## **Statistics**

#### 1.1 Basics

- $E(X) = \int x f(x)$  and  $Var(X) = E((X \mu)^2)$ .
- Sample mean:  $\overline{X} = \frac{1}{n} \sum X_i$ .
- Sample variance:  $S^2 = \frac{1}{n-1} \sum (X_i \overline{X})^2$ .
- Likelihood: Let  $X_i$  have joint pdf  $f(\mathbf{x}; \theta)$ . Given observed values  $x_i$  of  $X_i$ , the *likelihood* of  $\theta$  is  $L(\theta) = f(\mathbf{x}; \theta)$ . The maximum likelihood estimate  $\hat{\theta}(\mathbf{x})$  is the value of  $\theta$  that maximizes  $L(\theta)$ .

## 1.2 Convergence of Random Variables

RVs  $X_n$  with cdf  $F_n$ .

•  $X_n$  converges in distribution to X (weakly) if

$$\lim F_n(x) = F(x)$$

for all x at which F is continuous.

•  $X_n$  converges in probability to X if for all  $\epsilon > 0$ ,

$$\lim P(|X_n - X| > \epsilon) = 0.$$

•  $X_n$  converges almost surely to X (strongly) if

$$P\left(\lim X_n = X\right) = 1,$$

i.e., events for which  $X_n$  does not converge to X have probability 0.

#### 1.3 Parameter Estimation

An estimator is any statistic  $\hat{\theta} = \hat{\theta}(X)$  used to estimate  $\theta$ .

### 1.3.1 Properties of estimators

#### Mean squared error

•  $MSE(\hat{\theta}) = E\left[(\hat{\theta} - \theta)^2\right].$ 

#### Variance

•  $\operatorname{Var}(\hat{\theta}) = \operatorname{E}\left[(\hat{\theta} - \operatorname{E}(\hat{\theta}))^2\right].$ 

#### Bias

• Bias( $\hat{\theta}$ ) = E( $\hat{\theta}$ ) –  $\theta$ .

• Note that  $MSE(\hat{\theta}) = Var(\hat{\theta}) + Bias(\hat{\theta})^2$ .

• The unbiased estimator with the smallest variance is known as the *minimum-variance unbiased estimator* (MVUE).

#### Consistency

• An estimator  $T_n$  of  $\theta$  is weakly consistent if  $T_n$  converges in probability to  $\theta$ .

•  $T_n$  is strongly consistent if  $T_n$  converges almost surely to  $\theta$ .

#### Asymptotic normality

•  $T_n$  is asymptotically normal if

$$\sqrt{n}(T_n - \theta) \xrightarrow{D} N(0, V).$$

• Recall: CLT says that sample mean  $\overline{X}$  is asymptotically normal.

#### Efficiency

• The observed information is

$$J(\theta) = -\frac{d^2l}{d\theta^2}$$

for scalar parameter  $\theta$  or

$$J(\theta)_{ij} = -\frac{\partial^2 l}{\partial \theta_i \partial \theta_j}$$

for  $\theta = (\theta_1, \dots, \theta_p)$ .

• The larger  $J(\hat{\theta})$  is, the more concentrated  $l(\theta)$  is about  $\hat{\theta}$ .

• The expected or Fisher information is

$$I(\theta) = E\left(-\frac{d^2l}{d\theta^2}\right)$$

or

$$I(\theta)_{ij} = \mathrm{E}\left(-\frac{\partial^2 l}{\partial \theta_i \partial \theta_j}\right).$$

• Equivalently, define score to be gradient of  $l(\theta)$ , i.e.,

$$s(\theta) = \frac{\partial l}{\partial \theta}.$$

Then expected value of score at  $\theta$  is 0, i.e.,  $E(s \mid \theta) = 0$ . The Fisher information is defined to be the variance of the score  $Var(s(\theta)) = E(s(\theta)s(\theta)^T)$ .

• The Cramer-Rao bound says that for an unbiased estimator,

$$\operatorname{Var}(\hat{\theta}) \ge \frac{1}{I(\theta)}.$$

Generally, if  $E(\hat{\theta}) = \psi(\theta)$ , then

$$\operatorname{Var}(\hat{\theta}) \ge \frac{(\psi'(\theta))^2}{I(\theta)} = \frac{(1+b'(\theta))^2}{I(\theta)},$$

where b is the bias. In the unbiased multivariate case, the Cramer-Rao bound states that

$$cov(T(X)) \ge I(\theta)^{-1}$$
,

where  $M \geq 0$  means that M is positive semidefinite, i.e.,  $\mathbf{x}^T M \mathbf{x} \geq 0$  for all  $\mathbf{x}$ .

ullet The efficiency of an unbiased estimator is

$$e(T) = \frac{1/I(\theta)}{\text{Var}(T)}.$$

By the Cramer-Rao bound, we know  $e(T) \leq 1$ . An estimator is *efficient* if e(T) = 1 for all  $\theta$ , i.e., achieves Cramer-Rao bound for all  $\theta$ . Thus, an efficient estimator is the MVUE (but not necessarily conversely).

#### Properties of MLEs

Let  $\hat{\theta}$  be the MLE of  $\theta$ . Then

- $\sqrt{n}(\hat{\theta} \theta) \xrightarrow{D} N(0, I(\theta)^{-1})$ , so  $\hat{\theta}$  is asymptotically unbiased and asymptotically efficient.
- $\hat{\theta} \stackrel{(}{\to} P)\theta$ , i.e.,  $\hat{\theta}$  is consistent.
- Invariance property: the MLE of  $g(\theta)$  is  $g(\hat{\theta})$ .

## 1.4 Hypothesis Testing

 $X_i$  random sample from  $f(x;\theta)$ .

- Setup:
  - Null hypothesis  $H_0: \theta \in \Theta_0$
  - Alternative hypothesis  $H_1: \theta \in \Theta_1$
- Construct test statistic  $t(\mathbf{X})$  such that large values of  $t(\mathbf{X})$  cast doubt on  $H_0$ .

- Let  $t_{\text{obs}} = t(\mathbf{x})$ .
- The *p-value* or *significance level* is

$$p = P(t(\mathbf{X}) \ge t_{\text{obs}} \mid H_0).$$

- Small p-value = observed values unlikely under  $H_0$ .
- Can define critical region to be C such that we reject  $H_0$  if and only if  $\mathbf{x} \in C$ .

#### 1.4.1 Errors in hypothesis testing

Two types of error:

- Type I error: rejecting  $H_0$  when  $H_0$  is true
- Type II error: not rejecting  $H_0$  when  $H_0$  is false.

The *size* of a test is defined by

$$\alpha = P(\text{type I error})$$
  
=  $P(\text{reject } H_0 \mid H_0 \text{ true}).$ 

The *power* of a test is  $1 - \beta$ , where

$$\beta = P(\text{type II error})$$
  
=  $P(\text{don't reject } H_0 \mid H_1 \text{ true}).$ 

We want small size and high power.

For composite  $H_i$ , define the size

$$\alpha = \sup \theta \in \Theta_0 P(\text{reject } H_0 \mid \theta)$$

and the power function

$$w(\theta) = P(\text{reject } H_0 \mid \theta).$$

We want  $w \approx 0$  for  $\theta \in \Theta_0$  and  $w \approx 1$  for  $\theta \in \Theta_1$ .

#### Student's t-test

Consider statistic

$$t = \frac{\overline{x} - \mu_0}{s / \sqrt{n}},$$

which has t-distribution with n-1 degrees of freedom. Recall that a t-distribution is given by

$$T = \frac{Z}{\sqrt{V/\nu}},$$

where Z is standard normal and V is  $\chi^2$  with  $\nu$  degrees of freedom.

#### Noncentral t-distribution

- Noncentral distributions describe how a test statistic is distributed when the null hypothesis is false.
- Noncentral t-distribution is given by

$$T = \frac{Z + \mu}{\sqrt{V/\nu}},$$

where  $\mu$  is the noncentrality parameter.

• Under the alternative hypothesis, we get noncentral t-distribution with n-1 degrees of freedom and noncentrality parameter

$$\delta = \frac{\mu - \mu_0}{\sigma / \sqrt{n}}.$$

This allows us to set the power as

Power = 
$$P(\text{reject } H_0 \mid H_1)$$
.

Note that as n increases, power increases.

### 1.5 Bayesian Inference

- Unknown parameters are random variables.
- Probability model  $f(\mathbf{x} \mid \theta)$  that is conditional on the value of  $\theta$ .
- Prior density  $\pi(\theta)$ .
- After using observed data  $\mathbf{x}$ , get posterior density  $\pi(\theta \mid \mathbf{x})$ .
- We have

$$\pi(\theta \mid \mathbf{x}) \propto f(\mathbf{x} \mid \theta) \times \pi(\theta)$$
  
posterior  $\propto$  likelihood  $\times$  prior

#### 1.5.1 Prediction

- $X_{n+1}$  future observation and  $\mathbf{x} = (x_1, \dots, x_n)$  observed data.
- Assume, conditional on  $\theta$ , that  $X_{n+1}$  has density  $f(x_{n+1} \mid \theta)$  independent of  $X_i, \ldots, X_n$ .
- The density of  $X_{n+1}$  given  $\mathbf{x}$ , called the *posterior predictive density*, is a conditional density denoted  $f(x_{n+1} \mid \mathbf{x})$ .
- Then

$$f(x_{n+1} \mid \mathbf{x}) = \int f(x_{n+1}, \theta \mid \mathbf{x}) d\theta$$
$$= \int f(x_{n+1} \mid \theta, \mathbf{x}) \pi(\theta \mid \mathbf{x}) d\theta$$
$$= \int f(x_{n+1} \mid \theta) \pi(\theta \mid \mathbf{x}) d\theta$$

#### 1.5.2 Credible interval

- Bayesian alternative to confidence interval.
- A  $100(1-\alpha)\%$  credible set for  $\theta$  is  $jc \subset \Theta$  such that

$$P(\theta \in C \mid \mathbf{x}) = \int_C \pi(\theta \mid \mathbf{x}) d\theta = 1 - \alpha.$$

### 1.5.3 Hypothesis testing and Bayes factors

#### Setup

- Prior probabilities  $P(H_i)$  such that  $P(H_0) + P(H_1) = 1$ .
- A prior for  $\theta_i$  under  $H_i$ , denoted  $\pi(\theta_i \mid H_i)$ .
- A model for data **x** under  $H_i$ , denoted  $f(\mathbf{x} \mid \theta_i, H_i)$ .

#### **Bayes factor**

• Prior odds of  $H_0$  relative to  $H_1$  is

$$\frac{P(H_0)}{P(H_1)}$$

• Posterior odds of  $H_0$  relative to  $H_1$  is

$$\frac{P(H_0 \mid \mathbf{x})}{P(H_1 \mid \mathbf{x})}.$$

• Using Bayes' Theorem, get

$$\frac{P(H_0 \mid \mathbf{x})}{P(H_1 \mid \mathbf{x})} = \frac{P(\mathbf{x} \mid H_0)}{P(\mathbf{x} \mid H_1)} \times \frac{P(H_0)}{P(H_1)}$$

posterior odds = Bayes factor  $\times$  prior odds

where the Bayes factor of  $H_0$  relative to  $H_1$  is

$$B_{01} = \frac{P(\mathbf{x} \mid H_0)}{P(\mathbf{x} \mid H_1)}.$$

• The quantity

$$P(\mathbf{x} \mid H_i) = \int_{\Theta_i} f(\mathbf{x} \mid \theta_i, H_i) \pi(\theta_i \mid H_i) d\theta$$

is called the marginal likelihood for  $H_i$ .

# Linear Regression

• Observations

$$y_i = \beta_0 + \beta_1 x_{i1} + \dots + \beta_p x_{ip} + \epsilon_i = x_i \cdot \beta + \epsilon_i$$

i.e.,  $y = X\beta + \epsilon$ .

• Loss function

$$L(\beta) = \text{MSE}(\beta) = \frac{1}{n} ||y - X\beta||^2.$$

- Want  $\hat{\beta}$  that minimizes  $L(\beta)$ :
  - Find  $\hat{\beta}$  such that

$$\frac{\partial L}{\partial \beta} = 0.$$

- Project y onto Col(X).

Get  $\hat{\beta} = (X^T X)^{-1} X^T y$ . For simple linear regression, i.e.  $y = \beta_0 + \beta_1 x + \epsilon$ , we get

$$\hat{\beta}_0 = \overline{y} - \hat{\beta}_1 \overline{x},$$

$$\hat{\beta}_1 = \frac{\text{cov}(x, y)}{\text{Var}(x)} = \frac{\sum (x_i - \overline{x})(y_i - \overline{y})}{\sum (x_i - \overline{x})^2}.$$

- Residual is  $\hat{\epsilon} = y X\hat{\beta}$ . The residual/explained/total sum of squares is:
  - RSS =  $\|\hat{e}\|^2$ , i.e., variation in the error between the observed data and modeled values.
  - ESS =  $\sum (\hat{y}_i \overline{y}_i)$ , i.e., how much variation there is in the modeled values.
  - TSS =  $\sum (y_i \overline{y})$ , i.e., how much variation there is in the observed data.
  - In linear regression, TSS = RSS + ESS.
- Coefficient of determination is

$$R^2 = 1 - \frac{\text{RSS}}{\text{TSS}}.$$

Note that the baseline model always predicts  $\overline{y}$ , so  $R^2 = 0$ . Furthermore, adding features weakly decreases RSS, hence weakly increases  $R^2$ . Instead, we could use adjusted  $R^2$ :

$$\overline{R}^2 = 1 - \frac{\mathrm{RSS}/\mathrm{df_{res}}}{\mathrm{TSS}/\mathrm{df_{tot}}},$$

where df is degrees of freedom, so  $df_{res} = n - p - 1$  and  $df_{tot} = n - 1$ .

#### • Assumptions:

- Weak exogeneity: Predictor variables x can be treated as fixed values, and not random variables, i.e., x is error-free, i.e.,  $E(x\epsilon) = 0$ . So no confounding variables.
- **Linearity**: Mean of y is linear combination of parameters and x.
- **Independence**: Observations are independent of each other.
- **Homoscedasticity**: Constant variance of  $\epsilon$ .
- Normality:  $\epsilon$  follow a normal distribution.
- No multicollinearity: The independent variables are not highly correlated with each other.

# Natural Language Processing

- Solve sequence transduction, i.e., transforms an input sequence to an output sequence.
- Necessary to have some *memory*.

#### 3.1 RNN

- At each time step, receive two inputs: word embedding of current word and hidden state.
- Let  $x_t \in \mathbb{R}^n$  and  $h_{t-1} \in \mathbb{R}^m$ .
- Use weight matrix  $W_x \in \mathbb{R}^{m \times n}$  and hidden-state-to-hidden-state matrix  $W_h \in \mathbb{R}^{m \times m}$ .
- Then

$$o_{t} = W_{hh}h_{t-1} + W_{hx}x_{t} + b_{h}$$

$$h_{t} = \tanh(o_{t}) = \tanh(W_{h} \cdot [h_{t-1}, x_{t}] + b_{h})$$

$$y_{t} = g(W_{y}h_{t} + b_{y})$$

for some activation function g.

• We can show that

$$\nabla_{W_{hh}}(h_t) = \sum_{t'=1}^{t-1} h_{t'} \left( W_{hh}^{t-t'-1} \tanh'(o_{t'+1}) \cdot \dots \cdot \tanh'(o_t) \right).$$

So the influence of  $h_{t'}$  on  $h_t$  will be small if  $t' \ll t$  as  $\tanh(x)$  is small for |x| > 2, i.e., we have a vanishing gradient problem.

#### 3.2 LSTM

- Introduce cell state to RNN.
- Information is added or removed to the cell state through gates.
- Forget gate:
  - Input: previous cell state  $C_{t-1}$  and  $x_t$ .

- Output: number in [0,1] for each entry in  $C_{t-1}$ , i.e., how much to forget.
- So we have

$$f_t = \sigma(W_{hf}h_{t-1} + W_{xf}x_t + b_f)$$
  
=  $\sigma(W_f \cdot [h_{t-1}, x_t] + b_f),$ 

where  $\sigma$  is the sigmoid function applied element-wise.

#### • Input and update gate:

- Decide what and how much to add to the cell state.
- What to add:

$$\tilde{C}_t = \tanh(W_C \cdot [h_{t-1}, x_t] + b_C).$$

- How much to add:

$$i_t = \sigma(W_i \cdot [h_{t-1}, x_t] + b_i).$$

- Update cell state:

$$C_t = f_t \odot C_{t-1} + i_t \odot \tilde{C}_t$$

where  $\odot$  is element-wise multiplication.

#### • Output gate:

- Decide what parts of cell state to output:

$$o_t = \sigma(W_o \cdot [h_{t-1}, x_t] + b_o).$$

- Actual output:

$$h_t = o_t \odot \tanh(C_t)$$

• Drawbacks: computational complexity, overfitting, dropout harder to implement, sensitive to different random weight initializations, inability to handle temporal dependencies that are longer than a few steps,

### 3.3 Seq2seq

- Input  $\rightarrow$  Encoder  $\rightarrow$  Context Vector  $\rightarrow$  Decoder  $\rightarrow$  Output
- Encoder and decoder are both RNNs or LSTMs.
- Last hidden state of encoder is context vector, which initializes decoder RNN.
- At each time step, decoder receives previous token (input of RNN) and uses previous hidden state. The output of RNN is fed though FNN to get embedding of word.
- Stop when EOS token is outputted.
- For back-propagation, *teacher forcing* is used, i.e., plug in correct words to decoder and stop at correct length.
- Drawbacks:
  - Bottleneck problem: for long input sequences, information would tend to be lost.
  - For the decoder, different information may be relevant at different steps.

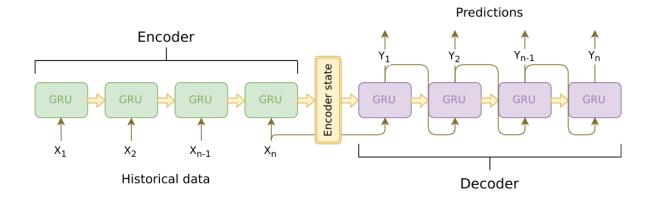


Image source: [5]

## 3.4 Seq2seq with Attention

- At each decoder step, it decides which source parts are more important
- Concretely, at each t,:
  - (i) Use previous token in RNN to update hidden state  $h_t$ .
  - (ii) Compute score( $h_t, s_k$ ) between decoder hidden state  $h_t$  and all encoder hidden states  $s_1, \ldots, s_m$ ;
  - (iii) Compute attention weights: softmax attention scores;
  - (iv) Compute attention vector  $a_t$ : weighted sum of encoder states.
  - (v) Use attention vector  $a_t$  and hidden state  $h_t$  in FNN to get output token.
- Popular score functions:
  - Dot-product:  $score(h_t, s_k) = h_t^T s_k$ .
  - Bilinear function:  $score(h_t, s_k) = h_t^T W s_k$ .
  - Multi-layer perceptron:  $score(h_t, s_k) = v^T \tanh(W[h_t, s_k]).$
- Drawback: RNN is difficult to parallelize.

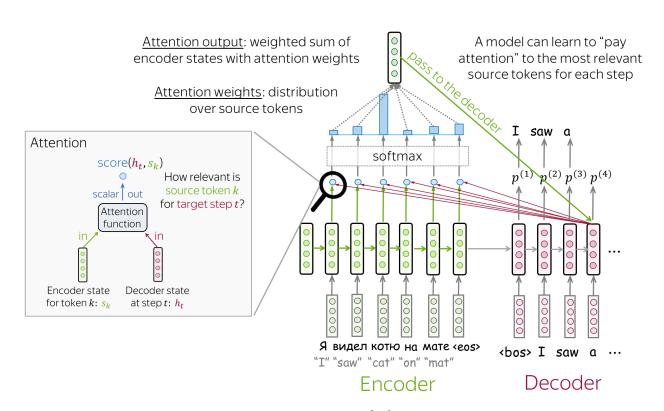


Image source: [15]

## **Transformers**

- Encoding component is a stack (six in original paper) of encoders with same structure but different weights. Decoding component is a stack of decoders of the same number.
- Encoders: Self-Attention  $\rightarrow$  FNN
  - First encoder receives embedding of words.
  - Other encoders receive output of encoder directly below.
- Decoders: Self-Attention  $\rightarrow$  Encoder-Decoder Attention  $\rightarrow$  FNN

### 4.1 Ingredients

#### 4.1.1 Self-attention

#### **Process**

- 1. Create three vectors from each of the encoder's input vectors: Query, Key, and Value vectors.
- 2. Note: in original paper, input vectors are in  $\mathbb{R}^{512}$ , and QKV vectors are in  $\mathbb{R}^{64}$ .
- 3. Self-attention score between i-th and j-th word is  $q_i \cdot k_j$  for all j.
- 4. Divide scores by square root of dimension (8; this gives more stable gradients) and softmax.
- 5. Multiply each value vector  $v_i$  by softmax scores.
- 6. Sum up the weighted value vectors. This is output of self-attention layer for i-th word.

In matrix form, suppose our input is  $X \in \mathbb{R}^{n \times d_{\text{model}}}$ , where n is the sequence length and  $d_{\text{model}}$  is the embedding dimension. (So rows of X are the inputs.) The QKV matrices are given by  $W_Q, W_K, W_V \in \mathbb{R}^{d_{\text{model}} \times d_k}$ . Then the QKV vectors

$$Q = XW_Q$$
$$K = XW_K$$
$$V = XW_V$$

are in  $\mathbb{R}^{n \times d_k}$  For each attention head, we have

$$\operatorname{Attention}(Q,K,V) = \operatorname{softmax}\left(\frac{QK^T}{\sqrt{d_k}}\right) \times V \in \mathbb{R}^{n \times d_k},$$

i.e., aggregate values weighted by attention.

#### 4.1.2 Multi-headed attention

Do h parallel attention heads, with different weight matrices for each head.

$$MultiHead(Q, K, V) = Concat(head_1, ..., head_h) \times W_O$$

where

$$head_i = Attention(Q_i, K_i, V_i)$$

is output of each attention head and  $W_O \in \mathbb{R}^{hd_k \times d_{\text{model}}}$ . Note that  $\text{MultiHead}(Q, K, V) \in \mathbb{R}^{n \times d_{\text{model}}}$ .

#### 4.1.3 Positional encoding

Transformers are permutation-invariant, so we add positional encoding to input embeddings.

#### 4.1.4 FNN

Output of attention fed into FNN:

$$FNN(x) = ReLU(xW_1 + b_1)W_2 + b_2.$$

#### 4.1.5 Residual connections & layer normalization

Each sub-layer (attention or feedforward) is followed by

$$LayerNorm(x + Sublayer(x)).$$

Given  $x \in \mathbb{R}^d$ , the layer norm is

$$LayerNorm(x) = \frac{x - \mu}{\sqrt{\sigma^2 + \epsilon}} \cdot \gamma + \beta,$$

where

$$\mu = \frac{\sum x_i}{d} \text{ (mean over features)}$$

$$\sigma^2 = \frac{\sum (x_i - \mu)^2}{d} \text{ (variance over features)}$$

 $\gamma, \beta$ : learned scaling and shifting parameters  $\epsilon$ : small constant to prevent division by zero.

## 4.2 Original Architecture

Encoder Layer:

- Input Embedding + Positional Encoding
- Multi-Head Self-Attention  $\rightarrow$  Residual + Layer Norm
- Feed-Forward Network  $\rightarrow$  Residual + Layer Norm

#### Decoder Layer:

- Masked Multi-Head Self-Attention (prevent peeking at future tokens)
- Multi-Head Encoder-Decoder Attention  $\rightarrow$  Residual + Layer Norm
- Feed-Forward Network  $\rightarrow$  Residual + Layer Norm

#### 4.3 Modern Architecture

- Input Embedding
- Layer Norm  $\rightarrow$  Self-Attention  $\rightarrow$  Residual
  - Grouped-Query Attention
  - Rotary Embeddings
- Layer Norm  $\rightarrow$  FNN  $\rightarrow$  Residual

#### 4.3.1 Grouped-query attention

- Use same  $W_K$  and  $W_V$  across heads, and each head has its own  $W_Q$ .
- $\bullet$  Better: divide heads into groups. Heads in the same group share  $W_K$  and  $W_V$

### 4.3.2 Rotary positional embeddings (RoPE)

- Limitations of absolute positional embedding:
  - Limited sequence length
  - Independence of positional embedding, e.g./ difference between position 1 and 2 is the same as the difference between position 1 and 500
- Alternative: relative positional embeddings [13]
  - Bias for positional offsets: use a bias to represent each possible positional offset
  - Relative attention becomes:

Attention
$$(Q, K, V)_i = \sum_{j=1}^n \operatorname{softmax}(e_{ij}) \times (x_j W_V + a_{ij}^V),$$

where

$$e_{ij} = \frac{QK^T + x_i W_Q(a_{ij}^K)^T}{\sqrt{d\iota}}.$$

- Here,  $a_{ij} \in \mathbb{R}^{1 \times d_a}$  is a vector of relative positional weights, i.e.,

$$a_{ij} = w_{\text{clip}(j-i,k)}$$
$$\text{clip}(x,k) = \max(-k, \min(k,x)).$$

Calculate w for both keys and values.

- Clipping allows scalability (i.e., arbitrarily long sequences)

- Limitations: slower; complicates key-value cache usage as each additional token changes the embedding for every other token.
- For **RoPE** [14], the intuition is to rotate each embedding by  $m\theta$ , where m is the position of the word in the sequence.
- Benefits:
  - Scalability: adding new words does not change the embedding of previous words
  - The dot product of the embeddings of two words does not depend on absolute position.
- Mathematically, we first work in  $\mathbb{C}^{d_k/2}$ . Let  $M_j$  be the rotation matrix by  $m\theta_j$ . Then the output of RoPE for the m-th word is just

$$Q_m \Theta_m = Q_m \times \begin{pmatrix} M_1 & & & \\ & M_2 & & \\ & & \ddots & \\ & & & M_{d_k/2} \end{pmatrix},$$

where  $Q_m$  is the m-th row of Q (i.e., query vector for m-th word).

• Do the same for key vector.

### 4.4 Other Techniques

#### 4.4.1 Sparse attention

Instead of global autoregressive self-attention, use local autoregressive self-attention in some layers.

#### 4.4.2 Mixture of experts (MOE)

- Instead of a single monolithic feedforward layer in the transformer block, use a set of expert networks, and route the input to only a few of them.
- A router (a smaller FNN) calculates which experts to turn on.
- Could do model merging by doing a weighted average of experts using weights from the router.

## 4.5 Complexity

#### 4.5.1 Number of parameters and FLOPs

Suppose input length is n and  $n_{\text{layer}}$  is the number of layers. Recall  $X \in \mathbb{R}^{n \times d_{\text{model}}}$ ;  $W_Q, W_K, W_V \in \mathbb{R}^{d_{\text{model}} \times d_k}$ ; and  $Q, K, V \in \mathbb{R}^{n \times d_k}$ . As in [9], we have the following steps in the transformer architecture:

- 1. Embed: token and positional embeddings.
  - Notation:
    - $-d_{\text{model}}$  is model dimension, i.e., dimension of the residual stream.
    - $-n_{\text{vocab}}$  is the number of tokens in the vocabulary.

- Parameters:  $n_{\text{vocab}} \times d_{\text{model}}$  for token embedding, and  $n \times d_{\text{model}}$  for positional embedding  $\Rightarrow (n_{\text{vocab}} + n)d_{\text{model}}$  parameters.
- FLOPs:  $O(d_{\text{model}})$  per token
- 2. Attention (QKV): project into QKV vectors.
  - Notation:
    - $-d_{\rm k}=d_{\rm v}$  is the dimension of the projections.
    - $-n_{\text{heads}}$  is the number of attention heads.
    - $-d_{\text{attn}} = d_k n_{\text{heads}}$  is the output of the multi-headed attention (usually =  $d_{\text{model}}$ ).
  - Parameters:  $3d_{\text{model}}d_{\mathbf{k}}$  per head per layer  $\Rightarrow 3n_{\text{layer}}d_{\text{model}}d_{\text{attn}}$  parameters.
  - FLOPs: we are doing three instances of  $[n, d_{\text{model}}] \times [d_{\text{model}}, d_{\text{k}}]$ , so get  $3 \times 2 \times nd_{\text{model}}d_{\text{k}}$  per head per layer  $\Rightarrow 6d_{\text{model}}d_{\text{attn}}$  per token per layer  $\Rightarrow 6n_{\text{layer}}d_{\text{model}}d_{\text{attn}}$  per token.
- 3. Attention (Mask): dot-product query and key, i.e.,  $QK^T$ .
  - FLOPs: doing  $[n, d_k] \times [d_k, n]$ , so get  $2n^2d_k$  per head per layer  $\Rightarrow 2nn_{\text{layer}}d_{\text{attn}}$  per token.
- 4. Attention (Project): project output of attention heads to  $d_{\text{model}}$  by multiplying with  $W_O \in \mathbb{R}^{d_{\text{attn}} \times d_{\text{model}}}$ 
  - Parameters:  $d_{\text{attn}}d_{\text{model}}$  per layer  $\Rightarrow n_{\text{layer}}d_{\text{attn}}d_{\text{model}}$  parameters.
  - FLOPs: doing  $[n, d_{\text{attn}}] \times [d_{\text{attn}}, d_{\text{model}}]$ , so get  $2nd_{\text{attn}}d_{\text{model}}$  per layer  $\Rightarrow 2n_{\text{layer}}d_{\text{attn}}d_{\text{model}}$  per token.
- 5. Feedforward: two linear layers in dense neural network.
  - Notation:
    - $d_{\rm ff}$  is output size of first linear layer (usually  $d_{\rm ff} = 4d_{\rm model}$ ).
  - Parameters: first weight matrix gives  $d_{\text{model}}d_{\text{ff}}$  and second gives  $d_{\text{ff}}d_{\text{model}}$  per layer  $\Rightarrow$   $2n_{\text{layer}}d_{\text{model}}d_{\text{ff}}$  parameters.
  - FLOPs: doing  $[n, d_{\text{model}}] \times [d_{\text{model}}, d_{\text{ff}}] \times [d_{\text{ff}}, d_{\text{model}}]$  per layer, so get  $4nd_{\text{model}}d_{\text{ff}}$  per layer  $\Rightarrow 4n_{\text{layer}}d_{\text{model}}d_{\text{ff}}$  per token.
- 6. De-embed: go back to vocabulary.
  - FLOPs: doing  $[n, d_{\text{model}}] \times [d_{\text{model}}, d_{\text{vocab}}]$ , so get  $2nd_{\text{model}}d_{\text{vocab}} \Rightarrow 2d_{\text{model}}d_{\text{vocab}}$  per token.

Note that we are doing FLOPs per token. So the total number of parameters is

$$2n_{\text{laver}}d_{\text{model}}(2d_{\text{attn}} + d_{\text{ff}}) + d_{\text{model}}(n_{\text{vocab}} + n) \approx 2n_{\text{laver}}d_{\text{model}}(2d_{\text{attn}} + d_{\text{ff}}) =: N$$

and the total FLOPs per token for the forward pass becomes

$$2N + 2d_{\text{model}}d_{\text{vocab}} + 2nn_{\text{laver}}d_{\text{attn}} \approx 2N + 2nn_{\text{laver}}d_{\text{attn}} =: C_{\text{forward}}$$

The FLOPs for the backwards pass is about twice the FLOPs for the forward pass, so the total number of FLOPs per token is

$$C \approx 6N$$
.

Then the total number of FLOPs is

6ND,

where D is the number of tokens in the corpus.

If we let  $\tau$  be the throughput, i.e., inputs processed per unit time (in FLOPs per second) and T be the total training time, then we get

 $\tau T \approx 6ND$ 

#### 4.5.2 Time complexity of self-attention

- Calculating Q, K, V are all  $O(nd_{\text{model}}d_k)$ .
- Calculating  $QK^T$  is  $O(n^2d_k)$ .
- Doing softmax is  $O(n^2)$ .
- Multiplying result with V is  $O(n^2d_k)$ .

Overall, the time complexity of the self-attention mechanism is  $O(nd_{\text{model}}d_k + n^2d_k)$ . If  $n \gg d_{\text{model}}, d_k$ , this reduces to just  $O(n^2)$ .

*Remark.* In [10], the authors show that the time complexity of self-attention is necessarily quadratic in the input length, unless the Strong Exponential Time Hypothesis (SETH) is false.

# Fine-Tuning Techniques

Full fine-tuning, of course, has the best performance but is computationally expensive. Need parameter-efficient fine-tuning (PEFT).

- Adapter-based methods
  - Adapters: small, trainable models inserted into pre-trained transformers.
  - Freeze original model weights.
- Instruction tuning

## 5.1 Low-Rank Adaptation (LoRA)

Introduced in [8].

- Intuition: only train low rank perturbations of the (selected) weight matrices.
- Let  $W_0 \in \mathbb{R}^{d \times k}$  be a pre-trained weight matrix.
- Update:  $W_0 + \Delta W = W_0 + BA$ , where  $B \in \mathbb{R}^{d \times r}, A \in \mathbb{R}^{r \times d}, r \ll \min(d, k)$ .
- Advantages:
  - Roughly converges to full fine-tuning as r increases.
  - No additional inference latency: when needing to switch to another downstream task, can recover  $W_0$  by subtracting BA, and then we can add a new B'A'.
- Notes [6]:
  - Optimal placement highly dependent on the dataset and model architecture.
  - For transformers, applying LoRA exclusively to attention layers provides the most stability and mitigates the risk of divergence.
  - For MoE, applying LoRA to each expert individually boosts performance, but significantly increases memory usage. Applying to router gives limited success.
  - Could optimize memory by using same B across different A.

### 5.2 QLoRA

Quantized LoRA, introduced in [4].

#### 5.2.1 4-bit NormalFloat Quantization

- Better quantization data type for normally distributed data.
- In general, for k-bit Normal Float, equally divide the normal distribution into  $2^k$  quantiles.
- Note: to ensure 0 has an exact representation, actually divide the negative part into  $2^{k-1}$  quantiles and the positive part into  $2^{k-1}+1$  quantiles, then remove one of the two overlapping zeros.
- Then

$$X^{\mathrm{NF4}} = \mathrm{round}\left(\frac{1}{\mathrm{absmax}(X^{\mathrm{FP32}})}X^{\mathrm{FP32}}\right) = \mathrm{round}(c^{\mathrm{FP32}}X^{\mathrm{FP32}}),$$

where  $c^{\text{FP}32}$  is the quantization constant.

#### 5.2.2 Block-wise Quantization

- One problem is quantization is that outliers severely impacts the scaling, and the full range of the lower precision data type may not be effectively used.
- Solution: Chunk the input tensor into blocks that are independently quantized, each with its own quantization constant.

#### 5.2.3 Double Quantization

- Helps reduce the memory footprint of quantization constants.
- Block-wise k-bit quantization for the quantization constants  $c^{\text{FP32}}$ .
- Gives  $c_2^{k\text{-bit}}$ , with second level of quantization constants  $c_1^{\text{FP32}}$ .

#### 5.2.4 Implementation

• Recall: LoRA has

$$Y = XW + XL_1L_2$$

for low rank matrices  $L_i$ .

• For QLoRA, do

$$Y^{\rm BF16} = X^{\rm BF16} \, {\rm double Dequant}(c_1^{\rm FP32}, c_2^{k\text{-bit}}, W^{\rm NF4}) + X^{\rm BF16} L_1^{\rm BF16} L_2^{\rm BF16},$$

where

$$\operatorname{doubleDequant}(c_1^{\operatorname{FP32}}, c_2^{k\operatorname{-bit}}, W^{\operatorname{NF4}}) = \operatorname{dequant}(\operatorname{dequant}(c_1^{\operatorname{FP32}}, c_2^{k\operatorname{-bit}}), W^{\operatorname{NF4}}) = W^{\operatorname{BF16}}(c_1^{\operatorname{FP32}}, c_2^{k\operatorname{-bit}}) = W^{\operatorname{BF16}}(c_1^{\operatorname{FP32}}, c_2^{k\operatorname{-bit}})$$

- Original paper [4] uses FP8 for  $c_2$ , block size of 64 for W, and block size of 256 for  $c_2$ .
- $\bullet$  Computations done in BF16 and storage of W in NF4.

### 5.2.5 $(IA)^3$

- Introduced in [12], Infused Adapter by Inhibiting and Amplifying Inner Activations is intended to improve over LoRA.
- Notation: if  $l \in \mathbb{R}^d$  and  $A \in \mathbb{R}^{m \times d}$ , then  $(l \odot A)_{ij} = l_j A_{ij}$ , i.e., element-wise multiplication of l and rows of A.
- Learned vectors  $l_{\rm v}, l_{\rm k}, l_{\rm ff}$ .
- Attention layer:

softmax 
$$\left(\frac{Q(l_k \odot K^T)}{\sqrt{d_k}}\right) \times (l_v \odot V)$$

• Feedforward layer:

$$(l_{\rm ff} \odot \gamma(W_1 x))W_2,$$

where  $\gamma$  is the FFN nonlinearity.

- Advantages:
  - Enables mixed tasks.
  - For single-task models, no additional computational cost compared to original, because one can simply replace W with  $l \odot W$ .

## Fourier Transform

• Let x(t) be a complex-valued function. Then its Fourier transform is

$$X(\omega) = \int_{-\infty}^{\infty} x(t)e^{-i\omega t} dt.$$

The inverse Fourier transform is

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{i\omega t} d\omega.$$

• For a sequence of discrete time signals, use discrete-time Fourier transform:

$$X(\omega) = \sum_{-\infty}^{\infty} x[n]e^{-i\omega n}.$$

The output is continuous in  $\omega$  and periodic. The inverse DTFT is:

$$x[n] = \frac{1}{2\pi} \int_{2\pi} X(\omega) e^{i\omega n} d\omega.$$

• Discrete Fourier transform converts a finite sequence of equally-spaced samples of a function into a same-length sequence of equally-spaced samples of the DTFT. It is given by

$$X[k] = \sum_{n=0}^{N-1} x[n]e^{-2\pi i k n/N}$$
$$= \sum_{n=0}^{N-1} x[n]W_N^{kn},$$

where  $W_N = e^{-2\pi i/N}$ . The inverse transform is given by

$$\frac{1}{N} \sum_{k=0}^{N-1} X[k] W_N^{-kn}.$$

• The fast Fourier transform is based on the idea that the N-th roots of unity  $W_N^k$  have nice properties when N is a power of 2. In vector form,

$$\mathbf{X}[k] = \mathbf{W_N} \mathbf{x}[n],$$

where  $W_N = V(W_N^0, \dots, W_N^{N-1})$  is a Vandermonde matrix. The computational complexity is  $O(N \log N)$  vs.  $O(N^2)$  of definition of DFT.

- Limitation of Fourier transform is that it lacks temporal resolution. A way to overcome this is the short-time Fourier transform.
  - Continuous case:

$$STFT\{x(t)\}(\tau,\omega) = \int_{-\infty}^{\infty} x(t)w(t-\tau)e^{-i\omega t} dt,$$

where w is the window function, usually Hann or Gaussian window.

- Discrete case:

$$STFT\{x[n]\}(k,\omega) = \sum_{n=-\infty}^{\infty} x[n]w[n-k]e^{-i\omega n}.$$

Uncertainty principle: there is a trade-off between temporal and frequency resolution.

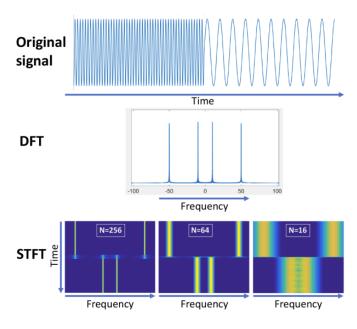


Image source: [7]

#### • Spectrograms:

- Divide a time-domain signal into segments of equal length.
- Apply FFT to each segment, transforming the data from the time domain to the frequency domain.
- Each segment corresponds to vertical line in spectogram.
- For window width w, spectogram $(t) = |\operatorname{STFT}(t)|^2$ .

#### • Nyquist-Shannon sampling theorem

- Sampling rate must be at least twice the bandwidth of the signal to avoid aliasing.
- Let x(t) have Fourier transform X(f). Suppose  $X_{1/T}(f)$  is the DTFT of sample sequence x[n].
- For DTFT, copies of  $X_f$  are shifted by multiples of the sampling rate  $f_s$  and added.

- $-\,$  If Nyquist–Shannon is not satisfied, copies will overlap.
- Any frequency component above  $f_s/2$  is indistinguishable from a lower-frequency component (i.e., alias).

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