

1st January. Measuring Instruments

Specification of Instrument

i) Accuracy

It is a measure of closeness to the true value of quantity.

$$\text{Ex- } 5A \quad \left\{ \begin{array}{l} \frac{4.9A}{4.7A} \leftarrow \text{better accuracy} \\ 4.6A \end{array} \right.$$

ii) Precision

It is the reproducibility of measures

Ex- $4.91, 4.92, 4.91, 4.91, 4.91$ high prec
 $4.92, 4.93, 4.94, 4.91, 4.96$ low !!

iii) Sensitivity

It is a ratio of magnitude of output signal to that of i/p signal.

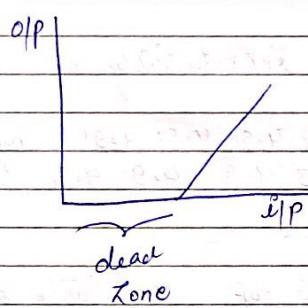
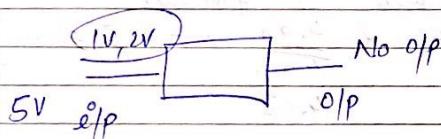
Magn. of o/p signal

4 " " i/p "

When there is small change in input there is a lot of change in output then that instrument is called highly sensitive instrument.

IV) Dead Zone

It is a largest ran changes of i/p quantity for which there is no change o/p of the instrument.



V) Dead time

It is the time req. by the instrument to response the changes in the i/p.

VI) Resolution

It is the smallest observable changes in the input that an ins. can response.

Error

Diffr. of true value & measured value.

1) Static Error

$\text{Measured value} - \text{True value}$

Error Correction,

$$\text{if } M.V = 4.5 \\ T.V = 5$$

$$E.C = -0.5$$

ii) Relative Static Error

$$\frac{\text{Measured value} - \text{True value}}{\text{True Value}}$$

Types of Error

i) Gross Error → Error due to Reading, Recording & calculation

ii) Systematic Error

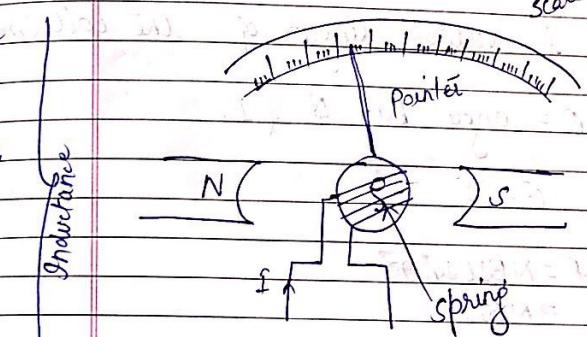
- a) Instrumental Error (Instrument)
- b) Environmental Error (temp, atm pres, wind)
- c) Observational Errors (Human eye)
- d) Random Error (Unknown)

PMMC - Permanent Magnet Moving coil
MI - Moving Iron

V/I	DC Bridge (Resistance)	AC Bridge (Capacitance Inductance)
1. PMMC (DC)	Ammeter / Voltmeter (Resistance)	Maxwell Induction Bridge
2. MI (AC/DC)	Potentiometer (Resistance)	Maxwell Inducto. Capacitance Bridge
3. Dymo Measuring Instrument (AC/DC)	Kelvin Bridge	Hay's Bridge
4. Electrostatic (AC/DC)	Substitution	Anderson Bridge
5. Thermal Instr/ Hot Wire (AC/DC)	Wheatstone Bridge	Owen Bridge
6. Rectifier (AC/DC)	Carry-Foster Slide-Wire Bridge	De-Sauls Bridge
7. Induction type (AC)	Direct Deflection	Schering Bridge
	Loss of charge Megohms	Megger

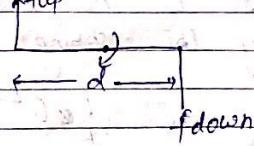
17th January

1. PMMC (Permanent Magnet moving coil)



When a current carrying coil is placed in a magnetic field, it experiences a force due to this field on the meter due to deflect. The deflection of the pointer is directly proportional to the amount of current passing through the coil. (Faraday's law).

Expression of torque in PMMC.



force on the coil = $NBIl \sin\theta$.

N = No. of turns of the coil
 B = Magnetic flux density (T wb/m²)

I = amount of current passing through the coil (A)

d = vertical length of the coil (cm)

θ = angle bet. B & I .

$\alpha = 90^\circ$

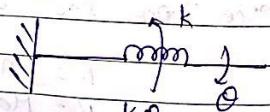
$$F = NBIL \sin 90^\circ \\ = NBIL$$

deflecting Torque due to both side of the coil
 = force \times \perp distance
 $= NBIL \times d$
 $= NBILd$
 $= NBIA$.
 ↑
 Area

Controlling torque (torque because of the spring)
 It is the restraining torque provided by the spring control.

$$T_c = k \theta \leftarrow \text{deflecting angle}$$

spring constant



At balance point or null deflection
 (steady state)

$$T_d = T_c$$

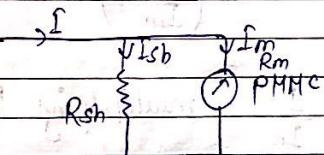
$$NBIA = k\theta$$

$$\theta = \frac{NBIA}{k}$$

$$= \frac{(NAB)}{k}$$

$[\theta \propto I]$ → scale is linear

Ammeter Shunt in PMMC.
 Connecting resistance in parallel with PMMC.



R_{sh} = Shunt Resistance

R_m = Meter Resistance

I = Current to be measured

I_m = full scale deflection current

I_{sh} = Shunt current.

$$V_{sh} = V_m$$

$$I_{sh} R_{sh} = I_m R_m$$

$$R_{sh} = \frac{I_m R_m}{I_{sh}}$$

$$I = I_{sh} + I_m$$

$$\boxed{I_{sh} = I - I_m}$$

$$R_{sh} = \frac{I_m R_m}{I - I_m}$$

divide num. & den. by I_m

$$R_{sh} = \frac{R_m}{\left(\frac{I}{I_m} - 1\right)}$$

$$\frac{I}{I_m} = (\text{multiplying Power}) m.$$

$$R_{sh} = \frac{R_m}{m-1}$$

Properties of ammeter shunt

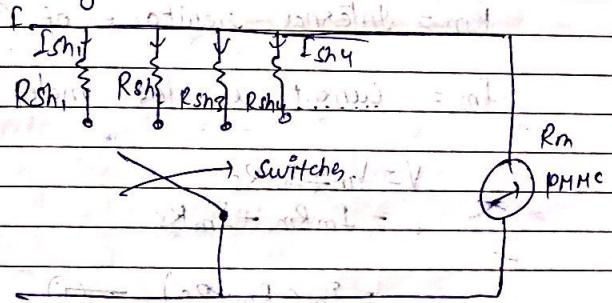
- The resistance of the shunt should not vary with time.

It should have a low temp. coeff. and size should be small.

It is usually made up of Manganin.

Multirange Ammeter

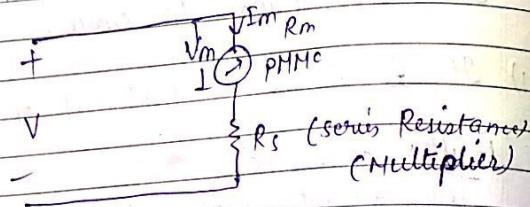
The current range of a dc-meter may be further extended by a no. of shunt resistances which are usually selected by a range switch which is called as Multirange Ammeter.



$$R_{sh} = \frac{R_m}{(m_i - 1)}$$

$$R_{sh_2} = \frac{R_m}{(m_2 - 1)}$$

PMMC as Voltmeter



V = full range voltage of the instrument.

R_m = internal resistance of the meter

I_m = current in the meter

$$\begin{aligned} V &= V_m + I_m R_c \\ &= I_m R_m + I_m R_s \\ &= I_m (R_m + R_s) \end{aligned}$$

$$V_m = I_m R_m$$

$$\textcircled{1} : \textcircled{11}$$

$$\sqrt{\frac{V}{V_m}} = \frac{R_m + R_s}{R_m}$$

Multiplying factor (m_v)

$$m_v = \frac{I_m R_s}{R_m}$$

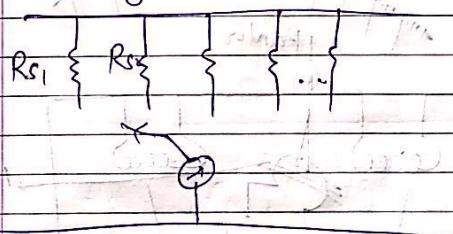
$$(m_v - 1) = \frac{R_s}{R_m}$$

$$R_s = R_m (m_v - 1)$$

Properties of Multiplier

- i) It is not changing with time.
- ii) It is non-inductive.
- iii) Variation of resist. with temp. should be small

Multirange Voltmeter



$$R_{s1} = (m_{v1} - 1) R_m$$

$$R_{s2} = (m_{v2} - 1) R_m$$

why it is "m"

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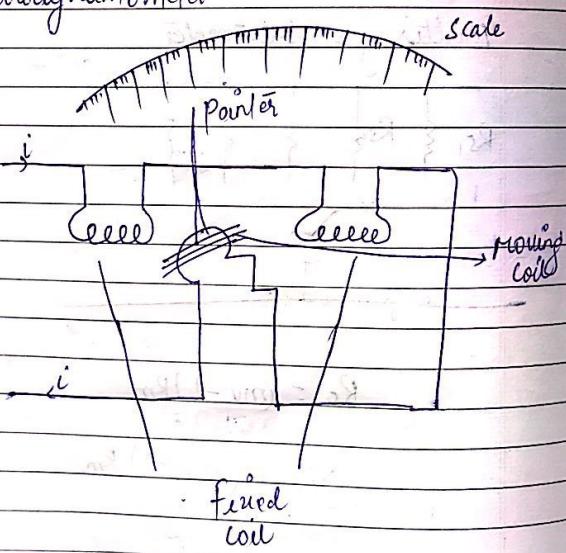
Some advantages of PMMC:

- i) Scale is linear.
- ii) Highly sensitive.
- iii) High accuracy.
- iv) Low power consumption.

disadvantage

- i) Meant for only dc purposes.
- ii) It is costly.

20th Jan Electrodynamometer



It consists of two part coil & fixed coil (FC) and moving coil (MC). Fixed coil is splitted into two part in order to have a uniform distribution of magnetic flux around the moving coil.

Fixed coil is used to provide the operating magnetic field and moving coil is the current carrying coil on which the deflection torque is produced.

Expression of Torque

Deflecting Torque in dynamometer is given by

$$T_d = i_1 i_2 \cdot \frac{dM}{d\theta}$$

i_1 = current in the fixed coil

i_2 = " " " moving "

M = Mutual inductance Effect bet FC & MC.

Controlling Torque

$$T_c = K\theta \leftarrow \text{deflecting angle}$$

Spring
cons.

Steady state / Null deflection

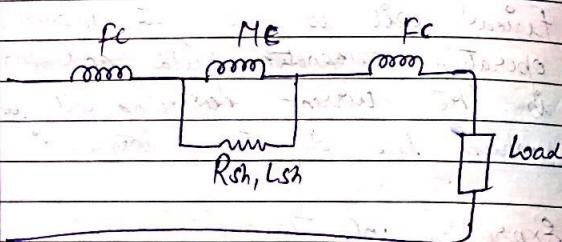
$$T_c = T_d$$

$$i_1 i_2 \frac{dM}{d\theta} = K\theta$$

$$\theta = \frac{i_1 i_2}{K} \frac{dM}{d\theta}$$

meter

a) Electrodynamometer type Ammeter



$$i_1 = i_2 = i \text{ (for balance condition)}$$

current in
in fc

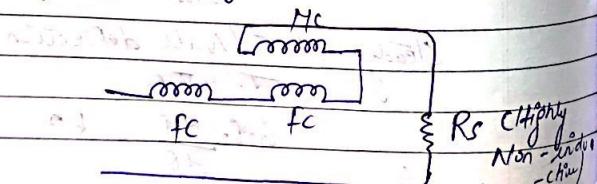
$$\theta = \frac{i^2}{K} \frac{dM}{d\theta}$$

$$\theta \propto i^2$$

Non-linear

Non-uniform

ii) Electrodynamometer type voltmeter



$$i_1 = i_2 = i \quad (\text{Null deflection/Balance condition})$$

$$T_d = \frac{i^2}{K} \frac{dM}{d\theta} \quad (i = \frac{V}{Z})$$

$$T_d = \left(\frac{V}{Z}\right)^2 \frac{dM}{d\theta}$$

$$T_c = K \theta$$

Steady State

$$T_d = T_c$$

$$\left(\frac{V}{Z}\right)^2 \frac{dM}{d\theta} = K \theta$$

$$\theta = \left(\frac{V}{Z}\right)^2 \frac{dM}{d\theta}$$

$$\theta \propto V^2$$

Non-linear

Non-uniform

Advantage

- i) High accuracy
- ii) free from hysteresis and eddy current losses.

Disadvantage

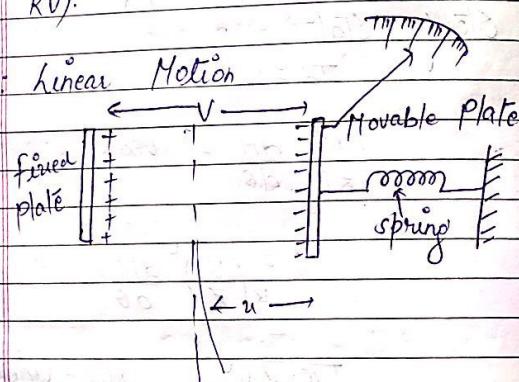
- i) scale is non-uniform.
- ii) friction loss are very high.
- iii) It is expensive.
- iv) High power consumption.

Electrostatic type Instrument
It is because of the static electric field.

In this type of instrument the deflecting torque is produced by the action of electric field on the charge conductor.

The voltage is very high (around KV).

Case I: Linear Motion



because the moving spring meter deflects.

deflecting force

$$f_d = \frac{V^2}{2} \left(\frac{dc}{du} \right)$$

V → Voltage C → Capacitance
 u → Movable distance.

Spring control force

$$f_c = Ku \quad K \rightarrow \text{spring constant}$$

Null deflection

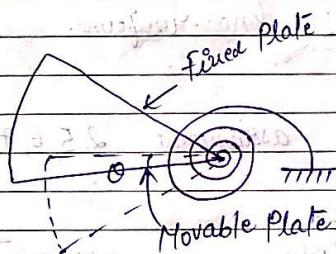
$$f_d = f_c$$

$$Ku = \frac{V^2}{2} \left(\frac{dc}{du} \right)$$

$$u = \frac{V^2}{2K} \left(\frac{dc}{du} \right)$$

Non-uniform.

Case II: Angular Motion



Deflecting Torque

$$T_d = \frac{1}{2} V^2 \frac{dc}{d\theta}$$

Spring control

$$T_c = Ku$$

Balance cont.

$$\Omega = \frac{V^2}{2k} \frac{dC}{d\theta}$$

$$10 \propto V^2$$

Advantage

1. We can measure high voltage using this instrument.
2. It is not affected by stray magnetic fields.

Disadvantage

1. Size is very large.
2. Scale is non-uniform.
3. Costly.

Q. Write assignment 2, 5, 6, 7 (from YB).

21st January

AC Bridge

for finding out unknown inductance, capacitance, frequency.

Types of Sources:

Low frequency :- Power line supply.

High frequency :- Oscillator

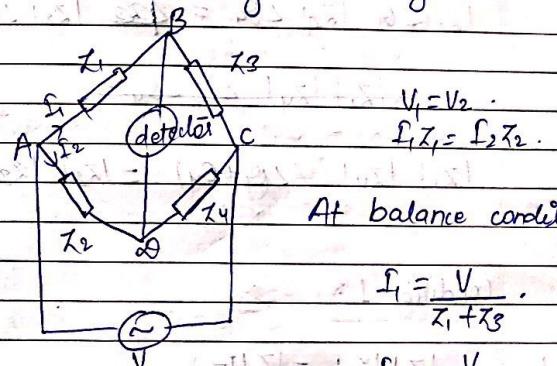
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Types of detector

- 1) Headphone: 250Hz to 4KHz (Sensitive)
- 2) Vibration Galvanometer: 5Hz to 1KHz (Audible range).
- 3) Tunable Amplifier detector: 10Hz to 100KHz.
- 4) CRO (Cathode Ray Oscilloscope): Above 5KHz

Condition for bridge balancing



At balance condition

$$\frac{VZ_1}{Z_1 + Z_3} = \frac{VZ_2}{Z_2 + Z_4}$$

$$\frac{I_1}{1 + \left(\frac{Z_3}{Z_1}\right)} = \frac{I_2}{1 + \left(\frac{Z_4}{Z_2}\right)}$$

$$\frac{I_1}{Z_2} = \frac{I_2}{Z_1} \Rightarrow \frac{Z_1}{Z_2} = \frac{I_2}{I_1} = \frac{Z_4 - Z_3}{Z_2}$$

$$Z_1 Z_4 = Z_2 Z_3$$

condn for bridge balance

$$Z_1 = |Z_1| \angle \theta_1$$

$$Z_2 = |Z_2| \angle \theta_2$$

$$Z_3 = |Z_3| \angle \theta_3$$

$$Z_4 = |Z_4| \angle \theta_4$$

$$|Z_1| \angle \theta_1 |Z_4| \angle \theta_4 = |Z_2| |Z_3| \angle (\theta_2 + \theta_3)$$

$$|Z_1| |Z_4| \angle (\theta_4 - \theta_1) =$$

$$|Z_1| |Z_4| \angle (\theta_4 + \theta_3) = |Z_2| |Z_3| \angle (\theta_2 + \theta_3)$$

Condition 1:

$$1. |Z_1| |Z_4| = |Z_2| |Z_3|$$

Product of magnitude of opp-arm
must be equal

$$2. \angle \theta_4 + \angle \theta_3 = \angle \theta_2 + \angle \theta_1$$

Sum of the far phase angle of
the opp-arm must be equal

Measurement of inductance

1. Maxwell Inductance Bridge

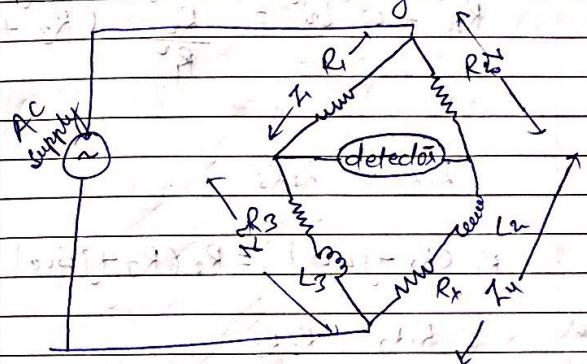
2. " " Capacitance Bridge

3. Hay's Bridge

4. Anderson's Bridge

5. Owen Bridge.

1. Maxwell Inductance Bridge



This bridge is for finding unknown medium inductance.

$$Z_1 = R_1$$

$$Z_4 = R_x + j \omega L_2$$

$$Z_2 = R_2$$

$$Z_3 = R_3 + j \omega L_3$$

$$|Z_1| |Z_4| = |Z_2| |Z_3|$$

$$|R_1| \sqrt{R_2^2 + \omega^2 L_2^2} = |R_2| \sqrt{R_3^2 + \omega^2 L_3^2}$$

$$\sqrt{\frac{R_x^2 + \omega^2 L_x^2}{R_3^2 + \omega^2 L_3^2}} = \frac{R_2}{R_1}$$

$$\frac{R_x^2 + \omega^2 L_x^2}{R_3^2 + \omega^2 L_3^2} = \frac{R_2^2}{R_1^2}$$

$$R_x^2 + \omega^2 L_x^2 = \frac{R_2^2}{R_1^2} (R_3^2 + \omega^2 L_3^2)$$

$$Z_1 Z_4 = Z_2 Z_3$$

$$R_1 (R_r + j\omega L_x) = R_2 (R_3 + j\omega L_3)$$

Real Part

$$R_1 R_x = R_2 R_3$$

$$R_x = \frac{R_2 R_3}{R_1}$$

Imag. Part

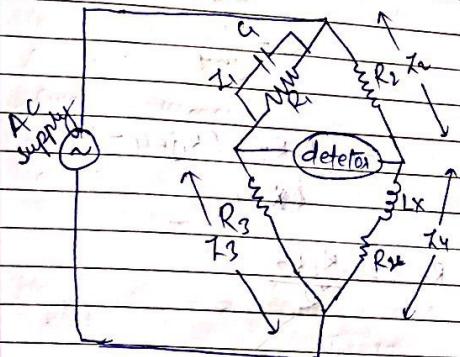
$$R_1 L_x \omega = R_2 \omega L_3$$

$$L_x = \frac{R_2 L_3}{R_1}$$

Quality factor

$$Q = \frac{\omega L_x}{R_x} = \frac{\omega L_3}{R_3}$$

ii) Maxwell Inductance Capacitance Bridge



$$Z_1 = R_1 + j\omega R_1$$

$$Z_1 = R_1 + j\omega Q$$

$$Z_2 = R_2$$

$$Z_3 = R_3$$

$$Z_4 = R_4 + j\omega L_4$$

$$Z_1 = Z_2 + Z_3 + Z_4$$

Bridge Balance

$$Z_1 Z_4 = Z_2 Z_3$$

$$(R_1 + j\omega Q) (R_4 + j\omega L_4) = R_2 R_3$$

$$R_x + j\omega L_x = \frac{R_2 R_3}{R_1 + \frac{1}{j\omega C}}$$

$$= \frac{R_2 R_3 (R_1 j\omega C + 1)}{R_1}$$

$$\text{Real Part } R_x = \frac{R_2 R_3}{R_1}$$

$$\text{Imp. Part } L_x = R_2 R_3 C$$

$$\phi = \frac{\omega L_x}{R_x}$$

$$\phi = \omega C R_1$$

Quality factor of the unknown arm

Medium

$$(\phi = 1 + 0/100)$$

must be in this range

Medium Qef

$R_1 R_4 + \frac{R_1}{j\omega C} + R_4 j\omega L + \frac{L}{C}$

$$= R_2 R_3$$

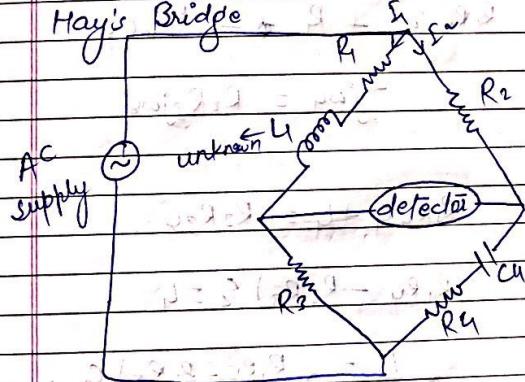
Advantage

- i) It is independent of frequency of excitation.
- ii) It can be measured at audio power freq.

Disadvantage

- i) Quality factor of low and high value is not suitable.
- ii) The capacitor used is very costly.

Hay's Bridge



$$Z_1 Z_4 = Z_2 Z_3$$

$$(R_1 + j\omega L_1) (R_4 + \frac{1}{j\omega C_1}) = R_2 R_3$$

$$R_1 R_4 + \frac{R_1}{j\omega C_1} + R_4 j\omega L_1 + \frac{L_1}{C_1}$$

$$R_1$$

$$(R_1 + j\omega L) \quad (R_4 + \frac{1}{j\omega C}) = R_2 R_3$$

$$R_1 + j\omega L = \frac{R_2 R_3}{R_4 + \frac{1}{j\omega C}}$$

$$= \frac{R_2 R_3 j\omega C}{R_4 j\omega C + 1}$$

$$R_1 R_4 j\omega C + R_1 = -\omega^2 R_4 C^2$$

$$+ j\omega L = R_2 R_3 j\omega C$$

$$R_1 R_4 C + L = R_2 R_3 C$$

$$(R_1 R_4 - R_2 R_3) C = L$$

$$L = (R_1 R_4 - R_2 R_3) C \quad \text{--- (I)}$$

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$$\omega^2 R_4 C^2 = j\omega L \cdot R_1$$

$$+ \omega^2 R_4 C^2 = \frac{L}{R_1} \quad \text{--- (II)}$$

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PAGE / /

$$C = \frac{\omega L}{R}$$

$$= \frac{\omega}{R_4 C}$$

$$= \frac{1}{\omega R_4 C}$$

from (I) & (II)

$$R_1 = (R_1 R_4 - R_2 R_3) C$$

$$R_1 = R_4 C \omega^2 (R_1 R_4 - R_2 R_3)$$

$$C = \frac{R_4 C \omega^2 (R_1 R_4 - R_2 R_3)}{R_4 C \omega^2}$$

from (I) & (II)

$$L = \frac{R_2 R_3 C}{1 + \omega^2 R_4^2 C^2}$$

$$C = \frac{\omega L}{R_1}$$

$$= \frac{\omega R_2 R_3 C}{R_1 + \omega^2 R_4^2 C^2}$$

$$C = \frac{\omega L}{R_1}$$

$$= \frac{\omega}{R_4 C \omega^2}$$

$$= \frac{1}{\omega R_4 C}$$

$$Q = \frac{1}{\omega R_0 C_0}$$

Quality factor depends upon ω , R_0 & C_0

Phasor diagram

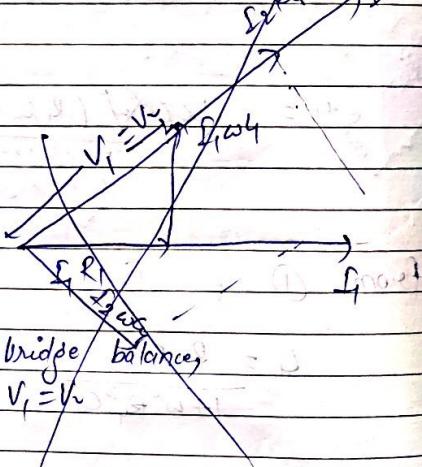
Let V_1 = voltage in arm 1

$$V_2 = " " " "$$

$$V_3 = " " " "$$

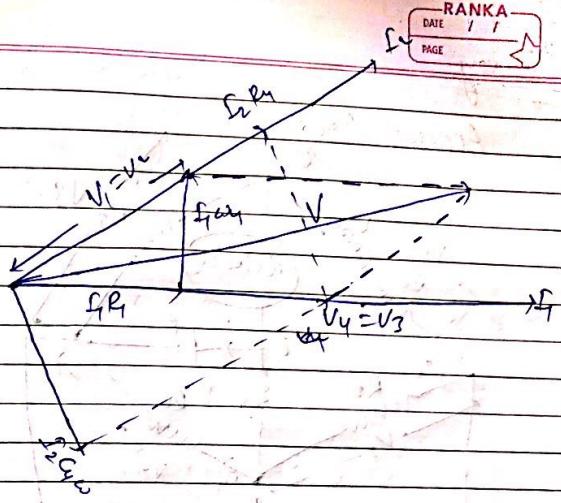
$$V_4 = " " " "$$

$$V_1 = V_2 = V_3 = V_4$$



At bridge balance,

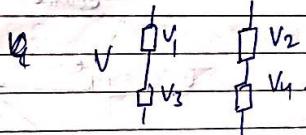
$$V_1 = V_2$$



At bridge balance

$$V_1 = V_2$$

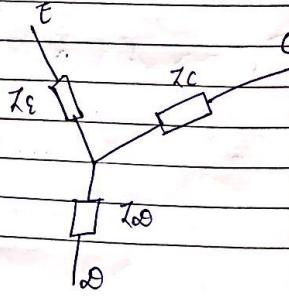
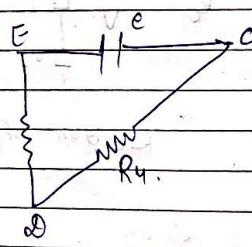
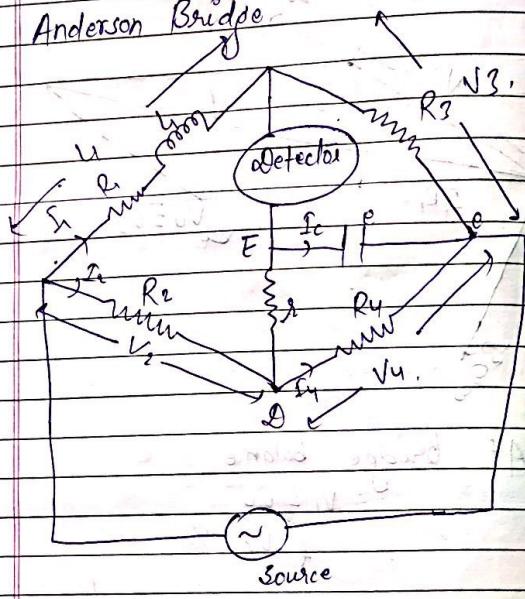
$$\& V_3 = V_4$$



Resultant of $V_1 + V_3$ or
 $V_2 + V_4$ is V .

24th January

Anderson Bridge



$$Z_E = \frac{g \times e}{jwC} \quad \text{where } g = \frac{1}{R_1 + R_2 + R_3}$$

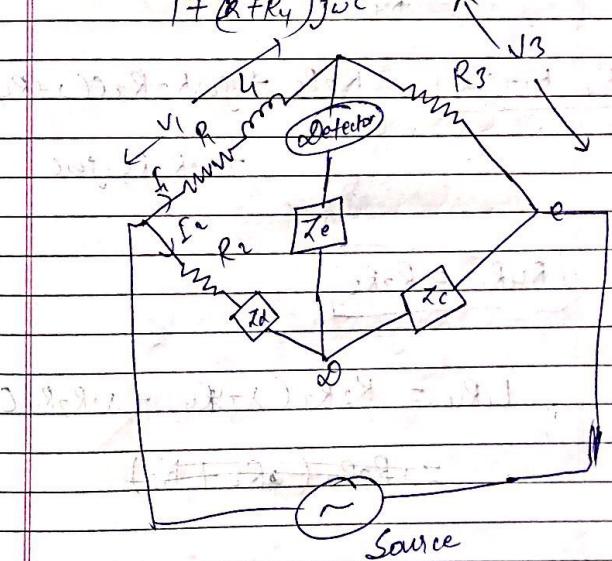
$$Z_C = \frac{R_4 \times e}{jwC} \quad \text{where } R_4 = \frac{1}{jwC + R_4}$$

$$Z_E = \frac{e}{jwCa + 1 + R_4 jwC}$$

$$Z_C = \frac{R_4 jwC}{1 + (R_4 + jwC) jwC}$$

$$Z_D = \frac{g \times R_4}{g + R_4 + 1 + (R_4 + jwC) jwC}$$

$$Z_E = \frac{g \times R_4 jwC}{1 + (R_4 + jwC) jwC}$$



$$R_1 + j\omega L = R_3$$

$$\frac{R_2 + j\omega R_4 \cdot j\omega C}{1 + j\omega C(R_2 + R_4)} = \frac{R_4 \times 1}{j\omega C} + j\omega C + R_4$$

$$(R_1 + j\omega L) (1 + j\omega C(R_2 + R_4))$$

$$R_2 + j\omega R_2 C (1 + R_4) + j\omega R_4 \cdot j\omega C$$

$$= \frac{R_3 j\omega C (1 + j\omega C(1 + R_4))}{R_4}$$

$$R_4 (R_1 + j\omega L) = R_3 R_2 + j\omega R_2 R_3 C (1 + R_4) + j\omega R_2 R_4 \cdot j\omega C$$

$$R_4 R_1 = R_3 R_2$$

$$L_1 R_4 = R_2 R_3 (1 + R_4) + j\omega R_3 R_4 C$$

$$= j\omega R_3 R_4 (j\omega R_2 + R_2)$$

$$L = R_2 R_3 (1 + 1) + j\omega R_3 C$$

$$= R_3 C \left[\frac{j\omega R_2 + 1 + j\omega}{R_4} \right]$$

$$= \frac{R_3 R_2 C}{R_4} + R_3 C + R_3 C$$

$$= \frac{R_4 R_1 C}{R_4} + R_3 C + R_3 C$$

$$= C (R_1 + R_3 + R_3)$$

$$= C [R_3 (1 + 1) + C R_1]$$

$$\vartheta = \frac{\omega L}{R_1}$$

$$= \frac{\omega C (R_1 + R_3 (1 + 1))}{R_1}$$

$$= \omega C + \omega C R_3 (1 + 1)$$

Advantage

1. low ϕ coil \Rightarrow $\omega L \ll 1$
2. low Quality coil $\Rightarrow Q \ll 1$

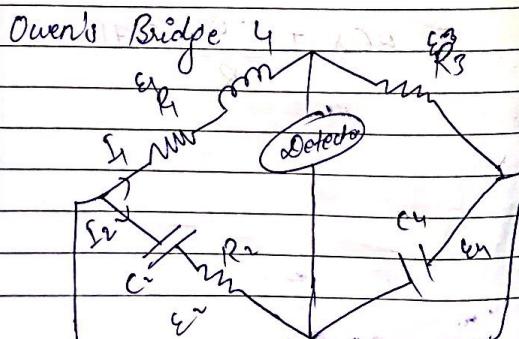
2. Cheaper

3. Accurate estimation of capacitance in terms of inductance is possible.

Disadvantage

1. Complex circuit
2. Balancing the eqn is difficult.

Phasor Diagram



$$\frac{R_1 + j\omega L}{R_2 + \frac{1}{j\omega C_2}} = \frac{R_3}{R_4}$$

$$Y = R_2 R_3 C_4$$

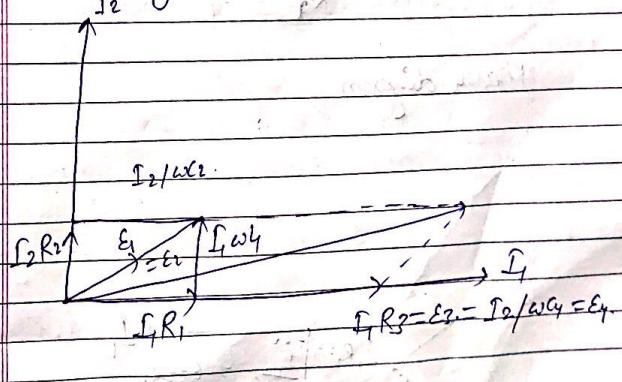
$$R_1 = R_3 \frac{C_4}{C_2}$$

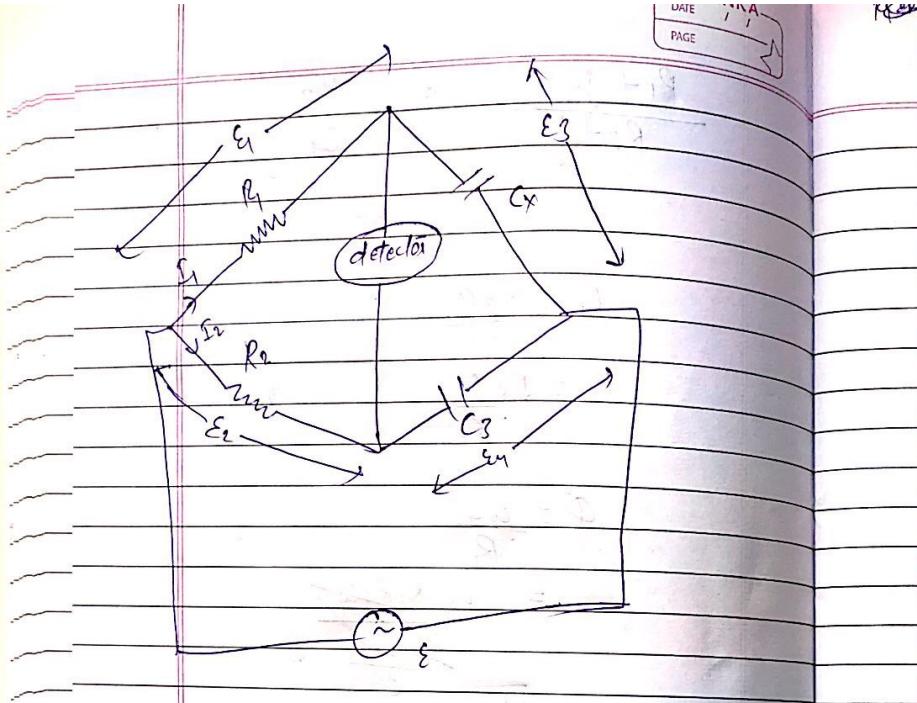
$$Q = \frac{\omega L}{R_1}$$

$$= \frac{\omega R_2 R_3 C_4}{R_2 C_4} = \frac{R_3 C_4}{C_2}$$

$$= \omega R_3 C_4$$

Phasor Diagram





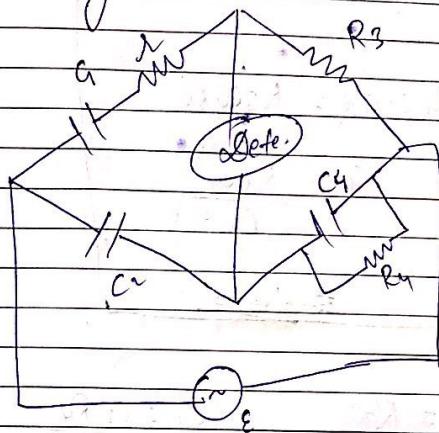
$C_x = \text{Unknown capacitance}$

$$C_x = \frac{R_2 C_3}{R_1}$$

Phasor diagram

$$\frac{\epsilon_1}{R_1} = \frac{\epsilon_2}{\omega C_x} = \frac{\epsilon_3}{R_3} = \frac{\epsilon_4}{\omega C_4}$$

De-Sauty



Balance cond'

$$(s + \frac{1}{j\omega q}) \times \left(\frac{R_4}{R_1} \parallel \frac{1}{j\omega q} \right) = R_3 \times \frac{1}{j\omega c_3}$$

$$(j\omega q + 1) \left(\frac{R_4}{R_1} \parallel \frac{1}{j\omega q} \right) = \frac{R_3}{C_3}$$

$$(1 + j\omega q) \left(\frac{R_4}{1 + j\omega q R_4 q} \right) = \frac{R_3}{C_3}$$

$$R_4 C_3 (1 + j\omega q) = R_3 (1 + j\omega R_4 q)$$

$$R_3 R_4 q = R_4 C_3 q R_3$$

cancel

$$R_4 C_3 = R_3$$

$$R_4 C_3 G_x = R_3 G_y R_4$$

$$G = \frac{R_3 G_y}{C_3 R_x}$$

$$= R_{\text{ext}}$$

$$Z = \frac{R_3 G_y}{C_2} \quad C_1 = \frac{C_2 R_4}{R_3}$$

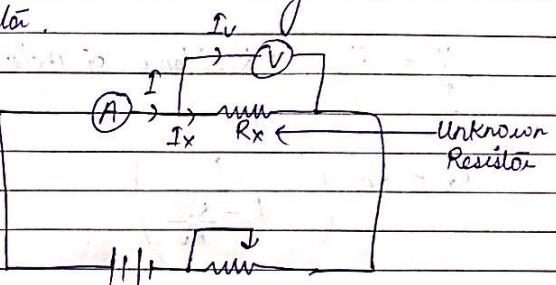
$$Q = \frac{V_C}{V_R} = \frac{1}{\omega R C}$$

4th feb DC Bridge (Resistance).

- | | | |
|---|------------------------|----------------------|
| Low Resistance (<1) | Medium (10 to 100Ω) | High |
| 1. Ammeter / Voltmeter | 1. Ammeter / Voltmeter | 1. Direct deflection |
| 2. Potentiometer | 2. Substitution | 2. Loss of charge |
| 3. Kelvin Bridge | 3. Wheatstone | 3. Megohm |
| 4. Carey-Foster
Slide wire
Bridge | 4. Megger | |

Ammeter/Voltmeter Method

Case 1 Voltmeter is directly connected across resistor.



Current through the ammeter = current through the unknown resistor (R_x) and current through the Voltmeter.

$$I = I_V + I_x \quad R_x = \frac{V}{I_x}$$

$$I_x = I - I_V$$

$$R_x = \frac{V}{I - I_V}$$

R_V = Internal resistance of the voltmeter.

$$I_V = \frac{V}{R_V}$$

$$R_x = \frac{V}{I - \frac{V}{R_V}} = \frac{VR_V}{IR_V - V}$$

$$= \frac{V}{I} \left(\frac{1}{1 - \frac{V}{IR_m}} \right)$$

$$\frac{V}{I} = R_m = \text{Resistance of meter}$$

$$R_x = R_m \left(\frac{1}{1 - \frac{R_m}{R_v}} \right)$$

$$= R_m \left(1 + \frac{R_m}{R_v} \right)^{-1}$$

Binomial Expansion

$$R_x = R_m \left(1 + \frac{R_m}{R_v} \right)$$

[As $R_v \gg R_m$]

Relative Error

$$= \frac{R_m - R_x}{R_x}$$

$$= \frac{R_m - R_m \left(1 + \frac{R_m}{R_v} \right)^{-1}}{R_m \left(1 + \frac{R_m}{R_v} \right)^{-1}}$$

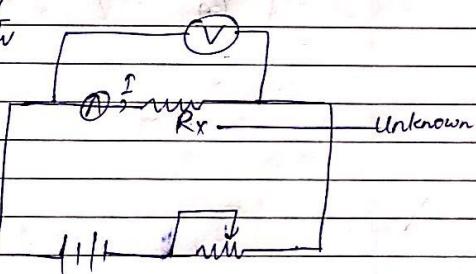
$$= \frac{R_m}{R_v} \cdot \frac{1}{1 + \frac{R_m}{R_v}}$$

$$= \frac{1}{R_v + R_m}$$

Case I Voltmeter is directly connected across
Resistor & Ammeter.

$$I_v = \frac{V}{R_v}$$

R_A = Internal
Resistance
of Ammeter



$$I_x = I + I_v$$

$$\therefore I_v = \frac{V}{R_v}$$

$$V = I R_A + I R_x$$

$$R_A + R_x = \frac{V}{I} = R_m$$

$$R_x = R_m - R_A$$

Relative Euro

$$\frac{R_m - R_x}{R_x}$$

$$= \frac{R_m - R_m + R_A}{R_m - R_A}$$

$$= \left(\frac{1}{\frac{R_m}{R_A} - 1} \right)$$