



Development of Batteries Smart Management System Using Machine Learning

Ahmed Shokry & Eric Moulines

Center for Applied Mathematics, Ecole Polytechnique, France

Outlines on work done/ in progress

- 1) **Charging Control Based on ML** (conference and journal paper)
- 2) **State Of-Health (SOH) Estimation** (standard battery tests, MIT & NASA datasets)
- 3) **SOH & Remaining Useful Life (RUL) Estimation** (field data, in progress)
- 4) **Report on SOH and RUL Estimation from field data** (in progress)



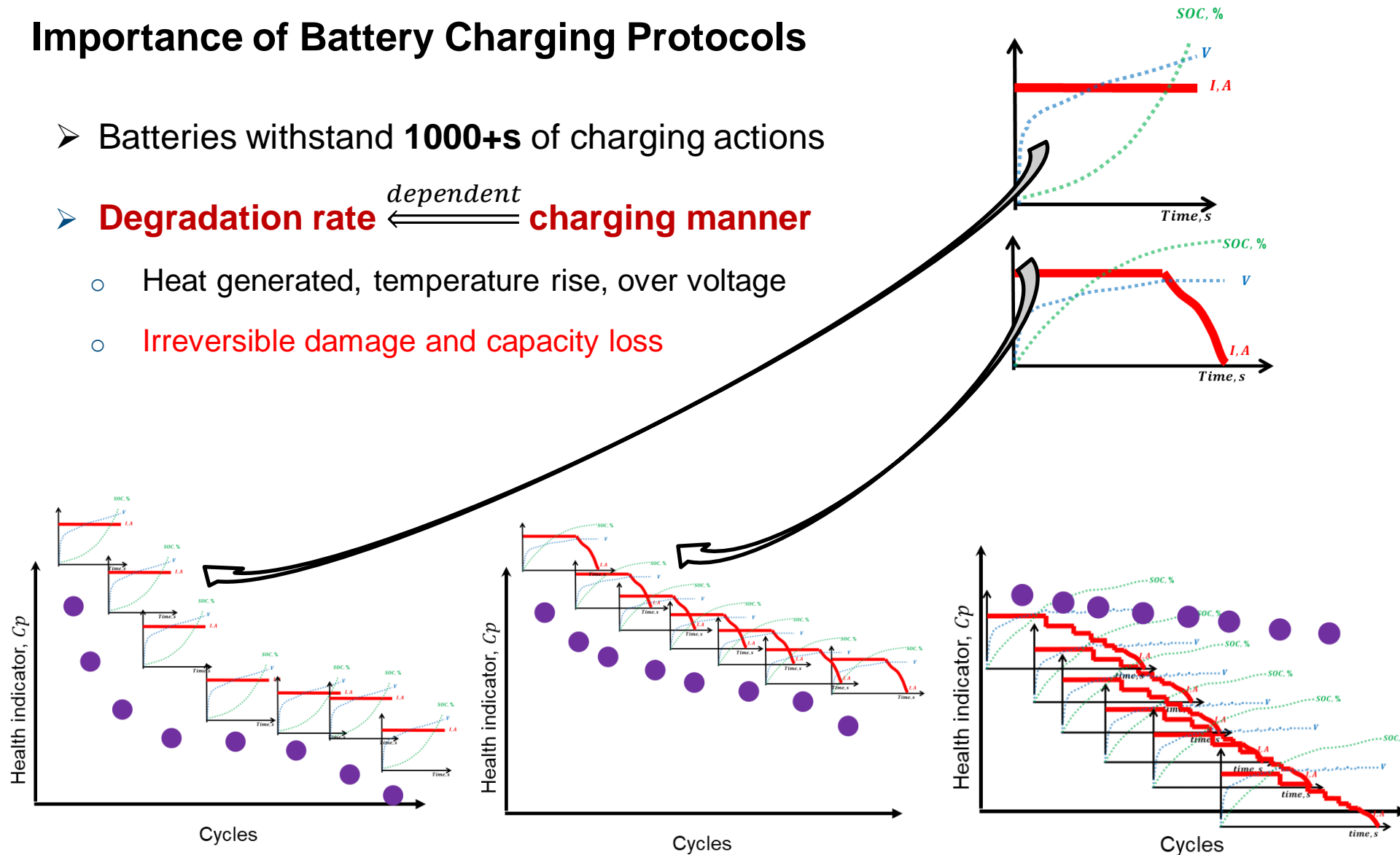
1) Charging Control Based on ML

1.1) Problem Context

3

Importance of Battery Charging Protocols

- Batteries withstand **1000+s** of charging actions
- **Degradation rate** $\xleftarrow{\text{dependent}}$ **charging manner**
 - Heat generated, temperature rise, over voltage
 - Irreversible damage and capacity loss





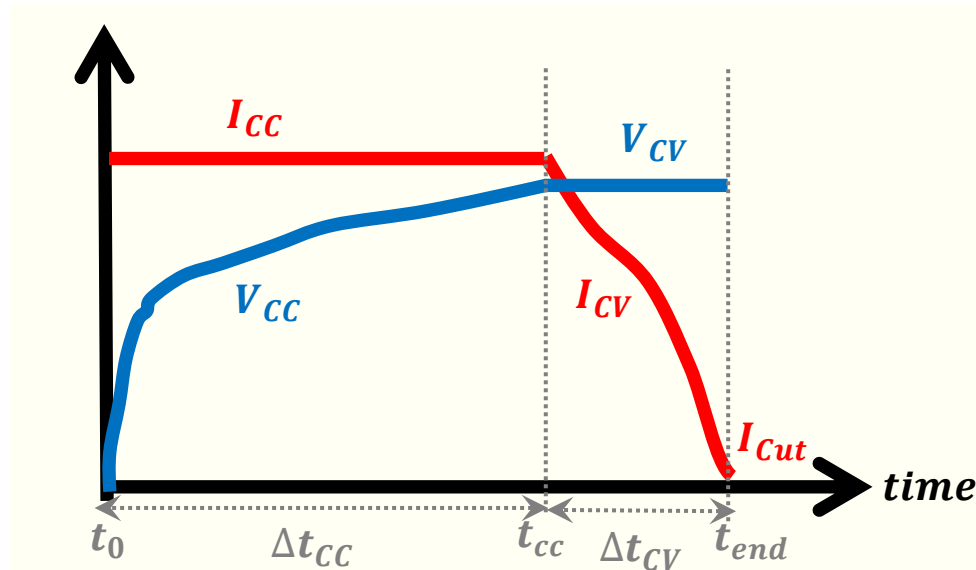
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4

1.1) Problem Context

- BMSs rely on empirical charging protocols, such as:

Constant Current-Constant Voltage (CC-CV)



- Experimentally designed protocols
- **Fair** performance (speed & safety), but **not optimal**, since
 - do not exploit knowledge about battery dynamic (electrochemical, thermal, etc.)
 - do not consider real-time conditions (initial SOC, temperature, etc.)



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5

1.1) Problem Context

- Advanced BMSs (ABMSs) rely on

**Real time application,
considering
disturbances**

**Battery Physics-
based Models**

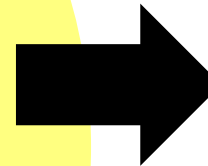
Electrochemical,
electrical, hydraulic
models

**Charging
setpoints &
Constraints**

(95%) & (cutoff voltages)

**Optimization
Algorithms**

Derivative based, in
most cases



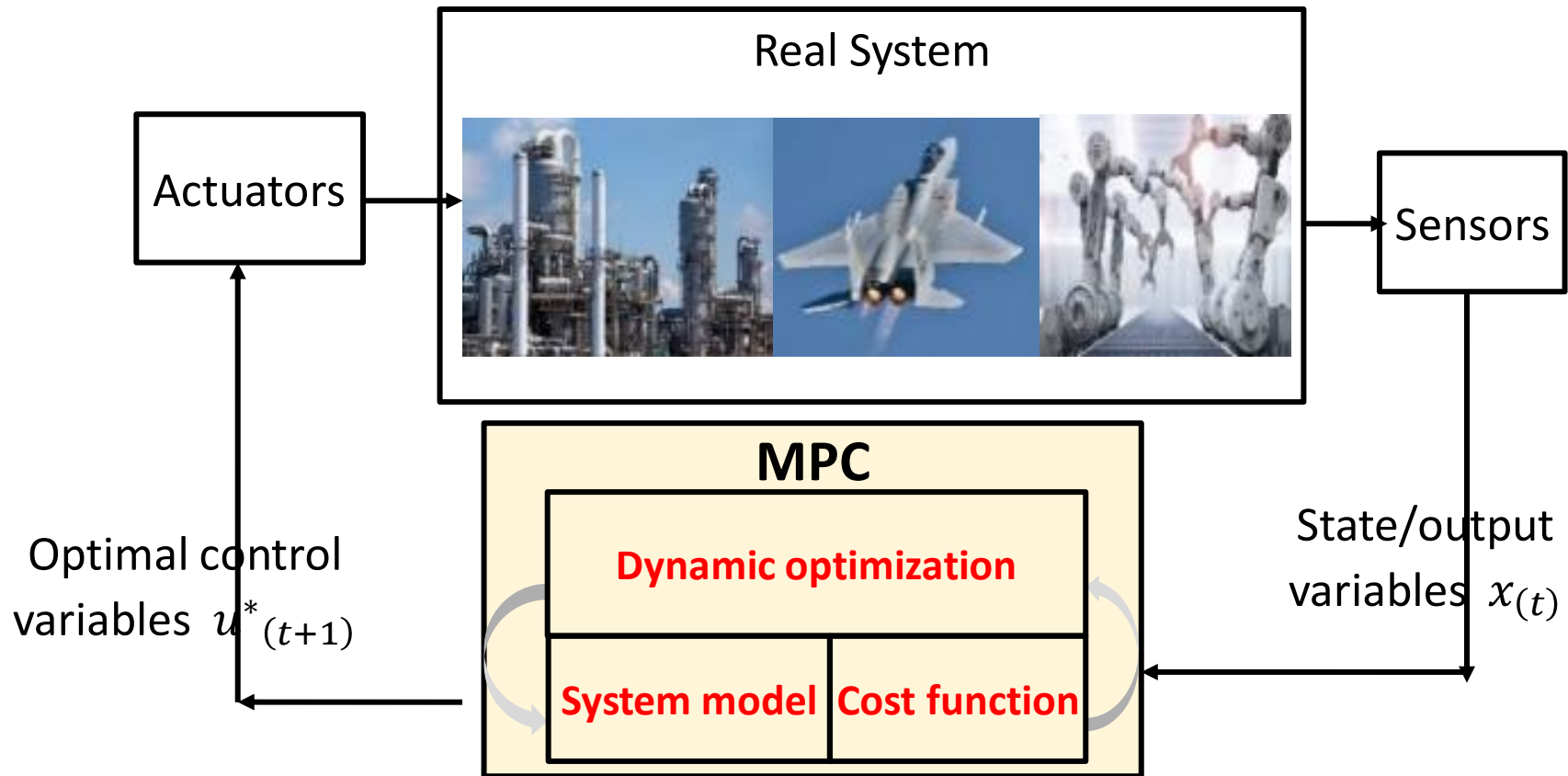
**Model Predictive
Control of Charging**



1) Charging Control Based on ML

1.2) Model Predictive Control (MPC)

6



Advantages:

- Multivariate nonlinear systems: complex Multiple-Inputs-Multiple Outputs (MIMO) relations
- Hard constraint on state and/or control variables

Perfectly suit for batteries (MIMO, nonlinear systems; operational, safety and health constraints)



1) Charging Control Based on ML

1.2) Model Predictive Control (MPC)

7

$$\min_{u_0, \dots, u_{N_p}} J = x'_{N_p} P x_{N_p} + \sum_{k=0}^{N_p-1} [(x_k - \check{r})' Q (x_k - \check{r}) + \Delta u'_k \mathcal{R} \Delta u_k]$$

S.T.:

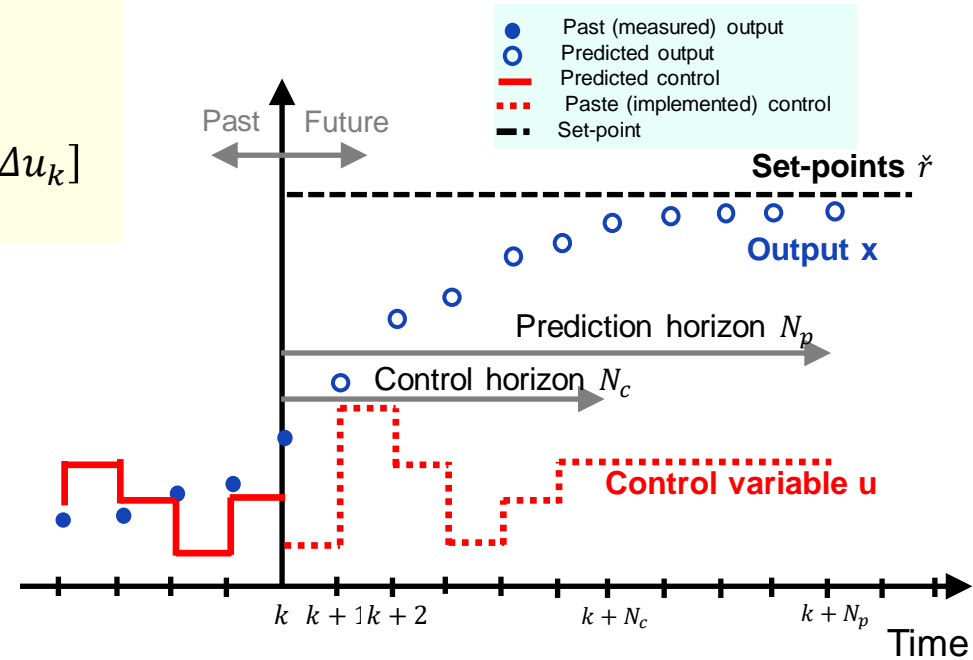
$$x_{\min} \leq x_k \leq x_{\max},$$

$$u_{\min} \leq u_k \leq u_{\max},$$

$$x_{k+1} = \mathbf{F}(x_k, u_k), \quad x \in R^m, u \in R^v,$$

$$g_l(x_k, u_k) \leq 0, \quad l = 1, 2, \dots, L$$

- x_k : State/output variables
- u_k : Control variables
- $\check{r} \in R^m$: Setpoints
- Δu_k : Control increment $\Delta u_k = u_k - u_{k-1}, k = 0:N_p$
- \mathbf{F} : System model
- g_l : Constraints
- Q, P, \mathcal{R} : coefficient matrices



Shortcomings:

- Very demanding computations
 - Repeated solution of an *optimal control optimization*
- Infeasible for operational implementation in many practical cases

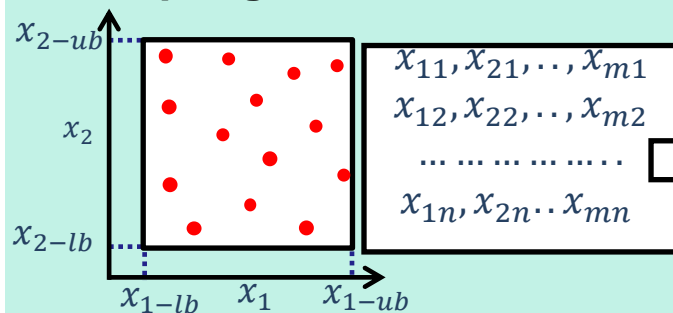


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1.3) Machine-learning-based Explicit MPC

Methodology 3: ML- based explicit MPC

1- Sampling over state domain



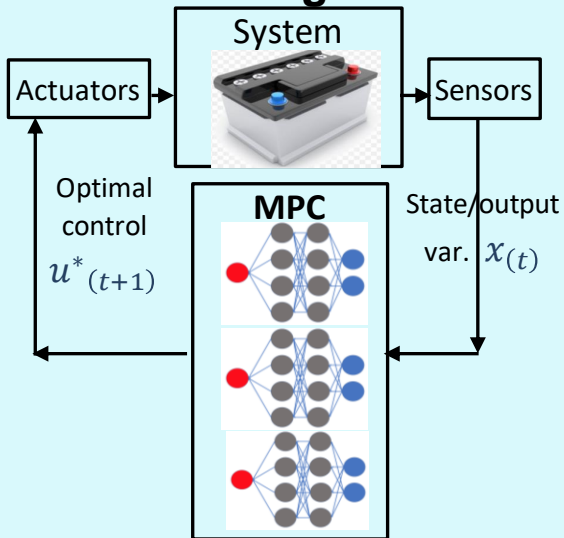
2- Data generation (via MPC)

$$\min_{u_k, \dots, u_{k+N_p}} J(u_k, \dots, u_{k+N_p}, x_k, \dots, x_{k+N_p}) = x'_{N_p} P x_{N_p} + \sum_{k=0}^{N_p-1} [(x_k - \check{r}) Q (x_k - \check{r}) + \Delta u'_k R \Delta u_k]$$

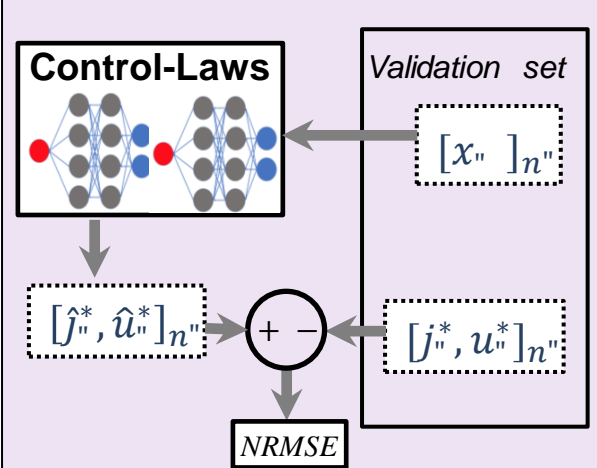
S.T.:

$$\begin{aligned} & j_1^*, u_{11}^*, u_{21}^*, \dots, u_{v1}^* \\ & j_2^*, u_{12}^*, u_{22}^*, \dots, u_{v2}^* \\ & \dots \dots \dots \\ & j_n^*, u_{1n}^*, u_{2n}^*, \dots, u_{vn}^* \end{aligned}$$

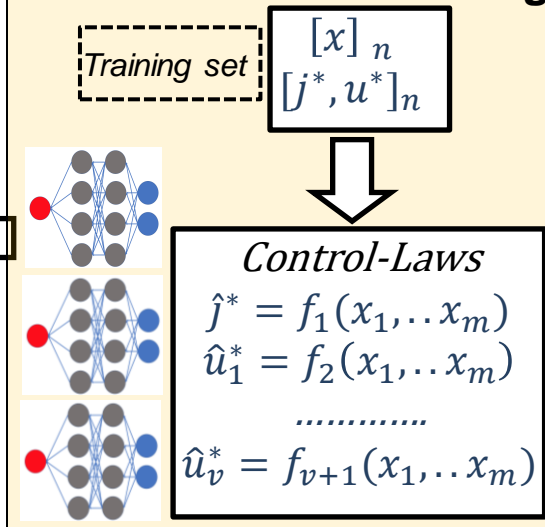
5- Online usage



4- Control Laws validation



3- Control Laws building

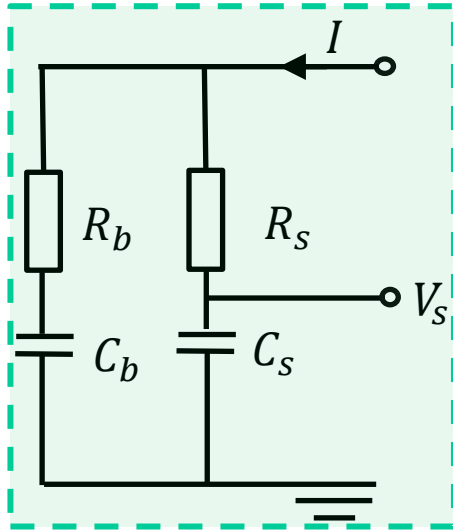




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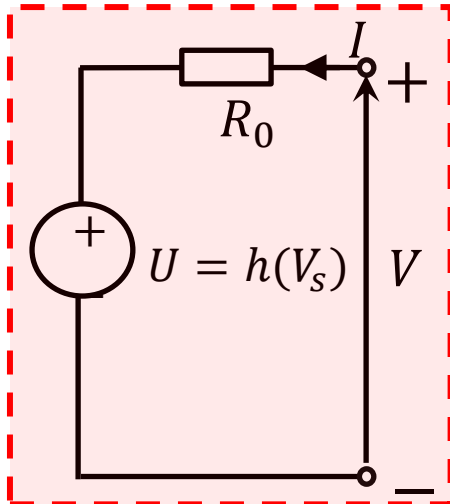
9

1.4) Applications : E-MPC of Nonlinear Double-Capacitors



First part

- Two capacitor-resistor circuits
- $R_s - C_s$: electrode surface.
- $R_b - C_b$: electrode's inner bulk.
- Parallel link \Rightarrow migration of charge through the electrode
- V_s and V_b : voltages across C_s and C_b



Second part

- $U = h(V_s)$: is the open-circuit voltages
- R_0 : internal resistance
- V : terminal voltage
- I : charging/discharging current

Mathematical model

$$\begin{bmatrix} \frac{dV_b(t)}{dt} \\ \frac{dV_s(t)}{dt} \end{bmatrix} = A \begin{bmatrix} V_b(t) \\ V_s(t) \end{bmatrix} + B I(t)$$

$$SOC(t) = \frac{C_b V_b(t) + C_s V_s(t)}{C_b + C_s}$$

$$V(t) = V_{oc}(t) + R_0 I(t)$$

$$R_0(t) = \beta_0 + \beta_1 \exp(-\beta_3(1 - SOC(t)))$$

$$V_{oc}(t) = h(V_s(t)) = \alpha_0 + \alpha_1 V_s(t) + \alpha_2 V_s(t)^2 + \alpha_3 V_s(t)^3 + \alpha_4 V_s(t)^4 + \alpha_5 V_s(t)^5$$

$$A = \begin{bmatrix} \frac{-1}{C_b(R_b + R_s)} & \frac{1}{C_b(R_b + R_s)} \\ \frac{1}{C_s(R_b + R_s)} & \frac{-1}{C_s(R_b + R_s)} \end{bmatrix}, B = \begin{bmatrix} \frac{R_s}{C_b(R_b + R_s)} \\ \frac{R_b}{C_s(R_b + R_s)} \end{bmatrix}$$



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10

1.4) Applications : E-MPC of Nonlinear Double-Capacitors

MPC

$$\min_{I_0, \dots, I_{N_p}} J = \sum_{k=0}^{N_p-1} [(SOC_k - \check{r}) Q (SOC_k - \check{r}) + \Delta I'_k \mathcal{R} \Delta I_k]$$

S.T.:

Operational constraints

$$SOC_0 = 20\%, \check{r} = 90\%$$

$$0 \leq I_k \leq 3 \text{ Amp},$$

$$V_k \leq 4.2 \text{ Volt},$$

Health constraints

$$V_{s,k} - V_{b,k} \leq -0.04 SOC_k + 0.08$$

Model

$$x_{k+1} = F(x_k, u_k),$$

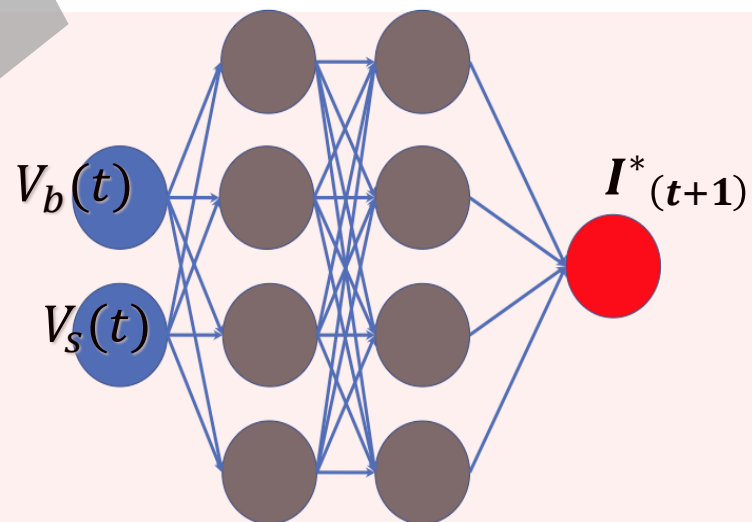
$$x \in R^{m=2}, u \in R^{v=1}$$

Coefficients

$$N_p = 10, N_u = 2, N_c = 1$$

$$Q = 1, \quad R = 0.1$$

Apply the
proposed
method



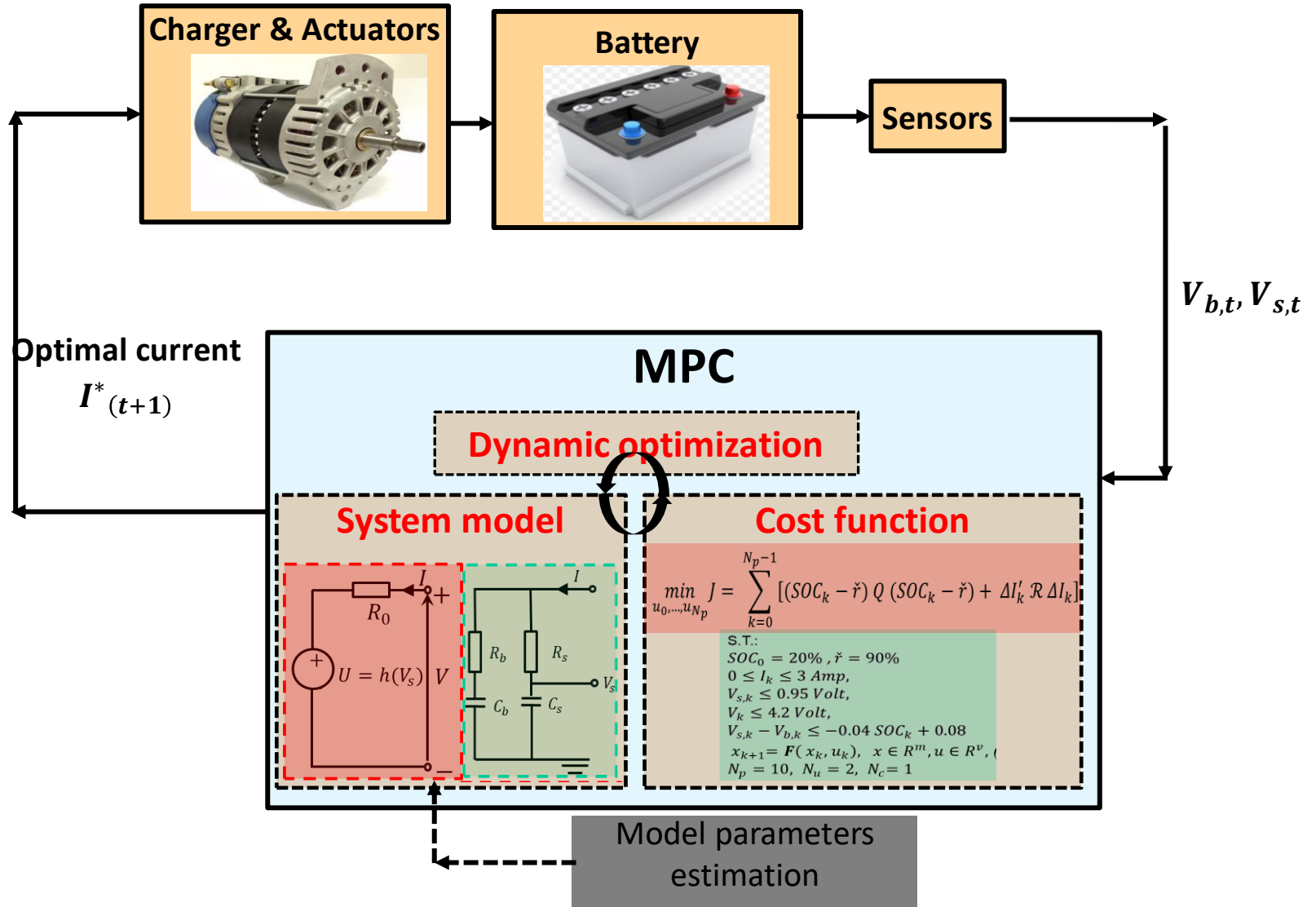


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11

1.4) Applications : E-MPC of Nonlinear Double-Capacitors

Application of the DNN-based control laws



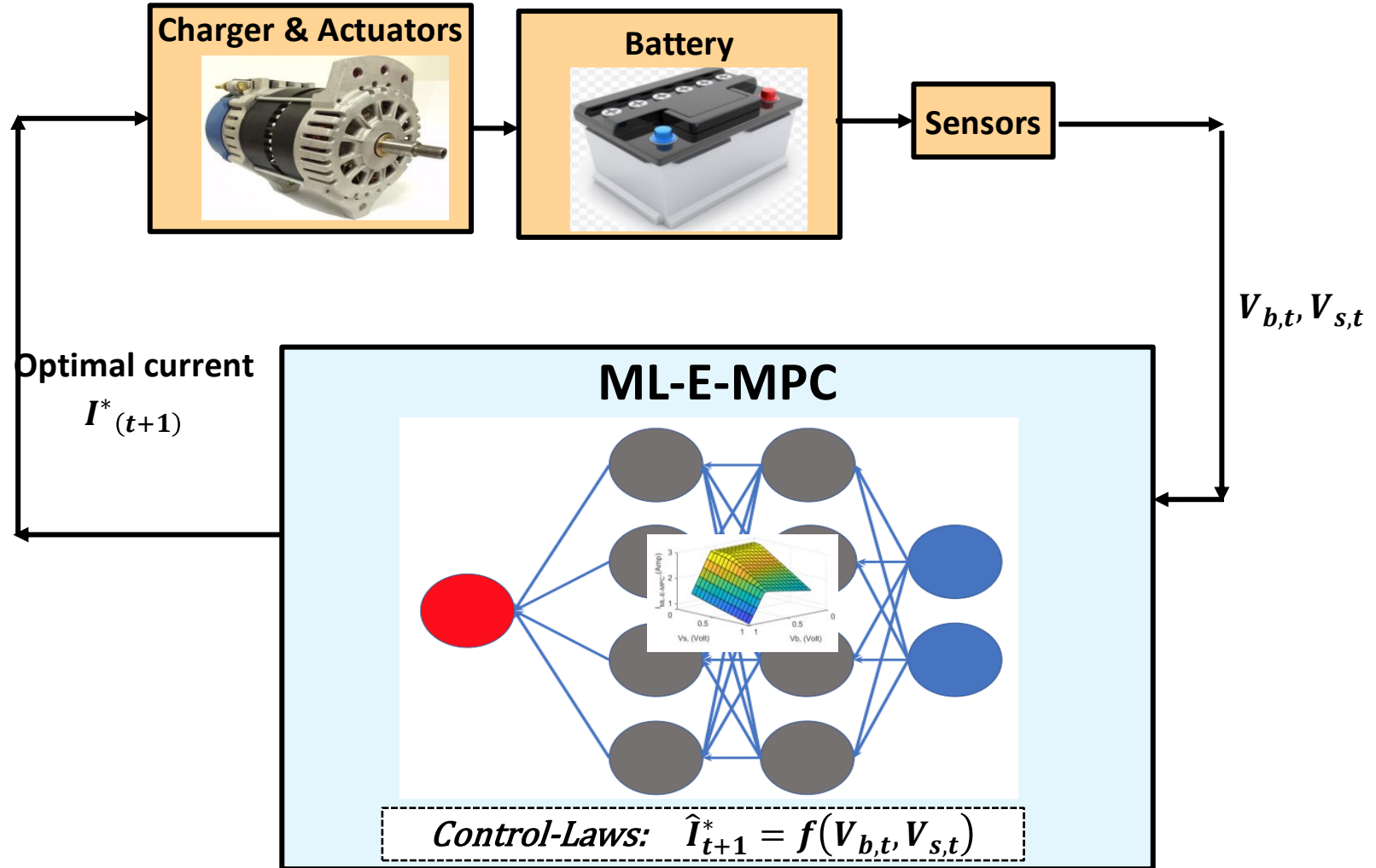


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12

1.4) Applications : E-MPC of Nonlinear Double-Capacitors

Application of the DNN-based control laws



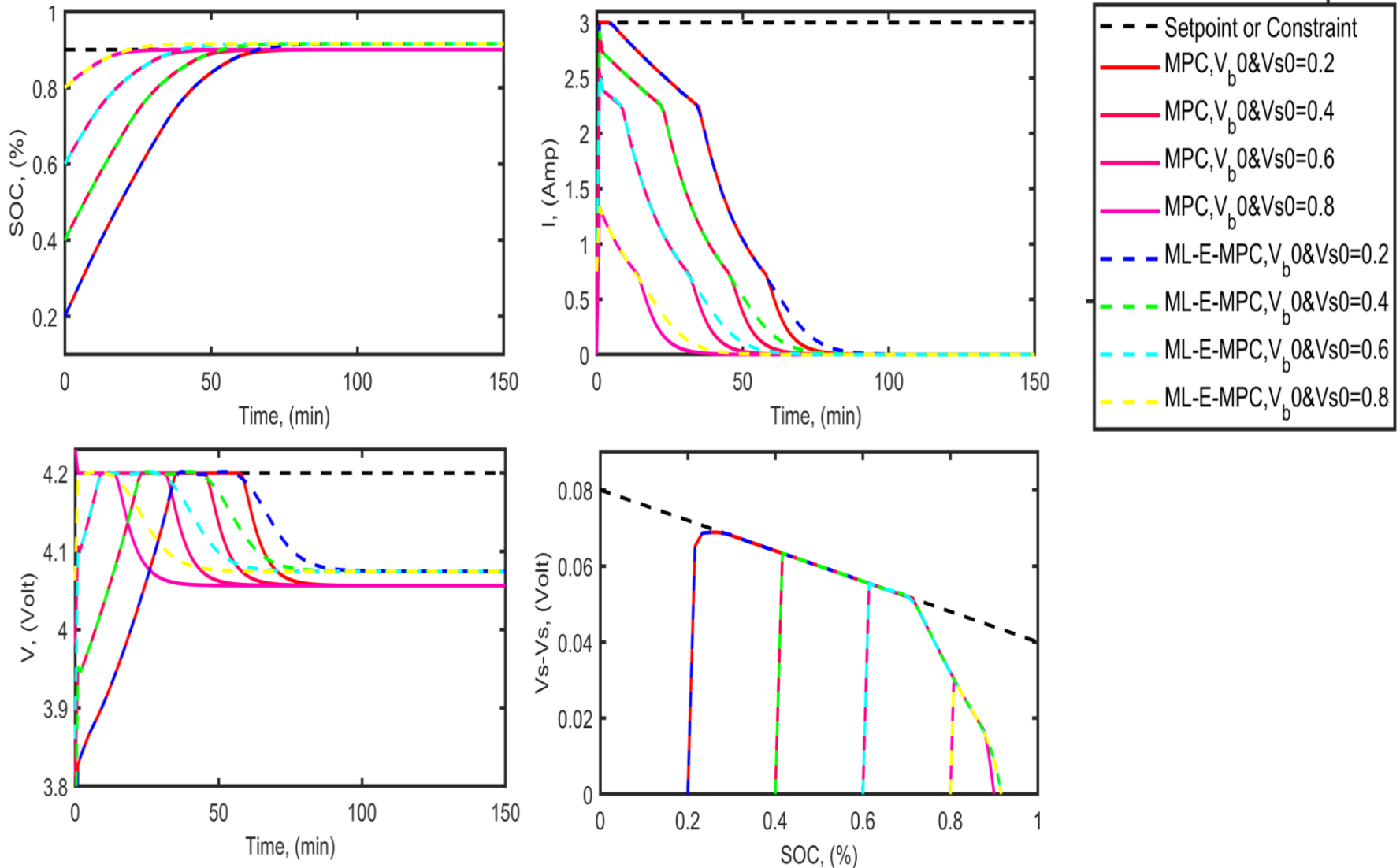


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13

1.4) Applications : E-MPC of Nonlinear Double-Capacitors

Application of the DNN-based control laws: different initial SOC



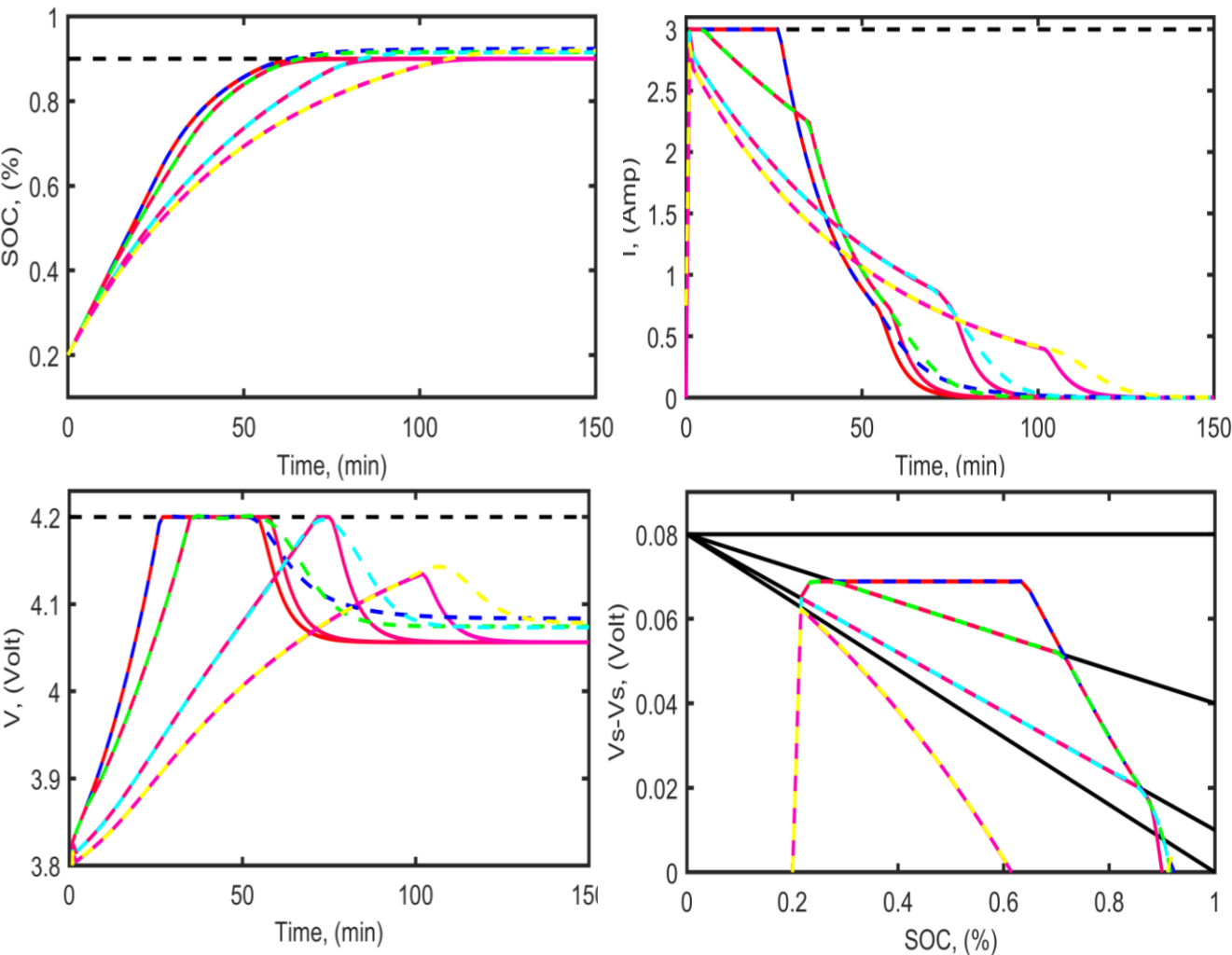


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14

1.4) Applications : E-MPC of Nonlinear Double-Capacitors

Application of the DNN-based control laws: different health-constraints adjustment



Performance

NRMSE (%)	Time (sec)*		Time saved (%)
	ML-E-MPC	MPC	
$\ll 1\%$	1.4	89.20	98.40