

Tutorial—Inverse Laplace Transform

Example 1. Find

$$\mathcal{L}^{-1} \left[\frac{s+4}{2s^2+5s+3} \right] \quad (1)$$

Hint: use partial fraction.

Example 2. Find

$$\mathcal{L}^{-1} \left[\frac{se^{-2s}}{s^2+2s+5} \right] \quad (2)$$

Hint: apply the Shifting Theorem to the part without the exponential term. The exponent corresponds is a time shift.

Example 3. Find

$$\mathcal{L}^{-1} \left[\frac{s+2}{s(s+1)^2(s+3)} \right] \quad (3)$$

Note: this equation has two simple roots and a double root (due to the $(s+1)^2$ term).

Example 4. Find

$$\mathcal{L}^{-1} \left[\frac{1}{3s^2(s^2+4)} \right] \quad (4)$$

Note: this function has one double root and two conjugate roots.

Example 5. A rectangular voltage pulse of unit height and duration T is applied to a series R - C combination at $t = 0$, see Fig. 1. Determine the voltage across the capacitor C as a function of time.

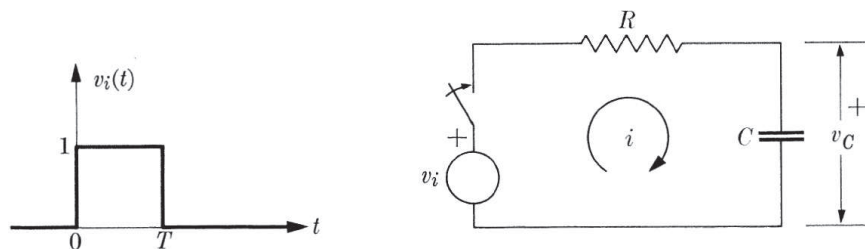


Figure 1: A rectangular pulse applied to an R - C circuit

Solution 1. First, we factor the denominator

$$B(s) = 2s^2 + 5s + 3 = 2 \left(s^2 + \frac{5}{2}s + \frac{3}{2} \right) = 2(s+1) \left(s + \frac{3}{2} \right) \quad (5)$$

the distinct roots are $s = -1$ and $s = -3/2$. Next, we expand the given function of s into partial fractions and determine the constants.

$$\frac{s+4}{2s^2+5s+3} = \frac{1}{2} \left[\frac{K_1}{s+1} + \frac{K_2}{s+3/2} \right] \quad (6)$$

then

$$K_1 = 2 \left[(s+1) \frac{s+4}{2(s+1)(s+3/2)} \right]_{s=-1} = \left[\frac{s+4}{s+3/2} \right]_{s=-1} = 6 \quad (7)$$

$$K_2 = 2 \left[(s+3/2) \frac{s+4}{2(s+1)(s+3/2)} \right]_{s=-3/2} = \left[\frac{s+4}{s+1} \right]_{s=-3/2} = -5 \quad (8)$$

hence

$$\frac{s+4}{2s^2+5s+3} = \frac{1}{2} \left[\frac{6}{s+1} - \frac{5}{s+3/2} \right] \quad (9)$$

Finally, we find the inverse transform

$$\mathcal{L}^{-1} \left[\frac{s+4}{2s^2+5s+3} \right] = \frac{1}{2} \mathcal{L}^{-1} \left[\frac{6}{s+1} - \frac{5}{s+3/2} \right] = \frac{1}{2} (6e^{-t} - 5e^{-3t/2}) \quad (10)$$

□

Solution 2. We know that multiplication by a factor e^{-2s} simply amounts to a shift in the independent variable from t to $t-2$. We can therefore first perform inverse Laplace transform without the shift and then re-apply it to the result.

First, we factor the denominator

$$B(s) = s^2 + 2s + 5, \Rightarrow s = \frac{1}{2} (-2 \pm \sqrt{4-20}) = -1 \pm j2 \quad (11)$$

The two distinct roots are $s_1 = -1 + j2$ and $s_2 = -1 - j2$. Next, we expand into partial fractions. Because of the conjugate roots, we have

$$\frac{d}{ds} B(s) = 2s + 2 = 2(s+1) \quad (12)$$

then

$$K_1 = \frac{s}{2(s+1)} \Big|_{s=s_1=-1+j2} = \frac{-1+j2}{j4} = \frac{1}{4}(2+j) \quad (13)$$

$$K_2 = \frac{s}{2(s+1)} \Big|_{s=s_2=-1-j2} = \frac{-1-j2}{-j4} = \frac{1}{4}(2-j) \quad (14)$$

so we have

$$\frac{s}{s^2 + 2s + 5} = \frac{1}{4} \left[\frac{2+j}{s+1-j2} + \frac{2-j}{s+1+j2} \right] \quad (15)$$

Now we find

$$\begin{aligned} \mathfrak{L}^{-1} \left[\frac{s}{s^2 + 2s + 5} \right] &= \frac{1}{4} \mathfrak{L}^{-1} \left[\frac{2+j}{s+1-j2} + \frac{2-j}{s+1+j2} \right] \\ &= \frac{1}{4} \left[(2+j)e^{(-1+j2)t} + (2-j)e^{(-1-j2)t} \right] \\ &= \frac{1}{2} e^{-t} (2 \cos 2t - \sin 2t) \end{aligned} \quad (16)$$

Finally, apply time shifting to obtain

$$\mathfrak{L}^{-1} \left[\frac{se^{-2s}}{s^2 + 2s + 5} \right] = \frac{1}{2} e^{-(t-2)} [2 \cos 2(t-2) - \sin 2(t-2)] U(t-2) \quad (17)$$

□

Solution 3. The denominator has four roots: simple roots $s = 0$, $s = -3$, double root $s = -1$. Using partial fraction, we have

$$\frac{A(s)}{B(s)} = \frac{s+2}{s(s+1)^2(s+3)} = \frac{K_1}{s} + \left(\frac{K_{22}}{(s+1)^2} + \frac{K_{21}}{s+1} \right) + \frac{K_3}{s+3} \quad (18)$$

Then

$$K_1 = s \frac{A(s)}{B(s)} \Big|_{s=0} = \frac{s+2}{(s+1)^2(s+3)} \Big|_{s=0} = \frac{2}{3} \quad (19)$$

$$K_{22} = (s+1)^2 \frac{A(s)}{B(s)} \Big|_{s=-1} = \frac{s+2}{s(s+3)} \Big|_{s=-1} = -\frac{1}{2} \quad (20)$$

$$K_{21} = \frac{d}{ds} \left(\frac{s+2}{s(s+3)} \right) \Big|_{s=-1} = \frac{s(s+3) - (s+2)(2s+3)}{s^2(s+3)^2} \Big|_{s=-1} = -\frac{3}{4} \quad (21)$$

$$K_3 = (s+3) \frac{A(s)}{B(s)} \Big|_{s=-3} = \frac{s+2}{s(s+1)^2} \Big|_{s=-3} = \frac{1}{12} \quad (22)$$

Hence

$$\frac{s+2}{s(s+1)^2(s+3)} = \frac{2}{3s} - \frac{1}{2(s+1)^2} - \frac{3}{4(s+1)} + \frac{1}{12(s+3)} \quad (23)$$

and

$$\mathfrak{L}^{-1} \left[\frac{s+2}{s(s+1)^2(s+3)} \right] = \frac{2}{3} - \frac{1}{2} \left(t + \frac{3}{2} \right) e^{-t} + \frac{1}{12} e^{-3t} \quad (24)$$

□

Solution 4. The denominator has four roots: one double root $s = 0$, two conjugate roots $s = j2$ and $s = -j2$. Furthermore, we keep the conjugate roots together as a unit. We write the partial fraction expansion as follows.

$$\frac{1}{3s^2(s^2 + 4)} = \frac{1}{3} \left[\frac{K_{12}}{s^2} + \frac{K_{11}}{s} + \frac{C_1s + C_2}{s^2 + 4} \right] \quad (25)$$

Note that from

$$\frac{K_2}{s - j2} + \frac{K_3}{s + j2} = \frac{(K_2 + K_3)s + j2(K_2 - K_3)}{s^2 + 4} \quad (26)$$

we can put $C_1 = K_2 + K_3$ and $C_2 = j2(K_2 - K_3)$. Then re-write Eq. 25 as

$$\begin{aligned} 1 &= K_{12}(s^2 + 4) + K_{11}s(s^2 + 4) + (C_1s + C_2)s^2 \\ &= (K_{11} + C_1)s^3 + (K_{12} + C_2)s^2 + 4K_{11}s + 4K_{12} \end{aligned} \quad (27)$$

Thus $K_{11} + C_1 = 0$, $K_{12} + C_2 = 0$, $4K_{11} = 0$, $4K_{12} = 1$, and $K_{12} = 1/4$, $K_{11} = 0$, $C_1 = 0$, $C_2 = -1/4$. Hence, from Eq. 25

$$\frac{1}{3s^2(s^2 + 4)} = \frac{1}{3} \left[\frac{1}{4s^2} - \frac{1}{4(s^2 + 4)} \right] \quad (28)$$

and then

$$\mathcal{L}^{-1} \left[\frac{1}{3s^2(s^2 + 4)} \right] = \frac{1}{12} \left(t - \frac{1}{2} \sin 2t \right) \quad (29)$$

□

Solution 5. First, let us write the expression for the input rectangular voltage pulse

$$v_i(t) = U(t) - U(t - T) \quad (30)$$

Next, we write the differential equation of this simple one-loop circuit using Kirchoff's voltage law. Since the desired response is $v_c = q/C$ and $i = dq/dt = C(dv_c/dt)$, we write the equation with the dependent variable

$$RC \frac{dv_c}{dt} + v_c = v_i(t) = U(t) - U(t - T) \quad (31)$$

We now apply Laplace transform $\mathcal{L}[v_c(t)] = V_c(s)$, then

$$RC [sV_c(s) - v_c(0^+)] + V_c(s) = \frac{1}{s} (1 - e^{-sT}) \quad (32)$$

With an initially uncharged C , $v_c(0^+) = 0$, hence

$$(RCs + 1)V_c(s) = \frac{1}{s} (1 - e^{-sT}), \Rightarrow V_c(s) = \frac{1 - e^{-sT}}{s(RCs + 1)} \quad (33)$$

Now we find the inverse transform of $V_c(s)$. Note that the factor e^{-sT} amounts to a simple shift in time T in the inverse transformation, we have

$$\frac{1}{s(RCs + 1)} = \frac{1}{RCs(s + 1/RC)} = \frac{1}{RC} \left[\frac{K_1}{s} + \frac{K_2}{s + 1/RC} \right] \quad (34)$$

where

$$K_1 = \left[\frac{1}{s + 1/RC} \right]_{s=0} = RC, \quad K_2 = \left[\frac{1}{s} \right]_{s=-1/RC} = -RC \quad (35)$$

Hence

$$\mathfrak{L} \left[\frac{1}{s} - \frac{1}{s + 1/RC} \right] = 1 - e^{-t/RC} \quad (36)$$

Finally, by considering the rectangular pulse (ceased after a time shift), then

$$\begin{aligned} v_c(t) &= \mathfrak{L}^{-1} \left[\frac{1 - e^{-sT}}{s(RCs + 1)} \right] \\ &= \left(1 - e^{-t/RC} \right) U(t) - \left(1 - e^{-(t-T)/RC} \right) U(t - T) \end{aligned} \quad (37)$$

A plot of $v_c(t)$ is shown in Fig. 2, the exact shape of $v_c(t)$ depends upon the relative values of T , R and C . \square

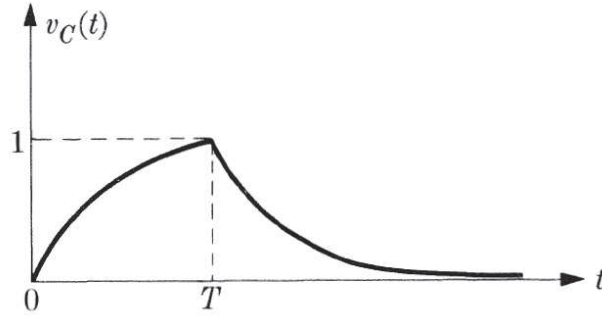


Figure 2: Output voltage of the R - C circuit