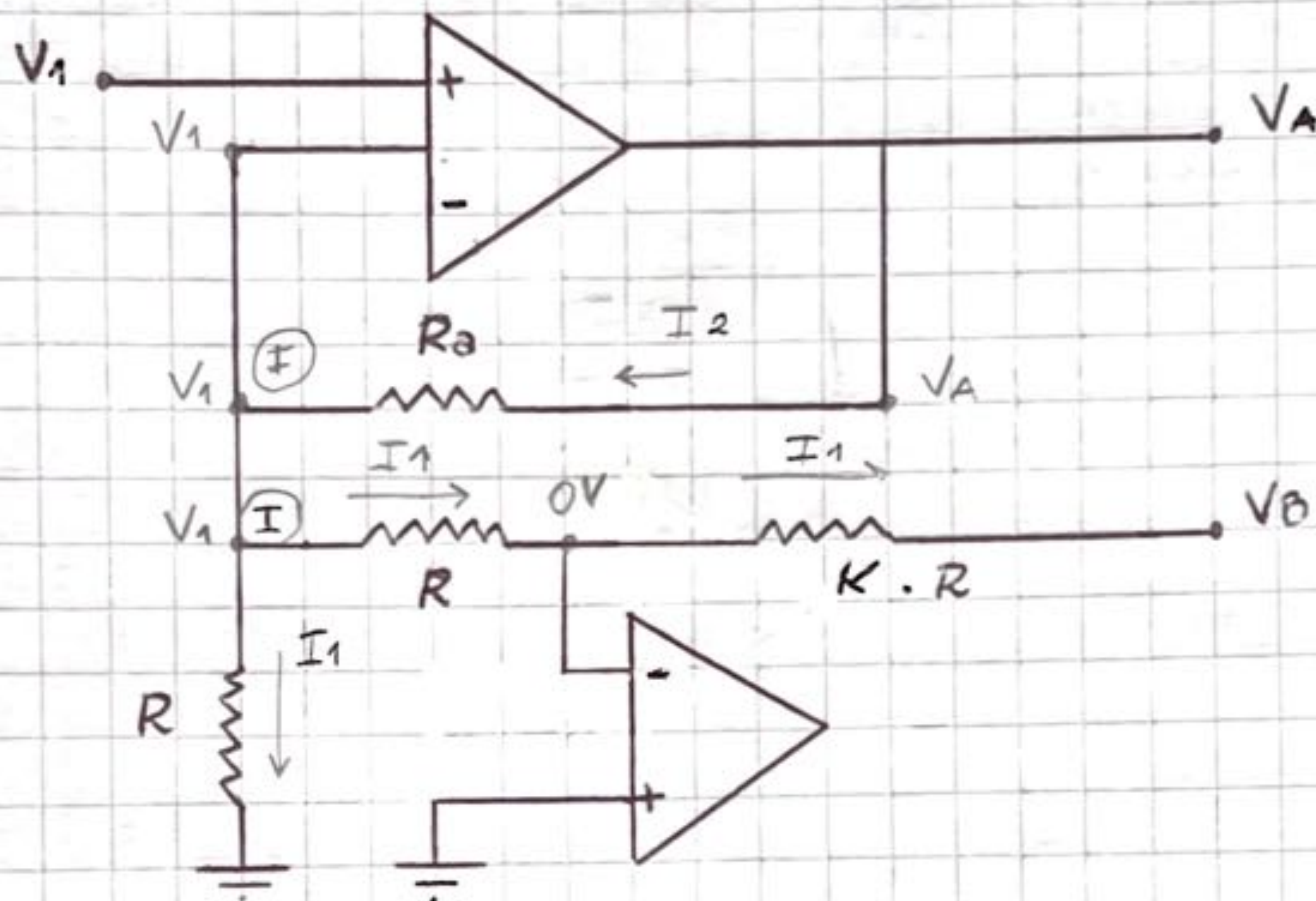


② Amplificador con salida diferencial. Obtener la expresión de  $V_{AB}$ . Utilizar el valor de  $R_2 = \frac{R}{2}(K-1)$ .



$$\bullet I_1 = \frac{V_1}{R} = \frac{-V_B}{K \cdot R}$$

$$\rightarrow \boxed{V_B = -V_1 K}$$

$$\bullet I_2 = \frac{V_A - V_1}{R_2}$$

Plantearmos nodos:

$$\textcircled{\text{I}} (V_A - V_1) \cdot G_2 = 2V_1 G$$

$$V_A \cdot G_2 - V_1 G_2 = 2V_1 G$$

$$V_A \cdot G_2 = 2V_1 (G + G_2)$$

$$V_A = 2V_1 \frac{(G + G_2)}{G_2}$$

$\rightarrow$

$$\boxed{V_A = 2V_1 \left( \frac{\frac{1}{R} + \frac{1}{\frac{R}{2}(K-1)}}{\frac{1}{\frac{R}{2}(K-1)}} \right)}$$

$$V_A = 2V_1 \left( \frac{\frac{R}{2}(K-1)}{1} \right) \left( \frac{1}{R} + \frac{1}{\frac{R}{2}(K-1)} \right)$$

$$V_{AB} = V_A - V_B = 2V_1 \left( \frac{K-1}{2} + 1 \right) + V_1 K = V_1 (K-1) + 2V_1 + V_1 K$$

$$V_{AB} = V_1 K - V_1 + 2V_1 + V_1 K = 2V_1 K + V_1 = V_1 (2K + 1)$$

$$\boxed{V_{AB} = V_1 (2K + 1)}$$



## ⑦ Desarrollo algebraico:

A)

$$\frac{V_1 - V_2}{R_1 + R_2} = \frac{V_1 - \frac{V_1 \cdot R_3}{R_3 + 1/CS}}{R_1} = \frac{1}{R_1} \left( V_1 - \frac{V_1 \cdot SCR_3}{SCR_3 + 1} \right) = \frac{V_1}{R_1} \left( 1 - \frac{SCR_3}{SCR_3 + 1} \right)$$

$$\frac{V_1 - V_2}{R_1 + R_2} = \frac{V_1}{R_1} \left( 1 - \frac{SCR_3}{CR_3 \left( S + \frac{1}{CR_3} \right)} \right) = \frac{V_1}{R_1} \left( 1 - \frac{S}{S + \frac{1}{CR_3}} \right) = \frac{V_1}{R_1} \left( \frac{1/CR_3}{S + \frac{1}{CR_3}} \right)$$

$$\frac{V_1 - V_2}{R_1 + R_2} = V_1 \cdot \frac{1}{CR_1 R_3} \cdot \frac{1}{S + \frac{1}{CR_3}}$$

Despejamos  $V_2$ .

$$-V_2 = \left( V_1 \cdot \frac{1}{CR_1 R_3} \cdot \frac{1}{S + \frac{1}{CR_3}} \right) (R_1 + R_2) - V_1$$

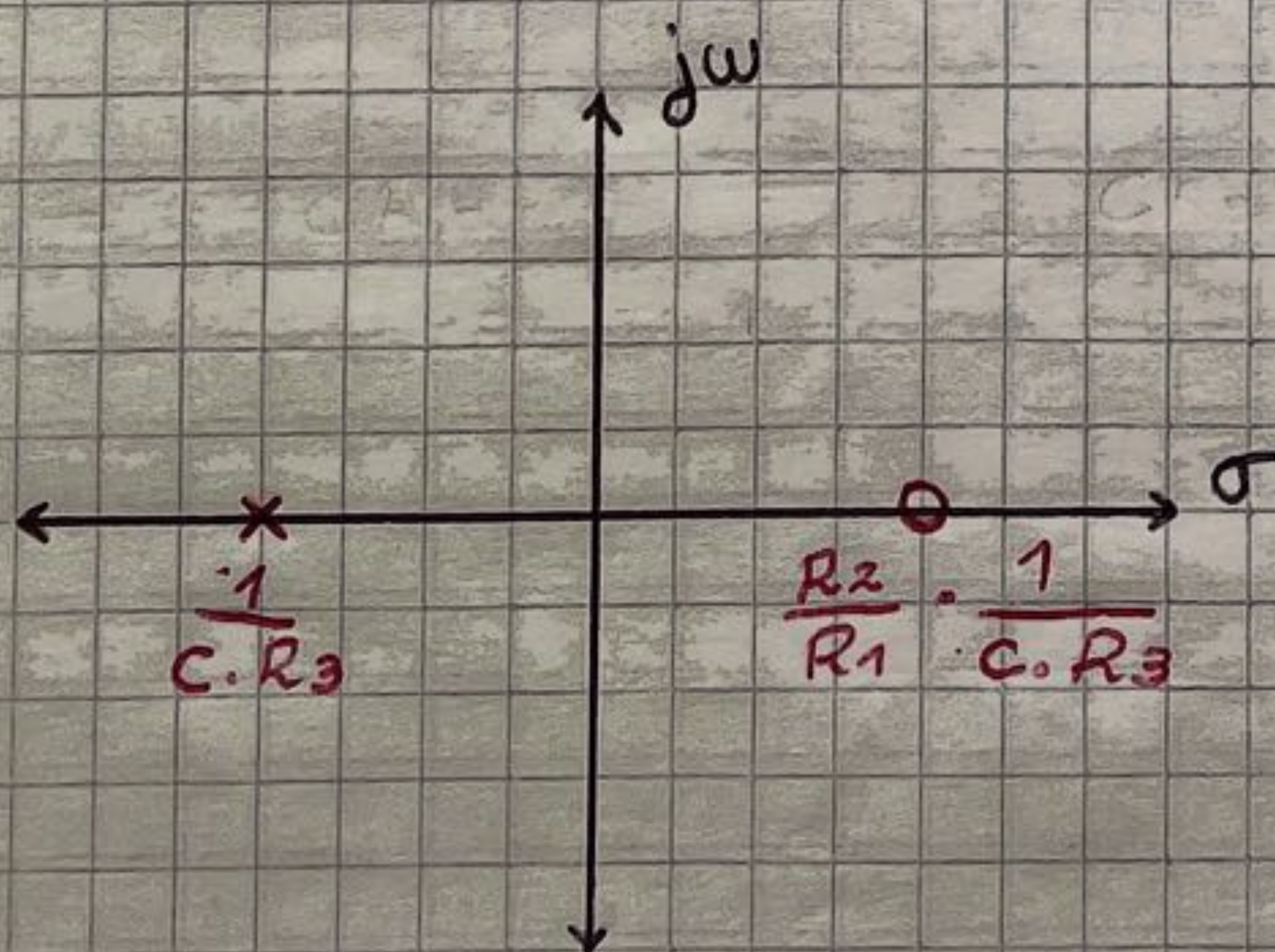
$$V_2 = V_1 \left( 1 - \frac{R_1 + R_2}{R_1 (SCR_3 + 1)} \right) = V_1 \left( \frac{R_1 (SCR_3 + 1) - R_1 - R_2}{R_1 (SCR_3 + 1)} \right)$$

$$V_2 = V_1 \left( \frac{SCR_3 + 1 - 1 - R_2/R_1}{SCR_3 + 1} \right)$$

Finalmente:

$$\frac{V_2}{V_1} = \frac{SCR_3 - R_2/R_1}{SCR_3 + 1} = \frac{CR_3 (S - R_2/R_1 CR_3)}{CR_3 (S + 1/CR_3)} = \frac{S - R_2/R_1 CR_3}{S + 1/CR_3}$$

Diagrama de polos y ceros:



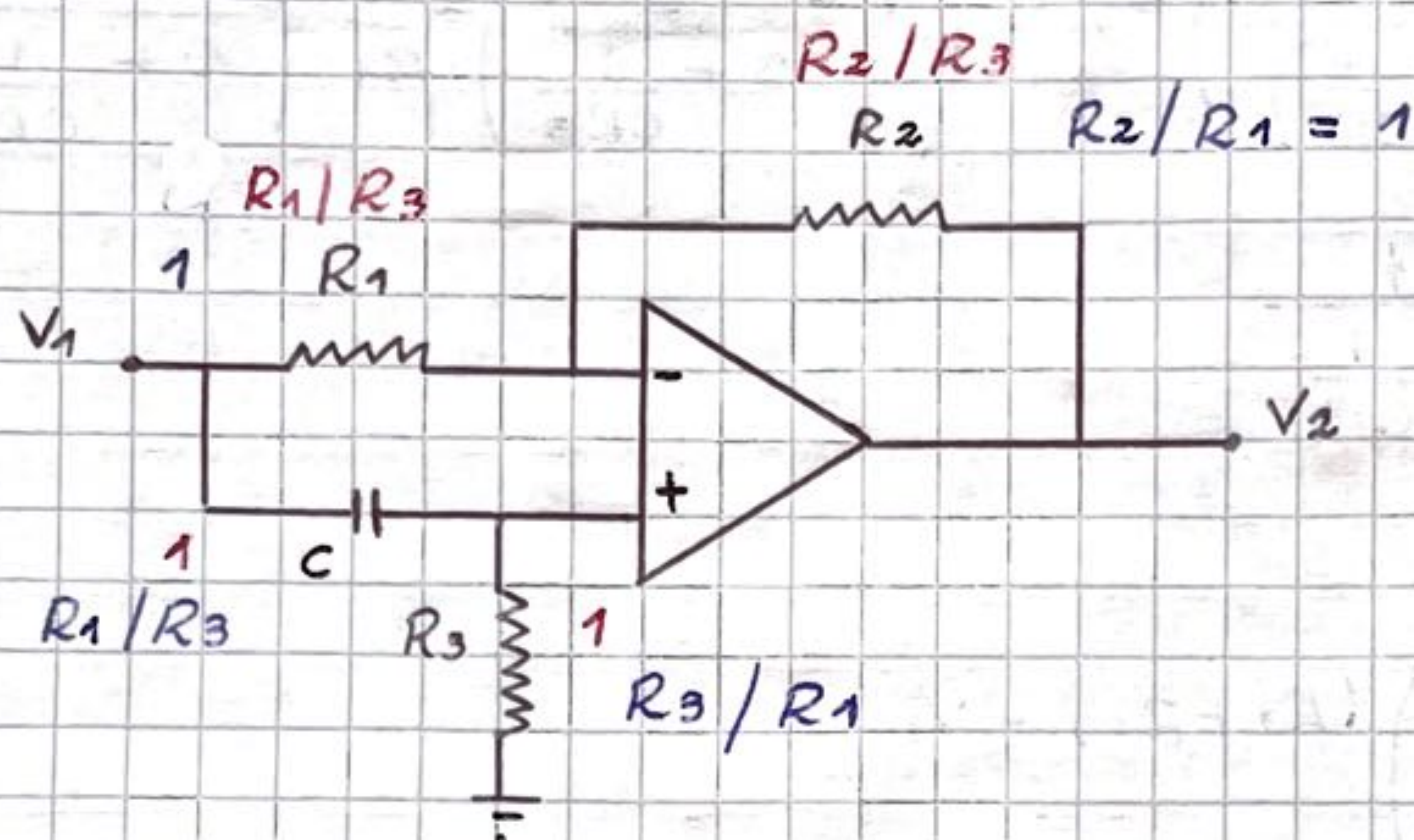


## Normalización:

$$\cdot \kappa_w = \frac{1}{R_3 \cdot C}$$

$$\cdot \kappa_z = \begin{matrix} \rightarrow R_1 \\ \rightarrow R_3 \end{matrix}$$

→ Recordamos que debemos multiplicar por estos parámetros para normalizar.



Recordamos que:

$$\frac{V_2}{V_1} = \frac{SCR_3 + R_2/R_1}{S \cdot CR_3 + 1}$$

Normalizamos:  $s' = s / \kappa_w \rightarrow s = s' \cdot \kappa_w = s' \cdot \frac{1}{R_3 \cdot C}$

$$\cdot \frac{V_2}{V_1} = \frac{s' - R_2/R_1}{s' + 1}$$

$$\gamma \quad c' = \kappa_w \cdot C \rightarrow c' = \frac{1}{R_3 \cdot C} \cdot C = \frac{1}{R_3}$$

$\gamma$  en impedancia:

$$\begin{cases} R_1' = R_1 / \kappa_z = R_1 / R_3 \\ R_2' = R_2 / \kappa_z = R_2 / R_3 \\ R_3' = R_3 / \kappa_z = 1 \\ C'' = c' \cdot \kappa_z = 1 / R_3 \cdot R_3 = 1 \end{cases}$$

$$\begin{cases} R_1' = R_1 / \kappa_z = 1 \\ R_2' = R_2 / \kappa_z = R_2 / R_1 \\ R_3' = R_3 / \kappa_z = R_3 / R_1 \\ C'' = c' \cdot \kappa_z = R_1 / R_3 \end{cases}$$



Expresión de módulo y fase:

Partimos de la transferencia:  $\frac{V_2}{V_1} = \frac{s - \frac{R_2}{R_1} \cdot \frac{1}{C \cdot R_3}}{s + \frac{1}{C \cdot R_3}}$

• Módulo:

$$\left. \frac{V_2}{V_1}(s) \right|_{s=j\omega} = \frac{j\omega - \frac{R_2}{R_1} \cdot \frac{1}{C \cdot R_3}}{j\omega + \frac{1}{C \cdot R_3}}$$

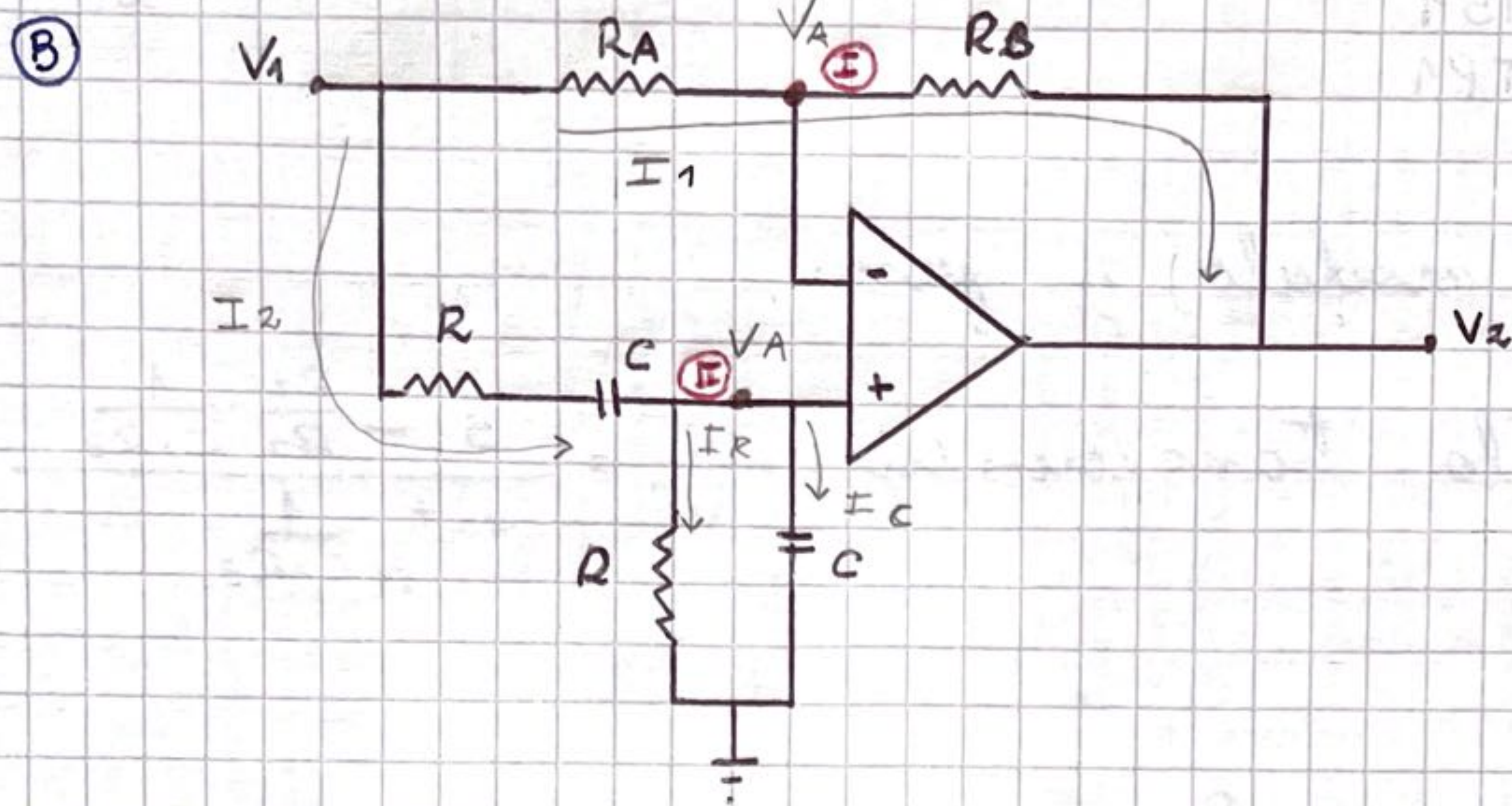
$$\left| \frac{V_2}{V_1}(j\omega) \right|^2 = \frac{\omega^2 + \left( \frac{R_2}{R_1} \cdot \frac{1}{C \cdot R_3} \right)^2}{\omega^2 + \left( \frac{1}{C \cdot R_3} \right)^2}$$

• Fase:  $\phi(\omega) = \sum_1^m \tan^{-1} \frac{\omega \pm \beta_z}{\alpha_z} - \sum_1^p \tan^{-1} \frac{\omega \pm \beta_p}{\alpha_p}$

$$\phi(\omega) = \tan^{-1} \left( \frac{\omega}{-\frac{R_2}{R_1} \cdot \frac{1}{C \cdot R_3}} \right) - \tan^{-1} \left( \frac{\omega}{\frac{1}{C \cdot R_3}} \right)$$

E) Este circuito es útil como rotador de fase de orden 1 si se pretende no modificar la amplitud de la señal.





$$\begin{cases} R_A/R_B = 5 \\ R = 1\text{ k}\Omega \\ C = 1\text{ }\mu\text{F} \end{cases}$$

Plantamos:

$$\cdot I_1 = (V_1 - V_A) G_A = (V_A - V_2) \cdot G_B$$

$$\cdot I_2 = (V_1 - V_A) \left( \frac{1}{R + \frac{1}{sC}} \right)^{Y_1} = (V_1 - V_A) \frac{sC}{sCR + 1}$$

$$\cdot I_R = V_A \cdot G$$

$$\cdot I_C = V_A \cdot sC^{Y_2}$$

Plantamos nodos:

$$\textcircled{\text{I}} (V_1 - V_A) G_A = (V_A - V_2) \cdot G_B$$

$$V_1 \cdot G_A - V_A \cdot G_A = V_A \cdot G_B - V_2 \cdot G_B$$

$$V_1 \cdot G_A + V_2 \cdot G_B = V_A \cdot G_B + V_A \cdot G_A$$

$$\rightarrow V_A = \frac{V_1 G_A + V_2 G_B}{G_B + G_A}$$

$$\textcircled{\text{II}} (V_1 - V_A) \left( \frac{sC}{sRC + 1} \right) = V_A \cdot G + V_A \cdot Y_2 = V_A (G + Y_2)$$

$$V_1 \left( \frac{sC}{sRC + 1} \right) - V_A \left( \frac{sC}{sRC + 1} \right) = V_A (G + Y_2)$$

$$V_1 \left( \frac{sC}{sRC + 1} \right) = V_A (G + Y_2) + V_A \left( \frac{sC}{sRC + 1} \right)$$

$$V_1 \cdot Y_1 = V_A [(G + Y_2) + Y_1]$$

Reemplazamos con  $V_A$

$$V_1 \cdot Y_1 = \left( \frac{V_1 G_A + V_2 G_B}{G_B + G_A} \right) (G + Y_2 + Y_1) = \left( \frac{V_1 G_A}{G_B + G_A} + \frac{V_2 G_B}{G_B + G_A} \right) (G + Y_1 + Y_2)$$



$$V_1 \cdot Y_1 = \left( \frac{V_1}{1 + \frac{G_B}{G_A}} + \frac{V_2}{1 + \frac{G_A}{G_B}} \right) (G + Y_1 + Y_2)$$

$$V_1 \cdot Y_1 = \left( V_1 \frac{1}{1 + \frac{G_B}{G_A}} + \frac{V_2}{1 + \frac{G_A}{G_B}} \right) \left( \frac{1 + SC}{R} \frac{SCR + 1}{SCR + 1} + SC \right)$$

$$\cdot \frac{1}{R} + \frac{SC}{SCR + 1} + SC = \frac{SCR + 1 + SCR + S^2 C^2 R^2 + SCR}{R(SCR + 1)} = \frac{S^2 C^2 R^2 + 3SCR + 1}{SCR^2 + R}$$

Entonces:

$$V_1 \cdot \frac{SC}{SCR + 1} = \frac{V_1}{1 + \frac{R_A}{R_B}} \left( \frac{S^2 C^2 R^2 + 3SCR + 1}{SCR^2 + R} \right) + \frac{V_2}{1 + \frac{R_B}{R_A}} \left( \frac{S^2 C^2 R^2 + 3SCR + 1}{SCR^2 + R} \right)$$

$$V_1 \left( \frac{SC}{SCR + 1} - \frac{S^2 C^2 R^2 + 3SCR + 1}{\left(1 + \frac{R_A}{R_B}\right)(SCR^2 + R)} \right) = \left( \frac{S^2 C^2 R^2 + 3SCR + 1}{\left(1 + \frac{R_B}{R_A}\right)(SCR^2 + R)} \right) V_2$$

$$\frac{SC \left(1 + \frac{R_A}{R_B}\right) R - (S^2 C^2 R^2 + 3SCR + 1)}{\left(1 + \frac{R_A}{R_B}\right)(SCR + 1) R} = \frac{SCR + SCR \left(\frac{R_A}{R_B}\right) - S^2 C^2 R^2 - 3SCR - 1}{\left(1 + \frac{R_A}{R_B}\right)(SCR + 1) R}$$

$$= \frac{-S^2 C^2 R^2 + SCR \left(\frac{R_A}{R_B} - 2\right) - 1}{\left(1 + \frac{R_A}{R_B}\right)(SCR^2 + R)}$$

Finalmente:

$$\frac{V_2}{V_1} = \frac{-S^2 C^2 R^2 + SCR \left(\frac{R_A}{R_B} - 2\right) - 1}{\left(1 + \frac{R_A}{R_B}\right)(SCR^2 + R)} \cdot \frac{\left(1 + \frac{R_B}{R_A}\right)(SCR^2 + R)}{S^2 C^2 R^2 + 3SCR + 1}$$

$$\frac{V_2}{V_1} = \frac{1 + R_B/R_A}{1 + R_A/R_B} \cdot \frac{-S^2 C^2 R^2 + SCR \left(\frac{R_A}{R_B} - 2\right) - 1}{S^2 C^2 R^2 + 3SCR + 1}$$

↓ Reemplazamos con  $R_A/R_B = 5$ .

$$\frac{V_2}{V_1} = \frac{-S^2 C^2 R^2 + SCR (5 - 2) - 1}{S^2 C^2 R^2 + 3SCR + 1} \cdot \frac{1 + 1/5}{1 + 5} = \frac{1}{5} \left( \frac{-S^2 C^2 R^2 + 3SCR - 1}{S^2 C^2 R^2 + 3SCR + 1} \right)$$



$$\frac{V_2}{V_1} = \left(-\frac{1}{5}\right) \cdot \left(\frac{S^2 C^2 R^2 - 3SCR + 1}{S^2 C^2 R^2 + 3SCR + 1}\right)$$

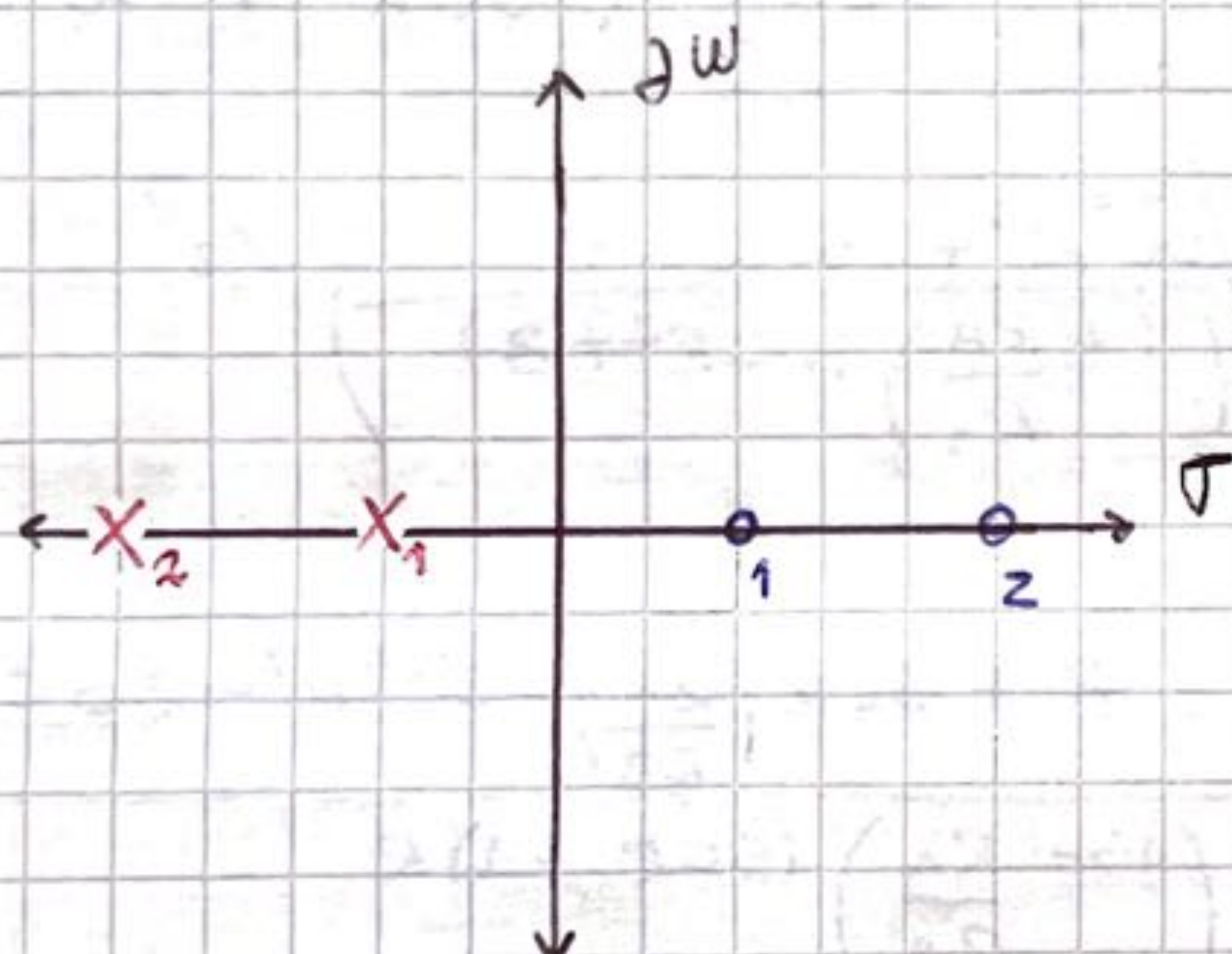
De forma mónica:

$$\frac{V_2}{V_1} = \left(-\frac{1}{5}\right) \left(\frac{S^2 - 3/CR + 1/C^2 R^2}{S^2 + 3/CR + 1/C^2 R^2}\right)$$

Reemplazando con valores:  $R = 1k\Omega$  y  $C = 1\mu F$ :

$$\frac{V_2}{V_1} = \left(-\frac{1}{5}\right) \left(\frac{S^2 - 3000S + 10^6}{S^2 + 3000S + 10^6}\right)$$

• Diagrama de polos y ceros:



$$o_1: 381,966$$

$$o_2: 2618,034$$

$$x_1: -381,966$$

$$x_2: -2618,034$$

• Módulo:  $\left.\frac{V_2(s)}{V_1}\right|_{s=j\omega} = \left(-\frac{1}{5}\right) \left(\frac{(j\omega)^2 - 3000j\omega + 10^6}{(j\omega)^2 + 3000j\omega + 10^6}\right) = \left(-\frac{1}{5}\right) \frac{-3000j\omega + (10^6 - \omega^2)}{3000j\omega + (10^6 - \omega^2)}$

$$\left|\frac{V_2(j\omega)}{V_1}\right|^2 = \frac{(10^6 - \omega^2)^2 + (3000\omega)^2}{(10^6 - \omega^2)^2 + (3000\omega)^2} \left(-\frac{1}{5}\right)^2$$

• Fase:  $\phi(\omega) = \sum_1^m \tan^{-1} \frac{\omega - \beta_z}{\alpha_z} - \sum_1^p \tan^{-1} \frac{\omega - \beta_p}{\alpha_p}$

$$\phi(\omega) = \left[\tan^{-1} \left(\frac{\omega}{381,966}\right) + \tan^{-1} \left(\frac{\omega}{2618}\right)\right] - \left[\tan^{-1} \left(\frac{\omega}{-381,966}\right) + \tan^{-1} \left(\frac{\omega}{-2618}\right)\right]$$



B) Normalización:

Partimos de:  $\frac{V_2(s)}{V_1} = \left(-\frac{1}{5}\right) \left(\frac{s^2 C^2 R^2 - 3sCR + 1}{s^2 C^2 R^2 + 3sCR + 1}\right)$

Entonces:  $s'^2 = s^2 C^2 R^2 \rightarrow s = \frac{s'}{CR} \rightarrow \omega = \frac{1}{CR}$

Adoptamos:  $\omega = \frac{1}{CR} \rightarrow \omega^2 = \frac{1}{C^2 R^2}$

Luego, nuestra función normalizada en frecuencia:

$$\frac{V_2(s)}{V_1} = \left(-\frac{1}{5}\right) \cdot \left(\frac{s'^2 - 3s' + 1}{s'^2 + 3s' + 1}\right)$$

La interpretación circuital es que es la pulsación de resonancia del circuito.

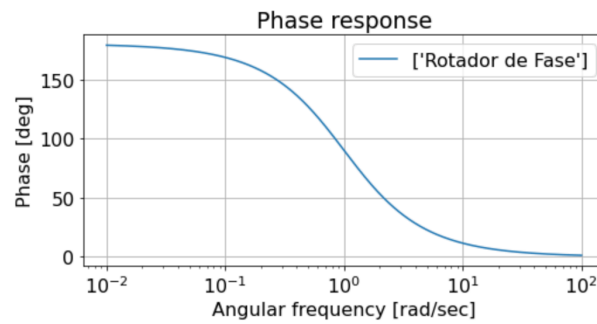
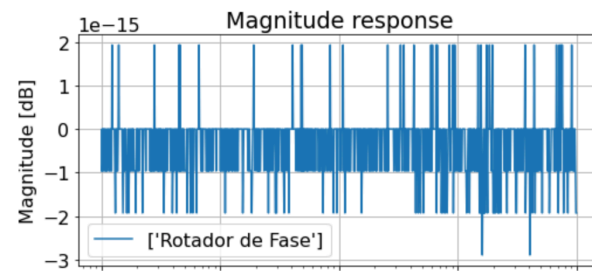
E) Este circuito podría ser útil si lo que se busca es un rotador de fase de  $180^\circ$  que a su vez atenúe a un 20% a la señal de entrada.



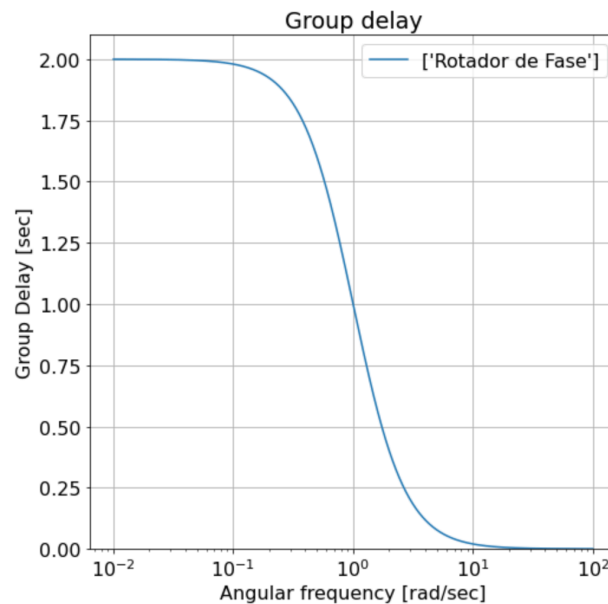
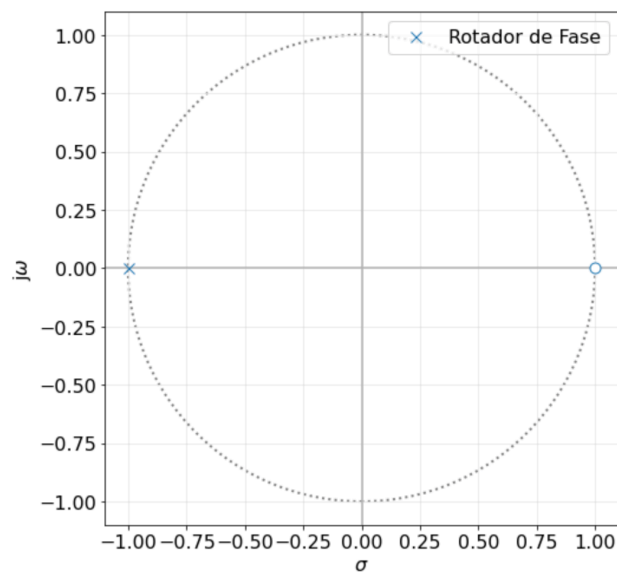
## Ejercicio 7) Circuito A) Punto c.

- Simule la función transferencia normalizada (Python, Matlab, etc.).

$$\frac{s+1}{s+1}$$



Poles and Zeros map





## Ejercicio 7) Circuito B) Punto c.

- Simule la función transferencia normalizada (Python, Matlab, etc.).

$$\frac{s^2 - 0.2 + s \ 0.6 + -0.2}{s^2 + s \ 3 + 1}$$

