2012-III: Transmission Lines and Antennas



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Reduced Maxwell's Equations

Assuming invariance of the medium along the z axis and a
wave solution in that direction, the homogeneous (no sources)
Maxwell's equations can be simplified by separating fields and
the del operator into transverse (i.e. Restricted to the xy plane)
and longitudinal/axial components (along z):

$$\mathbf{E}(x, y, z) = [\mathbf{e_t}(x, y) + \mathbf{\hat{z}}e_z(x, y)] e^{-j\beta z}$$

$$\mathbf{H}(x, y, z) = [\mathbf{h_t}(x, y) + \mathbf{\hat{z}}h_z(x, y)] e^{-j\beta z}$$

$$\nabla_{t} \times \mathbf{e_{t}} = -j\omega\mu_{0}\mathbf{\hat{z}}h_{z}$$

$$-\mathbf{\hat{z}} \times \nabla_{t}e_{z} - j\beta\mathbf{\hat{z}} \times \mathbf{e_{t}} = -j\omega\mu_{0}\mathbf{\hat{z}}h_{z}$$

$$\nabla_{t} \times \mathbf{h_{t}} = j\omega\epsilon_{0}\mathbf{\hat{z}}e_{z}$$

$$\nabla_{t} \times \mathbf{h_{t}} = j\omega\epsilon_{0}\mathbf{\hat{z}}e_{z}$$

$$\mathbf{\hat{z}} \times \nabla_{t}h_{z} + j\beta\mathbf{\hat{z}} \times \mathbf{h_{t}} = -j\omega\epsilon_{0}\mathbf{e_{t}}$$

$$\nabla_{t} \cdot \mathbf{h_{t}} = j\beta h_{z}$$

$$\nabla_{t} \cdot \mathbf{e_{t}} = j\beta e_{z}$$

TEM modes

$$\nabla_{t} \times \mathbf{e_{t}} = 0
\beta \hat{\mathbf{z}} \times \mathbf{e_{t}} = \omega \mu_{0} \mathbf{h_{t}}
\nabla_{t} \times \mathbf{h_{t}} = 0
\beta \hat{\mathbf{z}} \times \mathbf{h_{t}} = -\omega \epsilon_{0} \mathbf{e_{t}}
\nabla_{t} \cdot \mathbf{h_{t}} = 0
\nabla_{t} \cdot \mathbf{e_{t}} = 0$$

TE waves (H-waves)

• Also called H waves, these are obtained by setting $e_z = 0$ in the reduced equations:

$$\nabla_{t} \times \mathbf{e_{t}} = -j\omega\mu_{0}\hat{\mathbf{z}}h_{z}$$

$$-j\beta\hat{\mathbf{z}} \times \mathbf{e_{t}} = -j\omega\mu_{0}\mathbf{h_{t}}$$

$$\nabla_{t} \times \mathbf{h_{t}} = 0$$

$$\hat{\mathbf{z}} \times \nabla_{t}h_{z} + j\beta\hat{\mathbf{z}} \times \mathbf{h_{t}} = -j\omega\epsilon_{0}\mathbf{e_{t}}$$

$$\nabla_{t} \cdot \mathbf{h_{t}} = j\beta h_{z}$$

$$\nabla_{t} \cdot \mathbf{e_{t}} = 0$$

From Helmholtz equation:

$$\nabla_t^2 h_z + k_c^2 h_z = 0$$

$$\nabla_t^2 \mathbf{h_t} + k_c^2 \mathbf{h_t} = 0$$

$$k_c^2 \equiv k_0^2 - \beta^2$$

Curl of 3rd eq. at left, using then 5th at left and 2nd above:

$$\mathbf{h_t} = -\frac{j\beta}{k_c^2} \nabla_t h_z$$

2 x 2nd eq at left:

$$Z_{TE} \equiv Z_H \equiv Z_0 \frac{k_0}{\beta} > Z_0$$

TM waves (E-waves)

• Solutions are obtained by setting hz = 0, and are also called E waves. Previous solution can be reused via the duality transformations:

$$E \rightarrow H, H \rightarrow -E, \mu \rightarrow \epsilon, \epsilon \rightarrow \mu$$

$$\nabla_t^2 e_z + k_c^2 e_z = 0$$

$$\mathbf{e_t} = -\frac{j\beta}{k_c^2} \nabla_t e_z$$

$$\mathbf{h_t} = \frac{1}{Z_0} \frac{k_0}{\beta} \hat{\mathbf{z}} \times \mathbf{e_t} = \frac{1}{Z_{TM}} \hat{\mathbf{z}} \times \mathbf{e_t}$$

$$Z_{TM} \equiv Z_E \equiv Z_0 \frac{\beta}{k_0} \le Z_0$$

Computation of RLGC model parameters from Modal Fields (TEM lines)

$$L = \frac{\mu}{|I_0|^2} \int_S \mathbf{H} \cdot \mathbf{H}^* ds \quad (H/m)$$

$$C = \frac{\epsilon}{|V_0|^2} \int_S \mathbf{E} \cdot \mathbf{E}^* ds \quad (F/m)$$

$$R = \frac{R_s}{|I_0|^2} \int_{C_1 + C_2} \mathbf{H} \cdot \mathbf{H}^* dl \quad (\Omega/m)$$

$$G = \frac{\omega \epsilon''}{|V_0|^2} \int_S \mathbf{E} \cdot \mathbf{E}^* ds \quad (S/m)$$

Topics in guided propagation

- Power flow
- Conductor loss
- Dielectric loss
- Inhomogeneous filling

Loss in Transmission Lines and Waveguides

- Perturbation techniques can be applied to low loss transmission media such as TLs and waveguides:
 - Solve for the modal fields neglecting loss
 - Insert loss as a correction factor in the propagation constant (attenuation constant).
 - Various loss mechanisms are included additively in the attenuation constant

$$\alpha = \alpha_c + \alpha_d$$

$$\alpha_d = \frac{k^2}{2\beta} \tan \delta$$

$$\alpha_c = \frac{R_s Z_{\dagger}}{2} \frac{\int h_{tan}^2 dl}{\int \int e_t^2 ds}$$

$$R_s = \frac{1}{\sigma \delta_s}$$