

One-sided (unilateral) Laplace Transform  
Laplace Circuit Analysis  
Solving Differential Equations with LT

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# One-sided Laplace transform and its applications

$$X(s) = \int_0^{\infty} x(t) e^{-st} dt$$

- One-sided Laplace transform and its properties.
- Applications to initial value problems for ordinary differential equations.
- Applications to circuit analysis and, in particular, to transients.
- Transfer function.
- LTI systems in series and parallel. Inverse systems and linear feedback systems.
- Operational amplifiers, transfer functions for circuits with op amps.
- Bode plots, impulse response.
- Stability.

# One-sided Laplace transform

$$X(s) = \int_0^{\infty} x(t) e^{-st} dt \quad \text{unilateral (one - sided) Laplace transform}$$

$$x(t) \xrightarrow{L} X(s) \quad X(s) = L[x(t)]$$

Two-sided (bilateral) Laplace transform and its properties were studied within B39SA Signals and Systems

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt \quad \text{two - sided Laplace transform}$$

One-sided LT = two-sided LT applied to a causal signal (a signal that does not start before  $t=0$  is, i.e.  $x(t)=0$  for  $t<0$ .)

$$X(s) = \int_0^{\infty} x(t) e^{-st} dt = \int_{-\infty}^{\infty} x(t) u(t) e^{-st} dt$$

# Fourier and Laplace transforms

$$x(t) \xrightarrow{F} X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt \quad X(\omega) = F[x(t)]$$

$$x(t) = F^{-1}[X(\omega)] = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

It is not always possible to calculate the Fourier transform of a signal  $x(t)$  by integration. For example, if the signal is of finite power rather than finite energy the classical Fourier transform does not exist (the integral does not converge). A possible solution consists of multiplying  $x(t)$  by a convergence factor  $\exp(-\sigma t)$ :

$$x_{\sigma}(t) = x(t) e^{-\sigma t}$$

$$X_{\sigma}(\omega) = \int_{-\infty}^{\infty} x_{\sigma}(t) e^{-j\omega t} dt = \int_{-\infty}^{\infty} x(t) e^{-(\sigma + j\omega)t} dt = \int_{-\infty}^{\infty} x(t) e^{-st} dt$$

$$s = \sigma + j\omega$$

Motivation behind using bilateral (two-sided) Laplace transform

# Bilateral Laplace transform

$$x(t) \xrightarrow{L} X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt \quad X(s) = L[x(t)]$$

$$x_{\sigma}(t) = x(t) e^{-\sigma t}$$

$$X_{\sigma}(\omega) = \int_{-\infty}^{\infty} x_{\sigma}(t) e^{-j\omega t} dt = \int_{-\infty}^{\infty} x(t) e^{-(\sigma + j\omega)t} dt = \int_{-\infty}^{\infty} x(t) e^{-st} dt$$

$$s = \sigma + j\omega$$

$$x(t) = e^{\sigma t} x_{\sigma}(t) = e^{\sigma t} \frac{1}{2\pi} \int_{-\infty}^{\infty} X_{\sigma}(\omega) e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-\infty}^{\infty} X_{\sigma}(\omega) e^{(\sigma + j\omega)t} d\omega$$

$$= \frac{1}{2\pi j} \int_{\sigma - j\infty}^{\sigma + j\infty} X(s) e^{st} ds$$

inverse Laplace transform

# Bilateral Laplace transform: need for ROC

$$x(t) = e^{at} u(t)$$

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt = \int_0^{\infty} e^{(a-s)t} dt = \frac{1}{a-s} e^{(a-s)t} \Big|_0^{\infty} = \frac{1}{s-a}$$

ROC:  $\text{Re}(s) > \text{Re}(a)$  causal signal

$$x(t) = -e^{at} u(-t)$$

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt = \int_{-\infty}^0 -e^{(a-s)t} dt = -\frac{1}{a-s} e^{(a-s)t} \Big|_{-\infty}^0 = \frac{1}{s-a}$$

ROC:  $\text{Re}(s) < \text{Re}(a)$  anti-causal signal

These ambiguities can be removed if the signal  $x(t)$  is assumed to be *one-sided* or *causal*, i.e.,  $x(t) = 0$  if  $t < 0$ .

# One-sided Laplace transform

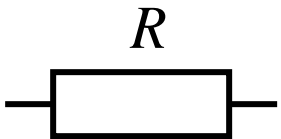
$$X(s) = \int_0^{\infty} x(t) e^{-st} dt \quad \text{unilateral (one - sided) Laplace transform}$$

$$x(t) \xrightarrow{L} X(s) \quad X(s) = L[x(t)]$$

# One-sided Laplace transform for circuit analysis

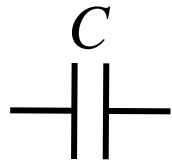
$$X(s) = \int_0^{\infty} x(t) e^{-st} dt \quad \text{unilateral (one - sided) Laplace transform}$$

$$L[dx/dt] = s X(s) - x(0)$$



$$v = Ri$$

$$V(s) = RI(s)$$



$$i = C \frac{dv}{dt}$$

$$V(s) = \frac{1}{sC} I(s) + \frac{v(0)}{s}$$



$$v = L \frac{di}{dt}$$

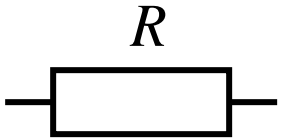
$$V(s) = sL(I(s) - i(0))$$



# One-sided Laplace transform for circuit analysis

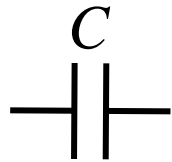
$$X(s) = \int_0^{\infty} x(t) e^{-st} dt \quad \text{unilateral (one - sided) Laplace transform}$$

$$\int_0^{\infty} \frac{dx}{dt} e^{-st} dt \stackrel{\text{by parts}}{=} \left[ x(t) e^{-st} \right]_0^{\infty} + s \int_0^{\infty} x(t) e^{-st} dt = sX(s) - x(0)$$



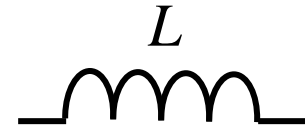
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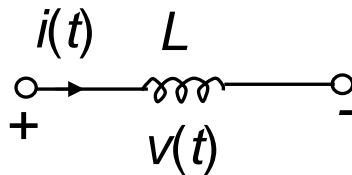
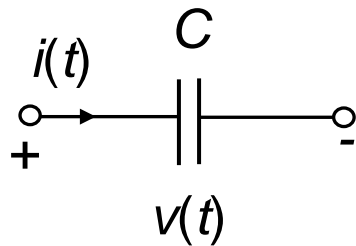
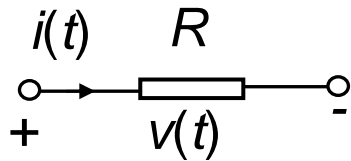


$$v = L \frac{di}{dt}$$

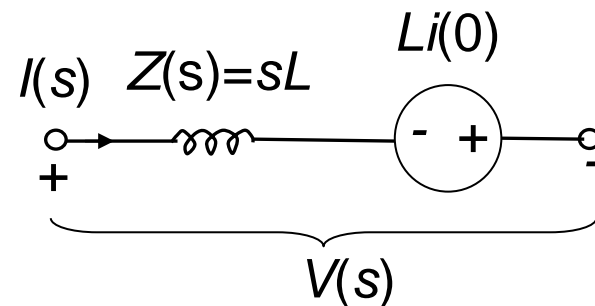
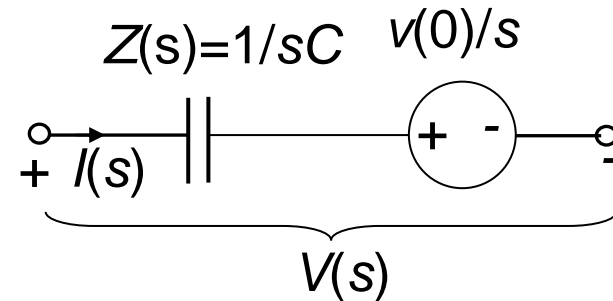
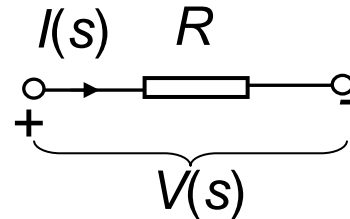
$$V(s) = L(sI(s) - i(0))$$

# Circuits in time domain and s-domain

Time domain  
circuit

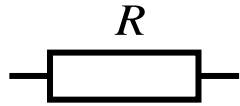


s-domain circuit



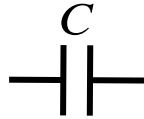
# Laplace transform for circuit analysis

$$X(s) = \int_0^{\infty} x(t) e^{-st} dt \quad L[dx/dt] = sX(s) - x(0)$$



$$v = Ri$$

$$V(s) = RI(s)$$



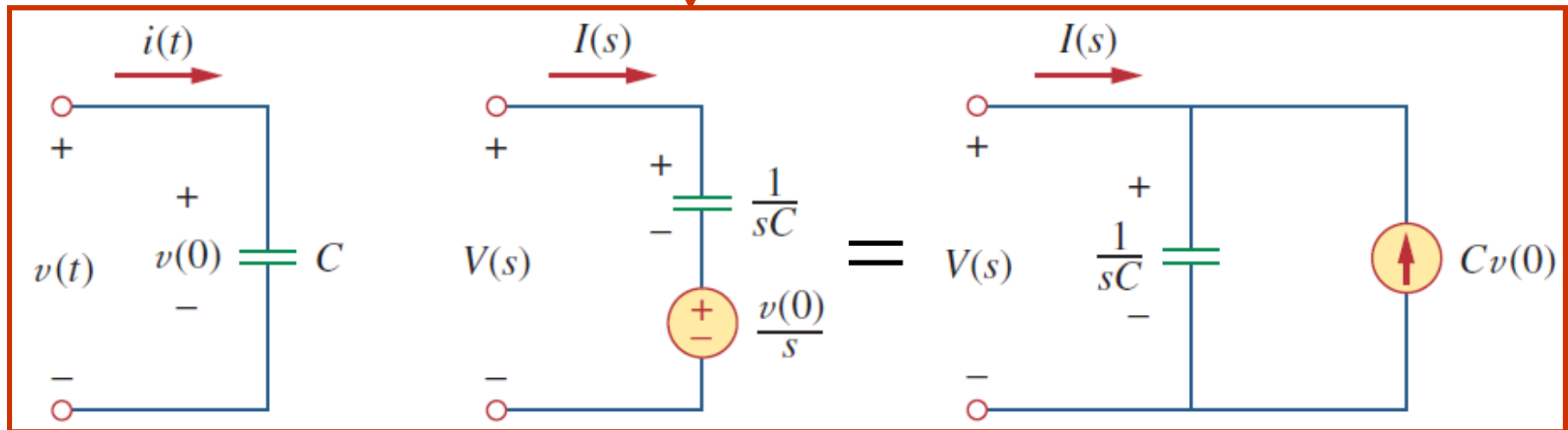
$$i = C \frac{dv}{dt}$$

$$V(s) = \frac{1}{sC} I(s) + \frac{v(0)}{s}$$



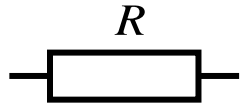
$$v = L \frac{di}{dt}$$

$$V(s) = L(sI(s) - i(0))$$



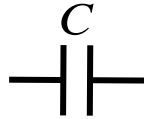
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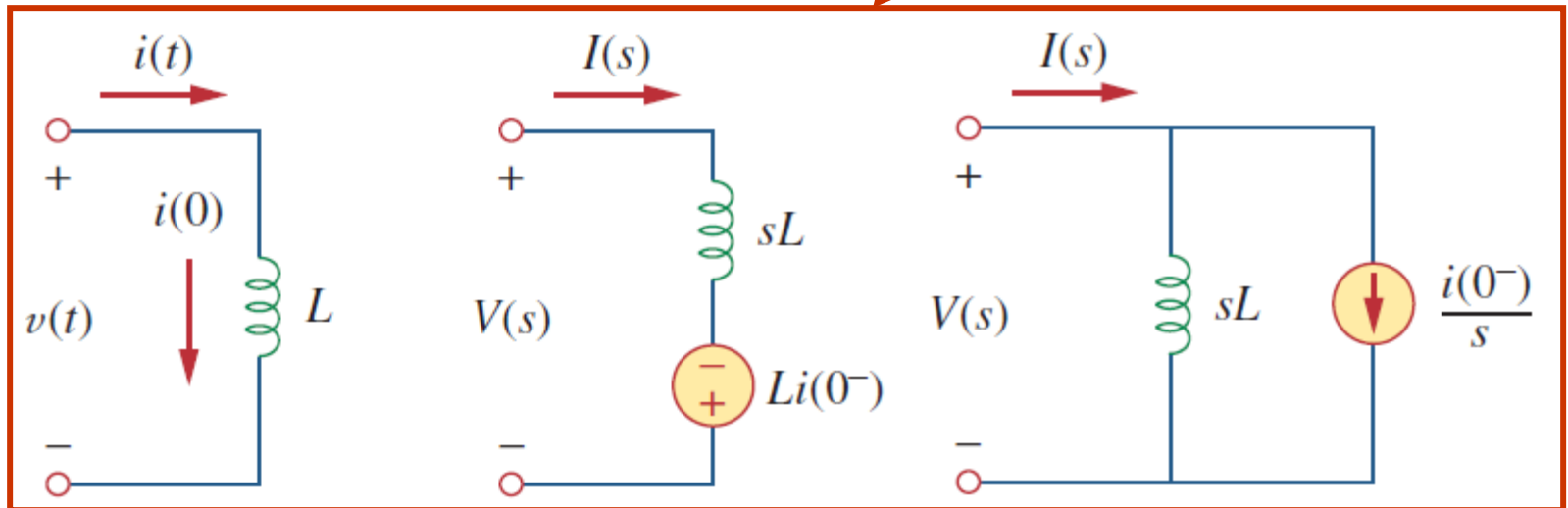
$$i = C \frac{dv}{dt}$$

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$$v = L \frac{di}{dt}$$

$$V(s) = L(sI(s) - i(0))$$



# One-sided Laplace transform pairs

Defined for  $t \geq 0$

It is assumed that  
 $x(t) = 0$  for  $t < 0$

$x(t)$	$X(s)$
$\delta(t)$	1
$u(t)$	$1/s$
$e^{-at}$	$1/(s + a)$
$t$	$1/s^2$
$t^n$	$n!/s^{n+1}$
$t^n e^{-at}$	$n!/(s + a)^{n+1}$
$\sin(\omega t)$	$\omega/(s^2 + \omega^2)$
$\cos(\omega t)$	$s/(s^2 + \omega^2)$
$e^{-at} \sin(\omega t)$	$\omega/[(s + a)^2 + \omega^2]$
$e^{-at} \cos(\omega t)$	$(s + a)/[(s + a)^2 + \omega^2]$

# Laplace and inverse Laplace transforms with matlab

```
>> syms t
>> x1 = 1/sqrt(t);
>> X1 = laplace(x1)
```

X1 =

$\pi^{1/2} / s^{1/2}$

```
>> syms s a
>> X = 1/(s+a);
>> x = ilaplace(X)
```

x =

$\exp(-a*t)$

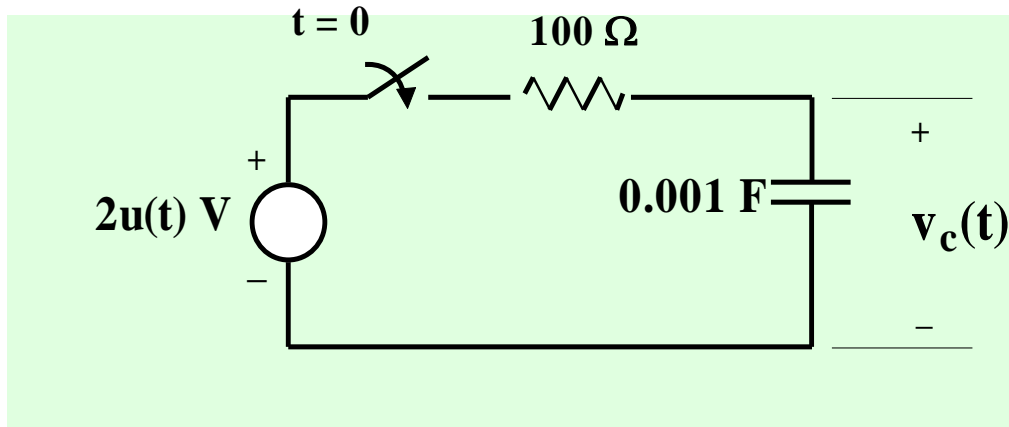
```
>> syms f(t) s
>> Df = diff(f(t),t);
>> laplace(Df,t,s)

ans =

s*laplace(f(t), t, s) - f(0)
```

# Laplace circuit analysis: Example 1

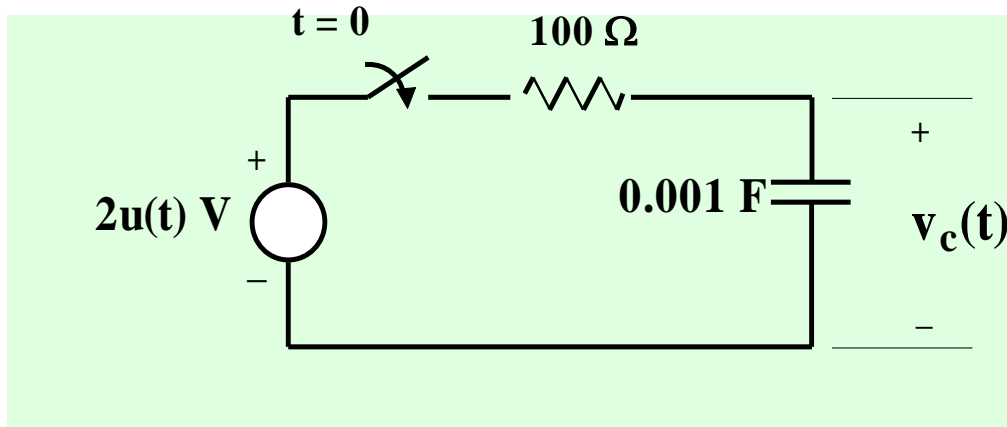
Assume  $v_c(0) = -4$  V. Use Laplace transform to find  $v_c(t)$ .



The time domain circuit

# Laplace circuit analysis: Example 1

Assume  $v_c(0) = -4$  V. Use Laplace transform to find  $v_c(t)$ .



$$i = C \frac{dv_c}{dt}$$

$$L\left[\frac{dx}{dt}\right] = sX(s) - x(0)$$

$$I(s) = C[sV_c(s) - v_c(0)]$$

$$V_c(s) = \frac{1}{s} \left[ \frac{I(s)}{C} + v_c(0) \right]$$



# Laplace circuit analysis: Example 1

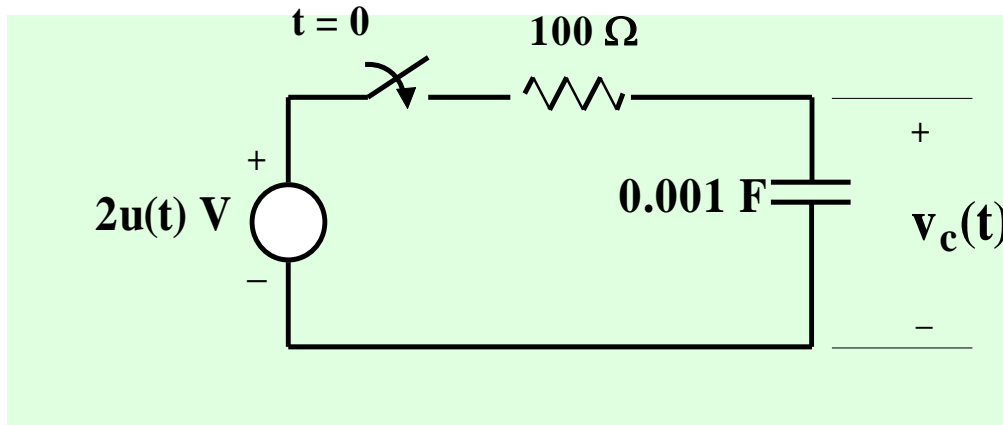
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$$i = C \frac{dv_c}{dt}$$

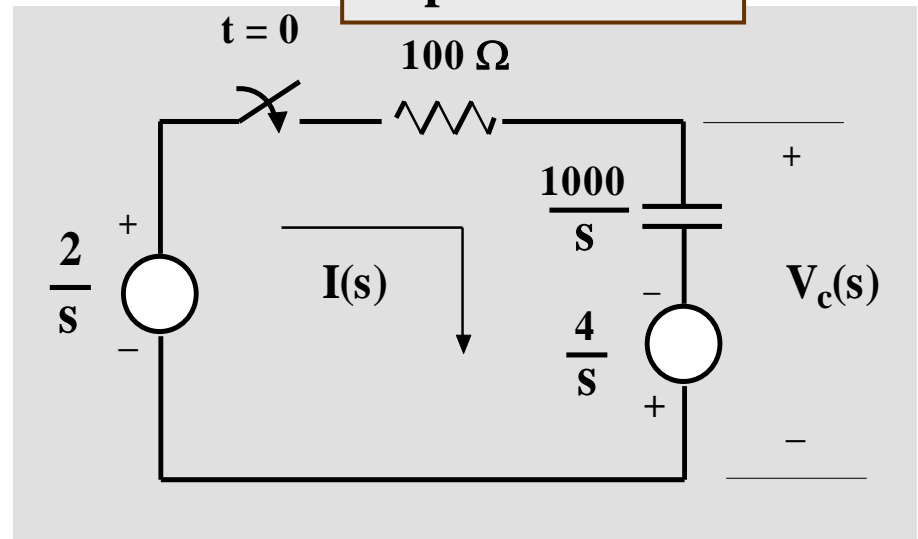


Kirchhoff voltage law in s-domain

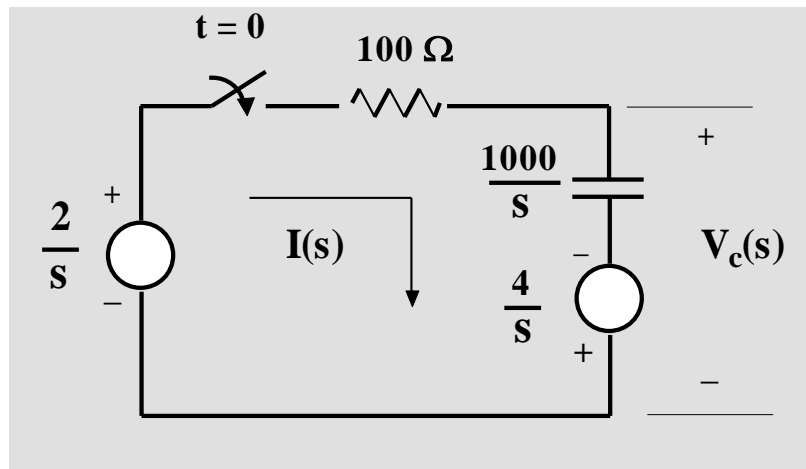
$$\frac{2}{s} + \frac{4}{s} = I(s) \left[ 100 + \frac{1000}{s} \right]$$

$$100I(s) = \frac{6}{s + 10}$$

**Laplace circuit:**



# Laplace circuit analysis: Example 1



①

$$\frac{2}{s} - 100I(s) - V_c(s) = 0$$

$$\frac{2}{s} - \frac{6}{s+10} = V_c(s)$$

②

$$V_c(s) = \frac{-4s + 20}{s(s+10)} = \frac{A}{s} + \frac{B}{s+10}$$

$$V_c(s) = \frac{2}{s} - \frac{6}{s+10}$$

$$v(t) = \left[ 2 - 6e^{-10t} \right] u(t)$$

**Check the boundary conditions**

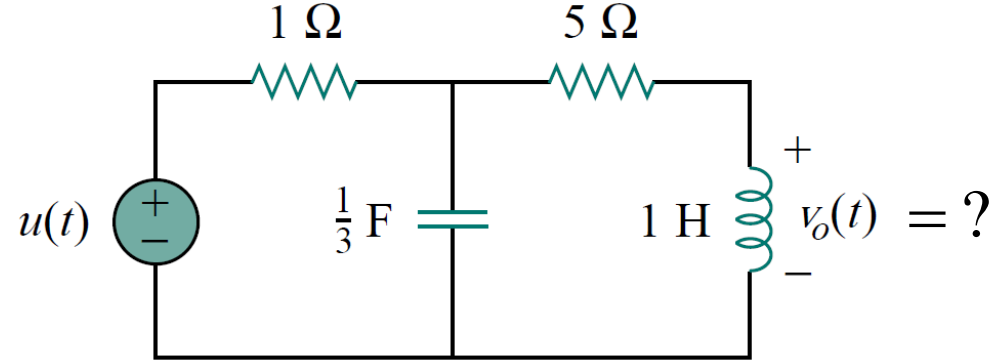
③

$$v_c(0) = -4 \text{ V}$$

$$v_c(\infty) = 2 \text{ V}$$

## Laplace circuit analysis: Example 2

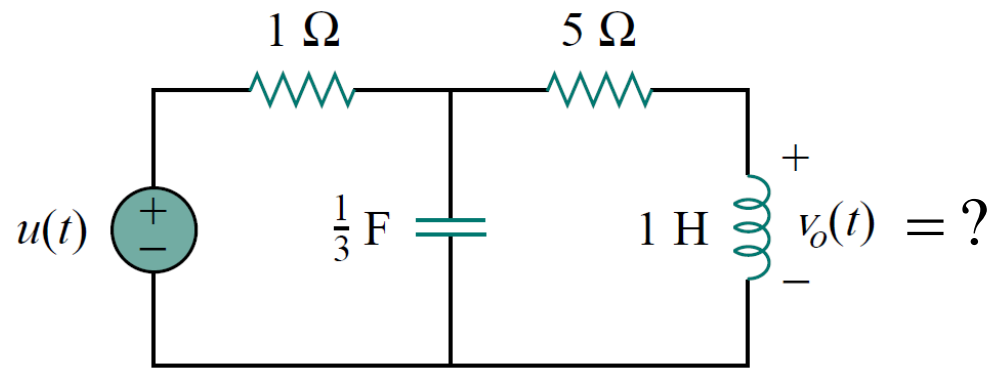
$$u(t) = \begin{cases} 1 & t > 0 \\ 0 & t < 0 \end{cases}$$



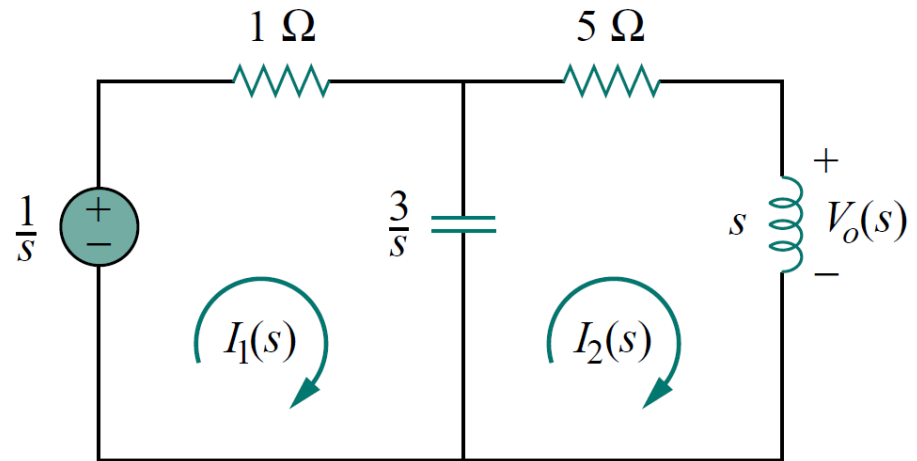
Find  $v_o(t)$ . Assuming no current in the circuit at  $t = 0$ .

# Laplace circuit analysis: Example 2

$$u(t) = \begin{cases} 1 & t > 0 \\ 0 & t < 0 \end{cases}$$



Let us use mesh analysis



$$I_2(s) = \frac{3}{s^3 + 8s^2 + 18s}$$

$$V_o(s) = sI_2(s)$$

$$v_o(t) = \frac{3}{\sqrt{2}} e^{-4t} \sin \sqrt{2} t \text{ V}$$

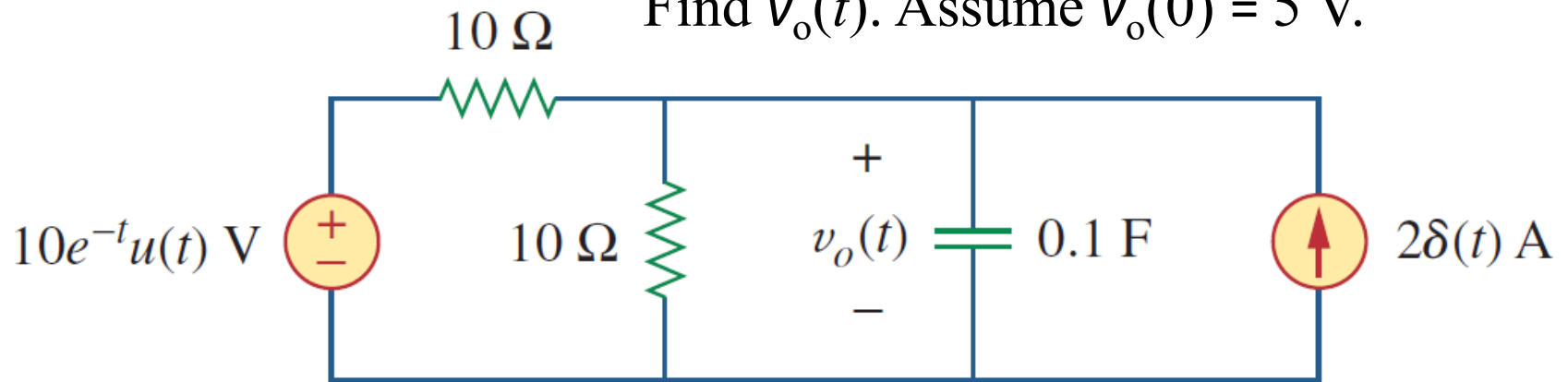
```
>> syms s
>> X = 3*s/(s^3+8*s^2+18*s);
>> x = ilaplace(X)

x =

(3*2^(1/2)*exp(-4*t)*sin(2^(1/2)*t))/2
```

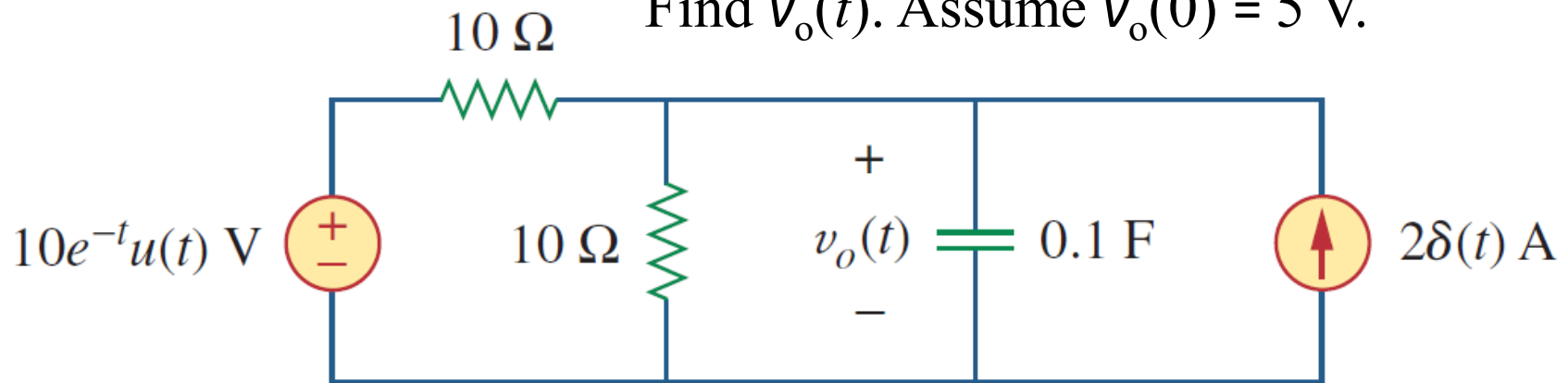
## Laplace circuit analysis: Example 3

Find  $v_o(t)$ . Assume  $v_o(0) = 5$  V.

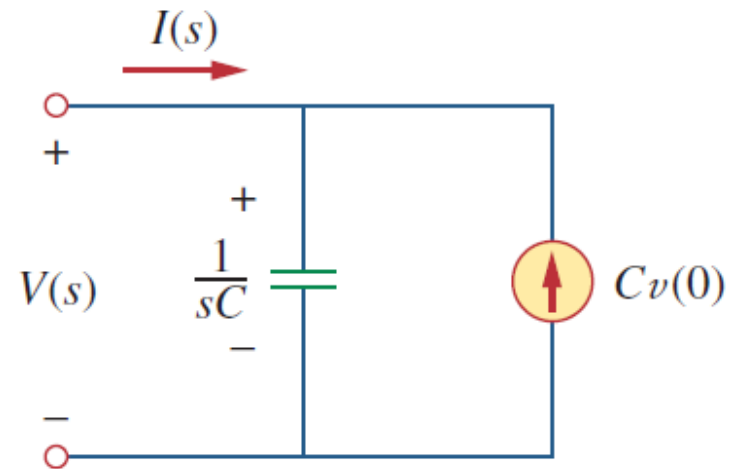
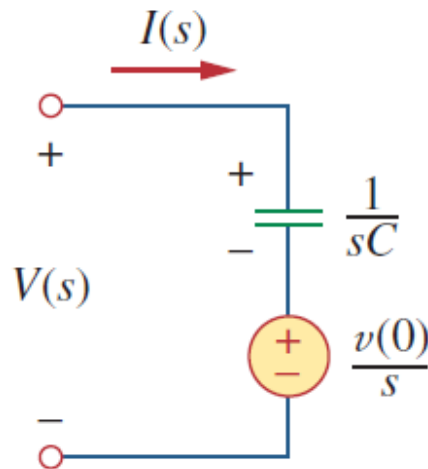
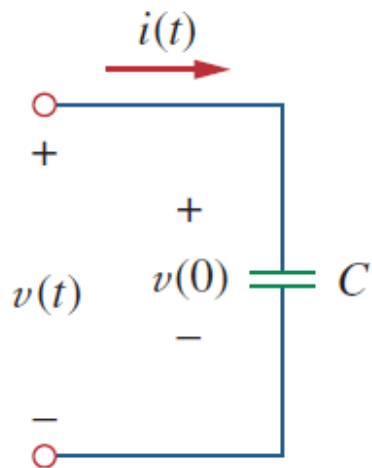


# Laplace circuit analysis: Example 3

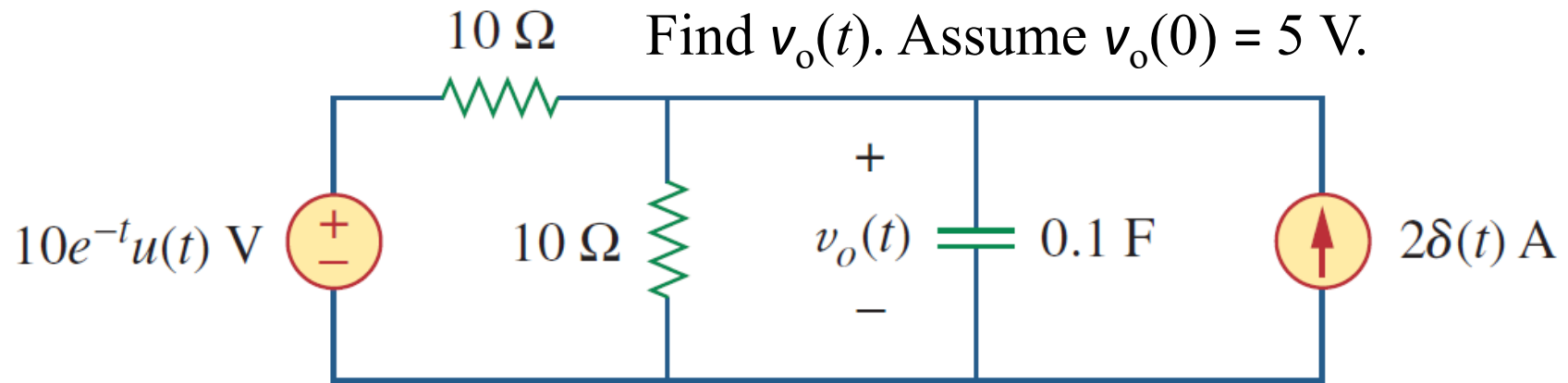
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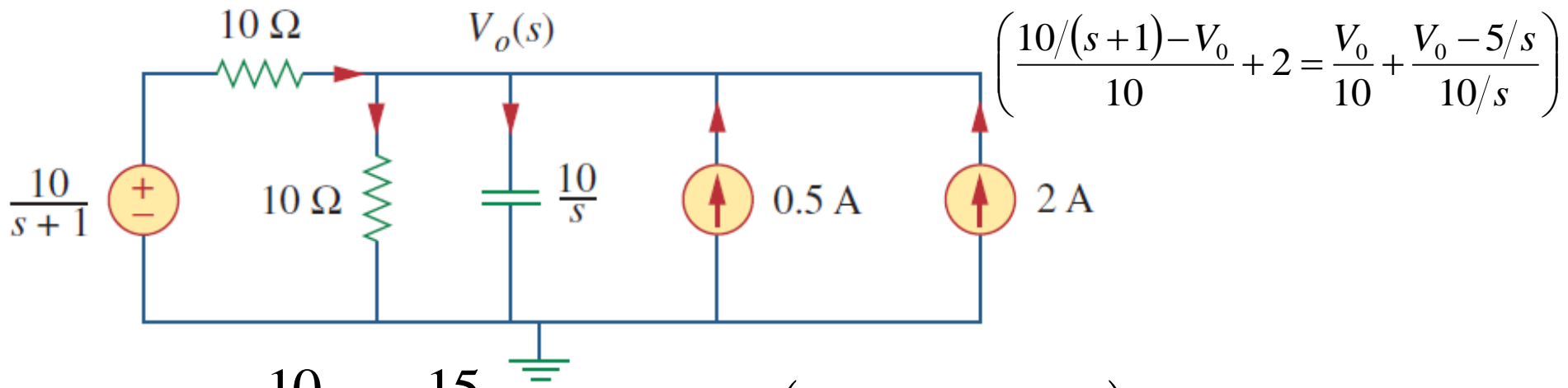
$$L\left[\frac{dx}{dt}\right] = sX(s) - x(0) \quad \text{---} \parallel \text{---} \quad i = C \frac{dv}{dt} \quad V(s) = \frac{1}{sC} I(s) + \frac{v(0)}{s}$$



# Laplace circuit analysis: Example 3

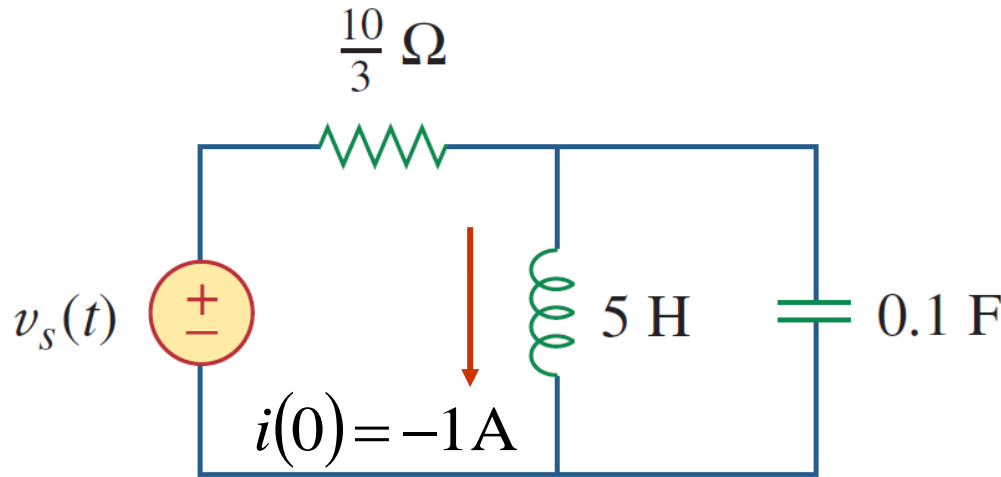


Let us use Nodal Analysis: 
$$\frac{10/(s+1) - V_0}{10} + 2 + 0.5 = \frac{V_0}{10} + \frac{V_0}{10/s}$$



$$V_o(s) = \frac{10}{s+1} + \frac{15}{s+2} \quad v_o(t) = (10e^{-t} + 15e^{-2t})u(t) \text{ V}$$

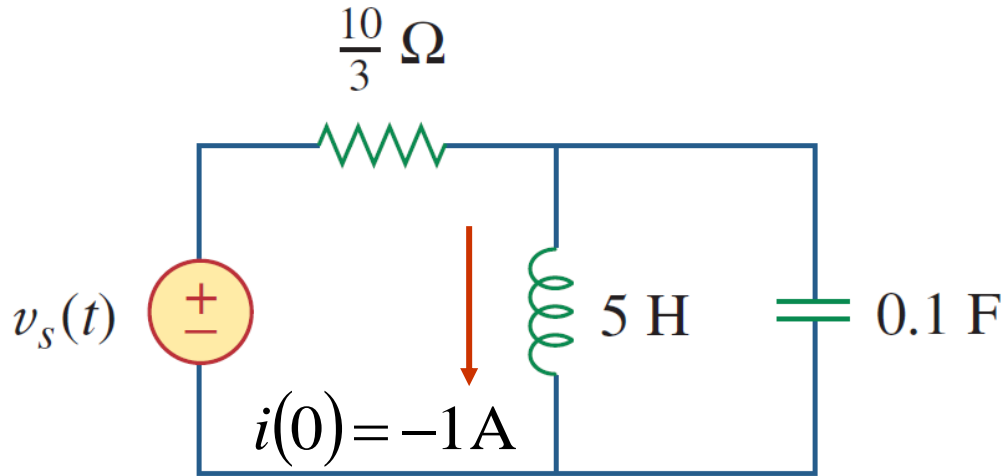
## Laplace circuit analysis: Example 4



Find the value of the voltage across the capacitor assuming that  $v_s(t) = 10u(t)$  V and assume that at  $t = 0$ , current -1 A flows through the inductor and +5 V is across the capacitor.



## Laplace circuit analysis: Example 4

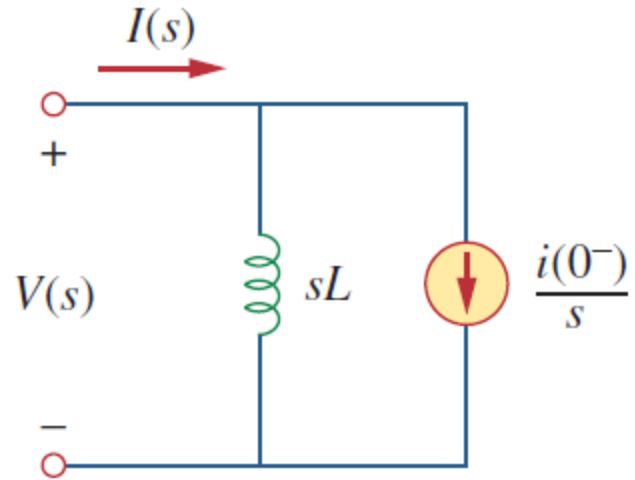
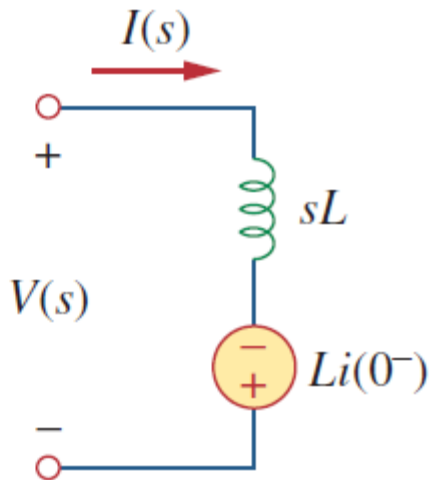
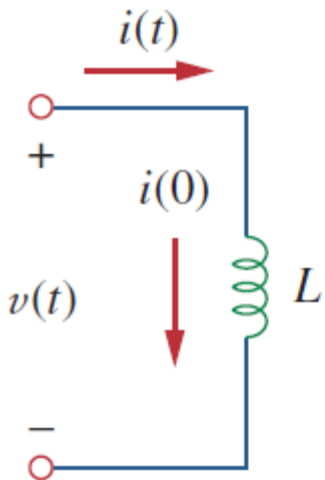


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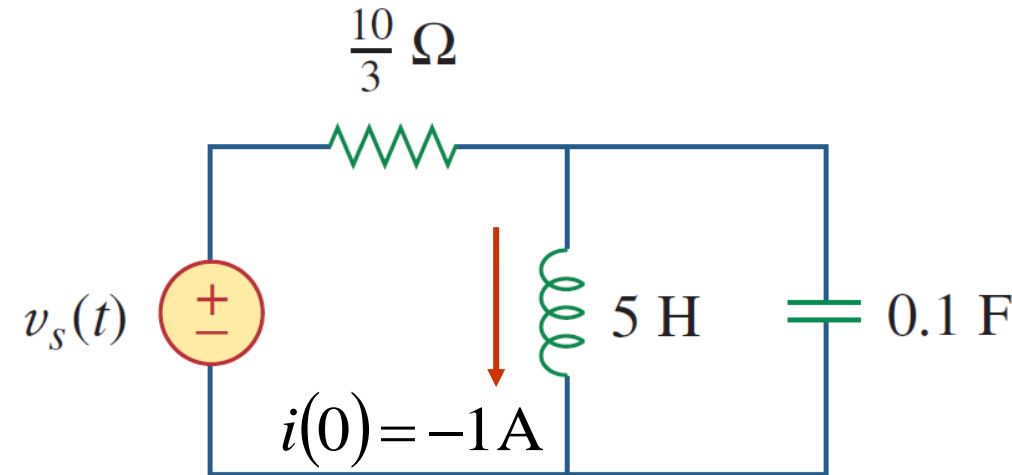
$$L\left[\frac{dx}{dt}\right] = s X(s) - x(0)$$

A diagram of an inductor, represented by a horizontal line with three loops. Above the inductor is the label  $L$ . To the right of the inductor is the equation  $v = L \frac{di}{dt}$ .

$$V(s) = L(sI(s) - i(0))$$

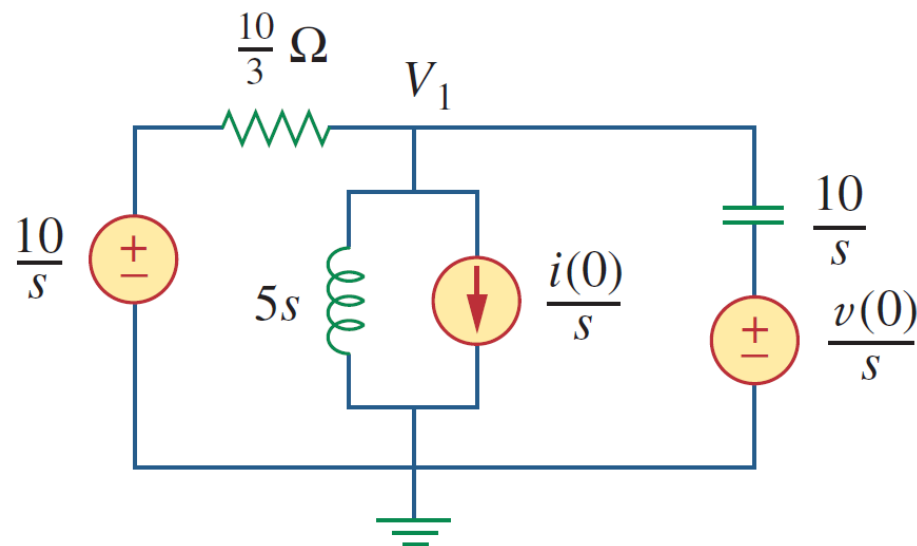


# Laplace circuit analysis: Example 4



Find the value of the voltage across the capacitor assuming that  $v_s(t) = 10u(t)$  V and assume that at  $t = 0$ , current  $-1$  A flows through the inductor and  $+5$  V is across the capacitor.

Convert the circuit to  $s$ -domain and use Nodal Analysis:



$$v_1(t) = (35e^{-t} - 30e^{-2t})u(t) \text{ V}$$

$$\frac{V_1 - 10/s}{10/3} + \frac{V_1 - 0}{5s} + \frac{i(0)}{s} + \frac{V_1 - v_C(0)/s}{10/s} = 0$$

$$\frac{1}{10} \left( s + 3 + \frac{2}{s} \right) V_1 = \frac{3}{s} + \frac{1}{s} + \frac{5}{10}$$

$$(s^2 + 3s + 2)V_1 = 40 + 5s$$

$$V_1 = \frac{40 + 5s}{(s+1)(s+2)} = \frac{35}{s+1} - \frac{30}{s+2}$$