

# ECE 322L

## Electronics 2

03/03/20- Lecture 12

Biasing the BJT 1

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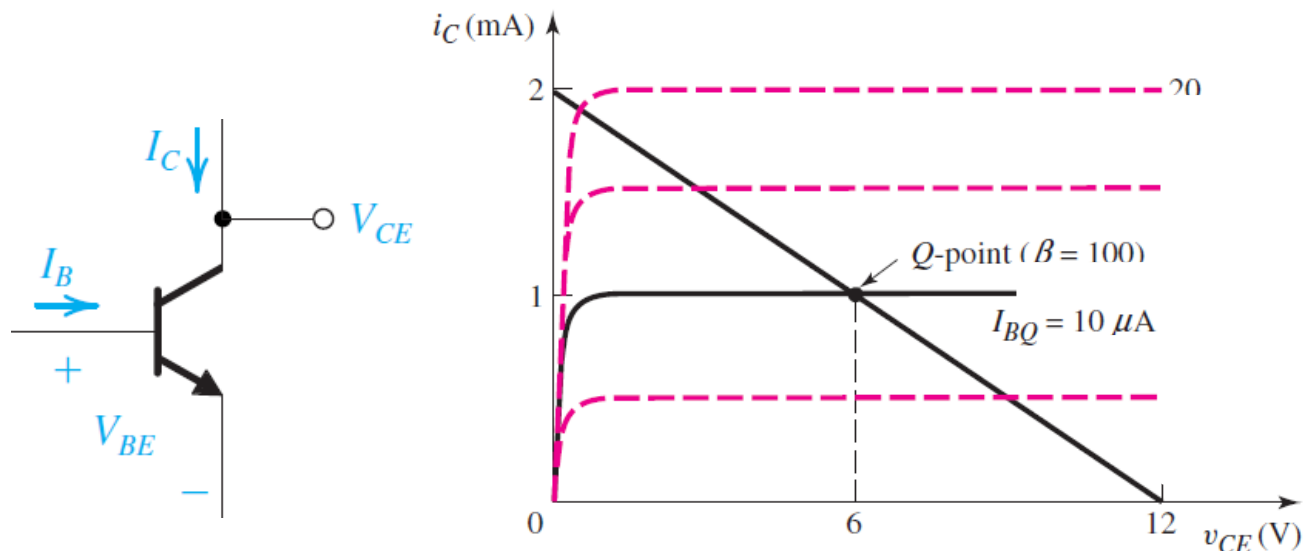
# Updates and Overview

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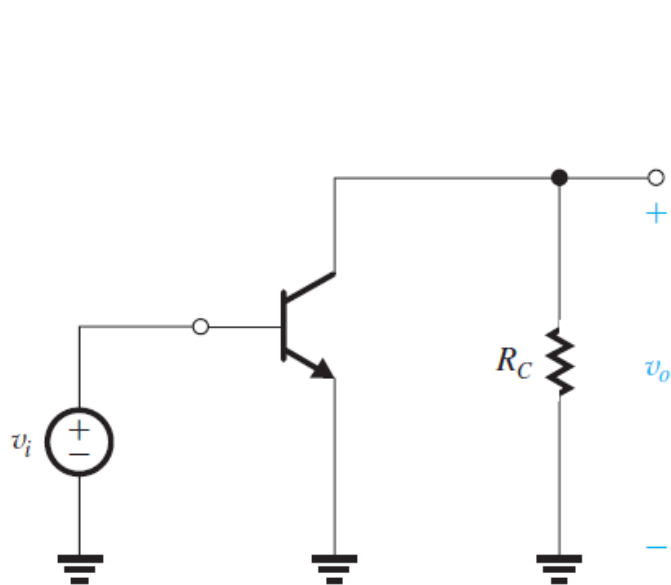
- Biasing the BJT (Neamen 5.2 and 5.4, S&S 6.7)

# Biassing the BJT

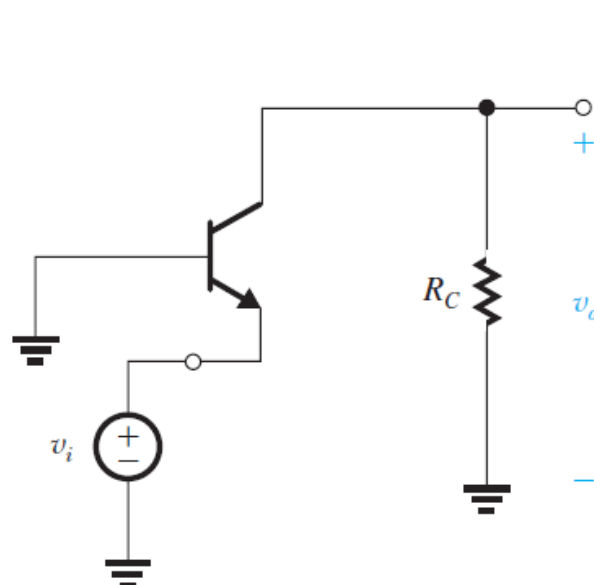
- Biasing: Establishing a Q point for the BJT
- The Q point is defined by the DC values of the input and output current and voltages. In CE configuration the latter are:  $I_{CQ}$ ,  $I_{BQ}$  ( $V_{BEQ}$ ), and  $V_{CEQ}$ .
- Desired characteristics of the Q point for an amplifier:
  - it is in the forward-active region;
  - it is insensitive to variations of temperature and  $\beta$ ;
  - it allows for maximum (or required) output signal swing and maximum (or required gain).



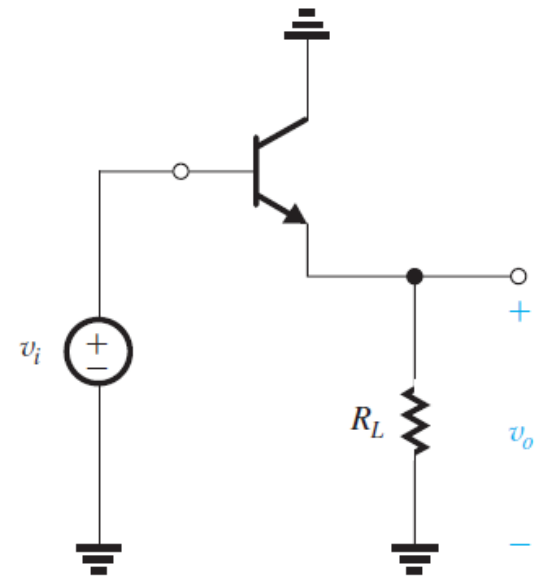
# BJT amplifier configurations



(a) Common-Emitter (CE)

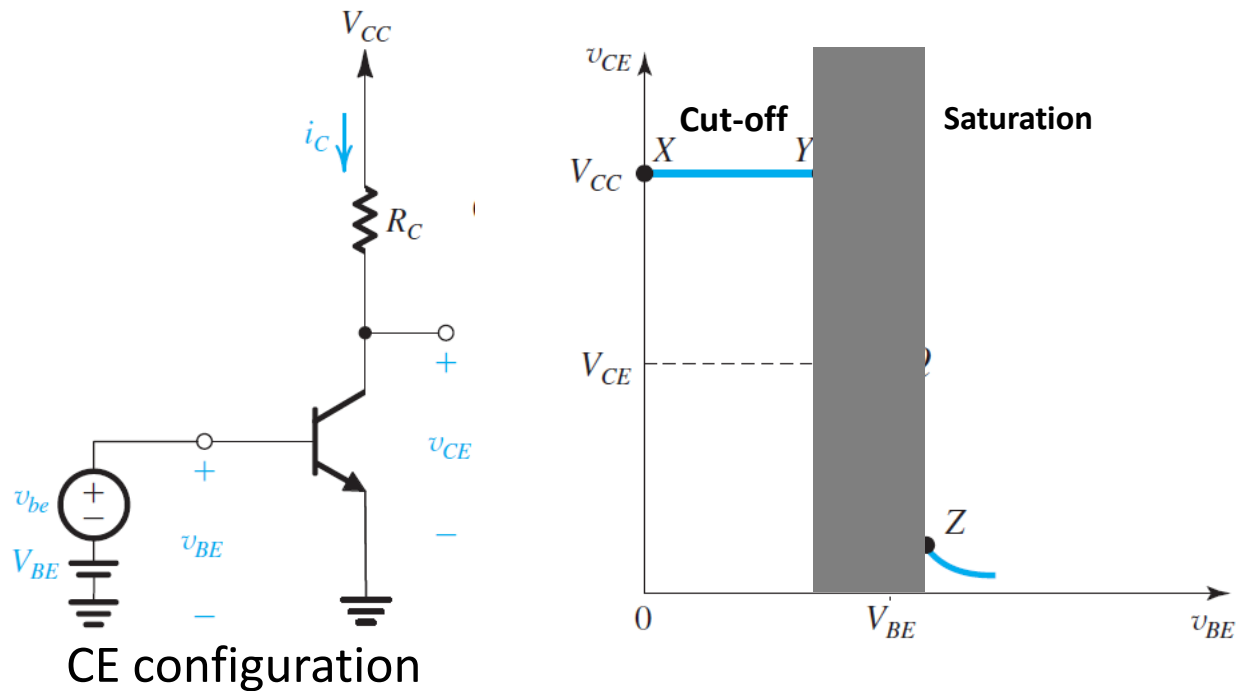


(b) Common-Base (CB)



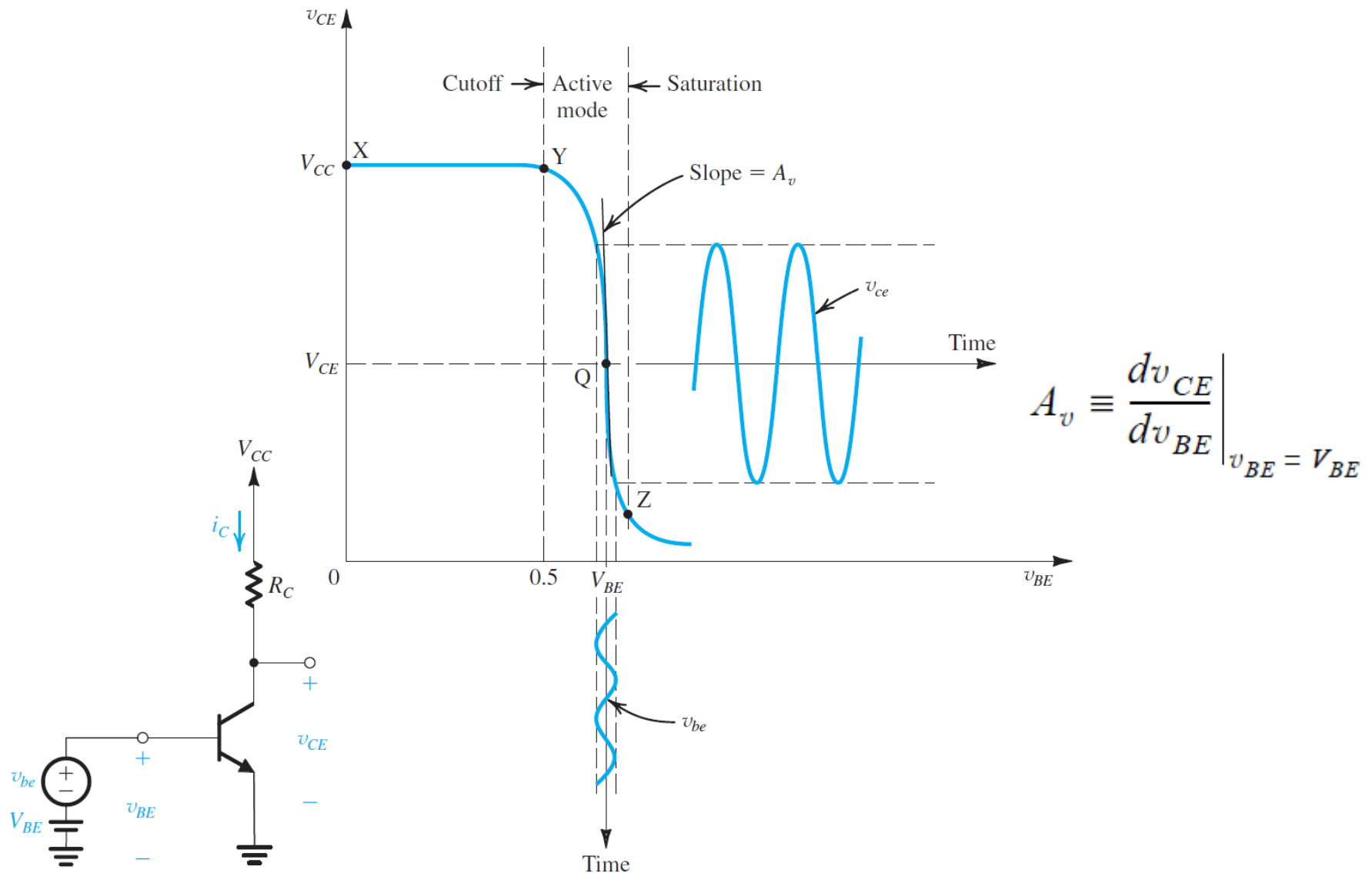
(c) Common-Collector (CC)  
or Emitter Follower

# Voltage-transfer characteristics (VTC)

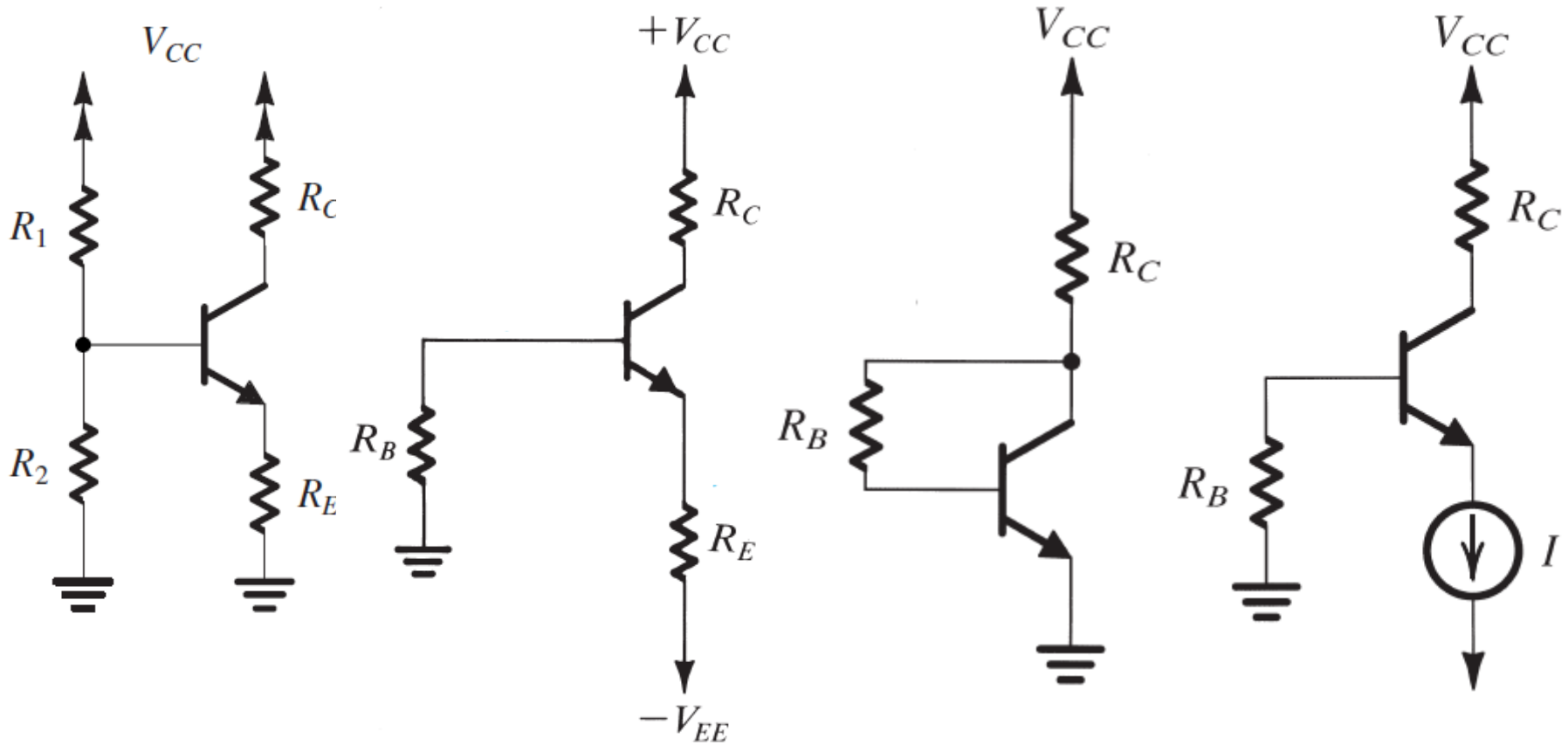


CE configuration: VTC is a  $V_{CE}$ - $V_{BE}$  plot  
CB configuration: VTC is a  $V_{CB}$ - $V_{EB}$  plot  
CC configuration: VTC is a  $V_{BC}$ - $V_{EC}$  plot

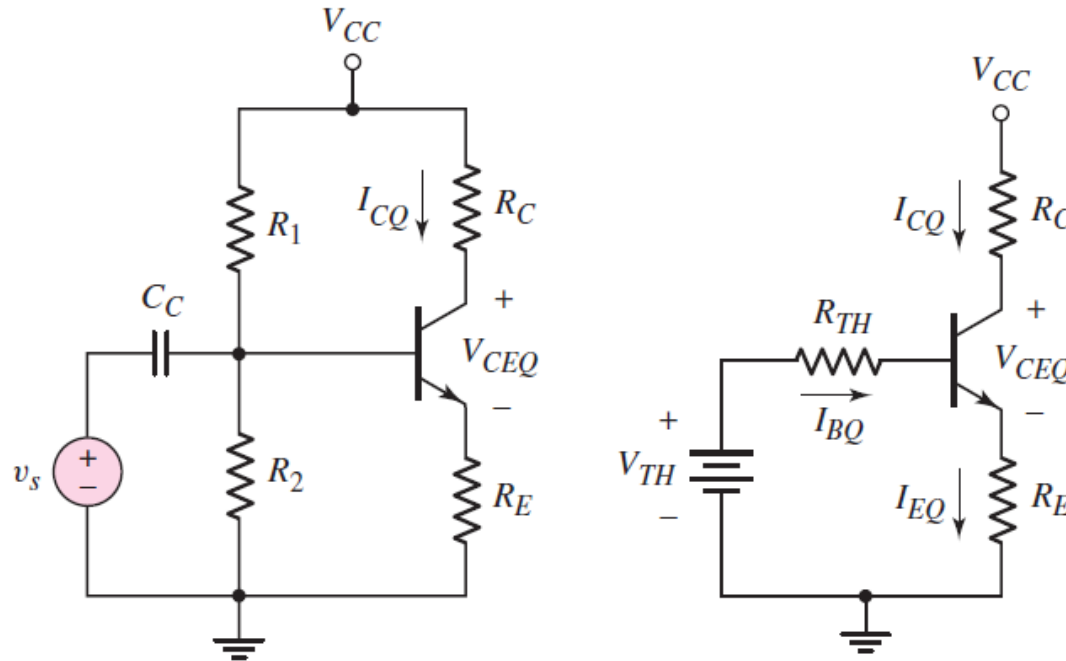
# Voltage-transfer characteristics (VTC)



# Biasing configurations



# Biasing configurations (1)



Thevenin equivalent of the voltage divider  $V_{TH} = [R_2/(R_1 + R_2)]V_{CC}$   $R_{TH} = R_1 \parallel R_2$

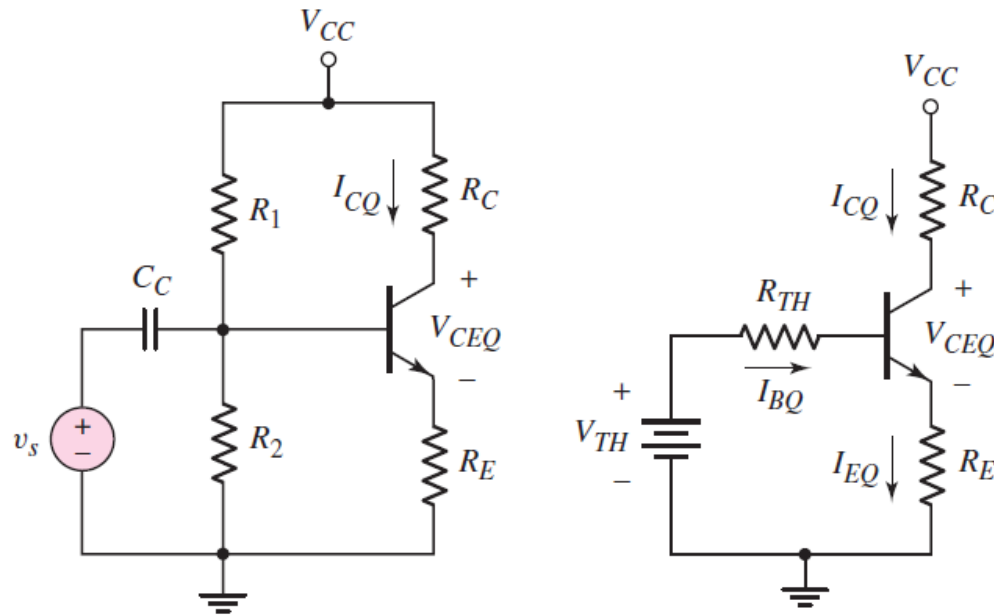
KVL on the input loop  $V_{TH} = I_{BQ}R_{TH} + V_{BE(on)} + I_{EQ}R_E$

Assuming operation in active region  $I_{EQ} = (1 + \beta)I_{BQ}$

$$I_{BQ} = \frac{V_{TH} - V_{BE(on)}}{R_{TH} + (1 + \beta)R_E} \quad I_{CQ} = \beta I_{BQ} = \frac{\beta(V_{TH} - V_{BE(on)})}{R_{TH} + (1 + \beta)R_E}$$



# Biasing configurations (1)



$$V_{TH} = [R_2 / (R_1 + R_2)] V_{CC} \quad R_{TH} = R_1 \parallel R_2$$

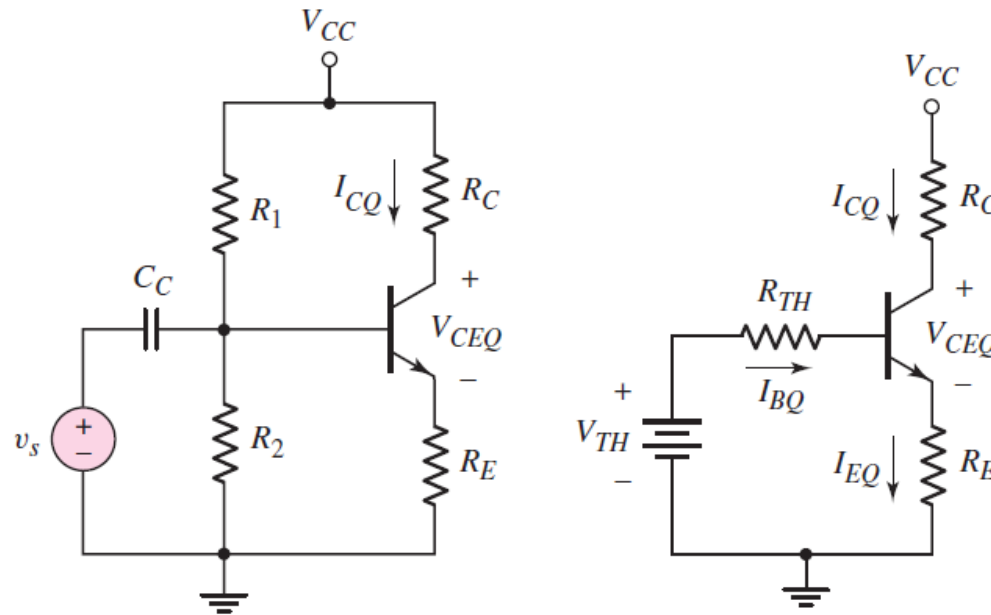
$$I_{CQ} = \beta I_{BQ} = \frac{\beta (V_{TH} - V_{BE(on)})}{R_{TH} + (1 + \beta) R_E}$$

$$R_{TH} \ll (1 + \beta) R_E \cdot I_{CQ} \cong \frac{\beta (V_{TH} - V_{BE(on)})}{(1 + \beta) R_E}$$

$$\beta \gg 1$$

$$\beta / (1 + \beta) \cong 1 \quad I_{CQ} \cong \frac{(V_{TH} - V_{BE(on)})}{R_E}$$

# Biassing configurations (1)

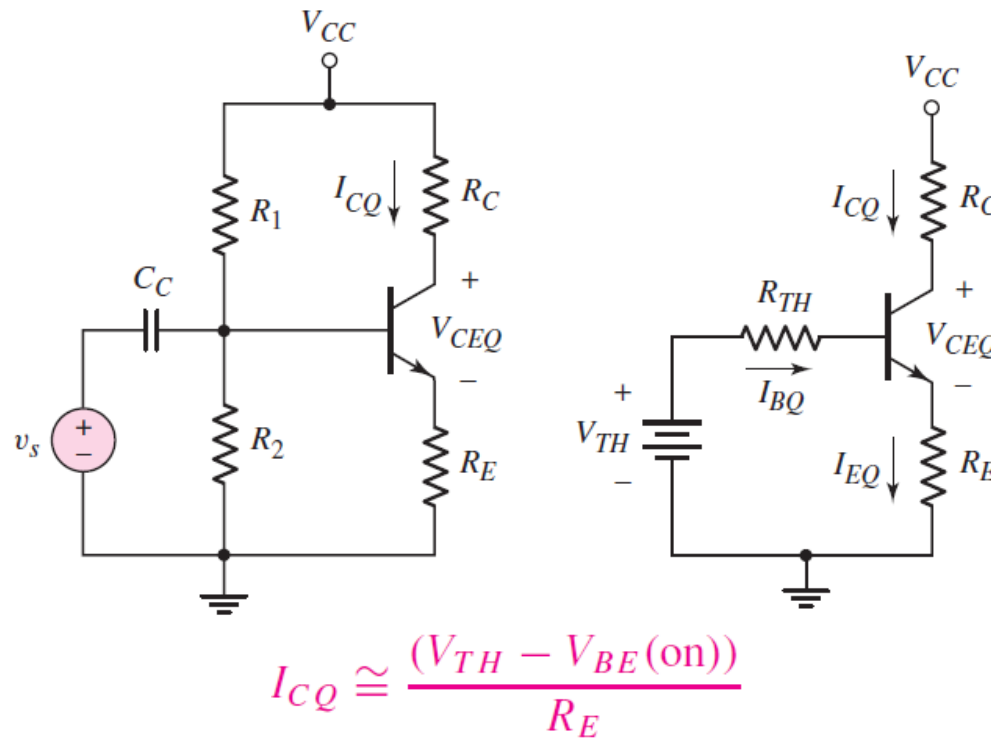


$$I_{CQ} \cong \frac{(V_{TH} - V_{BE(on)})}{R_E}$$

$V_{TH} \gg V_{BE(on)}$  guarantees stability of the Q point against temperature variation which reflects in a change of the  $V_{BE(on)}$ .

$R_{TH} \ll (1 + \beta)R_E$  guarantees stability of the Q point against variations of  $\beta$

# Biasing configurations (1)

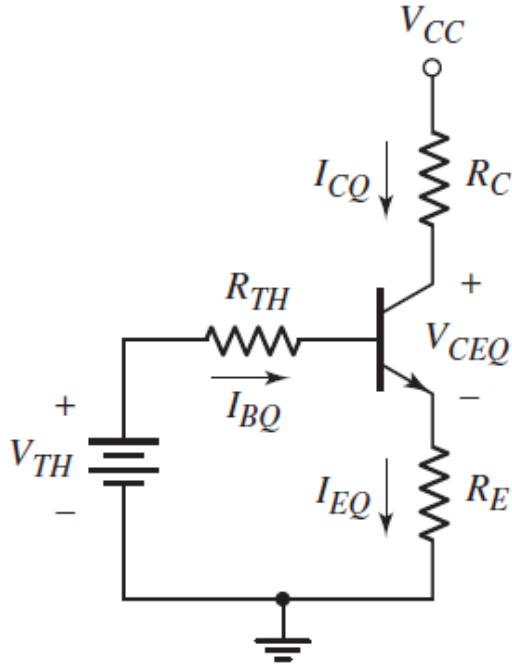


Mechanism by which  $R_E$  stabilizes  $I_C$  with respect to temperature:

- Increasing temperature will increase the emitter current.
- Increasing the emitter current will increase the voltage drop across  $R_E$  and decrease the  $V_{BE}$ .
- A decreasing  $V_{BE}$  will decrease the emitter current and the collector current.

# Biasing configurations (1)

## Examples of trade-off conditions



$V_{TH}$	Q point	Voltage gain	Output signal swing
High	Stable	Low	Limited by cut-off
Low	Not stable	High	Limited by saturation

$R_{TH}$	Q point	Input resistance
High	Not stable	High
Low	Stable	Low

Rule of thumb for a Stable Q point in the active region.

$$V_{TH} \cong \frac{1}{3} V_{CC} \quad R_{TH} \cong 0.1(1 + \beta)R_E$$

# In-class problem 1

In the circuit shown below, let  $V_{CC}=5$  V,  $R_E=0.2$  k $\Omega$ ,  $R_C=1$  k $\Omega$ ,  $\beta=150$ ,  $V_{BE(on)}=0.7$  V. Design a bias-stable circuit such that the Q point is in the center of the forward active region.

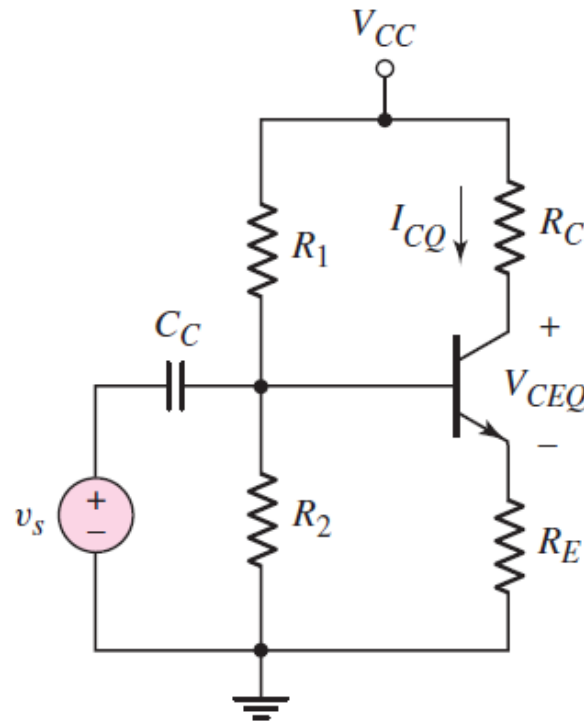
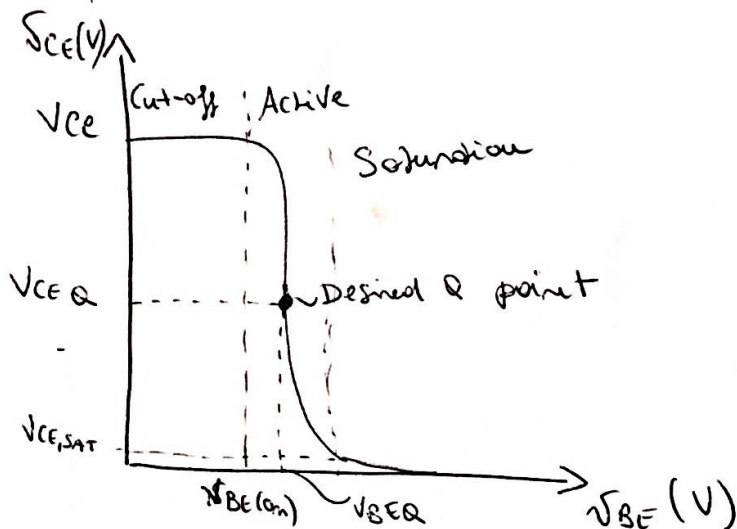
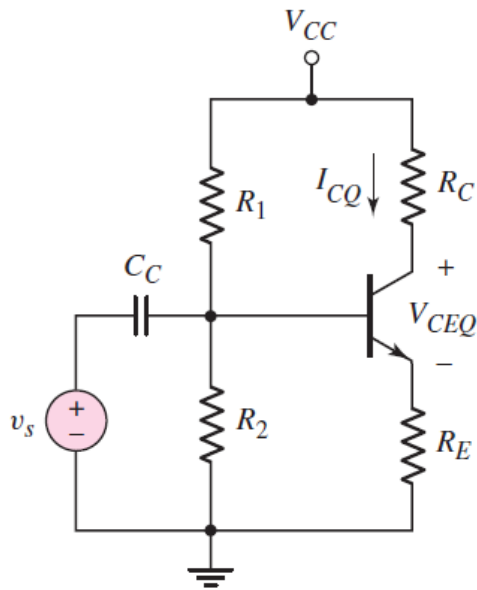


Figure 5.54 (a)

In addition sketch the input and output I-V characteristics and the input and output load lines.

# In-class problem 1, solution



$$V_{CEQ} = \left| \frac{V_{CC} - V_{CEsat}}{2} \right| + V_{CEsat}$$

$$V_{CEQ} = \left| \frac{5 - 0.2}{2} \right| + 0.2 = 2.6 \text{ V}$$

$$I_{CQ} = \frac{V_{CC} - V_{CEQ}}{R_E + R_C} = \frac{5 - 2.6}{(1 + 0.2) \text{ k}} = \frac{2.4}{1.2 \text{ k}} = 2 \mu\text{A}$$

$$I_{BQ} = \frac{I_{CQ}}{\beta} = \frac{2 \mu\text{A}}{150} = 13.3 \mu\text{A} \quad (0.0133 \mu\text{A})$$

$$R_{TH} = (0.1)(151)R_E = 0.1(151)(0.2) = 3.02 \text{ k}\Omega$$

$$V_{TH} = \frac{R_2}{R_1 + R_2} V_{CC} = \frac{1}{R_1} R_{TH} V_{CC} = \frac{1}{R_1} (3.02 \text{ k})(5)$$

KVL on the input loop:

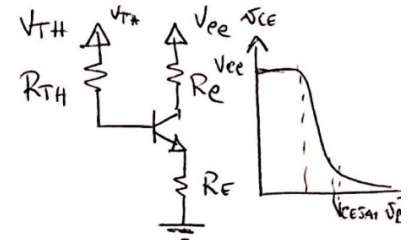
$$V_{TH} = I_{BQ} R_{TH} + V_{BE(on)} + \underbrace{(1 + \beta) I_{BQ} R_E}_{I_{CQ} R_E}$$

$$\frac{1}{R_1} (3.02 \text{ k})(5) = (0.0133)(3.02 \text{ k}) + 0.7 + (151)(0.0133)(0.2 \text{ k})$$

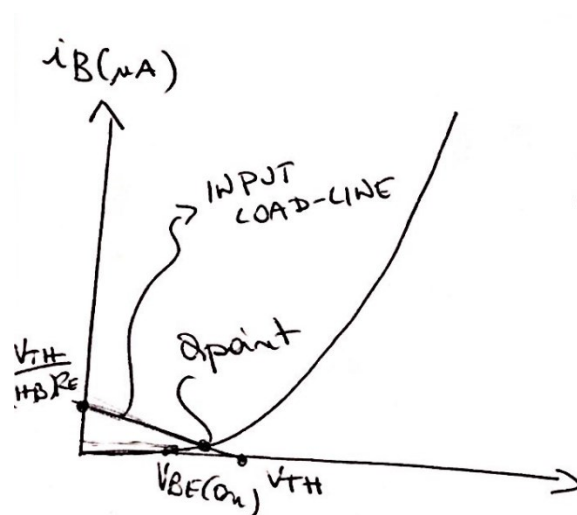
$$R_1 = 13 \text{ k}\Omega \quad R_{TH} = \frac{R_1 R_2}{R_1 + R_2} = 3.02 \text{ k}\Omega$$

$$13 \text{ k}\Omega R_2 = 13 \text{ k}\Omega \cdot 3.02 \text{ k}\Omega + R_2 \cdot 3.02 \text{ k}\Omega$$

$$9.98 \text{ k}\Omega R_2 = 39.26 \text{ M}\Omega \Rightarrow R_2 = 3.93 \text{ k}\Omega$$



# In-class problem 1, solution



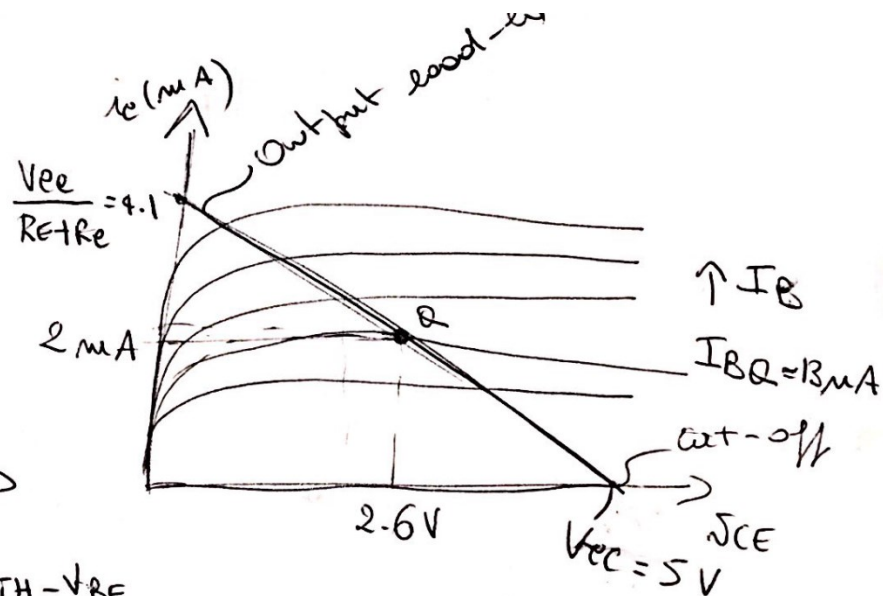
$$I_B = \frac{V_{TH} - V_{BE}}{R_{TH} + (1+\beta)R_E} \approx \frac{V_{TH} - V_{BE}}{(1+\beta)R_E}$$

$$V_{TH} = \frac{V_{CC} R_2}{R_1 + R_2} = \frac{5 \cdot 3.53K}{16.53K} = 1.16V$$

$$I_B = 0 \Rightarrow V_{BE} = V_{TH} = 1.16V$$

$$V_{BE} = 0 \Rightarrow I_B \approx \frac{V_{TH}}{(1+\beta)R_E} = \frac{1.16}{151 \cdot 0.2K} = 38\mu A$$

$$I_{BQ} = 13.3\mu A \text{ yields } V_{BEQ} \approx 0.71V$$



$$I_C = \frac{V_{CC} - V_{CE}}{R_E + R_C}$$

$$I_C = 0 \quad V_{CE} = V_{CC}$$

$$V_{CE} = 0 \quad I_C = \frac{V_{CC}}{R_E + R_C}$$

$$I_B = I_{BQ} = 13.3\mu A$$

# Overview of Lecture 13

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- Biasing the BJT 2 (Neamen 5.2 and 5.4, S&S 6.7)