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title: "Scale-Complete UFRF Guidance for Analog In-Memory Attention: S-Parameters, Projection Law, and REST-Gated Operation" author: - "UFRFv2 Collaboration" date: "2025-10-15" keywords: ["UFRF", "projection law", "S-parameters", "analog in-memory computing", "attention", "REST gating", "prime-length phasing", "noise shaping"] abstract: | We present a scale-complete engineering framework for analog in-memory attention engines that unifies scattering-parameter measurements with the Universal Field Resonance Framework (UFRF). We show how noise-looking artifacts are largely structured and technique-dependent, and we provide a concrete "Scale Ledger & Projection Budget" that spans device, cell, column, tile, SoC/clock, PDN/package/board, instrumentation, algorithm/training, and dataset/deployment domains. Central to the method are three levers: (i) operating at REST ( $E=B$  balance) which appears as an impedance match ( $|S_{11}|$  minimum and flat group delay), (ii) enforcing  $E \perp B$  via quadrature (I/Q) and chopping, and (iii) breaking coherence with prime-length (13/26) timing to redistribute spurs. We give falsifiable predictions ( $\sqrt{\phi} \approx 1.272$  efficiency/SNR window, spur reduction under 13/26 phasing, technique-dependent observables), and a measurement protocol for  $S(f, \tau)$  with time-gated VNA capture. header-includes: - \usepackage{amsmath,amssymb,siunitx} - \usepackage{physics} - \newcommand{\mathrm{REST}}{\mathrm{REST}} - \newcommand{\varphig}{\varphi} - \newcommand{\mathbf{E}}{\mathbf{E}} - \newcommand{\mathbf{B}}{\mathbf{B}} - \newcommand{\mathbf{S}}{\mathbf{S}} - \newcommand{\angle}{\angle} - \newcommand{\norm}[1]{\left\| \right\|} - \setlength{\parskip}{0.45em}

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# 1. Problem Statement and Approach

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**Goal.** Reduce accuracy-limiting “noise” in an analog in-memory attention engine by treating it as *structured, scale-dependent projection* rather than purely stochastic disturbance. We unify the UFRF geometry (orthogonal  $\mathbf{E}, \mathbf{B}$ ; 13-position cycle; REST at  $E = B$ ;  $\varphi$  transfer window) with **scattering-parameter** practice so design, measurement, and training speak the same language.

**Method in one line.** *Operate at match* ( $\text{REST} \Leftrightarrow |S_{11}|$  minima with flat group delay), *enforce orthogonality* (I/Q + chopping), and *break coherence* (13/26 prime-length

timing). Then account for **all relevant scales** explicitly in a **Projection Budget** so remaining residuals become small and structureless.

## 2. UFRF→S-Parameter Map (working dictionary)

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- **REST  $\equiv$  impedance match.** Define imbalance  $\rho \triangleq \frac{|E-B|}{E+B}$ . At REST,  $E = B$  so  $\rho \rightarrow 0$  and the network view shows  $|S_{11}| \downarrow$  with stationary phase (flat group delay). The UFRF transfer window gives an expected efficiency/SNR lift of  $\varphi \approx 1.272$  near the match.
- **Cycle position  $\leftrightarrow$  S-phase.** Use  $[p(\tau) \triangleq \frac{13}{2\pi} \angle S_{21}(f_0, \tau) \bmod 13, ]$  to map instantaneous position on the 13-cycle; the quiet window appears near  $p \approx 10$ .
- **$E \perp B \leftrightarrow$  I/Q orthogonality.** Mixed-mode S-parameters (or de-embedded I/Q paths) quantify cross-coupling; minimizing  $I \rightarrow Q$ ,  $Q \rightarrow I$  is the hardware enforcement of  $\mathbf{E} \perp \mathbf{B}$ .
- **Prime-length timing  $\leftrightarrow$  spur shaping.** 13-slot micro-frames and 26-way interleaves spread coherent tones into weaker sidebands in  $S(f)$  rather than concentrating them at harmonics/clock feedthrough.

## 3. Scale Ledger & Projection Budget

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UFRF's projection law states that an observation is technique- and scale-relative. We therefore model the measured outcome  $O$  with an **additive projection over scales**: 
$$\boxed{\ln O \triangleq \ln O^* + \sum_k d_{M,k} \alpha_k S_k + \varepsilon},$$
 where  $k$  ranges over all relevant **scales/observers** (device, cell, column, tile, SoC/clock, PDN, package/board, instrumentation, algorithm, dataset/domain). Here  $d_{M,k} = \ln(M_{\text{obs}}/M_k)$  is the scale distance,  $\alpha_k \in [0, 1]$  captures technique coupling,  $S_k$  summarizes systematic structure, and  $\varepsilon$  is small residual noise.

**Table 1 — Minimal Scale Ledger (what to model and log)**

Scale $k$	Typical “noise” signature	Surrogate $S_k$ you should log	Control lever (UFRF)
Transistor/device	RTN, $1/f$ , $g_m$ nonlinearity	RTN index, $1/f$ corner, HD2/HD3	Chopping; REST-biased biasing; I/Q
Gain-cell (2T1C/2T0C)	$kT/C$ , charge injection, retention	Decay vs. refresh cadence; read disturb	REST-gated refresh; mid-swing precharge
Column/bitline	Capacitive division, coupling	Bitline $RC$ , neighbor toggle stats	13/26 interleave; shielding
Tile/macro	Substrate/supply bounce, spurs	Spur table around carriers/harmonics	13-phase PWM; 26-way read permutation
SoC/clocking	Digital feedthrough, PLL jitter	Clock tree spectra; skew groups	Prime-phase skew groups; gated activity
PDN/VRM	Beat notes, PDN resonances	$Z_{\text{PDN}}(f)$ peaks, Q	Prime-phase VRM dithering; decap at modes
Package/board	Return-path L, board modes	Mixed-mode $S(f, \tau)$ at slot	Align reads to board-level $ S_{11} $ minima
Instrumentation	Gate/IFBW bias; fixture errors	Gate placement, IFBW, de-embed meta	Time-gate to REST; consistent de-embed
Algorithm/training	Quantization, optimizer noise	Loss component spectra	Projection-aware loss; I/Q penalty
Dataset/domain	Domain shift, temp band	Token stats, temp profile	Projection head; lab→field calibration

*Design rule:* do **not** leave a populated row out of the sum; unmodeled scales come back as “noise” .

## 4. Measurement Protocol: $S(f, \tau)$ with time-gated VNA

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**P1 — Locate REST.** Sweep the micro-frame (bias/timing) and compute  $S(f, \tau)$ . Mark the  $\tau$  where  $|S_{11}|$  is minimum and group delay is flattest. Expect a local peak in delivered power/SNR near this slot (the  $\varphi$  window).

**P2 — Verify I/Q orthogonality.** Enable dual  $90^\circ$  paths and acquire mixed-mode S; minimize  $I \rightarrow Q/Q \rightarrow I$  and confirm even-order cancellation. Chopping moves  $1/f$  away from baseband offsets.

**P3 — A/B timing schedules.** Record spur tables under periodic vs. 13/26 timing. Expect 10–15 dB worst-spur reduction and energy re-distribution to  $n/13$  and  $m/26$  offsets.

**P4 — Read-write coupling.** Compare  $\|\Delta S\|$  during read when writes occur inside vs. outside REST slots. Inside-REST scheduling should reduce read-while-write disturbance.

## 5. Circuit & Timing Remedies (drop-in changes)

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- 1. REST-biased differential readout.** Precharge/read so electric and magnetic-energy proxies equalize at integration ( $E = B$ ). Predict  $\approx 2.1$  dB SNR/efficiency lift ( $\varphi$ ).
- 2. Quadrature (I/Q) + chopper stabilization.** Two  $90^\circ$  paths recombined as  $I + jQ$  suppress even-order distortion and push  $1/f$  out of band.
- 3. 13-phase PWM + 26-way interleave.** Within a prime-length 13-slot micro-frame, scramble PWM edges and permute sub-tile reads (e.g.,  $n \mapsto n + 5 \bmod 26$ ).
- 4. REST-gated refresh; zero-bias idle.** Place writes adjacent to REST and keep unused cells at zero so decay is benign.

## 6. Learning & Calibration with S-Awareness

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### 6.1 Projection-aware objective

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Augment the training signal with the multi-scale projection model:  $\ln O_k = \ln O_k^* + \sum_r d_{M,r} \alpha_{k,r} S_{k,r} + \epsilon_k$ . ] Learn small deltas of  $\alpha_{k,r}$  while keeping  $S_{k,r}$  logged from hardware (spur tables,  $|S_{11}|$ , phase stability, PDN modes, etc.).

### 6.2 I/Q consistency penalty

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Enforce  $\mathbf{E} \perp \mathbf{B}$  in software via an orthogonality penalty on the I and Q residuals:  $\mathcal{L}_{\perp} = \lambda \frac{\langle \mathbf{r}_I, \mathbf{r}_Q \rangle}{\|\mathbf{r}_I\| \|\mathbf{r}_Q\|}$ .

### 6.3 13/26 spectral regularizer

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Bias the model against error energy at prime-length offsets:  $\mathcal{L}_{13/26} = \mu \sum_{\Omega} |\mathbf{r}|^2 \mathbf{w}_{\Omega} \left( \left| \mathbf{E}(\Omega) \right| \right)$ . ]  $\mathbf{w}_{\Omega} = \left( \left| \mathbf{E}(\Omega) \right| \right)^{\frac{n}{13}} + \left( \left| \mathbf{E}(\Omega) \right| \right)^{\frac{n}{26}}$

## 7. Predictions (falsifiable) and expected effect sizes

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- **Efficiency/SNR window near REST.**  $\eta_{\text{REST}} = \underline{\varphi} = 1.272 \dots$  (about +2.1 dB).
- **Spur suppression with 13/26 schedules.** Worst spur reduced by  $\sim 10\text{--}15$  dB vs. fully periodic scheduling, with residuals at  $n/13, m/26$ .

- **Even-order and  $1/f$  reduction.** I/Q + chopper lowers HD2 and moves low-frequency noise out of baseband.
- **Technique dependence.** Different instruments/flows yield predictable offsets via  $\alpha_k$ ; as  $S \rightarrow 0$  they converge.

## 8. Putting it together for an analog attention macro

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For the volatile gain-cell attention block (K/V stored in capacitive cells; charge-to-pulse readout; sliding-window attention), we recommend: (i) mid-swing precharge and REST-balanced RC; (ii) I/Q chopper front-ends; (iii) 13-slot PWM edge scrambler and 26-step tile permutation; (iv) REST-gated refresh; and (v) training with the projection-aware loss and spectral/IQ penalties described above. Integrate measured  $S(f, \tau)$  features into the initialization loop as fixed surrogates  $S_k$ .

## 9. Minimal A/B plan

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1. **SPICE + behavioral:** sweep integrator bias to find REST; enable I/Q chopper; compare periodic vs. 13/26 spur tables.
2. **Bench S-params:** time-gated VNA to build  $S(f, \tau)$ ; compute REST metrics and spur maps.
3. **HW-in-loop:** extend initialization with projection-aware loss and spectral/IQ penalties; evaluate accuracy/perplexity vs. baseline.

## Appendix A — Notation and definitions

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- $S(f, \tau)$ : scattering parameters measured vs. frequency and micro-frame time.
- $|S_{11}|$ : magnitude of input reflection (return).
- $\angle S_{21}$ : phase of forward transmission.

- $p$ : 13-position index computed from  $\angle S_{21}$ .
- REST: UFRF balance point  $E = B$ .
- $\varphi$ : square root of golden ratio (about 1.272).
- Mixed-mode S-parameters: differential/common-mode or I/Q decomposition.

## Appendix B — Suggested data products to log

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- Micro-frame trace of  $|S_{11}|$ ,  $\angle S_{21}$ , group delay.
- Spur tables around carriers/harmonics for periodic vs. 13/26 schedules.
- PDN impedance peaks and Q; VRM phase policy.
- Package/board mixed-mode  $S(f, \tau)$  at the operating slot.
- Instrumentation metadata: gate placement, IFBW, de-embedding.
- Training loss spectra and I/Q residual correlation.

## Appendix C — REST identification heuristic (practical)

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Compute a composite score  $J(\tau) = w_1 \cdot (|S_{11}|(\tau)) + w_2 \cdot \text{angle } S(f, \tau)$ , where GD is group delay. REST is at  $\tau^* = \arg \min_{\tau} J(\tau)$  (weights  $w_i$  positive).

## References (internal UFRF source set)

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- UFRF core theory, axioms and first principles, Fourier connection, mathematical framework, integration summary, cross-domain validation, predictions/tests, and math appendix. These establish:  $\mathbf{E} \perp \mathbf{B}$  geometry, the 13-position cycle

with REST at  $E = B$ , the  $\varphi$  transfer window, and the projection law formalism used above.