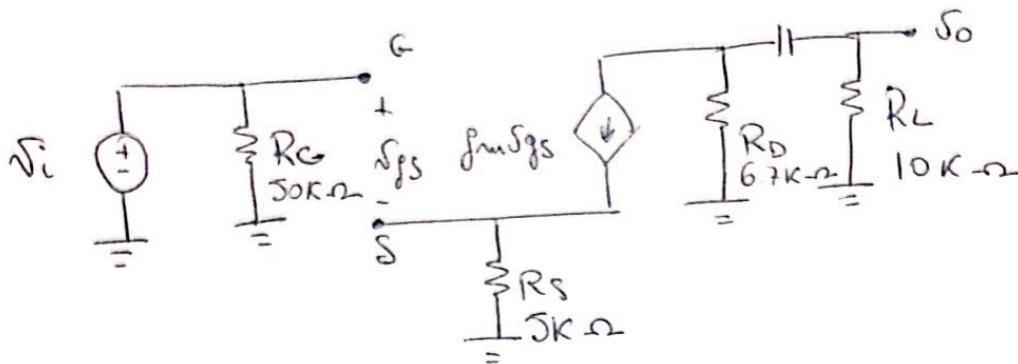


Corner frequencies of a CS amplifier with an output coupling capacitor

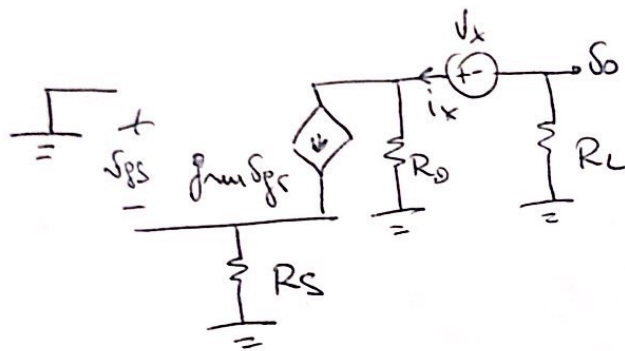
$$f_c = \frac{1}{2\pi \tau_c} ; \tau_c = C_c R_{eqc}$$

a) Sketch the low frequency small-signal circuit



b) Shut-down all the independent source and

c) replace the capacitor with a probe source $V_x(i_x)$.



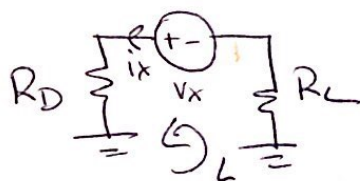
d) Calculate R_{eqc} as $\frac{V_x}{i_x}$

$$R_{eqc} = \frac{V_x}{i_x}$$

$$v_{gs} = 0 \Rightarrow v_{gs} = -v_s = -R_S g_m v_{gs}$$

$$v_s = R_S g_m v_{gs} \Rightarrow v_s = 0$$

Thus the circuit become



$$\text{KVL @ L: } -R_L i_x + V_x - R_D i_x = 0 \Rightarrow$$

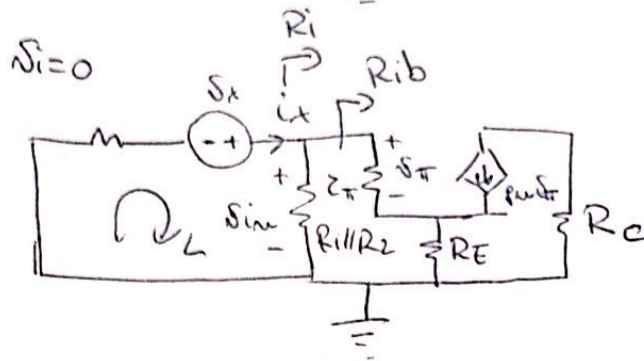
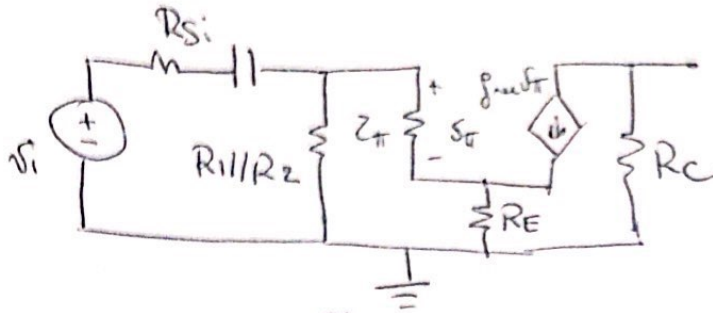
$$\Rightarrow \frac{V_x}{i_x} = R_D + R_L \Rightarrow R_{eqc} = R_D + R_L$$

$$f_c = \frac{1}{2\pi (R_D + R_L) C_c}$$

Cut-off frequencies for a CE amplifier with an input coupling capacitor and a load capacitor.

$$f_L = \frac{1}{2\pi \tau_c} = \frac{1}{2\pi R_{eq} C_c} \quad ; \quad f_H = \frac{1}{2\pi \tau_c} = \frac{1}{2\pi R_{eqL} C_L}$$

f_L



$$R_{eq} = \frac{v_x}{i_x}$$

KVL @ L

$$v_x = R_{Si} i_x + R_i i_x$$

$$R_{eq} = \frac{v_x}{i_x} = R_{Si} + R_i$$

$$R_i = R_1 || R_2 || R_{ib}$$

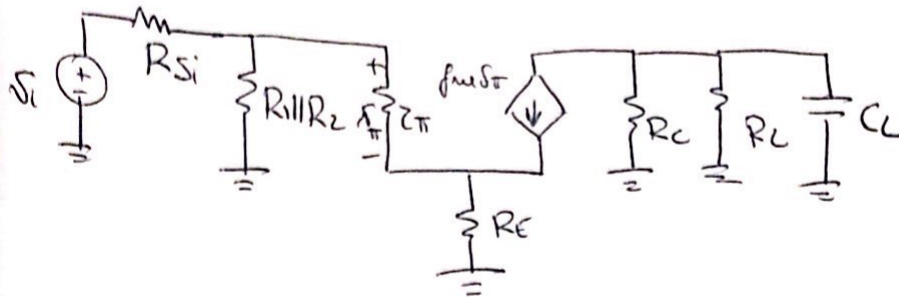
$$R_{ib} = Z_{\pi} + R_E (1 + \beta)$$

$$R_{eq} = R_{Si} + [R_1 || R_2 || [Z_{\pi} + R_E (1 + \beta)]]$$

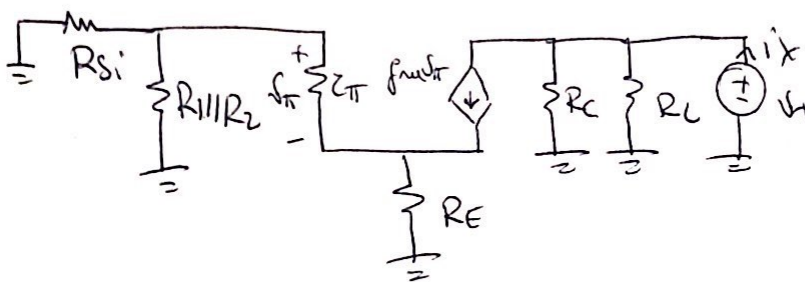
$$\tau_c = C_c [R_{Si} + R_1 || R_2 || [Z_{\pi} + R_E (1 + \beta)]]$$

$$f_c = f_L = \frac{1}{2\pi C_c \left[R_{Si} + \left[R_1 \parallel R_2 \parallel (Z_\pi + R_E(1+\beta)) \right] \right]}$$

$$f_H = \frac{1}{2\pi C_L} = \frac{1}{2\pi C_L R_{eqL}}$$



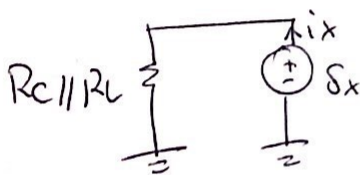
$$S_i = 0$$



$$\begin{aligned} V_\pi &= Z_\pi i_x = \\ &= Z_\pi \cdot g_m V_\pi \cdot \frac{R_E}{R_1 \parallel R_2 \parallel R_{Si} + Z_\pi + R_E} \end{aligned}$$

Hence $V_\pi = 0$ and $g_m V_\pi = 0$

As the dependent current source is determining the voltage drop that controls the current source, $g_m V_\pi = 0$. Thus, the circuit reduces to



$$R_{eqL} = R_C \parallel R_L$$

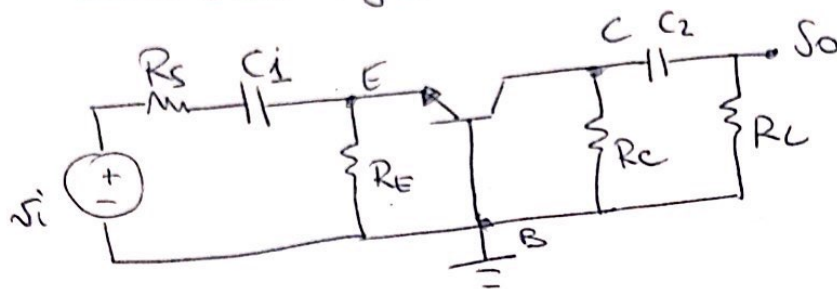
$$f_H = \frac{1}{2\pi C_L (R_C \parallel R_L)}$$

Cut-off frequency of a circuit with 2 coupling capacitors.

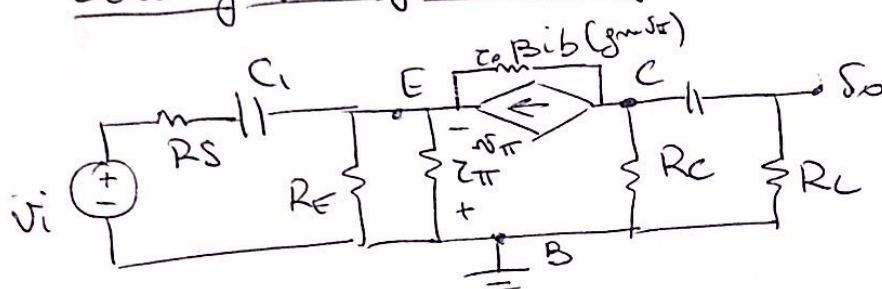
$$f_L = \frac{\omega_L}{2\pi} \quad \omega_L \approx \sum_{i=1}^n \frac{1}{R_i C_i} \quad \text{where } n \text{ is the}$$

number of capacitors affecting the frequency response at low frequency and $R_i C_i$ is the time constant associated with capacitor C_i when all other capacitors are replaced with short circuits

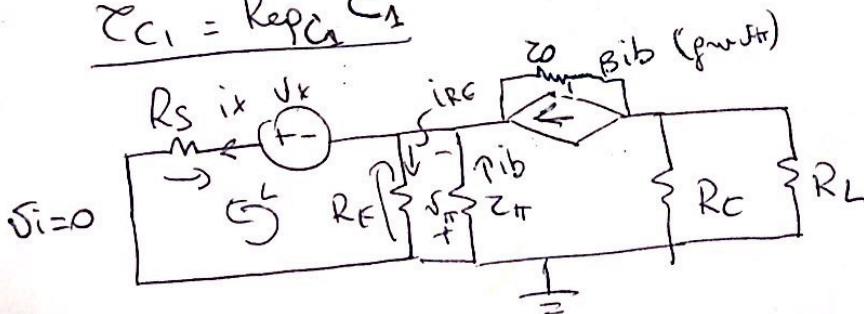
Low-frequency AC circuit



Low-frequency Small-signal circuit



$$Z_{C_1} = R_{eqC_1} C_1$$

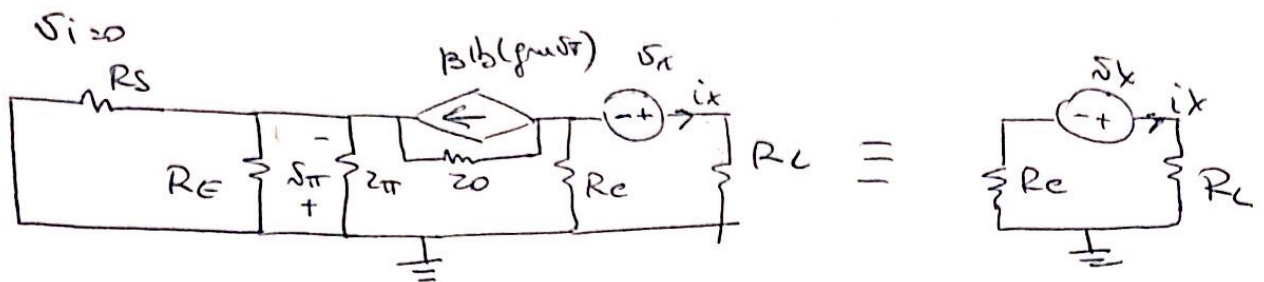


$$R_{eqC_1} = \frac{v_x}{i_x} \quad \text{see lecture 15}$$

$$R_{eqC_1} = R_s + R_{inCB} \approx R_s + \left(R_E \parallel \frac{r_{\pi}}{1+\beta} \right) \quad \text{see lecture 15}$$

$$Z_{C_1} = \left[R_s + \left(R_E \parallel \frac{r_{\pi}}{1+\beta} \right) \right] C_1$$

$$\underline{Z_{C2} = R_{eqC2} C_2}$$



$$R_{eqC2} = \frac{v_x}{i_x} = R_L + \underbrace{R_{outCB}}_{\text{See lecture 15}} \approx R_L + \underbrace{R_C}_{\text{See lecture 15}}$$

$$Z_{C2} = C_2 (R_L + R_C)$$

$$\omega_L \approx \frac{1}{\left[R_S + R_E \parallel \frac{z_{\pi}}{\beta + 1} \right] C_1} + \frac{1}{(R_L + R_C) C_2}$$

The smallest capacitor will dominate the low frequency response as all resistors have comparable values, typically.