Title

by

Author



Thesis presented in partial fulfilment of the requirements for the degree of Master of Engineering (Electronic) in the Faculty of Engineering at Stellenbosch University

Supervisor: Supervisor

December 2015

Declaration

By submitting this thesis electronically, I declare that the entirety of the work contained therein is my own, original work, that I am the sole author thereof (save to the extent explicitly otherwise stated), that reproduction and publication thereof by Stellenbosch University will not infringe any third party rights and that I have not previously in its entirety or in part submitted it for obtaining any qualification.

D /	date
Date:	

Copyright © 2015 Stellenbosch University All rights reserved.

Abstract

Title

Α.

Department of Electrical and Electronic Engineering, University of Stellenbosch, Private Bag X1, Matieland 7602, South Africa.

Thesis: MEng (Elec)
December 2015

abstract

Uittreksel

Title

("Title")

Α.

Departement Elektriese en Elektroniese Ingenieurswese, Universiteit van Stellenbosch, Privaatsak X1, Matieland 7602, Suid Afrika.

Tesis: MIng (Elek)
Desember 2015

abstrakte

Acknowledgements

I would like to express my sincere gratitude to the following people and organisations \dots

Dedications

Hierdie tesis word opgedra aan ...

Contents

De	claration	i
Αl	stract	ii
Ui	treksel	iii
A	knowledgements	iv
De	dications	\mathbf{v}
Co	ntents	vi
Li	t of Figures	vii
Li	t of Tables	⁄iii
No	menclature	ix
1	Discrete Element Method 1.1 Introduction	1 1
2	Literature review 2.1 Optimisation	2 2 2 3
Aı	pendices	8
A	Discrete Element Method Theory A.1 Ball elements	9
В	Activity Log B.1 Matlab	10 10
L.i.	t of References	11

List of Figures

A.1 Ball Element Parame	ers	
-------------------------	-----	--

List of Tables

Nomenclature

Constants

 $g = 9.81 \,\mathrm{m/s^2}$

Variables

Re_{D}	Reynolds number (diameter) []	
x	Coordinate	
\ddot{x}	Acceleration	!
θ	Rotation angle [rad]	
au	Moment N·m	1

Vectors and Tensors

 $\overrightarrow{\boldsymbol{v}}$ Physical vector, see equation ...

Subscripts

- a Adiabatic
- a Coordinate

Chapter 1

Discrete Element Method

1.1 Introduction

Introduction...

Chapter 2

Literature review

Process

- Keep problem statement in the foremost of the author's mind: Trust region use with SM for particular antenna design.
- Obtain reference material and understand it.
- Organise texts.
- Document relevant material using paragraphs in your own words citing references.

2.1 Optimisation

Describe the different optimisation problems that are required in a typical or subset of antenna design processes.

2.2 Space mapping

Introduction to what SM is and discuss the different types. Explain the basic SM techniques roughly and what they work for, details to follow. Problem is that they require a highly complex surrogate model. There are also more advanced algorithms that only require simple surrogate models. GSM, FDGSM.

- 2.2.1 Implicit space mapping (ISM)
- 2.2.2 Output space mapping (OSP)
- 2.2.3 Space mapping interpolating surrogates (SMISs)
- 2.2.4 Generalised space mapping (GSM)
- 2.2.5 Frequency-dependent generalised space mapping (FDGSM)

Space mapping

2.3 Actual paper summaries

2.3.1 A Space-Mapping Framework for Engineering Optimisation - Theory and Implementation: Koziel, Bandler and Madsen Koziel *et al.* (2006*a*)

Introduction and background

Space-mapping (SM) is a widely used engineering optimisation paradigm [citations]. SM uses two different models to run an optimisation on. The one model is a fine model that is a very good approximation of the 'real world' model. This is computationally expensive to run and a surrogate or coarse model is used that is cheaper to evaluate. There is a misalignment between these two models, but if they are still fairly good representation of the 'real world' model the misalignment should be small. If the alignment is indeed good then SM-based optimisation algorithms typically provide good results with few evaluations of the computationally expensive fine model.

SM is used extensively for models at microwave frequencies. The full-wave electromagnetic (EM) solvers take a long time to run [validate with own examples/citations if for own use], but they can be represented by physical based equivalent-circuit models that are cheap to simulate.

There are other engineering disciplines that have started using SM techniques. Bandler review [5](Bandler, Cheng, Dakroury, Mohamed, Bakr, Madsen and Sondergaard; Space mapping: the state of the art).

Mentioned SM techniques with references: Implicit space mapping (ISM)

Output space mapping (OSP)

Space mapping interpolating surrogates (SMISs)

Generalised space mapping (GSM) Frequency-dependent generalised space mapping (FDGSM)

The actual paper describes algorithms that exploit surrogate models based on OSM that force exact matching of responses and Jacobians between the fine and the surrogate model. Theoretical results are presented that show the influence of Jacobian matching on the convergence of the optimisation algorithm. Design-variable-dependent ISM is introduced to increase the flexibility of the surrogate model in a consistent way. Also proposed to make more accessible to engineers using particular tools.

Engineering design optimisation exploiting SM.

GISM to minimise misalignment between fine and course models.

Output SM to ensure the matching of response and first-order derivatives between mapped coarse and fine models at the current iteration point in the optimisation process.

Surrogate Optimisation

The optimisation problem is seen in Equation 2.1

$$\mathbf{R}_f: X_f \to R^m \underline{\subset} R^n \tag{2.1}$$

Where:

 \mathbf{R}_f is the response vector of the fine model. For example in microwave systems it could be the model evaluation of the scattering parameter $|S_{12}|$. m in the subscript of R_m represents the number of different frequency points. The goal of the optimisation is to solve

$$\boldsymbol{x}_{f}^{*} = \arg\min_{\boldsymbol{x} \in X_{f}} U\left(\mathbf{R}_{\mathbf{f}}(\boldsymbol{x})\right)$$
 (2.2)

where U is the objective function. It is assumed that the fine model is too computationally expensive to solve directly using the fine model. Instead inexpensive surrogate models are used. These are not as accurate as the fine model but are computationally cheap, which allows them to be run many times. An optimisation algorithms that generates a sequence of points $x^{(i)} \in X_f$, $i = 1, 2, \cdots$ and a family of surrogate models

$$\mathbf{R}_s^{(i)}: X_s^{(i)} \to R^m, i = 0, 1, \dots,$$
 (2.3)

so that

$$\boldsymbol{x}^{(i+1)} = \arg\min_{\boldsymbol{x} \in X_s^{(i)} \cap X_f} U\left(\mathbf{R}_s^{(i)}(\boldsymbol{x})\right)$$
 (2.4)

and $\mathbf{R}_s^{(i+1)}$ is constructed using a suitable matching condition with the fine model at previous point $\boldsymbol{x}^{(k)}, k = 1, \dots, i$.

SM uses that assumption that there is a coarse model $\mathbf{R}_c: X_c \to R^m, X_c \subseteq R^n$ that describes the same object as the fine model. A family of surrogate models is built up of a suitable distribution of \mathbf{R}_c such that matching conditions are satisfied.

Generalised Implicit Space-Mapping (GISM)

Enhancement of Space Mapping Optimization Algorithms for Engineering Design, Koziel 2006, Koziel et al. (2006b), using design-variable dependent ISM. Also recall Implicit Space Mapping Optimization Exploiting Preassigned Parameters, Bandler 2004, Bandler et al. (2004). Using a coarse model that depends on preassigned parameters,

$$\mathbf{R}_c: X_c \times X_p \to \mathbb{R}^m$$

where $X_p \subseteq \mathbb{R}^q$ is a domain with preassigned parameters.

The ISM optimisation algorithm adjusts the preassigned parameters x_p so that, at the current point $x^{(i)}$, the both the fine model and the coarse model response vectors match. The adjustment can be referred to as a preadjustment and that adjusted model becomes a surrogate. This in turn is used to obtain the next point $x^{(i+1)}$.

$$\mathbf{R}_{s}^{(i)}(\boldsymbol{x}) = \mathbf{R}_{c}\left(\boldsymbol{x}_{p}^{(i)}, \boldsymbol{x}\right) \tag{2.5}$$

where $\boldsymbol{x}_{p}^{(i)}$ is determines by solving the parameter-extraction problem of the for

$$\boldsymbol{x}_{p}^{(i)} = \arg\min_{\boldsymbol{x}} ||\mathbf{R}_{f}\left(\boldsymbol{x}^{(i)}\right) - \mathbf{R}_{c}\left(\boldsymbol{x}^{(i)}, \boldsymbol{x}\right)||$$
 (2.6)

Making preassigned parameters depend on design variables allows the number of degrees of freedom in the surrogate model 2.5 to be increased. In particular,

$$\mathbf{R}_s^{(i)}(\boldsymbol{x}) = \mathbf{R}_c \left(\boldsymbol{x}, \mathbf{G}^{(i)} \boldsymbol{x} + \boldsymbol{x}_p^{(i)} \right)$$
 (2.7)

where $\mathbf{G}^{(i)} \in M_{q \times n}$ and $\boldsymbol{x}_p^{(i)}$ are determined by solving the parameter-extraction problem

$$\left(\mathbf{G}^{(i)}, \boldsymbol{x}_{p}^{(i)}\right) = \arg\min_{\left(\mathbf{G}, \boldsymbol{x}\right)} \left|\left|\mathbf{R}_{f}\left(\boldsymbol{x}^{(i)}\right) - \mathbf{R}_{f}\left(\boldsymbol{x}^{(i)}, \mathbf{G} \cdot \boldsymbol{x}^{(i)} + \boldsymbol{x}\right)\right|\right|$$
(2.8)

Combining this with the GSM from Koziel *et al.* (2006*b*), a GISM framework can be defined with the family of surrogate models $\mathbf{R}_{s}^{(i)}$ defined as

$$\mathbf{R}_{s}^{(i)}(\boldsymbol{x}) = \mathbf{A}^{(i)} \cdot \mathbf{R}_{c} \left(\mathbf{B}^{(i)} \cdot \boldsymbol{x} + \mathbf{c}^{(i)}, \mathbf{G} \cdot \boldsymbol{x} + \boldsymbol{x}_{p}^{(i)} \right) + \mathbf{d}^{(i)} + \mathbf{E}^{(i)} \cdot \left(\boldsymbol{x} - \boldsymbol{x}^{(i)} \right)$$
(2.9)

where

$$\left(\mathbf{A}^{(i)}, \mathbf{B}^{(i)}, \mathbf{c}^{(i)}, \mathbf{G}^{(i)}, \boldsymbol{x}_{p}^{(i)}\right) = \arg\min_{(\mathbf{A}, \mathbf{B}, \mathbf{c}, \mathbf{G}, \boldsymbol{x}_{p})} \varepsilon^{(i)}(\mathbf{A}, \mathbf{B}, \mathbf{c}, \mathbf{G}, \boldsymbol{x}_{p})$$
(2.10)

$$\mathbf{d}^{(i)} = \mathbf{R}_f \left(\mathbf{x}^{(i)} \right) - \mathbf{A}^{(i)} \mathbf{R}_c \left(\mathbf{B}^{(i)} \cdot \mathbf{x}^{(i)} + \mathbf{c}^{(i)}, \mathbf{G}^{(i)} \cdot \mathbf{x} + \mathbf{x}_p^{(i)} \right)$$
(2.11)

$$\mathbf{E}^{(i)} = \mathbf{J}_{\mathbf{R}_f} \left(\boldsymbol{x}^{(i)} \right) - \mathbf{A}^{(i)} \cdot \mathbf{J}_{\mathbf{R}_c \cdot \boldsymbol{x}} \left(\mathbf{B}^{(i)} \cdot \boldsymbol{x}^{(i)} + \mathbf{c}^{(i)}, \mathbf{G}^{(i)} \cdot \boldsymbol{x} + \boldsymbol{x}_p^{(i)} \right) \cdot \mathbf{B}^{(i)}$$
$$- \mathbf{A}^{(i)} \cdot \mathbf{J}_{\mathbf{R}_c \cdot \boldsymbol{x}_p} \left(\mathbf{B}^{(i)} \cdot \boldsymbol{x}^{(i)} + \mathbf{c}^{(i)}, \mathbf{G}^{(i)} \cdot \boldsymbol{x} + \boldsymbol{x}_p^{(i)} \right) \cdot \mathbf{G}^{(i)} \quad (2.12)$$

From the above the following matrices

$$\mathbf{A}^{(i)} = diag\{a_1^{(i)}, \cdots, a_m^{(i)}\}$$

$$\mathbf{B}^{(i)} \in \mathbf{M}_{n \times n}$$

$$\mathbf{c}^{(i)} \in \mathbf{M}_{n \times 1}$$

$$\mathbf{G}^{(i)} \in \mathbf{M}_{q \times n}$$

as well as vector $x_p^{(i)}$ are found using parameter extraction applied to the matching condition $\varepsilon^{(i)}$.

Matrices

$$\mathbf{d}^{(i)} \in \mathbf{M}_{m \times 1}$$

and

$$\mathbf{E}^{(i)} \in \mathbf{M}_{m \times n}$$

are calculated using 2.11 and 2.12 respectively. $\mathbf{J}_{\mathbf{R}_c \cdot \boldsymbol{x}}$ and $\mathbf{J}_{\mathbf{R}_c \cdot \boldsymbol{x}_p}$ represent the Jacobian of the coarse model with respect to $\boldsymbol{x}^{(i)}$ and $\boldsymbol{x}_p^{(i)}$ respectively.

The Jacobian is not always available [should site this and explain why solvers do not typically expose this] because it is expensive to calculate and not often exposed by full wave solvers?? If the Jacobian information is not available, matrix $E^{(i)}$ can be estimated, for example, using the Broyden update.

Wiki - quasi-Newton method for finding roots. The idea is to compute the whole Jacobian only at the first iteration and to do rank-one update at the other iterations. This may not converge for non-linear systems.

Form of matching condition and description of w_k and v_k . Describes how the surrogate tries to match the fine model response at all previous $\boldsymbol{x}^{(k)}$ (including current point), but Jacobian matching is not exploited.

Input SM determined by matrices $\mathbf{B}^{(i)}$ and $\mathbf{c}^{(i)}$, the multiplication matrix $\mathbf{A}^{(i)}$, as well as the ISM parameters $\boldsymbol{x}_p^{(i)}$ and $\mathbf{G}^{(i)}$, can be considered as preconditioning of the coarse model that reduces the initial misalignment between the coarse and fine models over the neighbourhood of the current point $\boldsymbol{x}^{(i)}$. The terms $\mathbf{d}^{(i)}$ ensures perfect matching of responses at $\boldsymbol{x}^{(i)}$ (zero order consistency). $\mathbf{E}^{(i)}$ gives perfect matching of first-order derivatives at $\boldsymbol{x}^{(i)}$

Examples

- 1. Bandpass filter
- 2. Bandpass filter with double-coupled resonators
- 3. Seven-section impedance transformer

Conclusion

Appendices

Appendix A

Discrete Element Method Theory

A.1 Ball elements

A.1.1 Ball mass and inertia parameters

Consider a volume element dV with respect to a static base S of an arbitrary solid body with density ρ . The mass of the body is obtained by integrating over the volume of the body,

$$m = \int_{\text{body}} \rho \, dV \tag{A.1}$$

In figure A.1, a ball with radius R_i and uniform density ρ_i is depicted. The mass of the ball is after integration of equation (A.1)

$$m_i = \frac{4}{3}\pi \rho_i R_i^3. \tag{A.2}$$

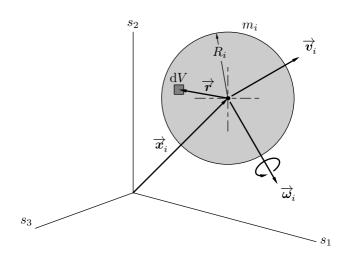


Figure A.1: Ball Element Parameters

Appendix B

Activity Log

B.1 Matlab

B.1.1 Parameter sweep

Recover lost work and get Windows machine up and running.

- 2016-01-22 Get back into the code and work out what is next.
- 2016-01-20

Add gradient for parameter space iterations for multiple frequencies. Problems with CADFEKO parameter sweep script -> log bug or follow up.

Validating responses, doesn't look right.

- 2016-01-29 Add 3D plots. Add FEKO path to matlab so it runs correctly.
- 2016-01-30

Visually validate S_{11} results and of gradient w.r.t. parameter space. Problem: Gradient seems to go bad after the first point.

B.1.2 Tex

- 2016-01-20 Create an activity log.
- 2016-01-31 Describe how the gradient here links up with the Broyden update and in turn into $E^{(i)}$.

List of References

- Bandler, J., Cheng, Q., Nikolova, N. and Ismail, M. (2004 Jan). Implicit space mapping optimization exploiting preassigned parameters. *Microwave Theory and Techniques*, *IEEE Transactions on*, vol. 52, no. 1, pp. 378–385. ISSN 0018-9480.
- Koziel, S., Bandler, J. and Madsen, K. (2006 Octa). A space-mapping framework for engineering optimization - theory and implementation. *Microwave Theory and Techniques*, *IEEE Transactions on*, vol. 54, no. 10, pp. 3721–3730. ISSN 0018-9480.
- Koziel, S., Bandler, J. and Madsen, K. (2006 Juneb). Space mapping optimization algorithms for engineering design. In: *Microwave Symposium Digest*, 2006. IEEE MTT-S International, pp. 1601–1604. ISSN 0149-645X.