**Features** 





### Microphone Amplifier with AGC and Low-Noise Microphone Bias

### **General Description**

The MAX9814 is a low-cost, high-quality microphone amplifier with automatic gain control (AGC) and lownoise microphone bias. The device features a low-noise preamplifier, variable gain amplifier (VGA), output amplifier, microphone-bias-voltage generator and AGC control circuitry.

The low-noise preamplifier has a fixed 12dB gain, while the VGA gain automatically adjusts from 20dB to 0dB, depending on the output voltage and the AGC threshold. The output amplifier offers selectable gains of 8dB, 18dB, and 28dB. With no compression, the cascade of the amplifiers results in an overall gain of 40dB, 50dB, or 60dB. A trilevel digital input programs the output amplifier gain. An external resistive divider controls the AGC threshold and a single capacitor programs the attack/release times. A trilevel digital input programs the ratio of attack-to-release time. The hold time of the AGC is fixed at 30ms. The low-noise microphone-biasvoltage generator can bias most electret microphones.

The MAX9814 is available in the space-saving 12-bump UCSP™ (1.5mm x 2mm) and 14-pin TDFN packages. This device is specified over the -40°C to +85°C extended temperature range.

### **Applications**

Digital Still Cameras Digital Video Cameras **PDAs** Bluetooth Headsets

Entertainment Systems (e.g., Karaoke)

Two-Way Communicators High-Quality Portable Recorders

IP Phones/Telephone Conferencing

#### **♦ Automatic Gain Control (AGC)**

- ◆ Three Gain Settings (40dB, 50dB, 60dB)
- **♦ Programmable Attack Time**
- **♦ Programmable Attack and Release Ratio**
- ♦ 2.7V to 5.5V Supply Voltage Range
- ♦ Low Input-Referred Noise Density of 30nV/√Hz
- ♦ Low THD: 0.04% (typ)
- **♦ Low-Power Shutdown Mode**
- ♦ Internal Low-Noise Microphone Bias, 2V
- ♦ Available in the Space-Saving 12-Bump UCSP (1.5mm x 2mm) and 14-Pin TDFN (3mm x 3mm) **Packages**
- ♦ -40°C to +85°C Extended Temperature Range

### **Ordering Information**

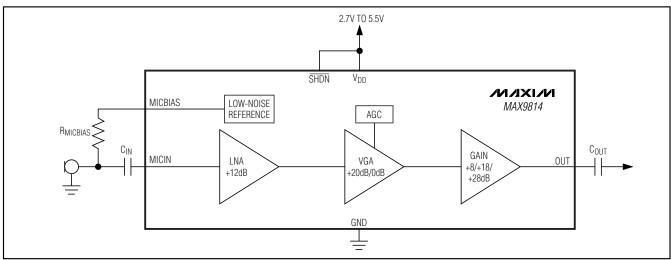
PART	TEMP RANGE	PIN- PACKAGE	PKG CODE	
MAX9814EBC+T	-40°C to +85°C	12 UCSP-12	B12-3	
MAX9814ETD+T	-40°C to +85°C	14 TDFN-14	T1433-2	

<sup>+</sup>Denotes a lead-free package.

#### Pin Configurations appear at end of data sheet.

UCSP is a trademark of Maxim Integrated Products, Inc.

### Simplified Block Diagram



Maxim Integrated Products 1

#### **ABSOLUTE MAXIMUM RATINGS**

V <sub>DD</sub> to GND	0.3V to +6V
All Other Pins to GND	$-0.3V$ to $(V_{DD} + 0.3V)$
Output Short-Circuit Duration	Continuous
Continuous Current (OUT, MICBIAS)	±100mA
All Other Pins	±20mA

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### **ELECTRICAL CHARACTERISTICS**

 $(V_{DD} = 3.3V, \overline{SHDN} = V_{DD}, C_{CT} = 470nF, C_{CG} = 2\mu F, GAIN = V_{DD}, T_{A} = T_{MIN}$  to  $T_{MAX}$ , unless otherwise specified. Typical values are at  $T_{A} = +25^{\circ}C$ .) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
GENERAL	•		· II			ч	
Operating Voltage	$V_{\mathrm{DD}}$	Guaranteed by PSRR test	2.7		5.5	V	
Supply Current	I <sub>DD</sub>			3.1	6	mA	
Shutdown Supply Current	ISHDN			0.01	1	μΑ	
Input-Referred Noise Density	en	BW = 20kHz, all gain settings		30		nV/√Hz	
Output Noise		BW = 20kHz		430		μV <sub>RMS</sub>	
Signal-to-Noise Ratio	SNR	BW = 22Hz to 22kHz (500mV <sub>RMS</sub> output signal)		61		dB	
		A-weighted		64			
Dynamic Range	DR	(Note 2)		60		dB	
Total Harmonic Distortion Plus	THD+N	$\begin{split} f_{IN} &= 1 \text{kHz, BW} = 20 \text{Hz to 20kHz,} \\ R_L &= 10 \text{k}\Omega, V_{TH} = 1 \text{V (threshold} = 2 \text{Vp-p), V}_{IN} = 0.5 \text{mV}_{RMS}, V_{CT} = 0 \text{V} \end{split}$		0.04		- %	
Noise	THD+N	$f_{IN}=1kHz$ , BW = 20Hz to 20kHz, $R_L=10k\Omega$ , $V_{TH}=0.1V$ (threshold = 200mV <sub>P-P</sub> ), $V_{IN}=30mV_{RMS}$ , $V_{CT}=2V$	0.2			/6	
Amplifier Input BIAS	VIN		1.14	1.23	1.32	V	
Maximum Input Voltage	V <sub>IN_MAX</sub>	1% THD		100		mV <sub>P-P</sub>	
Input Impedance	Z <sub>IN</sub>			100		kΩ	
		$GAIN = V_{DD}$	39.5	40	40.5		
Maximum Gain	А	GAIN = GND	49.5	50	50.6	dB	
		GAIN = unconnected	59.5	60	60.5		
		$GAIN = V_{DD}$	18.7	20	20.5		
Minimum Gain		GAIN = GND	29.0	30	30.8	dB	
		GAIN = unconnected	38.7	40	40.5		
Maximum Output Level	Vout_rms	1% THD+N, V <sub>TH</sub> = MICBIAS		0.707		V <sub>RMS</sub>	
Regulated Output Level		AGC enabled, V <sub>TH</sub> = 0.7V	1.26	1.40	1.54	V <sub>P-P</sub>	
AGC Attack Time	<sup>‡</sup> ATTACK	C <sub>T</sub> = 470nF (Note 3)		1.1		ms	
		A/R = GND 1:500			ms/ms		
Attack/Release Ratio	A/R	$A/R = V_{DD}$	1:2000				
		A/R = unconnected		1:4000			

#### **ELECTRICAL CHARACTERISTICS (continued)**

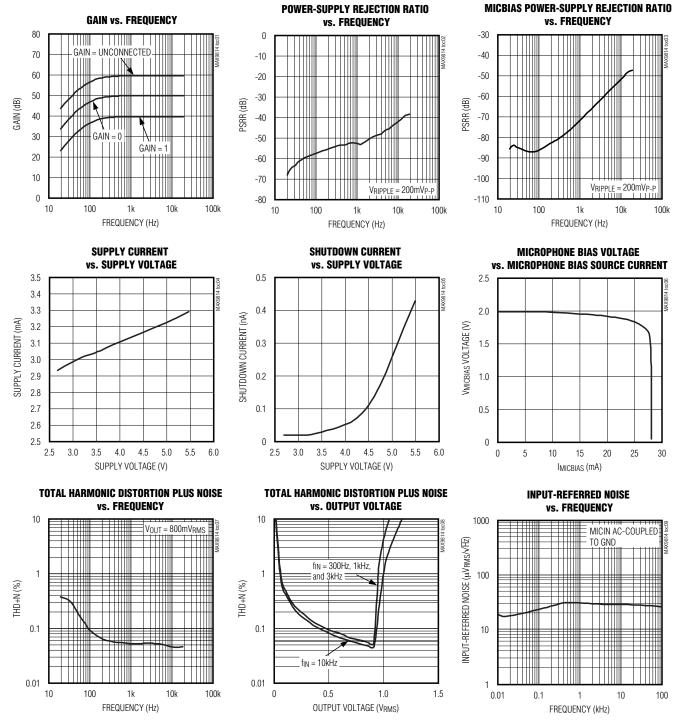
 $(V_{DD} = 3.3V, \overline{SHDN} = V_{DD}, C_{CT} = 470nF, C_{CG} = 2\mu F, GAIN = V_{DD}, T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise specified. Typical values are at  $T_A = +25^{\circ}C$ .) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
MICOUT High Output Voltage	VoH	IOUT sourcing 1mA		2.45		V
MICOUT Low Output Voltage	VoL	IOUT sinking 1mA		3		mV
MICOUT Bias		MICOUT unconnected	1.14	1.23	1.32	V
Output Impedance	Zout			50		Ω
Minimum Resistive Load	RLOAD_MIN			5		kΩ
Maximum Capacitive Drive	CLOAD_MAX			200		рF
Maximum Output Current	IOUT_MAX	1% THD, $R_L = 500\Omega$		1	2	mA
Output Short-Circuit Current	Isc		3	8		mA
		AGC mode; V <sub>DD</sub> = 2.7V to 5.5V (Note 4)	35	50		
Davier Cumply Dejection Datio	PSRR	f = 217Hz, V <sub>RIPPLE</sub> = 100mV <sub>P-P</sub> (Note 5)		55		4D
Power-Supply Rejection Ratio	PSRR	f = 1kHz, V <sub>RIPPLE</sub> = 100mV <sub>P-P</sub> (Note 5)		52.5		dB
		$f = 10kHz, V_{RIPPLE} = 100mV_{P-P}$ (Note 5)		43		
MICROPHONE BIAS						
Microphone Bias Voltage	VMICBIAS	IMICBIAS = 0.5mA	1.84	2.0	2.18	V
Output Resistance	RMICBIAS	IMICBIAS = 1mA		1		Ω
Output Noise Voltage	VMICBIAS_NOISE	IMICBIAS = 0.5mA, BW = 22Hz to 22kHz		5.5		μV <sub>RMS</sub>
		DC, V <sub>DD</sub> = 2.7V to 5.5V	70	80		
Power-Supply Rejection Ratio	PSRR	IMICBIAS = 0.5mA, VRIPPLE = 100mVp-p, fin = 1kHz		71		dB
TRILEVEL INPUTS (A/R, GAIN)						
Tri la continua di calcana Coment		A/R or GAIN = V <sub>DD</sub>	0.5V <sub>DD</sub> / 180kΩ	0.5V <sub>DD</sub> / 100kΩ	0.5V <sub>DD</sub> / 50kΩ	^
Tri-Level Input Leakage Current		A/R or GAIN = GND		0.5V <sub>DD</sub> / 100kΩ		mA
Input High Voltage	VIH		V <sub>DD</sub> x 0.	7		V
Input Low Voltage	V <sub>IL</sub>			\	/ <sub>DD</sub> x 0.3	V
Shutdown Enable Time	ton			60		ms
Shutdown Disable Time	toff			40		ms
DIGITAL INPUT (SHDN)						
SHDN Input Leakage Current			-1		+1	μΑ
Input High Voltage	VIH		1.3			V
Input Low Voltage	V <sub>IL</sub>				0.5	V
AGC THRESHOLD INPUT (TH)						
TH Input Leakage Current			-1		+1	μΑ

- Note 1: Devices are production tested at T<sub>A</sub> = +25°C. Limits over temperature are guaranteed by design.
- Note 2: Dynamic range is calculated using the EIAJ method. The input is applied at -60dBFS (0.707 $\mu$ V<sub>RMS</sub>), f<sub>IN</sub> = 1kHz.
- Note 3: Attack time measured as time from AGC trigger to gain reaching 90% of its final value.
- Note 4: CG is connected to an external DC voltage source, and adjusted until V<sub>MICOUT</sub> = 1.23V.
- Note 5: CG connected to GND with 2.2µF.

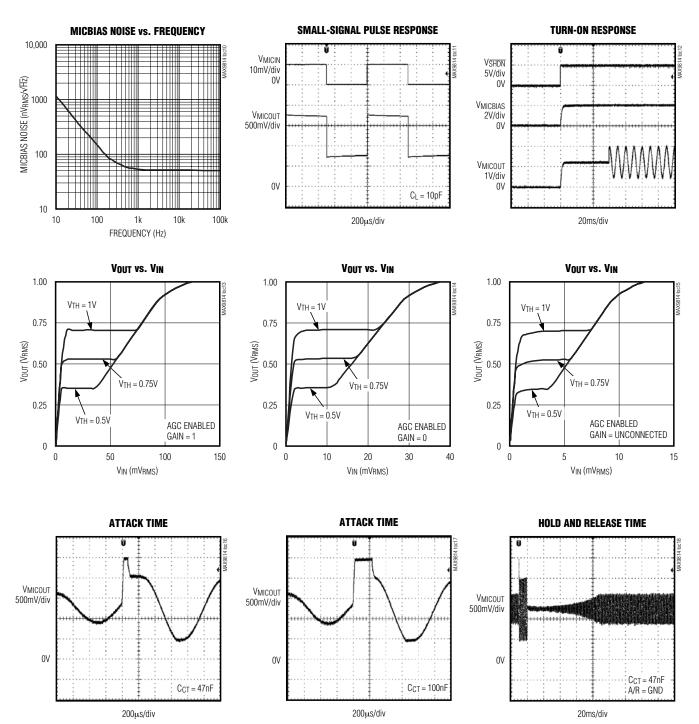
### **Typical Operating Characteristics**

 $(V_{DD}=5V, C_{CT}=470nF, C_{CG}=2.2\mu F, V_{TH}=V_{MICBIAS} \times 0.4, GAIN=V_{DD}$  (40dB), AGC disabled, no load,  $R_{L}=10k\Omega$ ,  $C_{OUT}=1\mu F$ ,  $T_{A}=+25^{\circ}C$ , unless otherwise noted.)



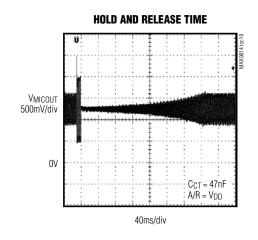
### Typical Operating Characteristics (continued)

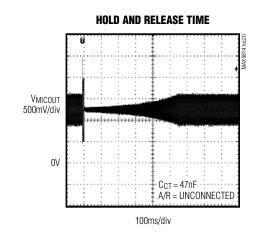
 $(V_{DD} = 5V, C_{CT} = 470nF, C_{CG} = 2.2\mu F, V_{TH} = V_{MICBIAS} \times 0.4, GAIN = V_{DD} (40dB), AGC disabled, no load, R<sub>L</sub> = 10k<math>\Omega$ , C<sub>OUT</sub> = 1 $\mu$ F, T<sub>A</sub> = +25°C, unless otherwise noted.)



#### Typical Operating Characteristics (continued)

 $(V_{DD} = 5V, C_{CT} = 470nF, C_{CG} = 2.2\mu F, V_{TH} = V_{MICBIAS} \times 0.4, GAIN = V_{DD}$  (40dB), AGC disabled, no load,  $R_L = 10k\Omega$ ,  $C_{OUT} = 1\mu F$ ,  $T_A = +25^{\circ}C$ , unless otherwise noted.)





### **Pin Description**

P	PIN		FUNCTION				
TDFN	UCSP	NAME	FUNCTION				
1	A1	CT	Timing Capacitor Connection. Connect a capacitor to CT to control the Attack and Release times of the AGC.				
2	B2	SHDN	Active-Low Shutdown Control				
3	A2	CG	Amplifier DC Offset Adjust. Connect a 2.2µF capacitor to GND to ensure zero offset at the output.				
4, 11	_	N.C.	No Connection. Connect to GND.				
5	А3	$V_{DD}$	Power Supply. Bypass to GND with a 1µF capacitor.				
6	A4	MICOUT	Amplifier Output				
7	B4	GND	Ground				
8	C4	MICIN	Microphone Noninverting Input				
9	В3	A/R	Tri-Level Attack and Release Ratio Select. Controls the ratio of attack time to release time for the AGC circuit.  A/R = GND: Attack/Release Ratio is 1:500  A/R = VDD: Attack/Release Ratio is 1:2000  A/R = BIAS: Attack/Release Ratio is 1:4000				
10	C3	GAIN	Tri-Level Amplifier Gain Control.  GAIN = V <sub>DD</sub> , gain set to 40dB.  GAIN = GND, gain set to 50dB.  GAIN = Unconnected, uncompressed gain set to 60dB.				
12	C2	BIAS	Amplifier Bias. Bypass to GND with a 0.47µF capacitor.				
13	C1	MICBIAS	Microphone Bias Output				
14	B1	TH	AGC Threshold Control. TH voltage sets gain control threshold. Connect TH to MICBIAS to disable the AGC.				

#### **Detailed Description**

The MAX9814 is a low-cost, high-quality microphone amplifier with automatic gain control (AGC) and a low-noise microphone bias. The MAX9814 consists of several distinct circuits: a low-noise preamplifier, a variable gain amplifier (VGA), an output amplifier, a microphone-bias-voltage generator, and AGC control circuitry.

An internal microphone bias voltage generator provides a 2V bias that is suitable for most electret condenser microphones. The MAX9814 amplifies the input in three distinct stages. In the first stage, the input is buffered and amplified through the low-noise preamplifier with a gain of 12dB. The second stage consists of the VGA controlled by the AGC. The VGA/AGC combination is capable of varying the gain from 20dB to 0dB. The output amplifier is the final stage in which a fixed gain of 8dB, 18dB, 20dB is programmed through a single tri-level logic input. With no compression from the AGC, the MAX9814 is capable of providing 40dB, 50dB, or 60dB gain.

#### **Automatic Gain Control (AGC)**

A device without AGC experiences clipping at the output when too much gain is applied to the input. AGC prevents clipping at the output when too much gain is applied to the input, eliminating output clipping. Figure 1 shows a comparison of an over-gained microphone input with and without AGC.

The MAX9814's AGC controls the gain by first detecting that the output voltage has exceeded a preset limit. The microphone amplifier gain is then reduced with a selectable time constant to correct for the excessive output-voltage amplitude. This process is known as the attack time. When the output signal subsequently lowers in amplitude, the gain is held at the reduced state for a short period before slowly increasing to the normal value. This process is known as the hold and release time. The speed at which the amplifiers adjust to changing input signals is set by the external timing capacitor CCT and the voltage applied to A/R. The AGC threshold can be set by adjusting VTH. Gain reduction is a function of input signal amplitude with a maximum AGC attenuation of 20dB. Figure 2 shows the effect of an input burst exceeding the preset limit, output attack, hold and release times.

If the attack and release times are configured to respond too fast, audible artifacts often described as "pumping" or "breathing" can occur as the gain is rapidly adjusted to follow the dynamics of the signal. For best results, adjust the time constant of the AGC to accommodate the source material. For applications in which music CDs are the main audio source, a 160µs attack time with an 80ms release time is recommended. Music applications typically require a shorter release time than voice or movie content.

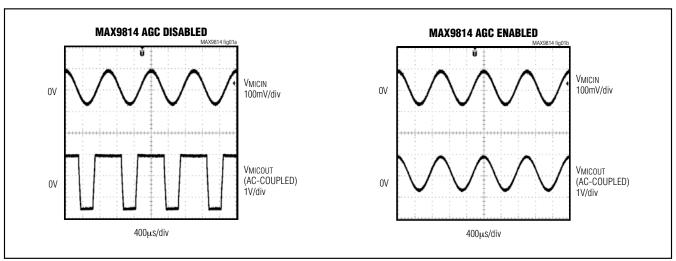


Figure 1. Microphone Input with and Without AGC

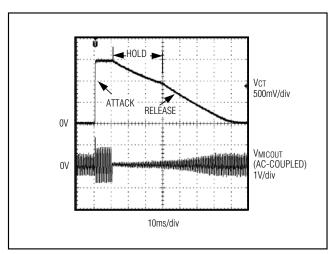


Figure 2. Input Burst Exceeding AGC Limit

#### Attack Time

The attack time is the time it takes for the AGC to reduce the gain after the input signal has exceeded the threshold level. The gain attenuation during the attack time is exponential, and defined as one-time constant. The time constant of the attack is given by 2400 x CCT seconds (where CCT is the external timing capacitor):

- Use a short attack time for the AGC to react quickly to transient signals, such as snare drum beats (music) or gun shots (DVD).
- Use a longer attack time to allow the AGC to ignore short-duration peaks and only reduce the gain when a noticeable increase in loudness occurs. Shortduration peaks are not reduced, but louder passages are. This allows the louder passages to be reduced in volume, thereby maximizing output dynamic range.

#### Hold Time

Hold time is the delay after the signal falls below the threshold level before the release phase is initiated. Hold time is internally set to 30ms and nonadjustable. The hold time is cancelled by any signal exceeding the set threshold level, and the attack time is reinitiated.

#### Release Time

The release time is how long it takes for the gain to return to its normal level after the output signal has fallen below the threshold level and 30ms hold time has expired. Release time is defined as release from a 20dB gain compression to 10% of the nominal gain setting after the input signal has fallen below the TH threshold and the 30ms hold time has expired. Release time is adjustable and has a minimum of 25s. The release time is set by picking an attack time using CCT

and setting the attack-to-release time ratio by configuring A/R as shown in Table 1:

- Use a small ratio to maximize the speed of the AGC.
- Use a large ratio to maximize the sound quality and prevent repeated excursions above the threshold from being independently adjusted by the AGC.

#### AGC Output Threshold

The output threshold that activates AGC is adjustable through the use of an external resistive divider. Once the divider is set, AGC reduces the gain to match the output voltage to the voltage set at the TH input.

#### **Microphone Bias**

The MAX9814 features an internal low-noise microphone bias voltage capable of driving most electret condenser microphones. The microphone bias is regulated at 2V to provide that the input signal to the low-noise preamplifier does not clip to ground.

#### **Applications Information**

#### **Programming Attack and Release Times**

Attack and release times are set by selecting the capacitance value between CT and GND, and by setting the logic state of A/R (Table 1). A/R is a tri-level logic input that sets the attack-to-release time ratio.

Table 1. Attack-and-Release Ratios

A/R ATTACK/RELEASE RATIO				
GND	1:500			
V <sub>DD</sub>	1:2000			
Unconnected	1:4000			

The attack and release times can be selected by utilizing the corresponding capacitances listed in Table 2.

Table 2. Attack-and-Release Time

		trelease (ms)				
ССТ	<sup>t</sup> ATTACK (ms)	A/R = GND	A/R = V <sub>DD</sub>	A/R = UNCONNECTED		
22nF	0.05	25	100	200		
47nF	0.11	55	220	440		
68nF	0.16	80	320	640		
100nF	0.24	120	480	960		
220nF	0.53	265	1060	2120		
470nF	1.1	550	2200	4400		
680nF	1.63	815	3260	6520		
1µF	2.4	1200	4800	9600		

#### **Setting the AGC Threshold**

To set the output-voltage threshold at which the microphone output is clamped, an external resistor-divider must be connected from MICBIAS to ground with the output of the resistor-divider applied to TH. The voltage VTH determines the peak output-voltage threshold at which the output becomes clamped. The maximum signal swing at the output is then limited to two times VTH and remains at that level until the amplitude of the input signal is reduced. To disable AGC, connect TH to MICBIAS.

#### **Microphone Bias Resistor**

MICBIAS is capable of sourcing 20mA. Select a value for RMICBIAS that provides the desired bias current for the electret microphone. A value of  $2.2k\Omega$  is usually sufficient for a microphone of typical sensitivity. Consult the microphone data sheet for the recommended bias resistor.

#### **Bias Capacitor**

The BIAS output of the MAX9814 is internally buffered and provides a low-noise bias. Bypass BIAS with a 470nF capacitor to ground.

#### **Input Capacitor**

The input AC-coupling capacitor (CIN) and the input resistance (RIN) to the microphone amplifier form a highpass filter that removes any DC bias from an input signal (see the *Typical Application Circuit/Functional Diagram*). CIN prevents any DC components from the input-signal source from appearing at the amplifier outputs. The -3dB point of the highpass filter, assuming zero source impedance due to the input signal source, is given by:

$$f_{-3dB\_IN} = \frac{1}{2\pi \times R_{IN} \times C_{IN}}$$

Choose C<sub>IN</sub> such that f<sub>-3dB\_IN</sub> is well below the lowest frequency of interest. Setting f<sub>-3dB\_IN</sub> too high affects the amplifier's low-frequency response. Use capacitors with low-voltage coefficient dielectrics. Aluminum electrolytic, tantalum, or film dielectric capacitors are good choices for AC-coupling capacitors. Capacitors with high-voltage coefficients, such as ceramics (non-COG dielectrics), can result in increased distortion at low frequencies.

#### **Output Capacitor**

The output of the MAX9814 is biased at 1.23V. To eliminate the DC offset, an AC-coupling capacitor (Cout) must be used. Depending on the input resistance (RL) of the following stage, Cout and RL effectively form a highpass filter. The -3dB point of the highpass filter, assuming zero output impedance, is given by:

$$f_{-3dB\_OUT} = \frac{1}{2\pi \times R_{I} \times C_{OUT}}$$

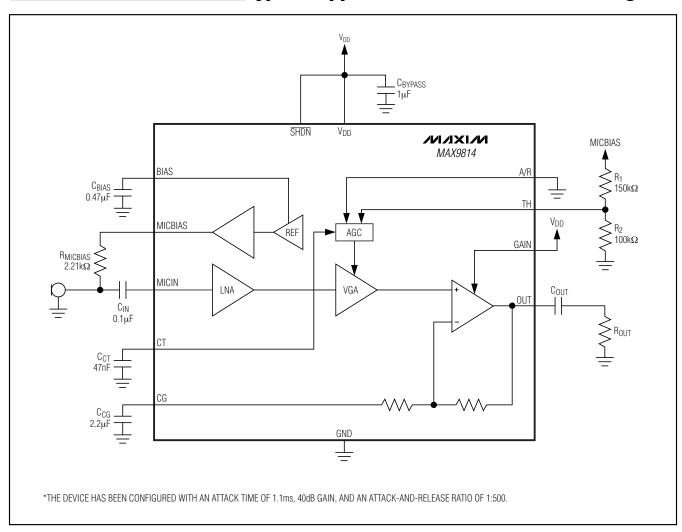
#### **Shutdown**

The MAX9814 features a low-power shutdown mode. When SHDN goes low, the supply current drops to 0.01µA, the output enters a high-impedance state, and the bias current to the microphone is switched off. Driving SHDN high enables the amplifier. Do not leave SHDN unconnected.

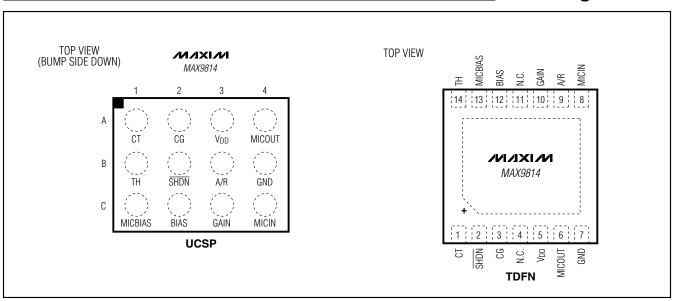
## Power-Supply Bypassing and PCB Layout

Bypass the power supply with a  $0.1\mu F$  capacitor to ground. Reduce stray capacitance by minimizing trace lengths and place external components as close to the device as possible. Surface-mount components are recommended. In systems where analog and digital grounds are available, connect the MAX9814 to analog ground.

### Typical Application Circuit/Functional Diagram



**Pin Configurations** 

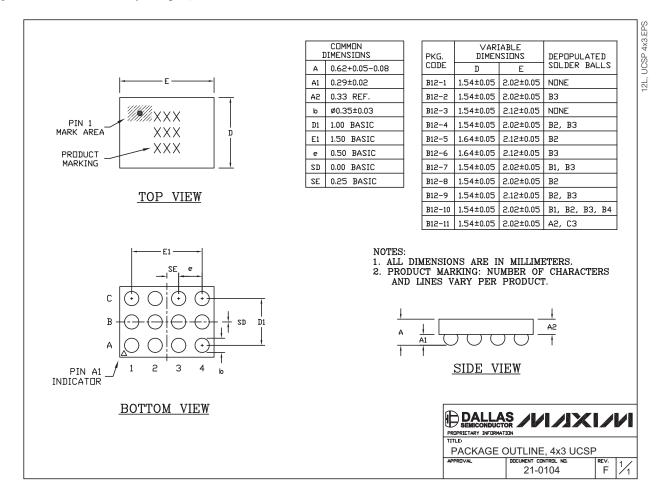


\_Chip Information

PROCESS: BiCMOS

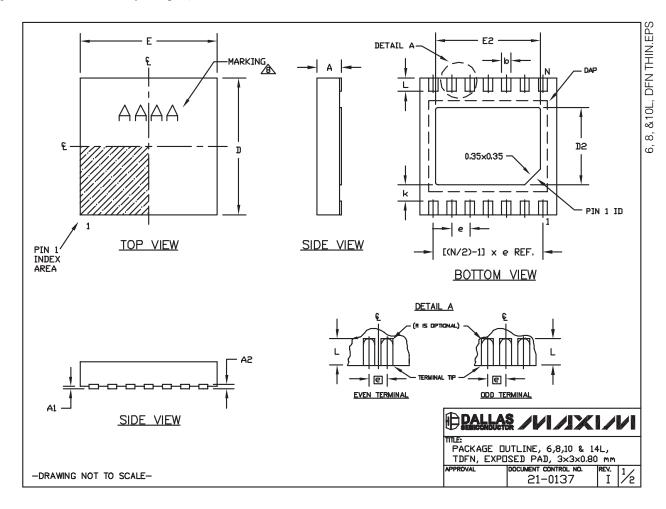
#### **Package Information**

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to <a href="https://www.maxim-ic.com/packages">www.maxim-ic.com/packages</a>.)



### **Package Information (continued)**

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to <a href="https://www.maxim-ic.com/packages">www.maxim-ic.com/packages</a>.)



#### Package Information (continued)

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to <a href="https://www.maxim-ic.com/packages">www.maxim-ic.com/packages</a>.)

COMMON DIMENSIONS					
SYMBOL MIN. MAX.					
Α	0.70	0.80			
D	2.90	3.10			
Е	2.90	3.10			
A1	A1 0.00 0.0				
L 0.20 0.40					
k 0.25 MIN.					
A2	0.20 REF.				

PACKAGE VARIATIONS								
PKG. CODE	N	D2	E2	е	JEDEC SPEC	b	[(N/2)-1] x e	
T633-2	6	1.50±0.10	2.30±0.10	0.95 BSC	MO229 / WEEA	0.40±0.05	1.90 REF	
T833-2	8	1.50±0.10	2.30±0.10	0.65 BSC	MO229 / WEEC	0.30±0.05	1.95 REF	
T833-3	8	1.50±0.10	2.30±0.10	0.65 BSC	MO229 / WEEC	0.30±0.05	1.95 REF	
T1033-1	10	1.50±0.10	2.30±0.10	0.50 BSC	MO229 / WEED-3	0.25±0.05	2.00 REF	
T1033-2	10	1.50±0.10	2.30±0.10	0.50 BSC	MO229 / WEED-3	0.25±0.05	2.00 REF	
T1433-1	14	1.70±0.10	2.30±0.10	0.40 BSC		0.20±0.05	2.40 REF	
T1433-2	14	1.70±0.10	2.30±0.10	0.40 BSC		0.20±0.05	2.40 REF	

#### NOTES:

- 1. ALL DIMENSIONS ARE IN mm. ANGLES IN DEGREES.
- 2. COPLANARITY SHALL NOT EXCEED 0.08 mm.
- 3. WARPAGE SHALL NOT EXCEED 0.10 mm.
- 4. PACKAGE LENGTH/PACKAGE WIDTH ARE CONSIDERED AS SPECIAL CHARACTERISTIC(S).
- 5. DRAWING CONFORMS TO JEDEC MO229, EXCEPT DIMENSIONS "D2" AND "E2". AND T1433-1 & T1433-2.
- 6. "N" IS THE TOTAL NUMBER OF LEADS.
- 7. NUMBER OF LEADS SHOWN ARE FOR REFERENCE ONLY.
- A MARKING IS FOR PACKAGE ORIENTATION REFERENCE ONLY.

TITLE:
PACKAGE DUTLINE, 6,8,10 & 14L,
TDFN, EXPOSED PAD, 3×3×0.80 mm

DOCUMENT CONTROL NO.

21-0137

-DRAWING NOT TO SCALE-

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.