

# Preliminaries

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### Geometric Series

A **geometric series** is the sum of a(n) (in)finite number of terms that have a constant ratio  $r$  between successive terms as

$$a + ar + ar^2 + \dots + ar^U = \sum_{n=0}^U ar^n$$

For  $r \neq 1$ , the sum  $S$  of  $U - L + 1$  terms of a geometric series is

$$\begin{aligned} S &= r^L + r^{L+1} + r^{L+2} \dots + r^U \\ rS &= r^{L+1} + r^{L+2} \dots + r^U + r^{U+1} \\ S - rS &= r^L - r^{U+1} \\ (1 - r)S &= r^L - r^{U+1} \\ S &= \frac{r^L - r^{U+1}}{1 - r} \end{aligned}.$$

The general formula is

$$\sum_{n=L}^U ar^n = a \left( \frac{r^L - r^{U+1}}{1 - r} \right).$$

In particular, when  $L = 0$ , the formula reduces to

$$\sum_{n=0}^U ar^n = a \left( \frac{1 - r^{U+1}}{1 - r} \right).$$

As  $U$  approaches infinity, the absolute value of  $r$  must be less than one for the series to converge. The sum then becomes

$$\sum_{n=0}^{\infty} ar^n = \frac{a}{1 - r}, \quad |r| < 1.$$

### Symmetric vs Periodic Windows

In the symmetric case, the second half of the window,  $M \leq n \leq N-1$ , is obtained by reflecting the first half around the midpoint. The symmetric option is the preferred method when using a window in FIR filter design.

The periodic window is constructed by extending the desired window length  $L$  by one sample to  $L = N + 1$ , constructing a symmetric window, and removing the last sample. The periodic version is the preferred method when using a window in spectral analysis *without zero-padding* because the discrete Fourier transform assumes periodic extension of the input vector. Zero-padding removes the advantage of symmetry.

## Fourier Transforms

### Discrete-Time Fourier Transform

The discrete-time Fourier transform (DTFT) of the sequence  $x(n)$  is

$$X(\tilde{\omega}) = \sum_{n=-\infty}^{\infty} x(n) e^{-j\tilde{\omega}n},$$

where  $\tilde{\omega} \triangleq \omega T_s \in [-\pi, \pi)$  is the *continuous* normalized radian frequency variable and  $T_s = \frac{1}{f_s}$  is the sampling period, the inverse of the sampling frequency  $f_s$ . Note that  $\omega = 2\pi f$  is the radian frequency in rad/s and  $f$  is the frequency in Hertz.

Therefore, the DTFT uses the continuous angular frequency variable  $\tilde{\omega}$ , whereas the discrete Fourier transform (DFT) is obtained from the DTFT by discretization of  $\tilde{\omega}$  as  $\omega T_s = \omega_k T_s$ , where  $\omega_k$  are the normalized discrete radian frequency given by

$$\omega_k = \frac{2\pi}{N}k, \text{ where } k = 0, 1, \dots, N - 1,$$

where  $N$  is the size of the DFT.

### Useful Relationships

| Variable                                    | Name                         | Unit          |
|---|------------------------------|---------------|
| $T$   | Period                       | s             |
| $f = \frac{1}{T}$                           | Frequency                    | Hz (cycle/s)  |
| $\omega = 2\pi f$                           | Angular frequency            | rad/s         |
| $f_s$                                       | Sampling frequency           | samples/s     |
| $\tilde{T} = \frac{1}{\nu} = \frac{f_s}{f}$ | Normalized period            | samples       |
| $\nu = \frac{f}{f_s}$                       | Normalized frequency         | cycles/sample |
| $\tilde{\omega} = 2\pi\nu$                  | Normalized angular frequency | rad/sample    |
| $k$   | Frequency bin                | samples       |
| $\frac{k}{N}$                               | Discrete frequency           | -             |

| Variable                      | Name                       | Unit    |
|-------------------------------|----------------------------|---------|
| $\omega_k = 2\pi \frac{k}{N}$ | Discrete angular frequency | radians |

Window size  $M = m \frac{f_s}{f_0}$  samples, where  $m$  is an integer number of periods.

Angular frequency  $\tilde{\omega} f_s = 2\pi f = \omega$  radians/s

Normalized frequency  $\frac{k}{N} = \frac{f}{f_s}$