

ELG713 Machine Learning for Adaptive and Intelligent Control Systems

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assignment GitHub: github.com/derekboase/Adaptive_Control_Code.git

```
In [1]: import matplotlib.pyplot as plt
import control as co
import numpy as np
import sympy as sp

from numpy import transpose as trans
from numpy import cos, sin, pi
from numpy.linalg import inv

Question 1
```

Question declarations:

```
In [2]: r1, s0, s1, a01 = sp.symbols('r_1, s_0, s_1, a_0_1')
a0, a1, a2, b0 = sp.symbols('a_0, \hat{a}_1, \hat{a}_2, \hat{b}_1, \hat{b}_2, \hat{b}_0')
# b, c, d, y, Ts = sp.symbols('b, c, d, y, Ts')
yddot, ydot = sp.symbols('yddot(y), ydot(y)')
y, u, uc = sp.symbols('y, u, u_c')
a_true, b_true, c_true, d_true = (1, 1, 2, 1/2)
G1 = (b/(s + a)).subs(a, 1).subs(b, 1)
G2 = c/(s + d)
G_sym = b0/(s**2 + a1*s + a2)
G = sp.Mul(G1, G2)
Bm = 1
Bm = ** 2 + 2*s + 1
G = sp.collect(sp.expand(G), s)
B_plus = 1
B_minus = c
```

Part 1

Using the compatibility conditions:

$$\begin{aligned} \deg A_m &= \deg A = 2 \\ \deg B_m &= \deg B = 0 \\ \deg A_0 &= \deg A - \deg B^+ - 1 = 1 \end{aligned}$$

The causality condition gives,

$$\deg A_m - \deg B_m^+ \geq \deg A - \deg B^+$$

Given that

$$\deg A_m = \deg A = 2$$

and

$$\deg B_m^+ = 0$$

then

$$\deg B^+ = 0$$

With this we can find,

$$\deg A_c = \deg A_c + \deg A_m + \deg B^+ = 3$$

It is then found that,

$$\deg R = \deg A_c - \deg A = 3 - 2 = 1$$

Using the minimum phase relationship,

$$\deg R = \deg T - \deg S = 1$$

The control parameters may then be found using the Diophantine equation,

$$AR' + B - S = A_0A_m$$
$$(s^2 + (a + d)s + ad)(s + r_1) + bc(s_0s + s_1) = (s + a_0)(s^2 + 2s + 1)$$

```
In [3]: R = s**2 + s
R_prime = R
S = s0*s + s1
A0 = (s + a0)
T = A0*Bm/B_minus

A_m = s**2 + 2s + 1
B_m = 1
R = r1 + s
S = s0*s + s1
T = (a0 + s)/c
```

```
In [4]: dio_RHS = sp.collect(sp.expand(A*R_prime + B_minus*S), s)
dio_RHS = sp.collect(sp.expand(A0*Bm), s)
dio_poly = sp.Poly(sp.collect(sp.expand(dio_RHS - dio_RHS), s), s)

var = [r1, s0, s1]
for idx in range(len(dio_poly.coeffs())):
    control_vars.append(sp.solve(dio_poly.coeffs()[idx], var[idx])[0])
r10_sym = control_vars[0]
s0c_sym = control_vars[1]
s1c_sym = control_vars[2]
```

The above equation is described in the estimators by,

$$r_1 = a_0 - d + 1$$

$$s_0 = \frac{2a_0 - dr_1 - d - r_1 + 1}{c}$$

$$s_1 = \frac{a_0 - dr_1}{c}$$

Part 2

The process is described by,

$$p^2y(t) + (d + 1)py(t) + dy(t) = cu(t)$$

Then we can define,

$$y_f(s) = H_f(s)Y(s)$$

and

$$u_f(s) = H_f(s)U(s)$$

where

$$H_f(s) = \frac{1}{Am(s)} \frac{1}{s^2 + 2s + 1}$$

The model then becomes,

$$\frac{p^2}{\lambda_m(s)}y(t) = -(d + 1)\frac{p}{\lambda_m(s)}y(t) - \frac{d}{\lambda_m(s)}y(t) + \frac{c}{\lambda_m(s)}u_f(t)$$

The RLS can be used on this model by defining,

$$y_n(t) = \frac{p^2}{\lambda_m(s)}y(t)$$

$$u_n(t) = \frac{p}{\lambda_m(s)}u(t)$$

With this choice of definition, we can reparameterize the model as,

$$y_n(t) = \phi(t)^T \theta$$

where,

$$\phi(t) = [-y_n(t) - y_n(t)u_n(t)]^T$$

$$\theta = [(d + 1)d \ c]^T$$

To begin to evaluate these parameters, the continuous time variables are estimated with the bilinear transformation. That is to say the following mapping is used,

$$s = \frac{z(1-q^{-1})}{T_s(1+q^{-1})}$$

where T_s is the sampling time

```
In [5]: bilinear = 2*(1-q**(-1))/(Ts*(1+q**(-1)))

Rf1 = (s/Am).subs(s, bilinear)
Rf1 = sp.collect(sp.simplify(sp.expand(Rf1)), q)

Rf0 = (1/Am).subs(s, bilinear)
Rf0 = sp.collect(sp.simplify(sp.expand(Rf0)), q)

bilinear = 2*(1-q**(-1))/(Ts*(1+q**(-1)))

bfl, afl = sp.fraction(Rf1)
afl_poly = sp.Poly(afl, q)
bfl_poly = sp.Poly(bfl, q)
afl_coeffs = []
bfl_coeffs = []
for n in range(len(afl_poly.coeffs())):
    afl_coeffs.append(sp.simplify(afl_poly.coeffs()[n]/afl_poly.coeffs()[0]))
for n in range(len(bfl_poly.coeffs())):
    bfl_coeffs.append(sp.simplify(bfl_poly.coeffs()[n]/afl_poly.coeffs()[0]))
bfl_coeffs.insert(1, 0)

Rf1 = (s/Am).subs(s, bilinear)
Rf1 = sp.collect(sp.simplify(sp.expand(Rf1)), q)

Rf0 = (1/Am).subs(s, bilinear)
Rf0 = sp.collect(sp.simplify(sp.expand(Rf0)), q)

bfo, afo = sp.fraction(Rf0)
afo_poly = sp.Poly(afo, q)
bfo_poly = sp.Poly(bfo, q)
afo_coeffs = []
bfo_coeffs = []
for n in range(len(afo_poly.coeffs())):
    afo_coeffs.append(sp.simplify(afo_poly.coeffs()[n]/afo_poly.coeffs()[0]))
for n in range(len(bfo_poly.coeffs())):
    bfo_coeffs.append(sp.simplify(bfo_poly.coeffs()[n]/afo_poly.coeffs()[0]))

The coefficients for the denominator of the sampled filter H_f1 are given as,
```

$$\text{den}[H_{f1}] = \left[\frac{2(T-2)}{T+2}, \frac{T^2-4T+4}{T^2+4T+4} \right]$$

The coefficients for the numerator of the sampled filter H_{f1} are given as,

$$\text{num}[H_{f1}] = \left[\frac{2T}{T^2+4T+4}, 0, -\frac{2T}{T^2+4T+4} \right]$$

The coefficients for the denominator of the sampled filtered H_{f0} are given as,

$$\text{den}[H_{f0}] = \left[\frac{2(T-2)}{T+2}, \frac{T^2-4T+4}{T^2+4T+4} \right]$$

The coefficients for the denominator of the sampled filtered H_{f0} are given as,

$$\text{num}[H_{f0}] = \left[\frac{T^2}{T^2+4T+4}, \frac{2T^2}{T^2+4T+4}, \frac{T^2}{T^2+4T+4} \right]$$

Attention is taken to two results:

1. The denominators are both the same, as they come from the reference model parameters
2. The order of the coefficients goes from $\theta_2 \rightarrow \theta_1$

The following difference equations are then found for y_n and u_n .

$$y_n(k) = \frac{2T}{T^2-4T+4}y(k) - \frac{2T}{T^2-4T+4}y(k-2) - \frac{2T-2}{T^2-4T+4}y(k-1) - \frac{T^2-4T+4}{T^2-4T+4}y_n(k)$$

$$u_n(k) = \frac{T^2-4T+4}{T^2-4T+4}y(k) + \frac{2T-2}{T^2-4T+4}y(k-1) + \frac{T^2}{T^2-4T+4}y(k-2) - \frac{2T-2}{T^2-4T+4}y_n(k-1) - \frac{T^2-4T+4}{T^2-4T+4}u_n(k-2)$$

$$u_n(k) = \frac{T^2-4T+4}{T^2-4T+4}u(k) + \frac{2T-2}{T^2-4T+4}u(k-1) + \frac{T^2}{T^2-4T+4}u(k-2) - \frac{2T-2}{T^2-4T+4}u_n(k-1) - \frac{T^2-4T+4}{T^2-4T+4}u_n(k-2)$$

Furthermore, rearranging the equation for $y_n(k)$ we find,

$$y_n(k) = \frac{T^2+4T+4}{4}y_n(k) + \frac{T^2-4}{2}y_n(k-1) + \frac{T^2-4T-4}{4}y_n(k-2) + 2y(k-1) - y(k-2)$$

To find the control action, $u(k)$, we look to the equation,

$$u(t) = \frac{T}{c}u_n(t) - \frac{\hat{c}}{R}(t)$$

The approach to find the discretized equation is the same as in the previous question so the details are neglected for readability.

$$u(k) = -\frac{T^2-2}{T_sT^2}u(k-1) - \frac{dT^2+2b_0}{T_sT^2}y(k) - \frac{dT^2-2b_0}{T_sT^2}y(k-1) + \frac{a_0T^2}{c(T_sT^2)}u_n(k) + \frac{a_0-1}{c(T_sT^2)}u_n(k-1)$$

Part 3

```
In [6]: def reference_signal(time=np.linspace(0, 1000, 1001)):
    sig = []
    rat = 2.0/40.0*pi
    rat = 1.0/25.0*pi
    for t in range(1000):
        if sin(10*rat*t) >= 0:
            sig.append(1)
        else:
            sig.append(0)
    return np.array(sig, dtype=float)

In [7]: uc = reference_signal()
lam = 0.01
Ts = 0.001
end = 1000

# Assignment, works
am1 = -2*np.exp(-2)
am2 = np.exp(-4)
a1_true, d_true, c_true = 1.5, 0.5, 2

# In class example, works perfectly!
# am1 = -1.3205
# am2 = 0.4966
# a1_true, a2_true, b0_true, b1_true = -1.6065, 0.6065, 0.1065, 0.0902

theta = np.array([2*true, d_true, b1_true, c_true]).reshape(3,1) # a1[0], a2[1], b0[2], b1[3]

In [8]: theta_hat = np.array([0.1, 0.1, 0.1]).reshape(theta.shape) # z d c
theta_hat_list = np.concatenate((theta_hat,
                                (theta_hat).reshape(3,1)), axis=1)

y = np.zeros(2)
u = np.zeros(2)

y2 = np.zeros(2)
y1 = np.zeros(2)
y0 = np.zeros(2)
u0 = np.zeros(2)
P = 10*np.eye(3)

for t in range(2, end):
    phi = np.array([-y1[1]-1, -y0[1]-1, u0[1]-1]).reshape(3,1)
    theta_hat = (theta_hat + Ts*(P @ phi + trans(phi) @ theta - trans(phi) @ theta_hat))
    theta_hat_list = np.concatenate((theta_hat,
                                    (theta_hat).reshape(3,1),
                                    (theta_hat_list + Ts*(P @ phi + phitrans(phi) @ theta_hat,
                                    axis=1)

    P = inv(inv(P) + phitrans(phi))
    K = P@phi

    r1 = (2-theta_hat[1])
    s0 = (3 - theta_hat[1]) + r1 - theta_hat[1] - r1
    s1 = (1 - theta_hat[1]*r1)/theta_hat[2]

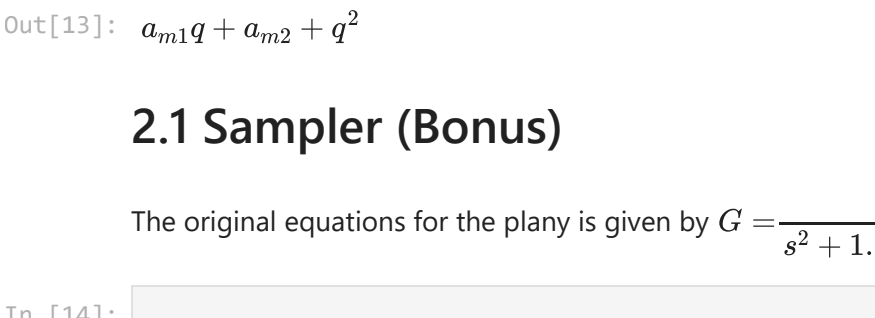
    y2 = np.concatenate((y2,
                        (trans(phi) @ theta).reshape(-1,)), axis=0)

    y = np.concatenate((y,
                        (0.025*y2[2] - 1.995*y2[2]-1 + 0.9025*y2[2]-2 + 2*y[1] - y[2]).reshape(-1,)), axis=0)

    u = np.concatenate((u,
                        (1/(Ts*r1 + 2)*(Ts*r1 - 2)*u[1]-1 - (Ts*a1 + 2*a0)*y[1] - (Ts*a1 - 2*a0)*y[1]-1 + (Ts*a1 - 2*a0)*y[1]-1)), axis=0)

    y1 = np.concatenate((y1,
                        (0.04535*y[1] - 0.0435*y[2]-1 + 1.8095*y[1]-1 + 0.8186*y[1]-2).reshape(-1,)), axis=0)
    y0 = np.concatenate((y0,
                        (0.00227*y[1] + 0.00454*y[1]-1 + 0.00227*y[2]-2 + 1.8095*y0[1]-1 - 0.8186*y0[2]-2)).reshape(-1,)), axis=0)
    u0 = np.concatenate((u0,
                        (0.00227*u[1] + 0.00454*u[1]-1 + 0.00227*u[2]-2 + 1.80952*u0[1]-1 - 0.8186*u0[2]-2)).reshape(-1,)), axis=0)

In [9]: a, b = theta_hat_list.shape
colours = ['blue', 'orange', 'green']
for idx, th in enumerate(theta_hat_list):
    plt.plot(th, label=colours[idx], c=colours[idx], '--')
plt.legend(['Parameters'])
plt.show()
```



Question 2

```
In [10]: import matplotlib.pyplot as plt
import control as co
import numpy as np
import sympy as sp

from numpy import transpose as trans
from numpy import cos, sin, pi
from numpy.linalg import inv

Question Parameters
```

```
In [11]: a1, a2, b0, b1 = sp.symbols('a1, a2, b0, b1')
r1, s0, s1, a0 = sp.symbols('r_1, s_0, s_1, a_0')
am1, am2 = sp.symbols('a_m1, a_m2')
y, u, uc = sp.symbols('y, u, u_c')
q = sp.symbols('q')
```

```
In [12]: # A = (q**2 - 0.503215*q + 0.0497871)
# B = (1.59831*q + 0.587984)
# A = (q**2 + a1*q + a2)
# B = (b0*q + b1)
B_poly = sp.Poly(B, q)

H = B/A
B_plus = (q + B_poly.coeffs()[1]/B_poly.coeffs()[0])
B_minus = B_poly.coeffs()[0]
```

$$\hat{b}_0q + \hat{b}_1$$

$$\hat{a}_1q + \hat{a}_2 + q^2$$

```
In [13]: # Am_test = co.sample_system(co.tf([1,1],[1,2,-1]), Ts=2, method='zoh')
# Am = co.tf([1,1],[1,2,-1], Ts=2, method='zoh')
# A = q**2 + a1*q + a2
# B = q**2 + b1*q + b2
```

$$a_{m1}q + a_{m2} + q^2$$

2.1 Sampler (Bonus)

The original equations for the plant is given by $G = \frac{2}{s^2 + 1.5s + 0.5}$

$$\text{co.sample_system}(co.tf([2,1],[1,1.5,0.5]), Ts=2, method='zoh')$$

$$\frac{1.598s + 0.588}{z^2 - 0.5032z + 0.04979} \quad dt = 2$$

2.2 Indirect Linear Controller Design Using MDPP

We begin by factorizing B as follows,

$$B = B(q)^+ B(q)^- = (q + 0.36788)(1.59831)$$

To cancel the zeros, we require that they be stable and well-damped. In this case the zero is $q = -0.36788$ and is thus stable (inside the unit circle) and well-damped (sufficiently far from the unit circle). It is then clear that we may cancel the zeros.

Using the compatibility conditions:

$$\begin{aligned} \deg A_m &= \deg A = 2 \\ \deg B_m &= \deg B = \deg B^+ = 1 \\ \deg A_0 &= \deg A - \deg B^+ - 1 = 0 \end{aligned}$$

With this we can find,

$$\deg A_c = \deg A_c + \deg A_m + \deg B^+ = 3$$

From the minimum phase conditions we then get that,

$$\deg R = \deg S = \deg T = \deg A_c - \deg A = 1$$

Noting that $\deg R = 1$ and $\deg B^+ = 1$ then it follows that

$$\deg R' = 0$$

Let $R' = 1$, then the left hand side (LHS) of the Diophantine equation becomes,

$$\text{LHS} = AR' + B - S$$

$$\text{LHS} = A + B_{\text{minus}}(s_0q + s_1)$$

From the compatibility conditions, we get the $\deg A_c = 0$, thus we choose it to be,

$$A_c = 1$$

The right hand side (RHS) of the Diophantine equation becomes,

$$\text{RHS} = A_cA_m$$

$$\text{RHS} = Am$$

$$\text{dio_poly} = \text{sp.Poly}(sp.collect(\text{LHS} - \text{RHS}, q), q)$$

To find s_0 we consider the first degree terms.

$$s_0_{\text{sym}} = \text{sp.solve}(\text{dio_poly.coeffs}()[0], s_0[0])$$

$$s_1_{\text{sym}} = \text{sp.solve}(\text{dio_poly.coeffs}()[1], s_1[0])$$

$$S = (s_0_{\text{sym}}q + s_1_{\text{sym}})$$

$$R = B_{\text{plus}} + S$$

$$T = \text{sp.Poly}(R, q)$$

To find s_1 we consider the equations that are constant in q .

$$s_1_{\text{sym}} = \text{sp.solve}(\text{dio_poly.coeffs}()[1], s_1[0])$$

$$S = (s_0_{\text{sym}}q + s_1_{\text{sym}})$$

$$R = B_{\text{plus}} + S$$

$$T = \text{sp.Poly}(R, q)$$

With this the S polynomial is,

$$S = q \left(-\frac{\hat{a}_1 + a_{m1}}{\hat{b}_0} \right) + \frac{-\hat{a}_2 + a_{m2}}{\hat{b}_0}$$

$$R = B^+ = q + \frac{\hat{b}_1}{\hat{b}_0}$$

The R polynomial is given as,

$$R = q + \frac{\hat{b}_1}{\hat{b}_0}$$

The T polynomial is found using,

$$T = A_cB_m$$

where $B_m = B - B_{\text{minus}} = A_m(q)^{n-d}$

```
In [21]: Bm = Am.subs(q, 1)*q
t0_sym = Bm/B_minus
T = t0_sym

Then T is,
```

$$T = \frac{q(a_{m1} + a_{m2} + 1)}{\hat{b}_0}$$

As a summary, the control polynomials are given by,

$$R = q + r_1 = q + \frac{\hat{b}_1}{\hat{b}_0}$$

$$S = s_0q + s_1 = \frac{q(-\hat{a}_1 + a_{m1})}{\hat{b}_0} + \frac{-\hat{a}_2 + a_{m2}}{\hat{b}_0}$$

$$T = tq = \frac{q(a_{m1} + a_{m2} + 1)}{\hat{b}_0}$$

These values are then used in the linear controller algorithm,

$$Ru(t) = Tu_c(t) - Sy(t)$$

$$u(t) = -r_1u(t-1) + \hat{b}_0u(t) - s_0y(t) - s_1y(t-1)$$

The output equation is given by,

$$y(t) = \phi(t-1)^T \theta$$

where,

$$\phi(t-1)^T = [-y(t-1) - y(t-2)u(t-1)u(t-2)]$$

$$\hat{\theta} = [\hat{a}_1 \hat{a}_2 \hat{b}_1 \hat{b}_0]^T$$

2.3-4 Minimal Degree Pole Placement Algorithm for Indirect Controller and Simulation

```
In [22]: def reference_signal(time=np.linspace(0, 100, 101)):
    sig = []
    rat = 2.0/40.0*pi
    rat = 1.0/25.0*pi
    for t in range(100):
        if sin(10*rat*t) >= 0:
            sig.append(1)
        else:
            sig.append(0)
    return np.array(sig, dtype=float)

def noise_generator(time=np.linspace(0, 100, 101)):
    noise = []
    for t in range(100):
        if t < 35:
            noise.append(0)
        else:
            noise.append(0.25)
    return np.array(noise)

We start by initializing the values of the variables,
```

```
In [23]: uc = reference_signal()
lam = 1

# Assignment, works
am1 = -2*np.exp(-2)
am2 = np.exp(-4)
a1_true, a2_true, b0_true, b1_true = -0.503215, 0.0497871, 1.59831, 0.587984

# In class example, works perfectly!
# am1 = -1.3205
# am2 = 0.4966
# a1_true, a2_true, b0_true, b1_true = -1.6065, 0.6065, 0.1065, 0.0902

theta = np.array([a1_true, a2_true, b0_true, b1_true]).reshape(4,1) # a1[0], a2[1], b0[2], b1[3]

In [24]: # Initializing the coefficients for A, B for t = 0
theta_hat = np.array([0, 0, 0, 0]).reshape(4,1)
a1[0], a2[1], b0[2], b1[3] += SET b0 TO 0.01 FOR CONV
a1_est, a2_est, b0_est, b1_est = theta_hat[0], theta_hat[1], theta_hat[2], theta_hat[3]
P = np.diag([10, 1, 100, 10]) # WORKS
B = (b0*q + b1)
B_poly = sp.Poly(B, q)

# Calculating the control parameters for t = 0
y = np.array([0])
y_hat = y
t0 = (am1 + am2 + 1)/theta_hat[2]
s0 = (am1 - theta_hat[0])/theta_hat[2]
u = t0*uc[0] - s0*y[0]

# Calculating the coefficients for A, B for t = 1
phi = np.array([-y[0], 0, u[0], 0]).reshape(4,1)
phi_hat = np.array([0, 0, 0, 0]).reshape(4,1)
K = P @ phi
phi_hat = inv(inv(phi) + trans(phi) @ phi) @ phi
theta_hat = (theta_hat + Ts*(K @ phi + trans(phi) @ theta_hat - trans(phi) @ theta_hat))
theta_hat_list = np.concatenate((theta_hat,
                                (theta_hat).reshape(3,1),
                                (theta_hat_list + Ts*(K @ phi + phitrans(phi) @ theta_hat,
                                axis=1)

P = inv(inv(P) + phitrans(phi))
K = P@phi

a1_est = np.concatenate((a1_est, theta_hat[0]), axis=0)
a2_est = np.concatenate((a2_est, theta_hat[1]), axis=0)
b0_est = np.concatenate((b0_est, theta_hat[2]), axis=0)
b1_est = np.concatenate((b1_est, theta_hat[3]), axis=0)

# Calculating the control parameters for t = i
r1 = theta_hat[3]/theta_hat[2]
t0 = (am1 + am2 + 1)/theta_hat[2]
s0 = (am1 - theta_hat[0])/theta_hat[2]
s1 = (am2 - theta_hat[1])/theta_hat[2]
u = np.concatenate((u,
                        (-s1*y[0] + t0*uc[1] - s0*y[1] - s1*y[0]))

for t in np.arange(2, 101):
    phi = np.array([-y[t-1], -y[t-2], u[t-1], u[t-2]]).reshape(4,1)
    y = np.concatenate((y,
                        (trans(phi) @ theta).reshape(-1,)), axis=0)

    K = P @ phi
    phi_hat = inv(inv(phi) + trans(phi) @ phi) @ phi
    theta_hat = (theta_hat + Ts*(K @ phi + trans(phi) @ theta_hat - trans(phi) @ theta_hat))
    theta_hat_list = np.concatenate((theta_hat,
                                    (theta_hat).reshape(3,1),
                                    (theta_hat_list + Ts*(K @ phi + phitrans(phi) @ theta_hat,
                                    axis=1)

    P = inv(inv(P) + phitrans(phi))
    K = P@phi

    a1_est = np.concatenate((a1_est, theta_hat[0]), axis=0)
    a2_est = np.concatenate((a2_est, theta_hat[1]), axis=0)
    b0_est = np.concatenate((b0_est, theta_hat[2]), axis=0)
    b1_est = np.concatenate((b1_est, theta_hat[3]), axis=0)

    # Calculating the control parameters for t = i
    r1 = theta_hat[3]/theta_hat[2]
    t0 = (am1 + am2 + 1)/theta_hat[2]
    s0 = (am1 - theta_hat[0])/theta_hat[2]
    s1 = (am2 - theta_hat[1])/theta_hat[2]
    u = np.concatenate((u,
                        (-s1*y[t-1] + t0*uc[t] - s0*y[t] - s1*y[t-1]))

In [25]: plt.plot(np.linspace(0, 100, 101), y)
plt.plot(np.linspace(0, 100, 101), uc)
plt.legend(['y', 'u_c'])
plt.title('Control Action vs. Time')
plt.show()
```

