

Summer 2016-2017 Final Exam

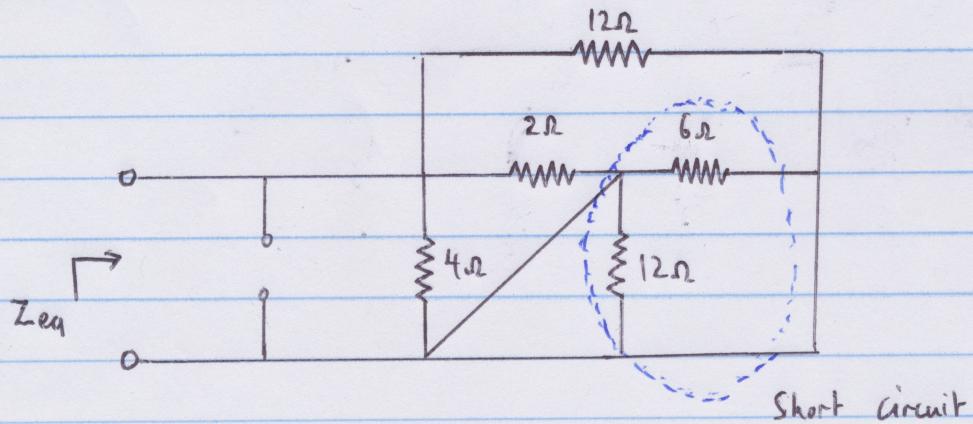
Question 1

i)

a) Under DC conditions

0.5F capacitor \Rightarrow Open circuit

4H inductor \Rightarrow Short circuit



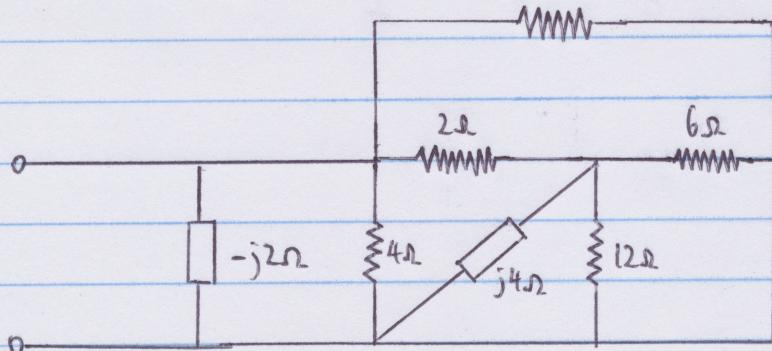
$$Z_{eq} = 12 // 2 // 4$$

$$= \left(\frac{1}{12} + \frac{1}{2} + \frac{1}{4} \right)^{-1} = 1.2 \Omega$$

b) $w = 1 \text{ rad/s}$

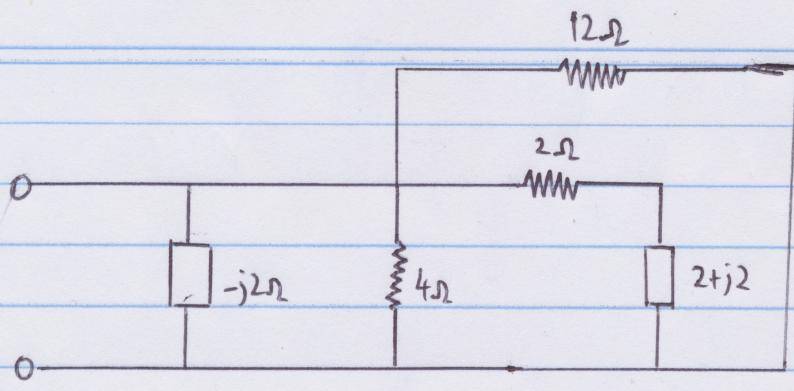
$$Z_{eq}(0.5\text{F}) = -\frac{j}{wC} = -\frac{j}{0.5} = -j2\Omega$$

$$Z_{eq}(4\text{H}) = jwL = j4\Omega$$



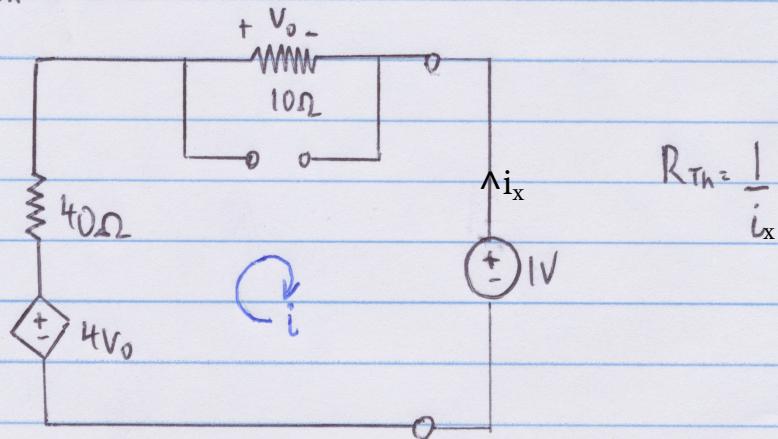
$$6 // 12 // j4 = \left(\frac{1}{6} + \frac{1}{12} + \frac{1}{j4} \right)^{-1}$$

$$= (2 + j2) \Omega$$



$$\begin{aligned}
 Z_{eq} &= -j2 // 4 // (4+j2) // 12 \\
 &= \left(\frac{1}{-j2} + \frac{1}{4} + \frac{1}{4+j2} + \frac{1}{12} \right)^{-1} = (1.2 - j0.9) \Omega
 \end{aligned}$$

ii) Finding R_{Th}



KVL in mesh:

$$-4V_o + 40i + V_o + 1 = 0$$

$$\text{But } V_o = 10i$$

$$40i - 3(10i) = -1$$

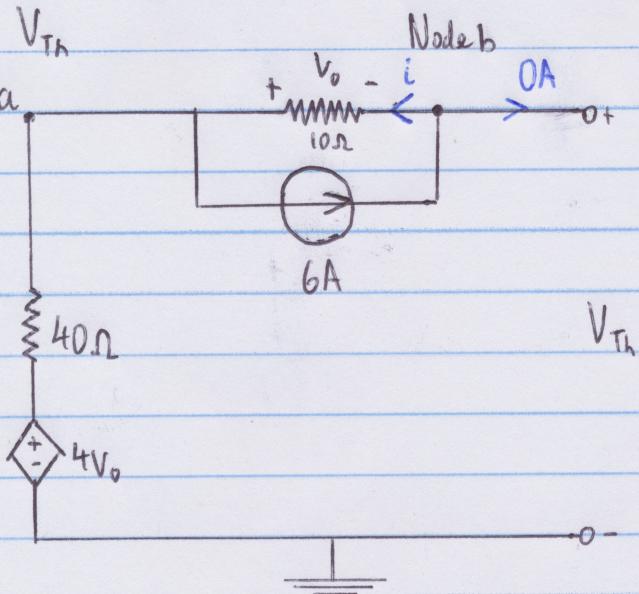
$$10i = -1$$

$$i = -0.1 \text{ A} \Rightarrow i_x = 0.1 \text{ A}$$

$$\therefore R_{Th} = 10\Omega$$

Finding V_{Th}

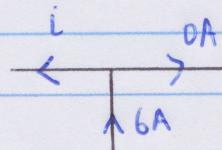
Node a



KCL at node b:

$$i + 0 = 6$$

$$\therefore i = 6A$$



$$\text{Also: } 6 = \frac{V_b - V_a}{10} \Rightarrow V_a - V_b = -60$$

KCL at node a:

$$\frac{V_a - (4 + 60)}{40} + 6 + \frac{V_a - V_b}{10} = 0$$

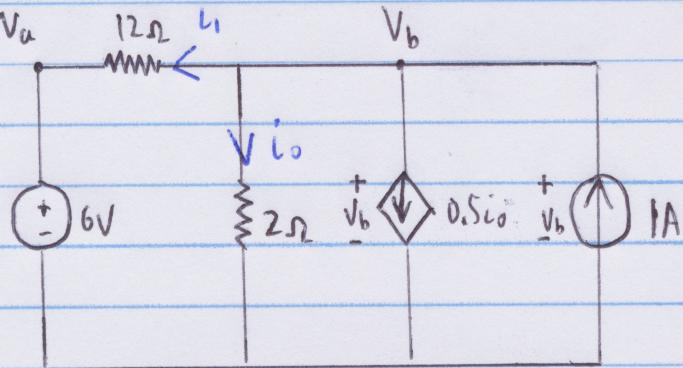
$$V_a + 240 + 240 + 4V_a - 4V_b = 0$$

$$5V_a - 4V_b = -480$$

$$\left(\begin{array}{cc|c} V_a & V_b & \\ 1 & -1 & -60 \\ 5 & -4 & -480 \end{array} \right) \quad R_2 = R_2 - 5R_1$$

$$\left(\begin{array}{cc|c} 1 & -1 & -60 \\ 0 & 1 & -180 \end{array} \right) \quad \therefore V_b = -180V$$
$$\therefore V_{Th} = V_b - 0 = -180V$$

iii) V_a 12Ω i_1



KCL at node b:

$$\frac{V_b - 6}{12} + \frac{V_b}{2} + \frac{1}{2} \left(\frac{V_b}{2} \right) = 1$$

$$V_b - 6 + 6V_b + 3V_b = 12$$

$$10V_b = 18$$

$$V_b = 1.8V$$

Finding i_1 :

$$i_1 = \frac{V_b - V_a}{12} = \frac{1.8 - 6}{12} = -0.35A$$

$$P_{6V} = -0.35 \times 6 = 2.1W \text{ supplied}$$

$$P_{1A} = 1.8 \times -1 = 1.8W \text{ supplied}$$

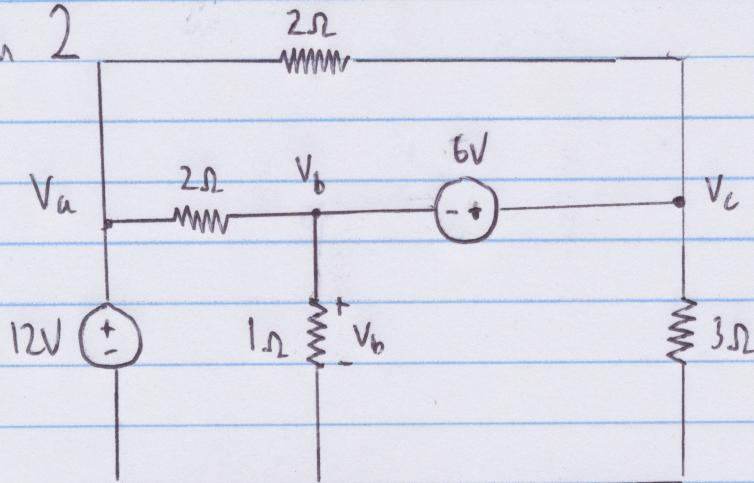
$$P_{0.5i_0} = \frac{1}{2} \left(\frac{1.8}{2} \right) \times 1.8 = 0.81W \text{ absorbed}$$

Explanation:

By sign convention, positive power absorbs and negative power supplies.

Question 2

i)



$V_a = 12V$ due to source

Supernode:

$$V_c - V_b = 6V \quad (1)$$

KCL at supernode:

$$\frac{V_b - 12}{2} + \frac{V_b}{1} + \frac{V_c - 12}{2} + \frac{V_c}{3} = 0$$

$$3V_b - 36 + 6V_b + 3V_c - 36 + 2V_c = 0$$

$$9V_b + 5V_c = 72 \quad (2)$$

$$\left(\begin{array}{cc|c} V_b & V_c \\ -1 & 1 & 6 \\ 9 & 5 & 72 \end{array} \right) \quad R_2 = R_2 + 9R_1$$

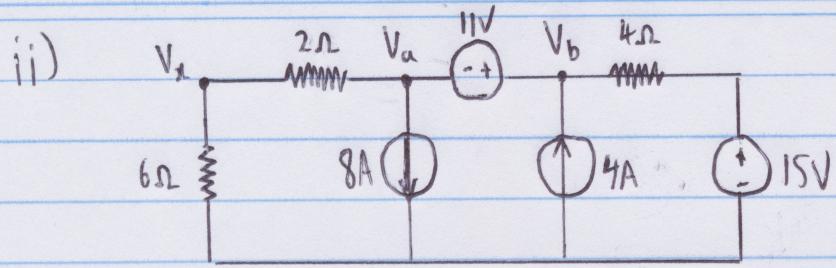
$$\left(\begin{array}{cc|c} -1 & 1 & 6 \\ 0 & 14 & 126 \end{array} \right)$$

$$\therefore V_c = \frac{126}{14} = 9V$$

$$V_b = -(6 - 9) = 3V$$

$$\therefore V_{1\Omega} = 3V$$

$$P_{3\Omega} = \frac{V_c^2}{R} = \frac{9^2}{3} = 27W \text{ absorbed}$$



KCL at node x :

$$\frac{V_x}{6} + \frac{V_x - V_a}{2} = 0$$

$$4V_x - 3V_a = 0 \quad \textcircled{1}$$

Supernode between nodes a and b :

$$V_b - V_a = 11V \Rightarrow V_a - V_b = -11 \quad \textcircled{2}$$

KCL at supernode:

$$\frac{V_a - V_n}{2} + 8 + \frac{V_b - 15}{4} = 4$$

$$\begin{aligned} 2V_a - 2V_n + 32 + V_b - 15 &= 16 \\ -2V_n + 2V_a + V_b &= -1 \quad \textcircled{3} \end{aligned}$$

$$\left(\begin{array}{ccc|c} V_n & V_a & V_b & \\ \hline 4 & -3 & 0 & 0 \\ 0 & 1 & -1 & -11 \\ -2 & 2 & 1 & -1 \end{array} \right) \quad R_3 = 2R_3 + R_1$$

$$\left(\begin{array}{ccc|c} 4 & -3 & 0 & 0 \\ 0 & 1 & -1 & -11 \\ 0 & 1 & 2 & -2 \end{array} \right) \quad R_3 = R_3 - R_2$$

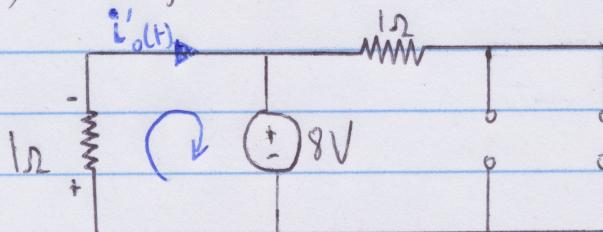
$$\left(\begin{array}{ccc|c} 4 & -3 & 0 & 0 \\ 0 & 1 & -1 & -11 \\ 0 & 0 & 3 & 9 \end{array} \right) \quad \therefore V_b = 3V$$

$$V_a = -8V$$

$$\underline{V_n = -6V}$$

iii) Must use superposition

Only turning on 8V source on



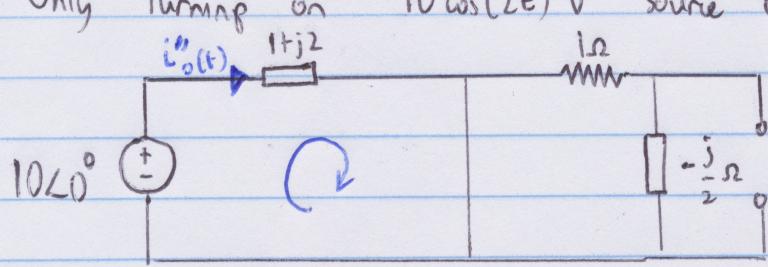
Inductor \Rightarrow Closed circuit
Capacitor \Rightarrow Open circuit

KVL:

$$i_o' + 8 = 0$$

$$i_o' = -8 \text{ A}$$

Only turning on $10\cos(2t)$ V source on



$$\omega = 2 \text{ rad/s}$$

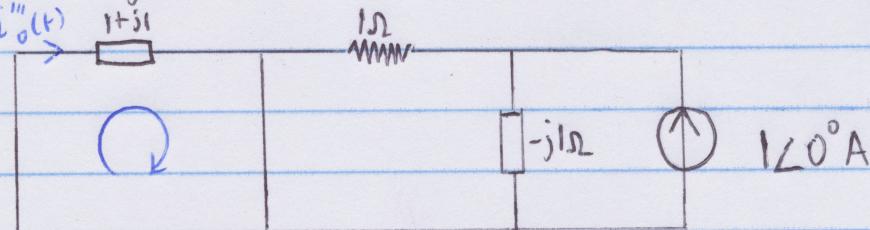
$$\text{Inductor} \Rightarrow j\omega L = j2 \Omega$$

$$\text{Capacitor} \Rightarrow -\frac{j}{\omega C} = -\frac{j}{2} \Omega$$

$$\text{KVL: } (1+j2) i_o'' - 10 = 0$$

$$i_o'' = \frac{10}{1+j2} = 4.472 \cos(2t - 63.4^\circ) \text{ A}$$

Only turning on $1\cos(t)$ A source on



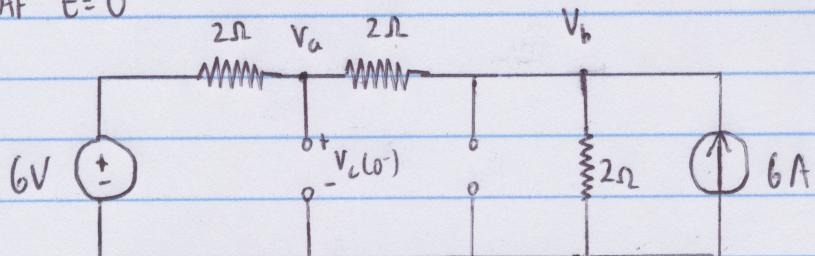
$$i_o'''(t) = 0 \text{ A due to short circuit}$$

\therefore By superposition:

$$i_o(t) = [-8 + 4.472 \cos(2t - 63.4^\circ)] \text{ A}$$

Question 3

i) At $t=0^-$



KCL at node a:

$$\frac{V_a - 6}{2} + \frac{V_a - V_b}{2} = 0$$

$$2V_a - V_b = 6 \quad (1)$$

KCL at node b:

$$\frac{V_b}{2} + \frac{V_b - V_a}{2} = 6$$

$$-V_a + 2V_b = 12 \quad (2)$$

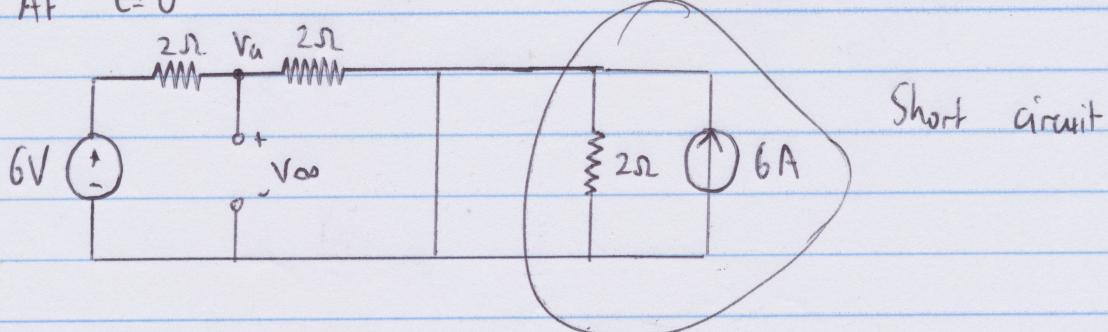
$$(2) \times 2 \Rightarrow -2V_a + 4V_b = 24 \quad (3)$$

$$(1) + (3) \Rightarrow 3V_b = 30$$

$$V_b = 10V$$

$$\therefore V_a = 8V = V_c(0^-)$$

At $t=0^+$



KCL at node a:

$$\frac{V_a - 6}{2} + \frac{V_a}{2} = 0$$

$$\therefore V_a = 3V$$

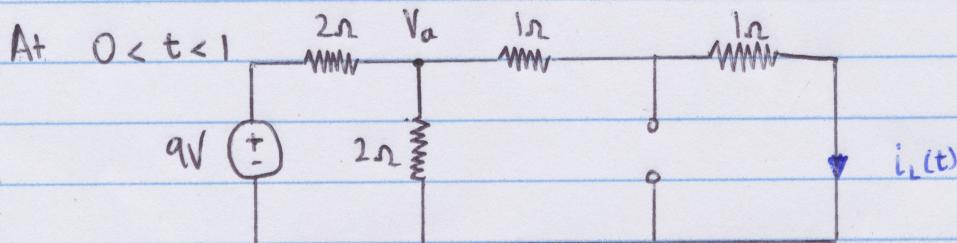
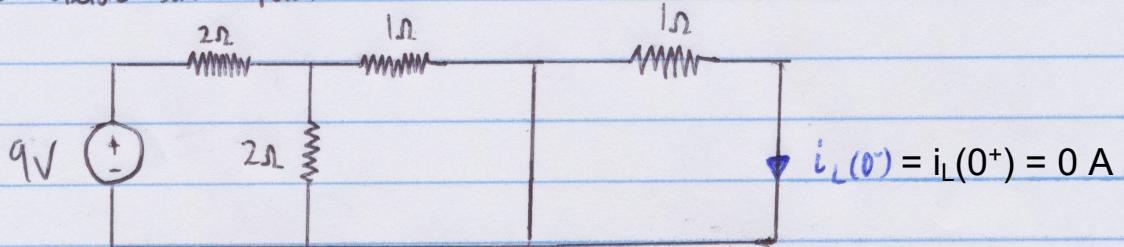
$$R_{Th} = 2/1/2 \Rightarrow T = RL$$

$$= 1\Omega \quad = 10 \text{ mS} = 10 \times 10^{-3}$$

$$V_L(t) = V_\infty + (V_L(0^-) - V_\infty) e^{-\frac{t}{T}} \text{ V}$$

$$= 3 + 5e^{-100t} \text{ V} \quad \text{for } t > 0$$

ii) At $t = 0^-$ (before switch opens)



KCL at node a:

$$\frac{V_a - 9}{2} + \frac{V_a}{2} + \frac{V_a}{2} = 0$$

$$3V_a = 9$$

$$V_a = 3V$$

$$\therefore i_L(t) = \frac{3}{2} A = i_L(\text{infinity})$$

Assuming that the circuit remains in this condition for a long time

$$R_{Th} = 2 + 2/1/2 = 3\Omega$$

$$T = \frac{L}{R_{Th}} = \frac{4}{3} \text{ s}$$

$$i_L(t) = 1.5 - 1.5e^{-\frac{3}{4}t} \text{ A} \quad \text{for } 0 < t < 1$$

At $t > 1$ (Circuit is the same as first one in question)

$$i_{\infty} = 0 \text{ A} \quad i(1) = 1.5(1 - e^{-0.75}) = 0.79 \text{ A}$$

$$R_{Th} = 1 \Omega$$

$$T = \frac{L}{R} = 4 \text{ s}$$

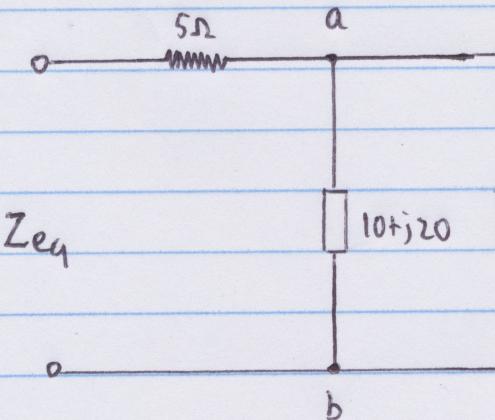
$i(1)$ is used as initial value in the equation instead of $i(0^+)$ since the switch changes at $t = 1$ s.

$$\therefore i_L(t) =$$

$$0.79 e^{-\frac{(t-1)}{4}} \text{ A} \quad \text{for } t > 1$$

$$i_L(t) = \begin{cases} 1.5(1 - e^{-\frac{3}{4}t}) \text{ A} & \text{for } 0 < t < 1 \\ 0.79 e^{-\frac{(t-1)}{4}} \text{ A} & \text{for } t > 1 \end{cases}$$

iii)



$$Z_1 = a + jb$$

$$Z_{eq}(2H) = j(2)(10) = j20 \Omega$$

$$Z_{eq} = 5 + Z_1 / (10 + j20)$$

For Z_{eq} to be purely resistive

$$\operatorname{Im}(Z_1 / (10 + j20)) = 0$$

$$Z_1 / (10 + j20) = \frac{(a + jb)(10 + j20)}{(a + 10) + j(b + 20)} = \frac{(10a - 20b) + j(20a + 10b)}{(a + 10) + j(b + 20)} = \underline{r_1 \angle \theta_1} \quad \underline{r_2 \angle \theta_2}$$

If $\theta_1 = \theta_2$ the circuit will be purely resistive:

$$\frac{20a + 10b}{10a - 20b} = \frac{b + 20}{a + 10}$$

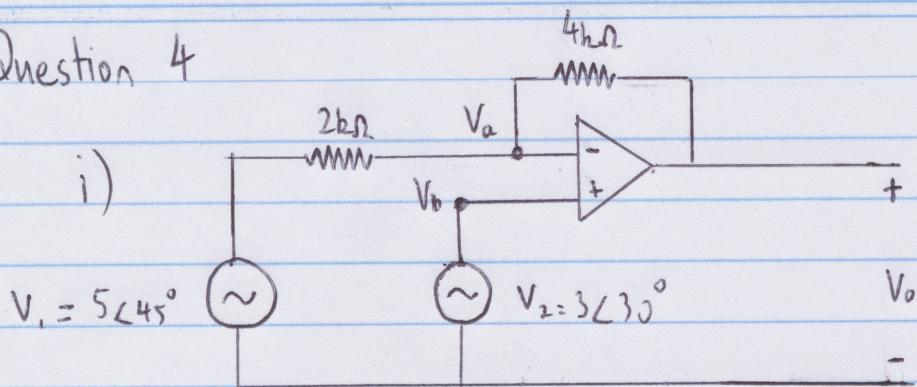
$$20a^2 + 20b^2 + 50ab = 0 \Rightarrow a^2 + b^2 + 25b = 0 \Rightarrow a^2 + b(b + 25) = 0$$

$$b = -25 \quad a = 0$$

$$\therefore Z_1 = -j25 = -\frac{j}{10} = -\frac{j}{10} \quad \Rightarrow \quad C = \frac{1}{250} = 4 \text{ mF}$$

Question 4

i)



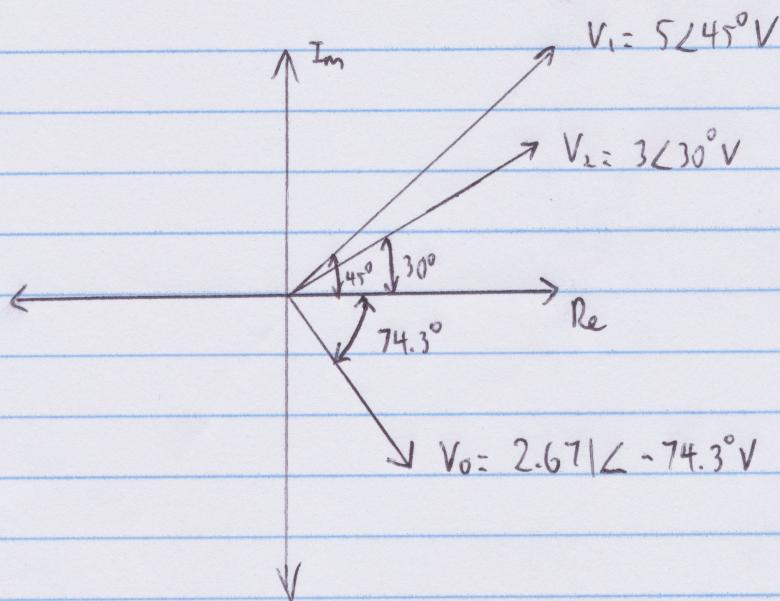
$$V_a = V_b = 3\angle 30^\circ$$

KCL at node a:

$$\frac{3\angle 30^\circ - 5\angle 45^\circ}{2,000} + \frac{3\angle 30^\circ - V_0}{4,000} = 0$$

$$\therefore V_0 = 4,000 \left(\frac{3\angle 30^\circ - 5\angle 45^\circ}{2,000} + \frac{3\angle 30^\circ}{4,000} \right)$$

$$= 2.671 \angle -74.3^\circ \text{ V}$$

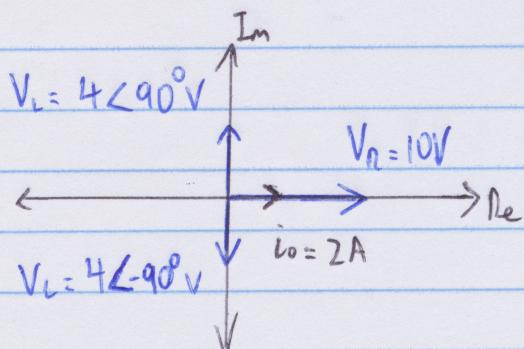


ii) $W = 1 \text{ rad/s}$

$$Z_{eq}(2H) = j2\Omega$$

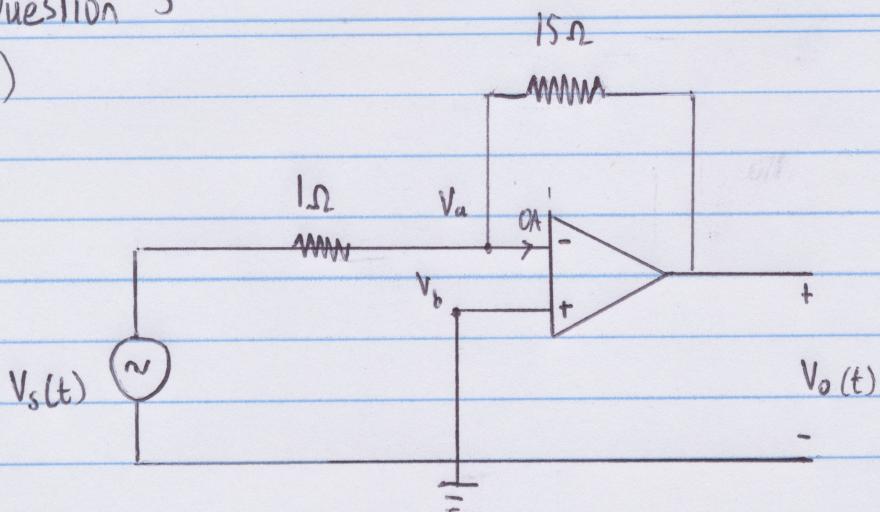
$$Z_{eq}(0.5F) = -j2\Omega$$

$$i_o = \frac{10\angle 0^\circ}{j2} = 2A$$



Question 5

i)



$$V_b = 0V$$

$V_b = V_a$ (Principle of Op amp)

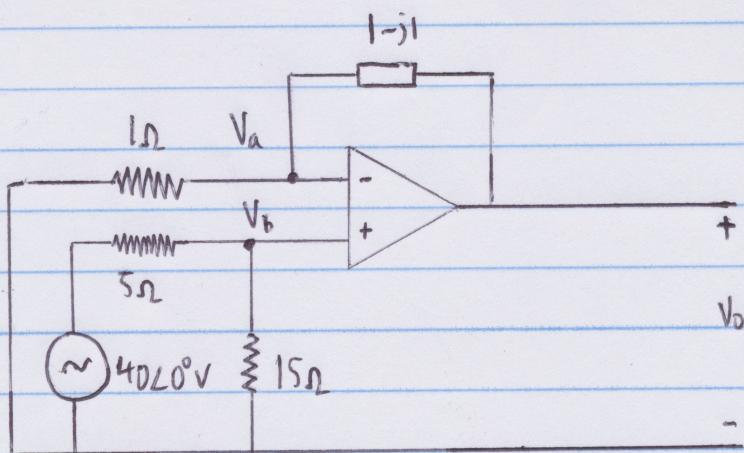
KCL at node a

$$\frac{0 - V_s(t)}{1} + \frac{0 - V_{o(t)}}{15} = 0$$

$$-V_s(t) = \frac{V_o(t)}{15}$$

$$\therefore V_o(t) = -15V_s(t)$$

ii)



KCL at node b:

$$\frac{V_b - 40\angle 0^\circ}{5} + \frac{V_b}{15} = 0$$

$$3V_b - 120 + V_b = 0$$

$$V_b = 30 \text{ V} = V_a \text{ (Op amp principle)}$$

KCL at node a:

$$\frac{V_a}{1} + \frac{V_a - V_b}{1-j} = 0$$

$$40 + \frac{40}{1-j} = \frac{V_b}{1-j}$$

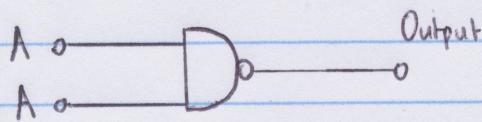
$$\begin{aligned} \therefore V_b &= (1-j) \left(40 + \frac{40}{1-j} \right) \\ &= 80 - j40 = 89.44 \angle -26.57^\circ \text{ V} \end{aligned}$$

$$\begin{aligned} \text{i)} \quad \text{a) } Z &= \overline{\overline{A+B}} + \overline{\overline{B+C}} \\ &= \overline{\overline{A+B}} \cdot \overline{\overline{B+C}} = (A+\bar{B}) \cdot (\bar{B}+C) \\ &= A \cdot B + A \cdot C + \cancel{\bar{B} \cdot B} + \cancel{\bar{B} \cdot C} \\ &= A \cdot B + A \cdot C + \bar{B} \cdot C \end{aligned}$$

b)	A	B	C	A.B	A.C	$\bar{B}.C$	Z
	0	0	0	0	0	0	0
	0	0	1	0	0	1	1
	0	1	0	0	0	0	0
	0	1	1	0	0	0	0
	1	0	0	0	0	0	0
	1	0	1	0	1	1	1
	1	1	0	1	0	0	1
	1	1	1	1	1	0	1

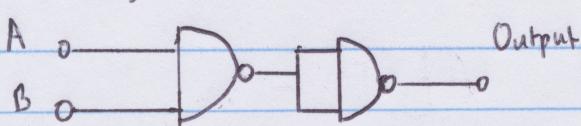
c)

i) NOT gate



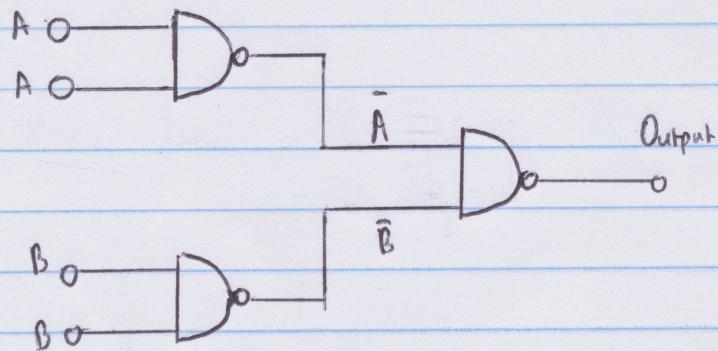
$$\text{Output} = \overline{A} \cdot A = \overline{A} + \overline{A} = \overline{A}$$

ii) AND gate



$$\text{Output} = \overline{\overline{A} \cdot B} = A \cdot B$$

iii) OR gate



$$\text{Output} = \overline{\overline{A} \cdot \overline{B}} = \overline{\overline{A}} + \overline{\overline{B}} = A + B$$