



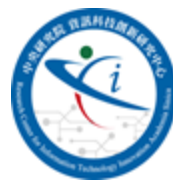
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Intelligent Reflecting Surfaces and Classical Relays: Coexistence and Co-Design

Te-Yi Kan¹, Ronald Y. Chang¹, and Feng-Tsun Chien²

¹Research Center for Information Technology Innovation, Academia Sinica, Taiwan

²Institute of Electronics, National Yang Ming Chiao Tung University, Taiwan



What is *Intelligent Reflecting Surfaces (IRSs)* ?

passive element



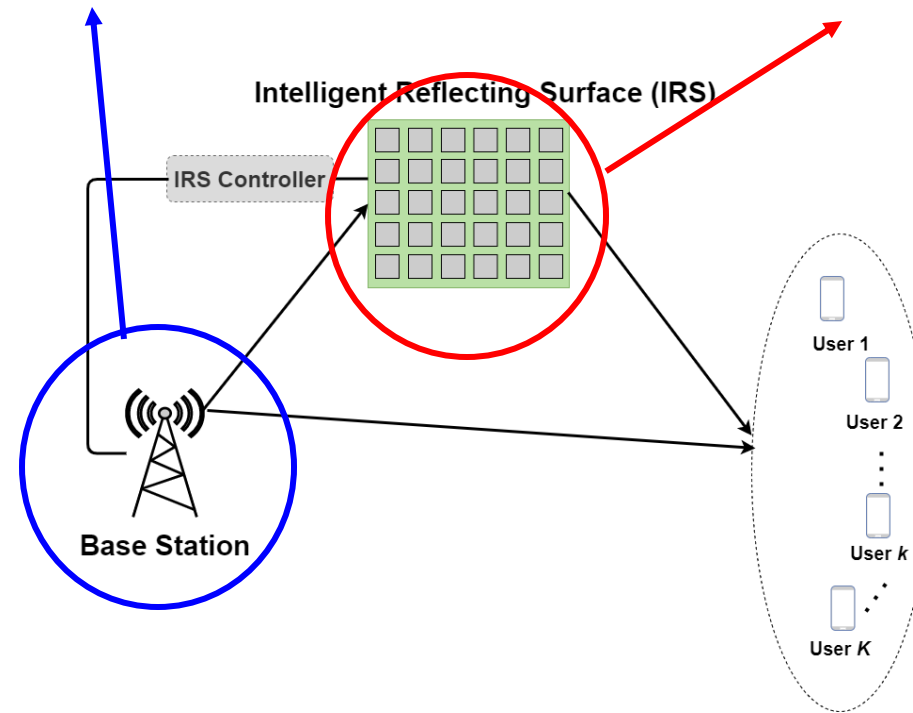
- IRSs employ low-cost ***passive*** elements to ***reflect*** signals.
- IRSs operate in a ***full-duplex*** manner.
- Operating IRSs eases ***signal processing*** and ***interference management*** requirements.



Key Limitations – existing works

Joint Beamforming Design

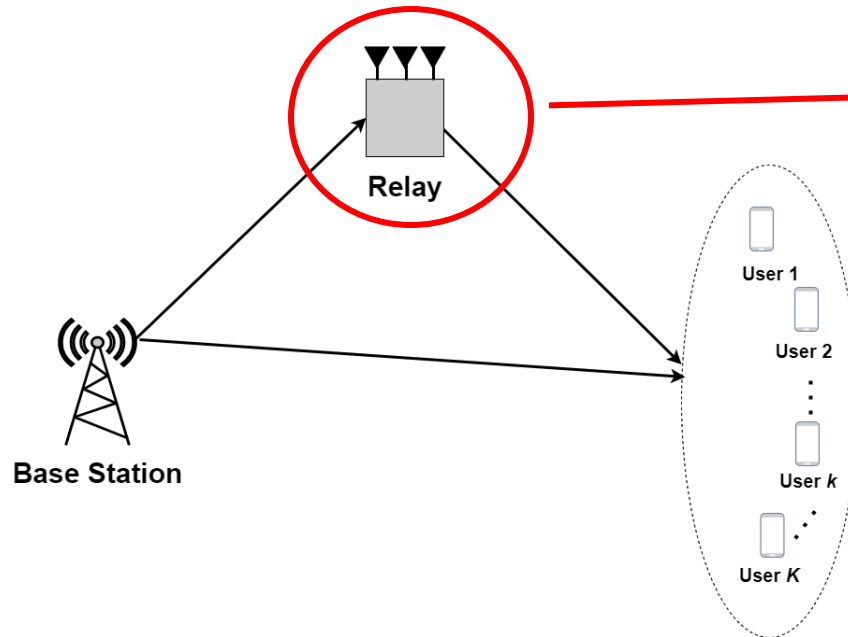
transmit beamforming & IRS phase shifts



Key Limitations – existing works (cont.)

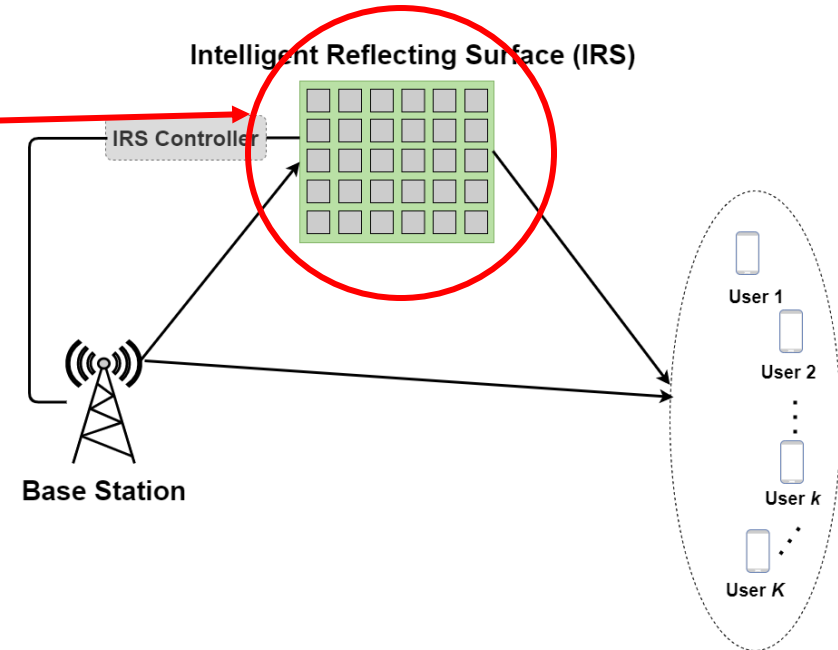
Differences? Similarities?

Relay-assisted system

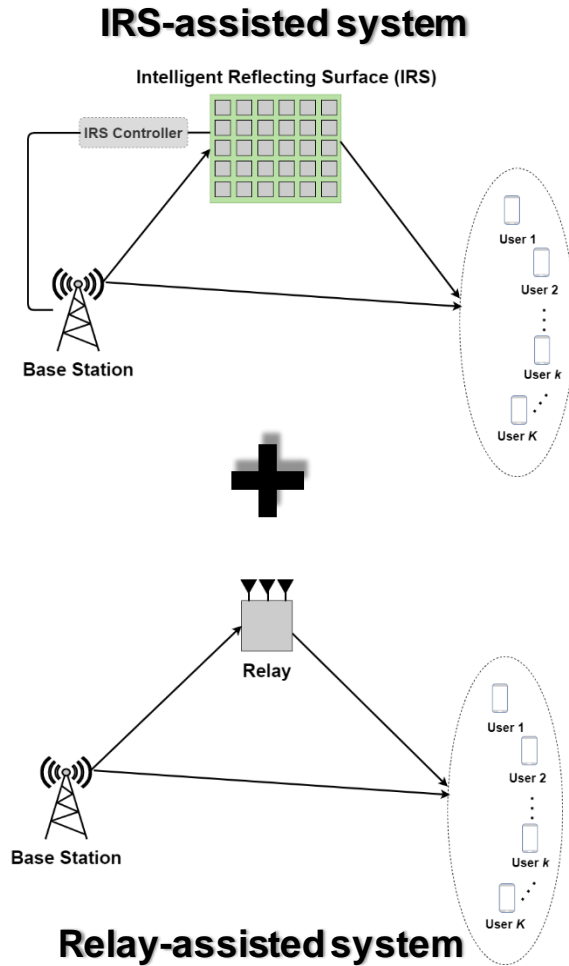


VS

IRS-assisted system



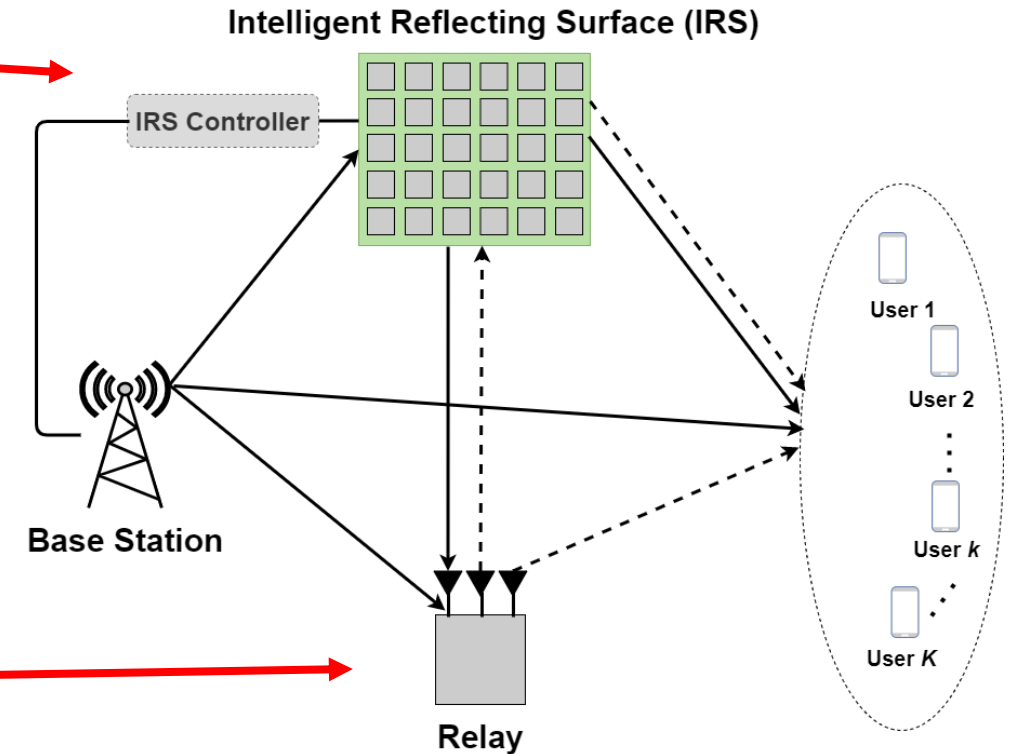
Motivations



*Sounds interesting!
Give it a try!*

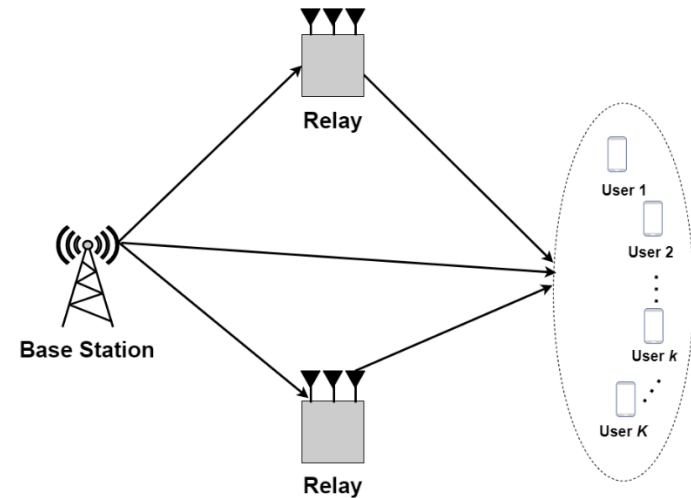


Coexisting IRS-relay assisted system



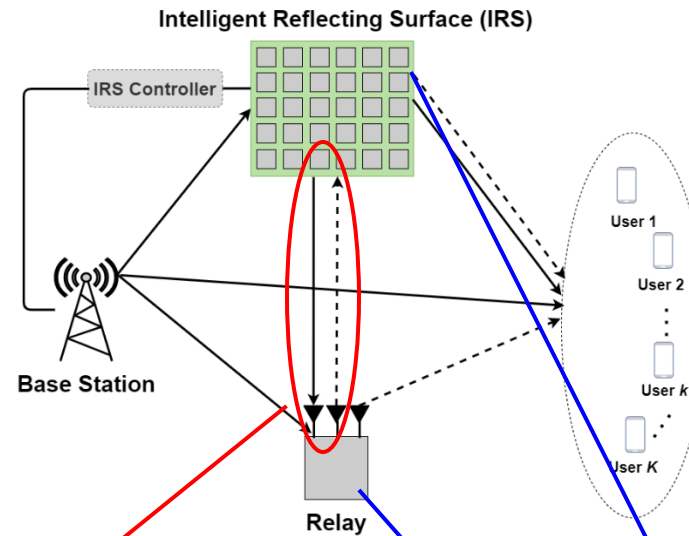
What's new?

multi-relay system



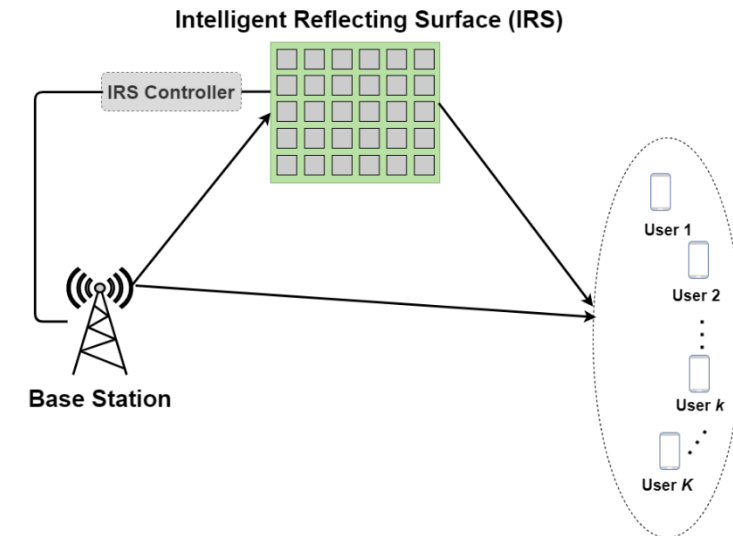
VS

coexistence system



VS

IRS-assisted system

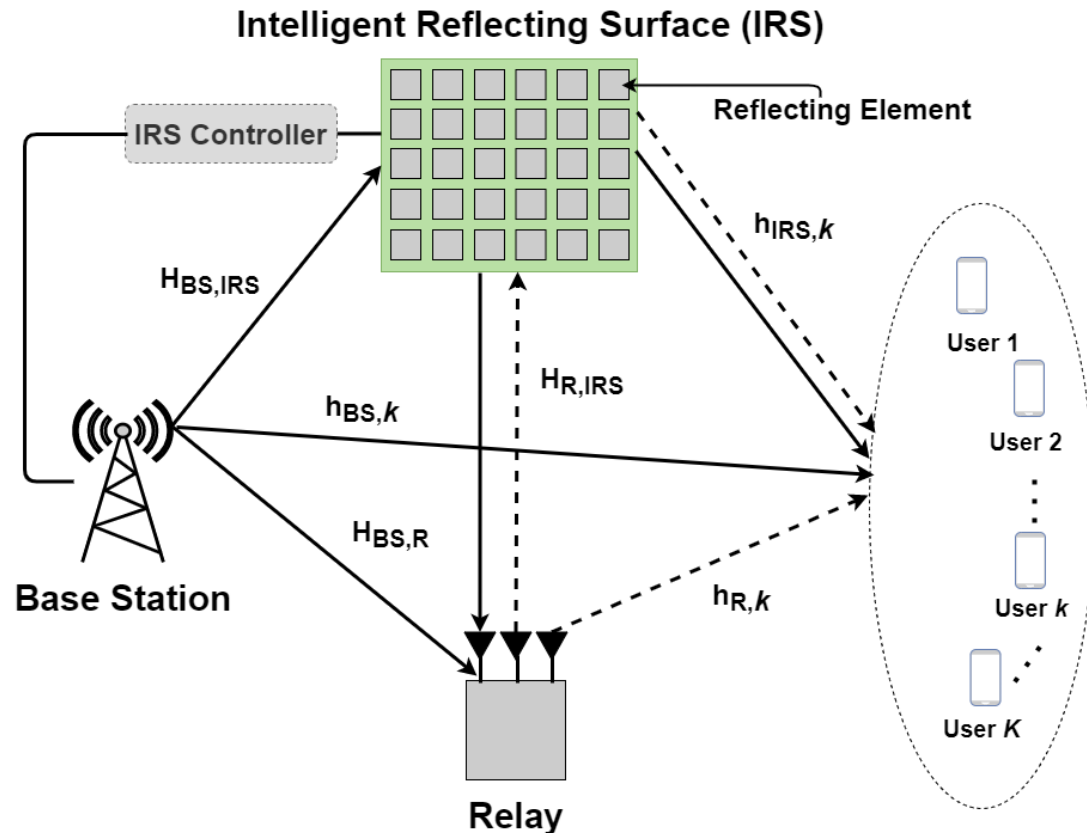


interaction between IRS and relay

*full-duplex relaying
+
half-duplex relaying*



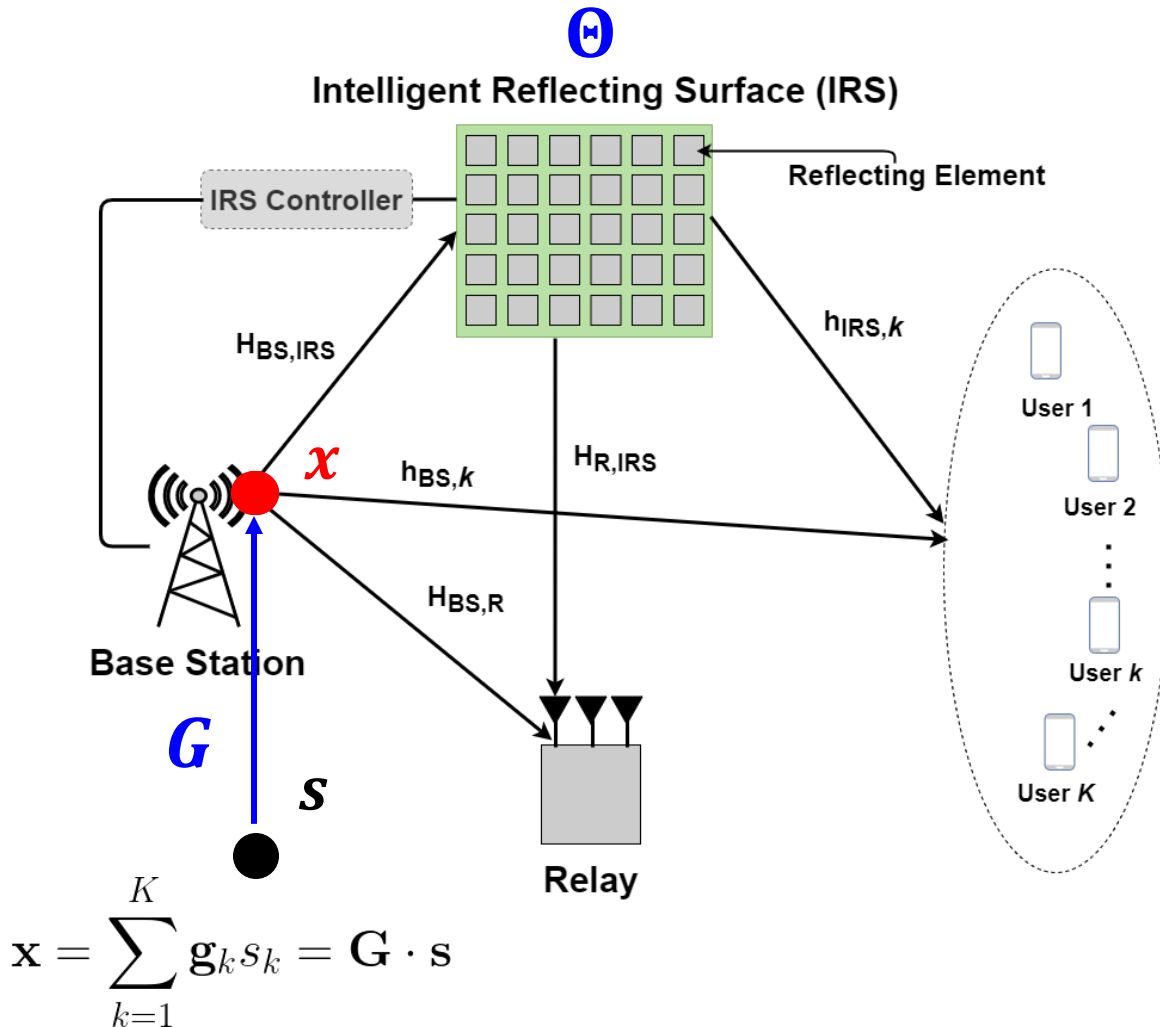
Coexisting IRS and relay assisted system



- Base station (BS) with M antennas
- DF relay with L antennas
- Ideal IRS with N reflecting elements
- K single-antenna end users
- IRS controller



Transmission – Phase I



first phase received signal

$$y_k^I = \mathbf{h}_{\text{IRS},k} \Theta \mathbf{H}_{\text{BS,IRS}} \mathbf{x} + \mathbf{h}_{\text{BS},k} \mathbf{x} + w_k^I$$

IRS-assisted link

direct link

noise

Relay received signal

$$y_R = \mathbf{H}_{\text{R,IRS}}^H \Theta \mathbf{H}_{\text{BS,IRS}} \mathbf{x} + \mathbf{H}_{\text{BS,R}} \mathbf{x} + \mathbf{w}_R$$

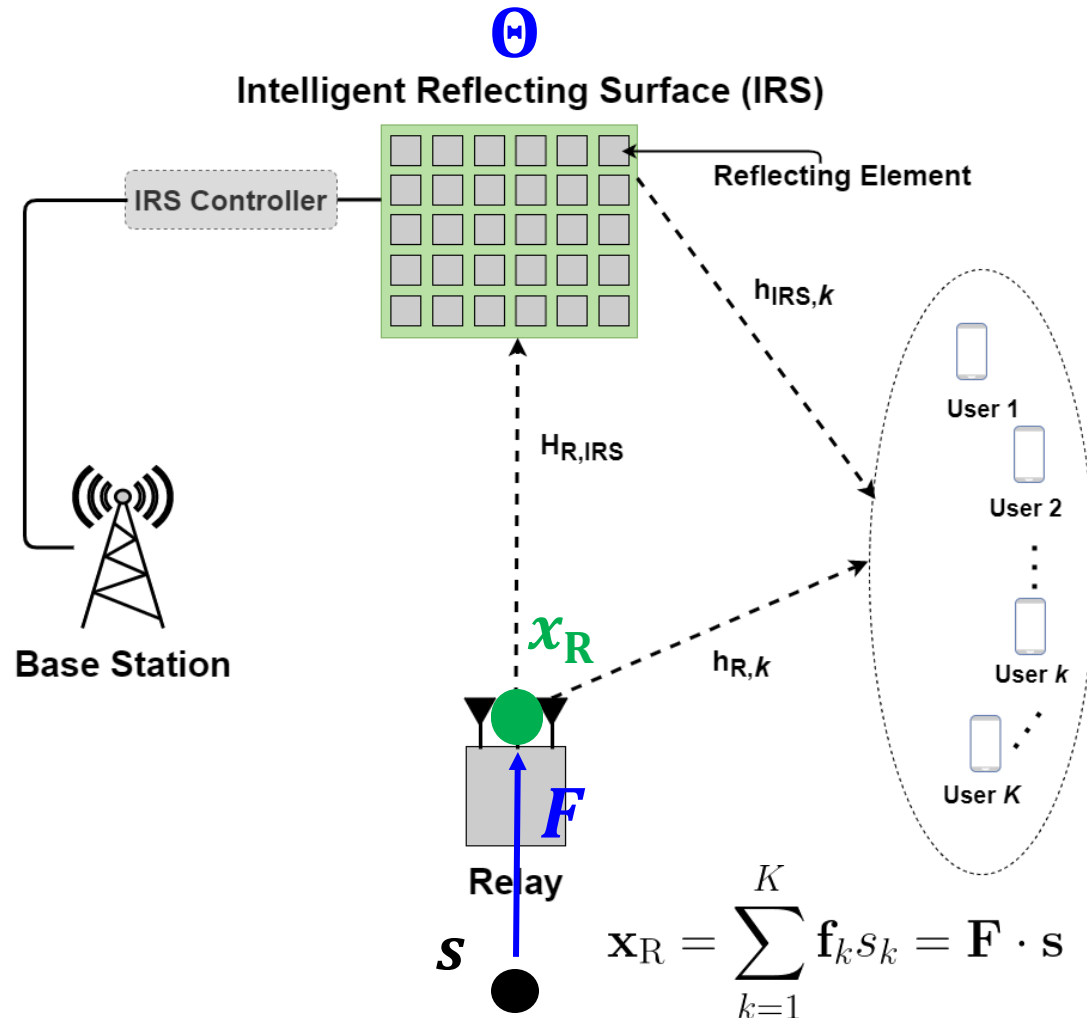
IRS-assisted link

direct link

noise



Transmission – Phase II



Second phase received signal

$$y_k^{\text{II}} = \mathbf{h}_{\text{IRS},k} \mathbf{\Theta} \mathbf{H}_{\text{R,IRS}} \mathbf{x}_R + \mathbf{h}_{\text{R},k} \mathbf{x}_R + w_k^{\text{II}}$$

$$= \mathbf{h}'_{\text{R},k} \mathbf{x}_R + w_k^{\text{II}}$$

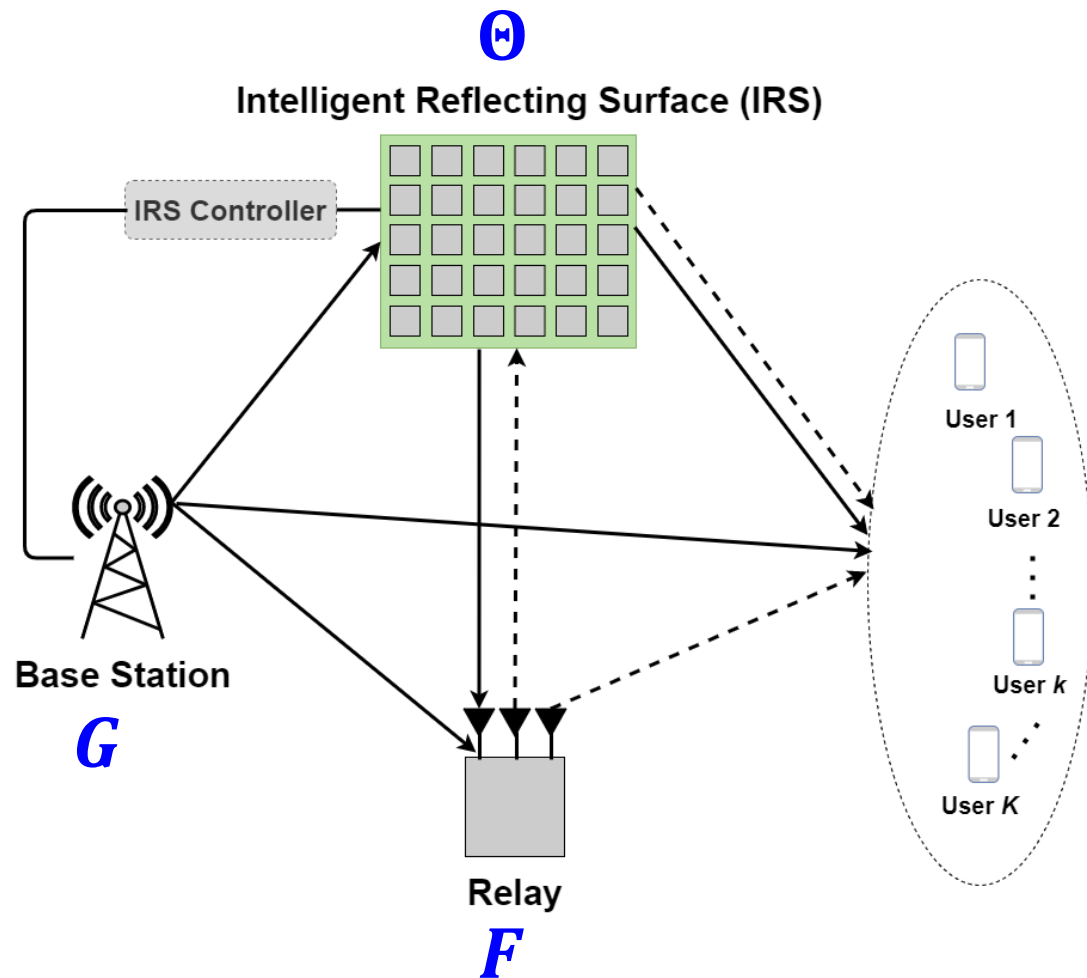
IRS-assisted link

direct link

noise



Joint beamforming design Problem



Sum-rate maximization:

$$\max \sum_{k=1}^K \log_2 (1 + \gamma_k), \quad \gamma_k = \underbrace{\gamma_k^I}_{k\text{-th user's SINR}} + \underbrace{\gamma_k^{\text{II}}}_{\text{first phase SINR}} \quad \leftarrow \text{second phase SINR}$$

Variables:

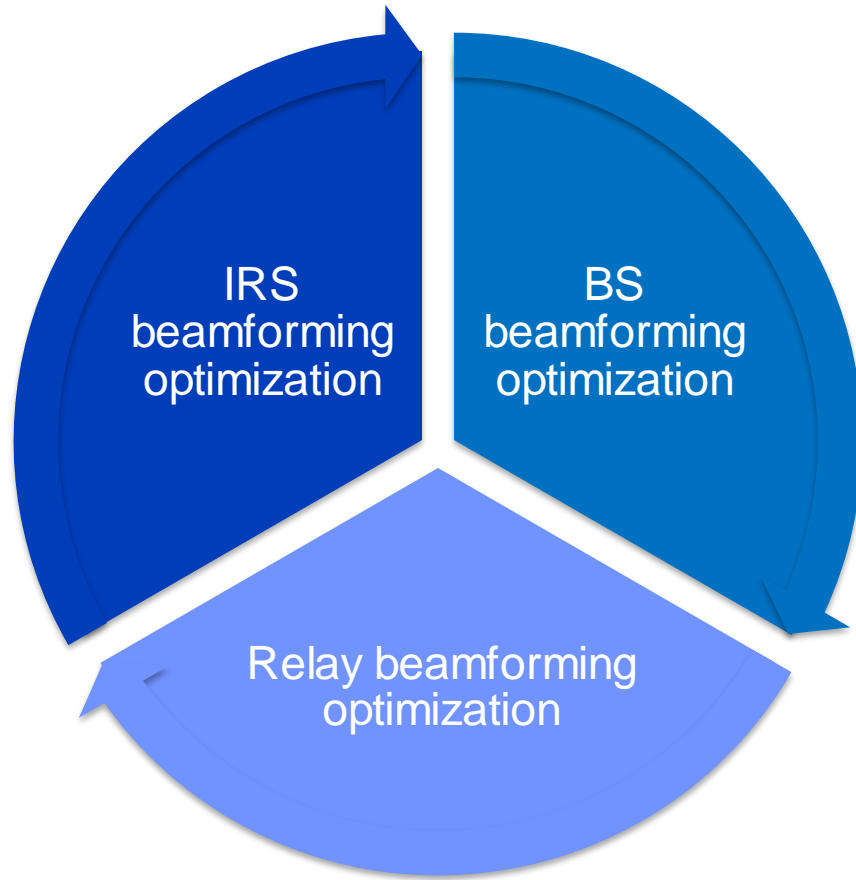
- BS beamformer \mathbf{G}
- Relay beamformer \mathbf{F}
- IRS beamformer Θ

Constraints:

- BS power constraint
- Relay power constraint
- IRS reflection constraints $\text{IRS}_{\text{ideal}} \triangleq \{|\theta_n| \leq 1\}$
- Decoding SINR requirements (matched filter)

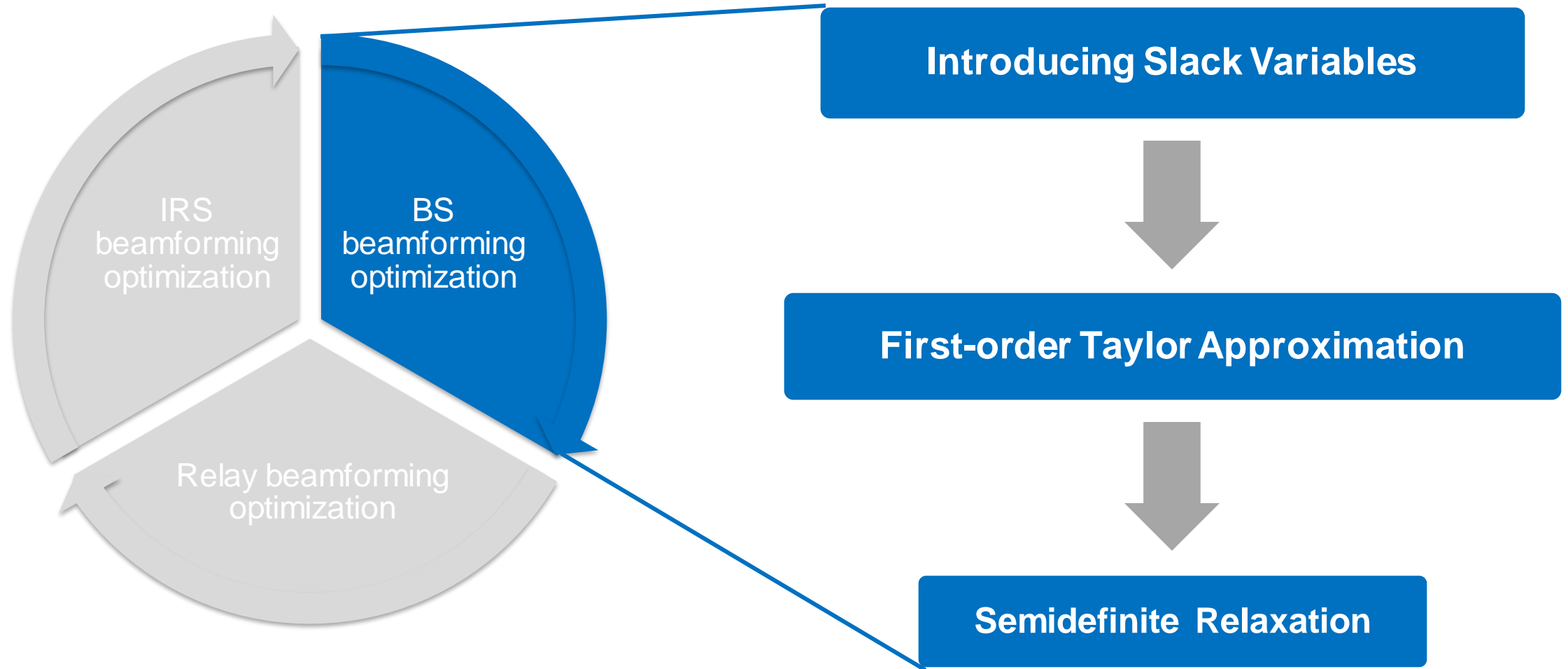


Proposed method



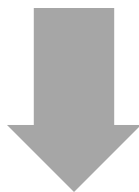
- Alternating optimization (AO)-based method
- Joint beamforming design is decoupled into 3 subproblems
 - BS beamforming
 - ◆ Optimizing BS beamformer \mathbf{G}
 - ◆ BS power constraint & decoding SINR requirements
 - Relay beamforming
 - ◆ Optimizing relay beamformer \mathbf{F}
 - ◆ relay power constraint
 - IRS beamforming
 - ◆ Optimizing IRS beamformer $\boldsymbol{\theta}$
 - ◆ IRS reflection constraints & decoding SINR requirements

BS Beamforming Optimization



Introducing Slack Variables

Introducing Slack Variables



First-order Taylor Approximation



Semidefinite Relaxation

$$\max \sum_{k=1}^K \log_2 \left(1 + \frac{|\mathbf{h}'_{\text{BS},k} \mathbf{g}_k|^2}{\sum_{j=1, j \neq k}^K |\mathbf{h}'_{\text{BS},k} \mathbf{g}_j|^2 + \sigma_k^2} + \gamma_k^{\text{II}} \right)$$



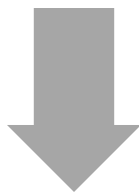
$$\mathbf{G}_k = \mathbf{g}_k \mathbf{g}_k^H$$

$$\max \sum_{k=1}^K \log_2 \left(1 + \frac{\text{tr}(\mathbf{h}'_{\text{BS},k} \mathbf{h}'_{\text{BS},k}^H \mathbf{G}_k)}{\sum_{j=1, j \neq k}^K \text{tr}(\mathbf{h}'_{\text{BS},k} \mathbf{h}'_{\text{BS},k}^H \mathbf{G}_j) + \sigma_k^2} + \gamma_k^{\text{II}} \right)$$



Introducing Slack Variables (cont.)

Introducing Slack Variables



First-order Taylor Approximation



Semidefinite Relaxation

$$\max \sum_{k=1}^K \log_2 \left(1 + \frac{\text{tr}(\mathbf{h}'_{\text{BS},k} \mathbf{h}'_{\text{BS},k} \mathbf{G}_k)}{\sum_{j=1, j \neq k}^K \text{tr}(\mathbf{h}'_{\text{BS},k} \mathbf{h}'_{\text{BS},k} \mathbf{G}_j) + \sigma_k^2} + \gamma_k^{\text{II}} \right)$$



$$\max \sum_{k=1}^K R_{1,k}$$

$$R_{1,k} \leq \log_2 \left(1 + \frac{1}{\mathcal{S}_{1,k} \mathcal{I}_{1,k}} + \gamma_k^{\text{II}} \right)$$

$$\frac{1}{\mathcal{S}_{1,k}} \leq \text{tr}(\mathbf{h}'_{\text{BS},k} \mathbf{h}'_{\text{BS},k} \mathbf{G}_k)$$

$$\mathcal{I}_{1,k} \geq \sum_{j=1, j \neq k}^K \text{tr}(\mathbf{h}'_{\text{BS},k} \mathbf{h}'_{\text{BS},k} \mathbf{G}_j) + \sigma_k^2$$



First-order Taylor Approximation

Introducing Slack Variables

Decoding SINR requirements

$$R_{1,k} \leq \log_2 \left(1 + \frac{1}{\mathcal{S}_{1,k} \mathcal{I}_{1,k}} + \gamma_k^{\text{II}} \right)$$

$$\frac{1}{\mathcal{S}_{R,k} \mathcal{I}_{R,k}} \geq \gamma_R^{\text{th}}$$

Convex function

First-order Taylor Approximation

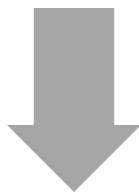
Semidefinite Relaxation

- A convex function is **lower bounded** by its first-order Taylor approximation.
- We **replace convex functions with their first-order Taylor approximation** to convexify constraints.



Semidefinite Relaxation

Introducing Slack Variables



First-order Taylor Approximation



Semidefinite Relaxation

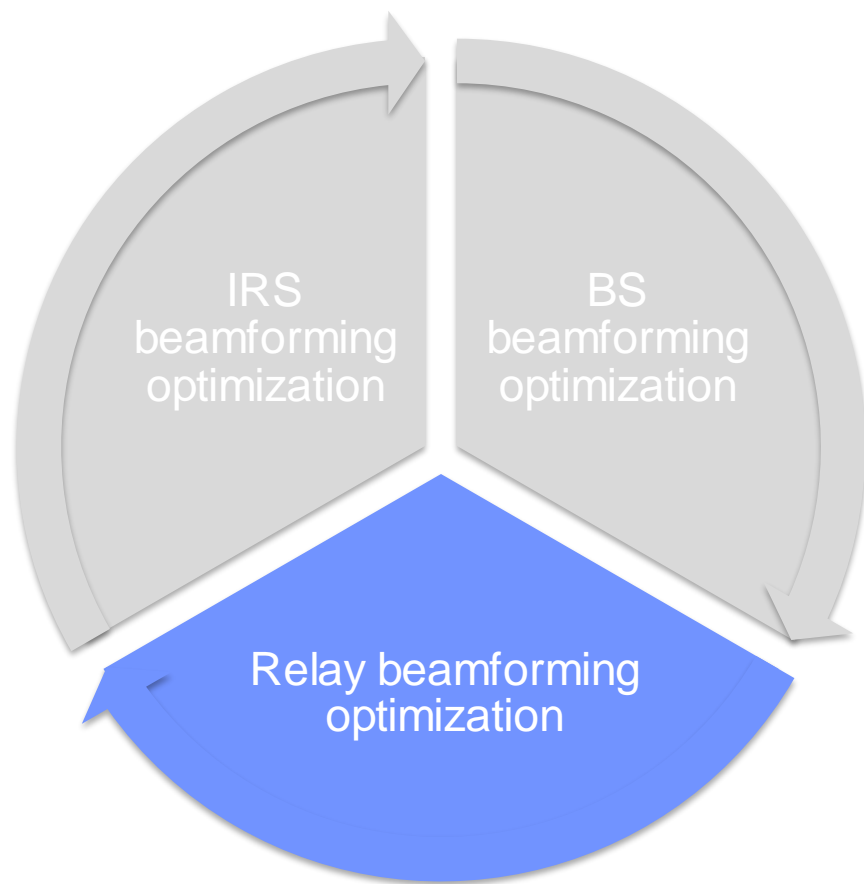
rank-one positive semidefinite matrix

$$\mathbf{G}_k = \mathbf{g}_k \mathbf{g}_k^H$$

- The rank-one constraints are relaxed.
- The resulting SDP problem can be solved by CVX.
- Rank-one solution can be obtained by using the Randomization procedure.



Relay Beamforming Optimization



Introducing Slack Variables



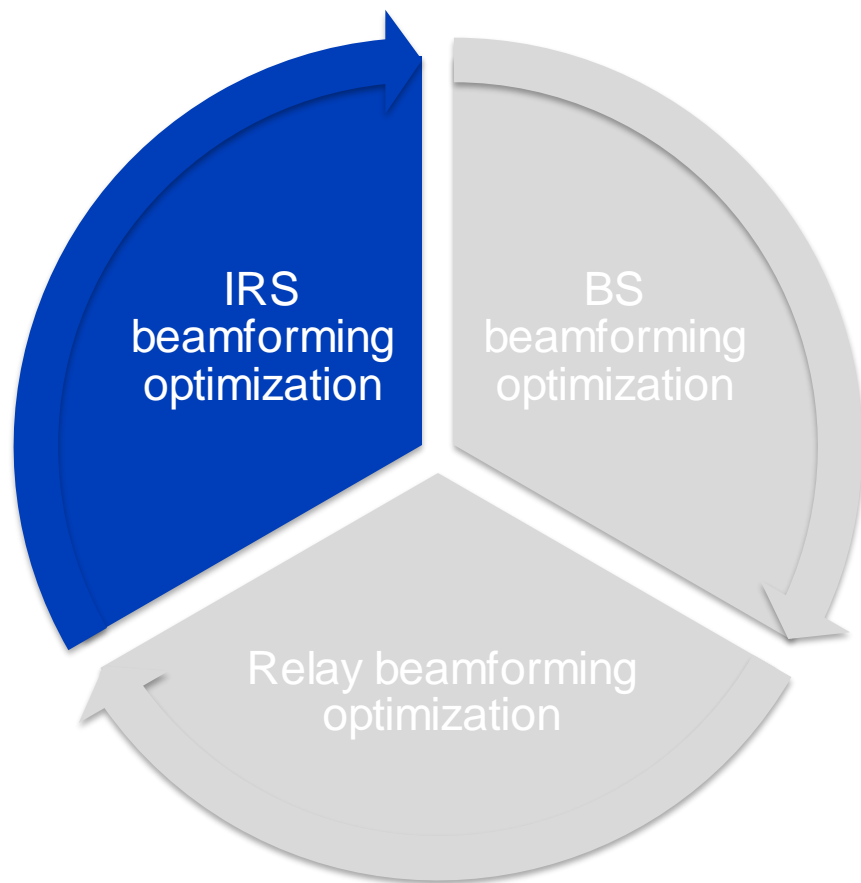
First-order Taylor Approximation



Semidefinite Relaxation



IRS Beamforming Optimization



Introducing Slack Variables



First-order Taylor Approximation

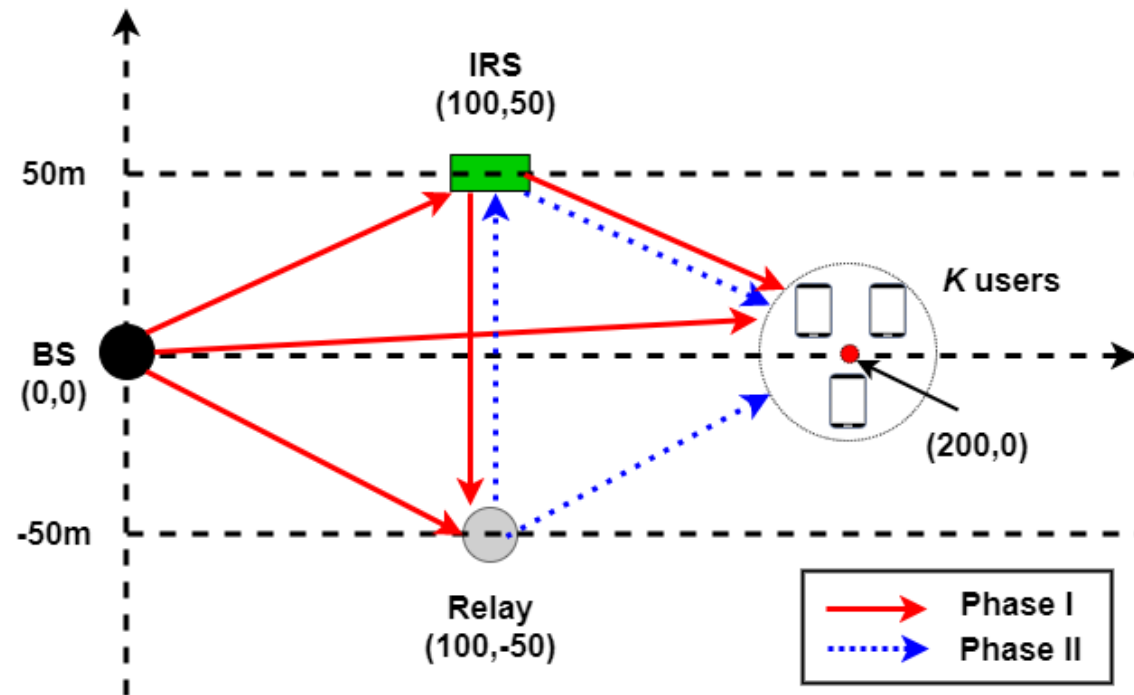


Semidefinite Relaxation



Simulation Setup

System Topology



Channel Model

All channels follow Rayleigh flat-fading

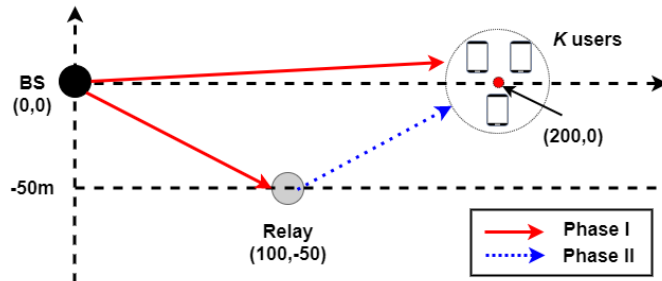
Small-scale fading: $\mathcal{CN}(0, 1)$

Large-scale fading: $\kappa(d/d_0)^{-\varrho}$

	fading constant κ	path-loss exponent ϱ
$\mathbf{H}_{BS,R}$	10^{-4}	3.5
$\mathbf{h}_{BS,k}$	10^{-4}	3.5
$\mathbf{h}_{R,k}$	10^{-4}	3.5
$\mathbf{H}_{BS,IRS}$	$10^{-0.5}$	2
$\mathbf{H}_{R,IRS}$	$10^{-0.5}$	2
$\mathbf{h}_{IRS,k}$	$10^{-0.5}$	2

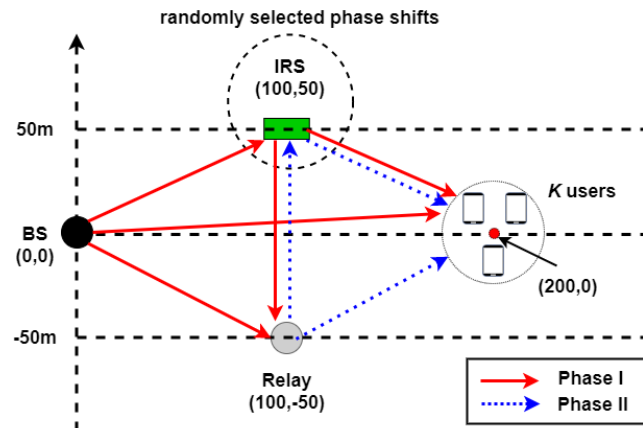
Benchmarks

Relay only scheme



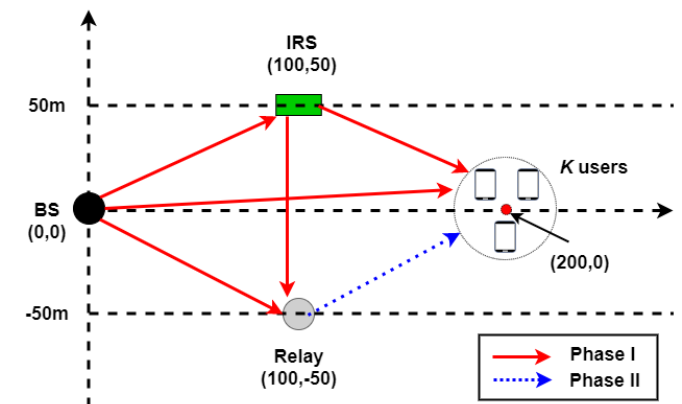
- There is **no IRS**.
- BS and relay beamforming are optimized.

Random scheme



- IRS adopts **random phase shifts** and **fixed amplitudes of one**.
- BS and relay beamforming are optimized.

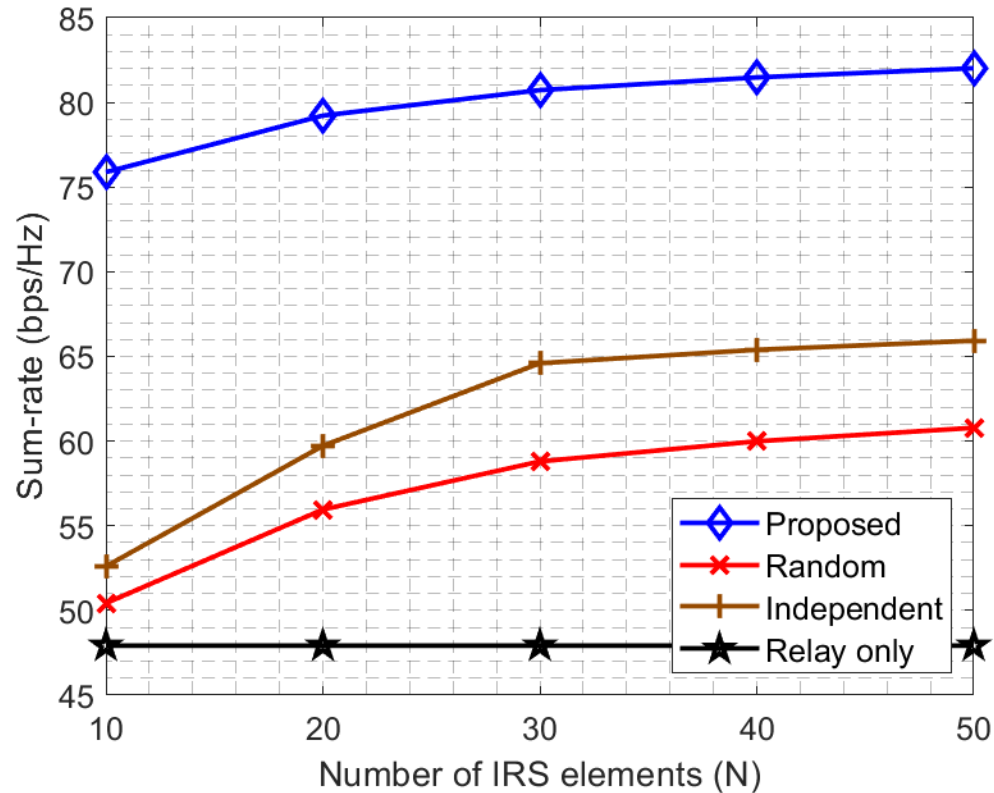
Independent scheme



- IRS is **turned off in the second phase**.
- BS, relay, and IRS beamforming are optimized in this case.



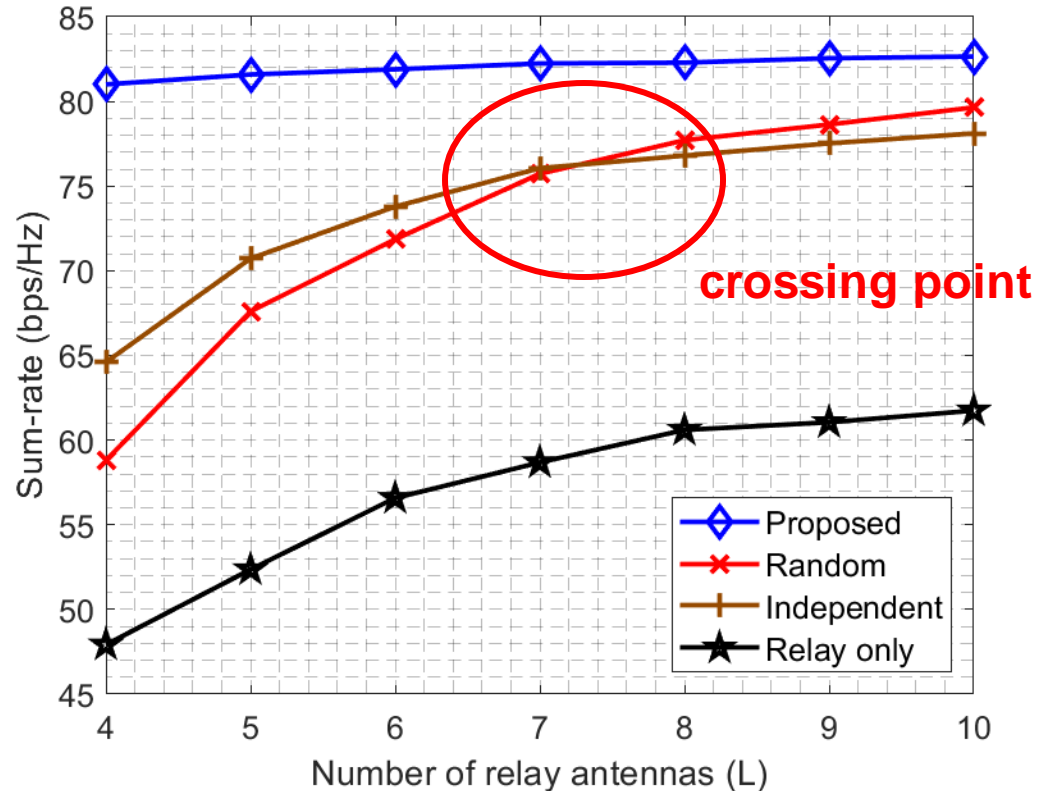
Impact of IRS elements



- The sum-rate of all IRS-assisted schemes increases as N increases.
- The proposed scheme provides the highest sum-rate.
- The Independent scheme achieves better performance than the Random scheme.
- The IRS-assisted schemes outperform the Relay only scheme.

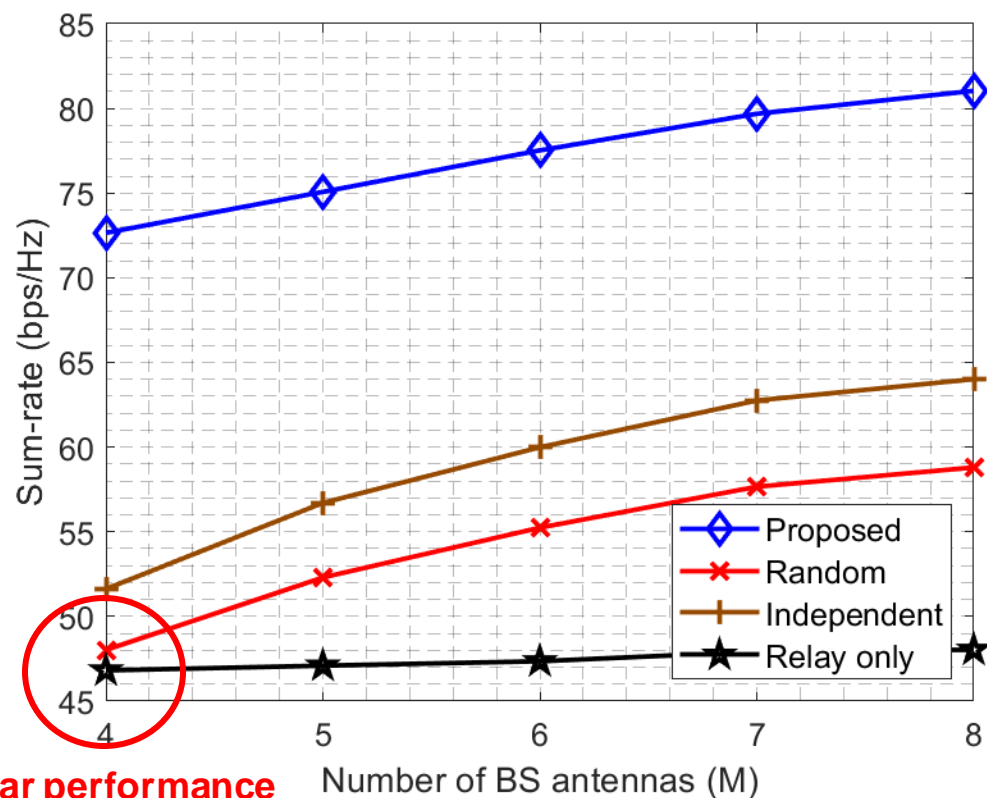


Impact of relay antennas



- The sum-rates of all schemes increases with L .
- The performance gains of the proposed scheme **diminish** as L increases.
- The independent and Random schemes exhibit a **crossing point**.

Impact of BS antennas



- The sum-rate of all schemes increases with M .
- The effect of M on sum-rate is **moderate** for the Relay only scheme.
- The Independent scheme achieves better performance than the Random scheme.
- The Random scheme achieves only a **small performance gain** in the small M region.



Key Takeaways

1. We have presented a joint beamforming design for a multiuser MISO system in which **an IRS and a decode-and-forward relay coexist and assist downlink transmission simultaneously.**
2. The proposed scheme significantly improves the sum-rate performance of end users by **jointly optimizing BS beamforming, relay beamforming, and IRS beamforming.**
3. **The interaction between the IRS and relay** brings great merits.
4. The coexistence of an IRS and a relay **requires a proper co-design** to benefit communications.



Thanks for your attention!

Q&A



Te-Yi Kan¹, Ronald Y. Chang¹, and Feng-Tsun Chien²

¹Research Center for Information Technology Innovation (CITI), Academia Sinica, Taiwan

²Institute of Electronics, National Yang Ming Chiao Tung University (NYCU), Taiwan

E-mail: [tyk, rchang]@citi.sinica.edu.tw, and ftchien@mail.nctu.edu.tw

