

# REPUBLIC OF IVORY COAST



**UNION - DISCIPLINED - WORK** 

Ministry of Higher Education and Scientific Research (MESRS)



#### Institut National Polytechnique

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# **GROUP 6: ANALOGUE TRANSMISSION PROJECT 2021**

**TOPIC:** QoS DESIGN OF LEO SATELLITE RECEIVER

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# **PART ONE:**

#### ❖ Find the Q₃ uplink without coding

$$(\frac{c}{N})_{\text{uplink}} = \text{EIRP} + (Gr - Ts) + M + 228.6 - (Lp + Lx + BRF)$$

• 
$$g_r = \eta \left(\frac{\pi D}{\lambda}\right)^2 \Rightarrow Gr = 10\log\left[\eta \left(\frac{\pi D}{\lambda}\right)^2\right]$$
 with  $\lambda = \frac{C}{f}$ 

$$Gr = 10log \left[ \eta \left( \frac{\pi f D}{c} \right)^2 \right]$$

$$G_R = 10 \log \left[ \frac{60.5}{100} \left( \frac{\pi \times 5.79 \times 10^9 \times 2.22}{3 \times 10^8} \right)^2 \right] = 40.39875687 dB$$

• 
$$(\frac{c}{N})_u = \text{EIRP} + (\text{Gr} - \text{Ts}) + \text{M} + 228.6 - (\text{Lp} + \text{Lx} + \text{BRF})$$
  
 $\text{Ts} = \text{EIRP} + \text{GR} - (\frac{c}{N})_{\text{uplink}} + \text{M} + 228.6 - (\text{Lp} + \text{Lx} + \text{BRF})$ 

Lp = 
$$92.44 + 20 \log_{10} (fGHz) + 20 \log_{10} (d_{km})$$
  
=  $92.44 + 20 \log_{10} (5.79) + 20 \log_{10} (1107)$ 

#### Lp = 168.57652369212 dB

Lx = rain attenuation + atmospheric losses + poynting losses + polarization losses + other losses

$$Lx = 2.95 + 9.095 + 4.47 + 3.93 + 7.75$$

Lx = 28.195 dB

BRF = 
$$10\log_{10}$$
 (brf) =  $10\log_{10}$  (20 x  $10^6$ )

BRF = 73.01029996 dB

Ts = 48.975 + 40.39875687 - 15.75 + 3 + 228.6 - (168.57652369212 + 28.195 + 73.01029996)

Ts = 35.44193322 dBK

 $Ts = 10^{3.544193322} K$ 

Ts = 3501.00976420312 K

$$Ts = (Fs - 1)T_0$$

$$Fs = 1 + \frac{Ts}{T_0}$$

$$Fs = 1 + \frac{3501.00976420312}{290}$$

#### Fs = 13.07244746277

By taking:

1 => Antenna 7 => IF AMP1

 $2 \Rightarrow \text{Feeder}$   $8 \Rightarrow \text{Mixer } 2$ 

3 = > RF AMP 9 = > IF BPF2

 $4 \Rightarrow RF BPF$   $10 \Rightarrow IF AMP2$ 

 $5 \Rightarrow Mixer 1$   $11 \Rightarrow DEMOD$ 

 $6 \Rightarrow IF BPF1$   $12 \Rightarrow LPF$ 

$$\begin{split} \mathsf{FS} &= F_1 + \frac{F_2 - 1}{g_1} + \frac{F_3 - 1}{g_1 \, g_2} + \frac{F_4 - 1}{g_1 \, g_2 g_3} + \frac{F_5 - 1}{g_1 \, g_2 g_3 \, g_4} + \frac{F_6 - 1}{g_1 \, g_2 g_3 \, g_4 \, g_5} + \frac{F_7 - 1}{g_1 \, g_2 g_3 \, g_4 \, g_5 g_6} \\ &\quad + \frac{F_8 - 1}{g_1 \, g_2 g_3 \, g_4 \, g_5 g_6 g_7} + \frac{F_9 - 1}{g_1 \, g_2 g_3 \, g_4 \, g_5 g_6 g_7 g_8} + \frac{F_{10} - 1}{g_1 \, g_2 g_3 \, g_4 \, g_5 g_6 g_7 g_8 g_9} + \\ &\quad \frac{F_{11} - 1}{g_1 \, g_2 g_3 \, g_4 \, g_5 g_6 g_7 g_8 g_9 g_{10}} + \frac{F_{12} - 1}{g_1 \, g_2 g_3 \, g_4 \, g_5 g_6 g_7 g_8 g_9 g_{10} g_{11}} \end{split}$$

- $G_1 = 1$
- 2; 4; 5; 6; 8; 9; 11; 12 are passive components So their gain  $g = \frac{1}{L}$  and their F = L

$$T_{1} = (F_1 - 1) T_0 = > F_{1} = 1 + \frac{T_1}{T_0}$$

$$\mathsf{Fs} = \mathsf{F}_1 + \mathsf{L}_2 - \mathsf{1} + (\mathsf{F}_3 - \mathsf{1}) \, \mathsf{L2} + \frac{(L_4 - \mathsf{1})L_2}{g_3} + \frac{(L_5 - \mathsf{1})L_2 x \, L_4}{g_3} + \frac{(L_6 - \mathsf{1})L_2 x \, L_4 \, x \, L_5}{g_3} + \frac{(F_7 - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6}{g_3} + \frac{(L_9 - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8}{g_3 g_7} + \frac{(F_{10} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x \, L_6 x \, L_8 x \, L_9}{g_3 g_7} + \frac{(L_{11} - \mathsf{1})L_2 x \, L_4 \, x \, L_5 x$$

$$\frac{(F_{12}-1)L_2 \times L_4 \times L_5 \times L_6 \times L_8 \times L_9 \times L_{11}}{g_3 g_7 g_{10}}$$

With:

$$L2 = 5.045$$

$$L2 \times L4 = 5.045 \times 10^{0.778}$$

$$L2 \times L4 \times L5 = 5.045 \times 10^{1.465}$$

$$L2 \times L4 \times L5 \times L6 = 5.045 \times 10^{2.262}$$

$$L2 \times L4 \times L5 \times L6 \times L8 = 5.045 \times 10^{3.068}$$

$$L2 \times L4 \times L5 \times L6 \times L8 \times L9 = 5.045 \times 10^{3.856}$$

$$L2 \times L4 \times L5 \times L6 \times L8 \times L9 \times L11 = 5.045 \times 10^{4.981}$$

$$g_7 = 10^{2.2}$$

$$\mathbf{g}_{10} = 10^{2.025}$$

$$\mathbf{Q}_{7x}\mathbf{Q}_{10} = 10^{4.225}$$

Fs = F<sub>1</sub> + L<sub>2</sub>F<sub>3</sub> - 1 + 
$$\frac{5.045}{g_3}$$
 [ (L<sub>4</sub> - 1) + (L5 - 1) 10<sup>0.778</sup> + (L<sub>6</sub> - 1) x 10 <sup>1.465</sup> + (F<sub>7</sub> - 1) x 10<sup>2.262</sup> + (L<sub>8</sub> - 1) x 10<sup>2.262-2.025</sup> + (L9 - 1) x 10<sup>3.068-2.025</sup> + (F<sub>10</sub> - 1) x 10<sup>3.856-2.025</sup>

+ (L<sub>11</sub>- 1) x 
$$10^{3.856-4.225}$$
 + (F<sub>12</sub> - 1) x  $10^{4.981-4.225}$ 

$$g_3 = \frac{5.045}{Fs - \frac{T_1}{Tr_0} - 5.045F3} \left[ (L_4 - 1) + (L_5 - 1) \times 10^{0.778} + (L_6 - 1) \times 10^{1.465} + (F_7 - 1) \times 10^{2.262} + (L_8 - 1) \times 10^{1.465} \right]$$

$$)x10^{-0.237} + (L_9 - 1) \times 10^{1.043} + (F_{10} - 1) \times 10^{1.043} + (L_{11} - 1) \times 10^{-0.369} + (F_{12} - 1) \times 10^{0.756}$$

#### $q_3 = 7081.74503826386$

$$G_3 = 10 \log_{10} (g_3) = 10 \log (7081.74503826386)$$

#### ❖ Find G₃ with coding

$$16 \text{ QAM} \Leftrightarrow 2^4 \text{ QAM} => n = 4$$

$$R_b = \frac{n \times brf}{1+\alpha} = \frac{4 \times 20}{1.22} \text{ Mbps}$$

$$R_b = \frac{4000}{61} \text{ Mbps}$$

We have 
$$\frac{eb}{n_0} = \left(\frac{C}{n}\right) \times \frac{brf}{R_b}$$
 with Brf = 20 MHz and R<sub>b</sub> =  $\frac{4000}{61}$  Mbps

So 
$$(\frac{Eb}{N0})$$
 uncoded =  $(\frac{C}{N})$  uncoded +  $10\log_{10}(\frac{brf}{Rb})$ 

$$\left(\frac{Eb}{N0}\right)_{\text{uncoded}} = 15.75 + 10\log_{10}\left(\frac{61 \times 20}{4000}\right)$$

$$(\frac{Eb}{N0})_{uncoded} = 10.59299843 \text{ dB}$$

Then we get BER by:

BER =4(1 - 
$$\frac{1}{\sqrt{M}}$$
) Q  $\left[ \sqrt{\frac{3log_2M}{M-1}} \left( \frac{eb}{n_0} \right) \text{ uncoded} \right]$  and M = 16 and

$$\left(\frac{eb}{n0}\right)$$
 uncoded =  $10^{\frac{(Eb)}{N0}$  uncoded

BER =3Q 
$$\left(\sqrt{\frac{4}{5}} \left(\frac{eb}{n0}\right) \text{ uncoded}\right)$$

$$G_c = 20log_{10} \left[ erfc^{-1}(2BER_{coded}) \right] - 20log_{10} \left[ erfc^{-1}(2BER_{uncoded}) \right] + 10log_{10}(r_c)$$

$$G_c = 20log_{10}$$
  $\left[ 5.75426103121299990263 \right] - 20log_{10} \left[ 1.894516994431603057406 \right] +$ 

 $10log_{10}(\frac{7}{8})$ 

 $G_c = 9.06990160145 dB$ 

We have  $(\frac{Eb}{N_0})_{coded} = (\frac{Eb}{N_0})_{uncoded} - G_c$ , in DB, then we get it in linear value by :

$$\left(\frac{Eb}{N_0}\right)$$
 coded =  $\left(\frac{Eb}{N_0}\right)$  uncoded — GC

$$\left(\frac{Eb}{N_0}\right)$$
 coded = 10.59299843 - 9.06990160145

$$\left(\frac{Eb}{N_0}\right) coded = 1.52309682855 dB$$

 $R_{bcoded} = R_b x r_c$ 

#### $R_{bcoded} = 57.37704875 \text{ Mbps}$

$$\left(\frac{C}{N}\right) \text{ coded} = \left(\frac{Eb}{N_0}\right) coded + 10 \log\left(\frac{(R_b)coded}{B_n}\right)$$

$$\left(\frac{C}{N}\right)_{\text{coded}} = 1.52309682855 + 10\log\left(\frac{57.37704875}{20}\right)$$

# $(\frac{C}{N})_{\text{coded}} = 6.10017893273 \text{ dB}$

$$\left(\frac{C}{N}\right)_{\text{coded}} = \text{EIRP} + \left(\frac{G}{T}\right)_{\text{coded}} + 228.6 - L_p - L_x - 10\log (Brf) + M$$

$$Ts = EIRP + GR - (\frac{c}{N})_{coded} + M + 228.6 - (Lp + Lx + BRF)$$

Ts = 48.975 + 40.39875687 - 6.10017893273 + 3 + 228.6 - (168.57652369212 + 28.195 + 73.01029996)

Ts = 45.09175428515 dBK

 $Ts = 10^{4.509175428515} K$ 

Ts = 32297.98500096295 K

$$Ts = (Fs - 1) T0$$

$$Fs = 1 + \frac{Ts}{T_0}$$

$$Fs = 1 + \frac{32297.98500096295}{290}$$

Fs = 112.37236207229

$$g_3 = \frac{5.045}{Fs - \frac{T_1}{T_0} - 5.045F3} \left[ (L_4 - 1) + (L_5 - 1) \times 10^{0.778} + (L_6 - 1) \times 10^{1.465} + (F_7 - 1) \times 10^{2.262} + (L_8 - 1) \times 10^{1.465} \right]$$

$$)x10^{-0.237} + (L_9 - 1) \times 10^{1.043} + (F_{10} - 1) \times 10^{1.831} + (L_{11} - 1) \times 10^{-0.369} + (F_{12} - 1) \times 10^{0.756}$$

 $g_3 = 43.37332905242$ 

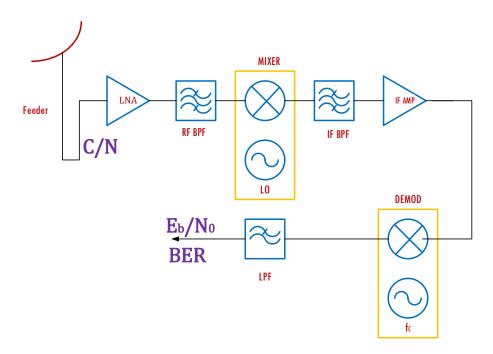
$$G_3 = 10 \log_{10} (g_3) = 10 \log (43.37332905242)$$

 $G_3 = 16.3722 dB$ 

#### CONCLUSION

The gain of the RF Amplifier decreases when a coding channel is used. So with coding we can low the noise figure efficiently even with a low RF Amplifier gain

# **PART TWO:**



1) Using the figure below and the information in table 1, calculate the system *Noise Temperature.* 

By taking:

$$1 => Antenna$$
  $5 => IF BPF$ 

$$2 \Rightarrow LNA$$
  $6 \Rightarrow IFAMP$ 

$$3 => RF BPF$$
  $7 => DEMOD$ 

$$4 => Mixer$$
  $8 => LPF$ 

$$F_{S} = F_{1} + \frac{F_{2}-1}{g_{1}} + \frac{F_{3}-1}{g_{1}g_{2}} + \frac{F_{4}-1}{g_{1}g_{2}g_{3}} + \frac{F_{5}-1}{g_{1}g_{2}g_{3}g_{4}} + \frac{F_{6}-1}{g_{1}g_{2}g_{3}g_{4}g_{5}} + \frac{F_{7}-1}{g_{1}g_{2}g_{3}g_{4}g_{5}g_{6}} + \frac{F_{8}-1}{g_{1}g_{2}g_{3}g_{4}g_{5}g_{6}g_{7}}$$

With:  $g_1 = 1$ 

The components 3; 4; 5; 7 and 8 are passive so their  $g=rac{1}{L}$  and their F=L

So

$$F_{S} = F_{1} + \frac{F_{2}-1}{1} + \frac{L_{3}-1}{1 \cdot g_{2}} + \frac{L_{3}(L_{4}-1)}{1 \cdot g_{2}} + \frac{L_{4}L_{3}(L_{5}-1)}{1 \cdot g_{2}} + \frac{L_{5}L_{4}L_{3}(F_{6}-1)}{1 \cdot g_{2}} + \frac{L_{5}L_{4}L_{3}(L_{7}-1)}{g_{2} \cdot g_{6}} + \frac{L_{7}L_{5}L_{4}L_{3}(L_{8}-1)}{g_{2} \cdot g_{6}}$$

With:

With: 
$$L_3 = 10^{0.525}$$
 
$$L_4L_3 = 10^{0.525} \times 10^{0.737} = 10^{1.262}$$
 
$$L_5L_4L_3 = 10^{1.262} \times 10^{0.595} = 10^{1.857}$$
 
$$L_7L_5L_4L_3 = 10^{1.857} \times 10^{1.203} = 10^{3.06}$$
 
$$g_2 = 10^3$$
 
$$g_2 \ g_6 = 10^3 \times 10^{2.5} = 10^{5.5}$$
 
$$F_S = F_1 + F_2 - 1$$
 
$$+ \frac{1}{g_2} \left[ L_3 - 1 + L_3(L_4 - 1) + L_4L_3(L_5 - 1) + L_5L_4L_3(F_6 - 1) \right]$$
 
$$+ \frac{1}{g_2 \ g_6} \left[ L_5L_4L_3(L_7 - 1) + L_7L_5L_4L_3(L_8 - 1) \right]$$

Therefore:

$$\begin{aligned} &\mathsf{F_s} = 1 + \tfrac{606.1}{290} + \ 1.98 - 1 \, + \, \tfrac{1}{10^3} \left[ \ 10^{0.525} - 1 \, + \, 10^{0.525} (10^{0.737} - 1) \, + \\ & 10^{1.262} \left( 10^{0.595} - 1 \right) + 10^{1.857} \left( 2.75 - 1 \right) \right] & + & \tfrac{1}{10^{5.5}} \left[ \\ & 10^{1.857} \left( 10^{1.203} - 1 \right) + \ 10^{3.06} (10^{0.555} - 1) \right] \end{aligned}$$

Fs = 4.27965262699

That gives us  $T_s = (F_s - 1) To = (4.27965262699 - 1) \times 290$ 

2) Using the information in tables 1 and 2, find the  $\left(\frac{G}{T}\right)$  of the satellite receiver, when no Channel *coding Scheme* is applied.

$$\left(\frac{c}{N}\right)$$
 = EIRP + G<sub>R</sub> - T<sub>s</sub>+M + 228.6 - L<sub>p</sub> - L<sub>X</sub> - 10log<sub>10</sub>(brf)

$$G_R = (\frac{c}{N}) + T_s + 10 Log_{10}(brf) - EIRP - 228.6 + (L_X + L_p) - M$$

$$\mathsf{Lp} = 92.44 + 20 \ Log_{10}(1027) + 20 Log_{10}(2.4)$$

Lp = 160.27563370618 dB

 $L_X$  = rain attenuation + atmospheric loss + poynting loss + polarization loss + other loss  $L_X$  = 26.37 dB

$$\left(\frac{c}{n}\right)_D = \frac{1}{\left(\frac{c}{n}\right)_T - \frac{1}{\left(\frac{c}{n}\right)_U}} = \frac{1}{\frac{1}{10^{1.225} - \frac{1}{10^{1.575}}}} = 30.34075972$$

$$\left(\frac{c}{N}\right)_D = 14.82 \text{ dB}$$

$$G_{R \text{ uncoded}} = 14.82 + 29.78225844607 + 160.27563370618 + 26.37 + 10 Log_{10}(6 \times 10^6) - 39.685 - 228.6 - 3$$

$$g_r = 10^{\frac{GR \text{ uncoded}}{10}}$$

$$g_{r \, uncoded} = 594.89520112628$$

3) The uncoded *Data Rate* depends on the *Digital Modulation* type and given as:

$$R_b = \frac{B_n \times (Number\ of\ bits/symbol)}{(1+\alpha)(Hz.s/symbol)} \rightarrow bits/s$$

Where  $B_n$  and  $\alpha$  are the **Noise Bandwidth** in Hz and the **Roll-off factor**, respectively,

Find the data rate in Mbps.

$$R_b = \frac{B_n * N}{1 + \alpha} = \frac{6Mhz * N}{1 + 0.25}$$

Where  $M = 2^N = >$  N = 4 bits /Symbol

$$R_b = \frac{6*4}{1,25}$$

$$R_b = \frac{96}{5}$$
 Mbps

4) Find  $\left(\frac{e_b}{n_0}\right)$  at the output of the digital demodulator and derive the *BER* (probability of error) at the output of the digital demodulator

We have 
$$\frac{eb}{n_0} = (\frac{C}{n}) \times \frac{brf}{R_b}$$
 with  $b_{rf} = 6$  Mhz and  $R_b = \frac{96}{5}$  Mbps

So 
$$\left(\frac{Eb}{N_0}\right) = \left(\frac{C}{N}\right) + 10Log_{10}\left(\frac{brf}{R_b}\right)$$

$$\left(\frac{Eb}{N_0}\right) = 14.82 + 10 Log_{10}\left(\frac{5 \times 6}{96}\right)$$

$$(\frac{Eb}{N_0})_{uncoded}$$
 = 9.7685002168 dB

Then we get BER by:

$$BER = 4(1 - \frac{1}{M}) \quad Q \left[ \int_{\frac{3log_2M}{M-1}}^{\frac{E_b}{N_0}} \left( \frac{E_b}{N_0} \right) \right] \quad and \quad M = 16 \text{ and } \left( \frac{e_b}{n_0} \right)_{uncoded} = 10^{\frac{\left( \frac{E_b}{N_0} \right) uncoded}{10}}$$

$$BER = 3Q \left( \sqrt{\frac{4}{5}} \left( \frac{e_b}{n_0} \right)_{uncoded} \right)$$

BER=3Q(2.75403848203)

5) Find the *Coding Gain* using the result obtained in 4) and the *Required BER* in table 2.

Gc = 
$$20log_{10} \left[ erfc^{-1}(2BER_{coded}) \right]$$
 -  $20log_{10} \left[ erfc^{-1}(2BER_{uncoded}) \right]$  +  $10log_{10}(r_c)$ 

$$Gc=20log_{10}$$
  $\left[5.303968158477018302364\right]$  -

$$20log_{10} \left[ 1.677739629706683560762 \right] + 10log_{10}(\frac{5}{6})$$

$$Gc = 9.20571444178 dB$$

6) Calculate the new value of  $\left(\frac{e_b}{n_0}\right)$  at the output of the coder.

We have 
$$(\frac{Eb}{N_0})_{coded} = (\frac{Eb}{N_0})_{uncoded} - G_{c}$$
, in DB, then we

get it in linear value by:

$$(\frac{eb}{n_0})_{\text{coded}} = 10^{\frac{\left(\frac{Eb}{N_0}\right)coded}{10}} = 10^{\frac{\left(\frac{Eb}{N_0}\right)uncoded - Gc}{10}} = 10^{\frac{9.7685002168 - 9.20571444178}{10}}$$

$$(\frac{eb}{n_0})_{coded} = 1.13835724912$$

7) Find the data rate at the output of the coder

$$R_{bcoded} = R_b x r_c$$

$$R_{bcoded} = 16 Mbps$$

8) Derive the value of the coded  $\left(\frac{c}{n}\right)$ ,

$$\left(\frac{c}{n}\right)_{\text{coded}} = \left(\frac{eb}{n0}\right)_{\text{coded}} \times \frac{(R_b)coded}{B_n} = 1.13835724912 \times \frac{16}{6}$$

$$(\frac{c}{n})_{\text{coded}} = 3.03561933099$$

9) let find the new  $(\frac{G}{T})$ 

$$\left(\frac{C}{N}\right)_{\text{coded} = \text{EIRP}} + \left(\frac{G}{T}\right)_{\text{coded}} + 228.6 - L_p - L_x - 10\log (Brf) + M$$

So 
$$G_{r=}(\frac{C}{N})_{coded} + T_{s} + -EIRP - 228.6 + L_{p} + L_{x} + 10log (Brf) - M$$

=  $10\log (3.03561933099) + 29.78225844607 - 39.685 - 228.6 + 160.27563370618 + 26.37 + 10\log (6 x 10^6) - 3$ 

 $G_{r \, coded} = 17.74687775367 \, dB$ 

$$g_r = 10^{\frac{17.74687775367}{10}}$$

$$g_{r coded} = 59.52340617434$$

10) With the gain of antenna parabolic

$$g = \frac{4\pi\eta a_{phy}}{\lambda^2} = (\frac{\pi D}{\lambda})^2 \eta$$

i) Let show that D = 
$$\sqrt{\frac{4a_{phy}}{\pi}}$$

According to the formula of g, we have

$$\frac{4\pi\eta a_{phy}}{\lambda^2} = \left(\frac{\pi D}{\lambda}\right)^2 \eta$$
 then we get

$$\left(\frac{\pi D}{\lambda}\right)^2 = \frac{4\pi \eta a_{phy}}{\lambda^2} \times \frac{1}{\eta}; \ (\pi D)^2 \times \frac{4\pi \eta a_{phy}}{\lambda^2} \times \frac{1}{\eta} \times \lambda^2$$

$$= 4a_{phy}\pi$$

$$D^2 = \frac{{}^{4}a_{phy}}{\pi} \quad \text{so } D = \sqrt{\frac{{}^{4}a_{phy}}{\pi}}$$

$$D = \sqrt{\frac{\lambda^2 g}{\pi^2 \eta}} = \frac{\lambda}{\pi} \sqrt{\frac{guncoded}{\eta}} = \frac{c}{\pi f} \sqrt{\frac{guncoded}{\eta}}$$

$$D = \frac{3 \times 10^8}{\pi \times 2.4 \times 10^9} \sqrt{\frac{594.89520112628}{0.6075}}$$

ii) Let calculate D after coding

$$D = \frac{C}{\pi f} \sqrt{\frac{g_{coded}}{\eta}}$$

$$D = \frac{3 \times 10^8}{\pi \times 2.4 \times 10^9} \sqrt{\frac{59.52340617434}{0.6075}}$$

$$D = 0.39385003361 \text{ m}$$

11)

# **CONCLUSION**

The diameter of the antenna without coding is 1.24510859758 m which is difficult to achieve even impossible with small receivers such as mobile phone laptop.

The code allowed to go from a diameter of  $1.24510859758\ m$  to a one of  $0.39385003361\ m$  which more practical. So the code make it possible to carry out the antenna.