# EE 230 Experiment 1

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## 1 Introduction

Simulation of various RC circuits is done using NGSPICE in this experiment. The results are then compared with the theoretical values.

## 2 Procedure

The various circuits and their simulation results are described below,

#### 1. RC Integrator

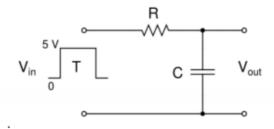


Figure 1: RC Integrator Circuit

The SPICE code for simulation of a RC Integrator circuit with  $T=0.1\tau$  is given below.

```
Dhruv Shah 190020039 RC Integrator Circuit
*Circuit Description
r1 1 2 10k
c1 2 0 0.1u
*Voltage Description
*v1n 1 0 pulse(0 5 delay rise_time fall_time T Period)
v1n 1 0 pulse(0 5 0 0 0 10m 20m)
```

- \*transient Analysis
  \*.tran step 3\*period
- .tran 0.1m 60m
- .control

run

```
*Plotting plot v(1) v(2) set hcopypscolor = 1 hardcopy integrator_10tau.ps v(1) v(2) .endc .end
```

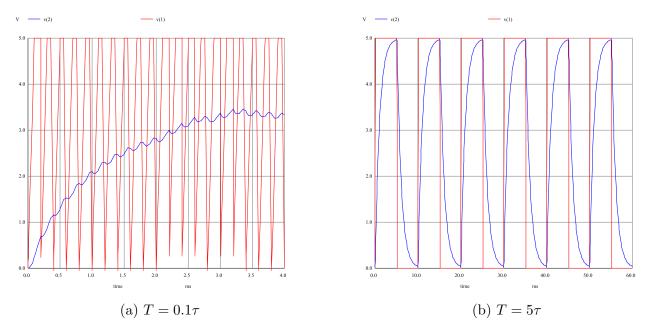


Figure 2: Waveforms for RC Integrator Circuit

It is noted that the Integrator performs better when  $T \gg \tau$ .

#### 2. RC Differentiator

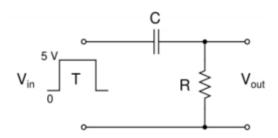


Figure 3: RC Differentiator Circuit

The SPICE code for simulation of a RC Differentiator circuit with  $T=0.1\tau$  is given below.

Dhruv Shah 190020039 RC Differentiator Circuit
\*Circuit Description
r1 2 0 10k
c1 1 2 0.1u
\*Voltage Description
\*v1n 1 0 pulse(0 5 delay rise\_time fall\_time T Period)

```
v1n 1 0 pulse(0 5 0 0 0 0.0001 0.0002)
*transient Analysis
*.tran step 3*period

.tran 1u 4m
.control
run

*Plotting
plot v(1) v(2)
set hcopypscolor = 1
hardcopy diff_pt1tau.ps v(1) v(2)
.endc
.end
```

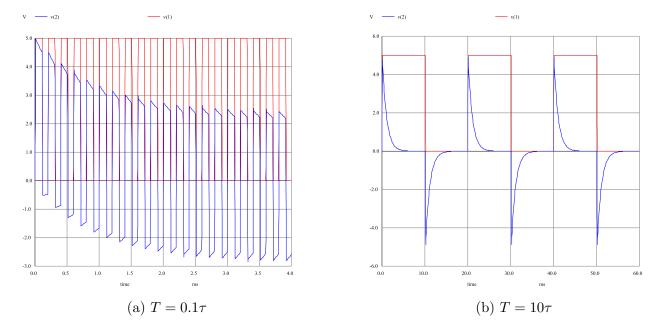


Figure 4: Waveforms for RC Differentiator Circuit

It is noted that the Differentiator performs better when  $T \ll \tau$ .

#### 3. RC Lowpass Filter

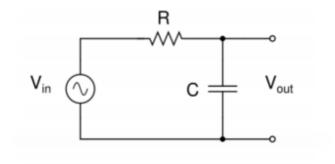


Figure 5: RC Lowpass Filter Circuit

The SPICE code for simulation of a RC Lowpass filter is given below.

```
Dhruv Shah 190020039 RC Lowpass Filter
*Circuit Description
r1 1 2 10k
c1 2 0 0.1u
*Voltage Description
vin 1 0 dc 0 ac 1
*AC analysis for 1 Hz to 1MHz, 10 points per decade
. ac dec 10 1 1Meg
. control
run
*Magnitude dB plot for v(2) on log scale
plot vdb(2) xlog
set hcopypscolor = 1
hardcopy low_pass.ps vdb(2)
.endc
. end
```

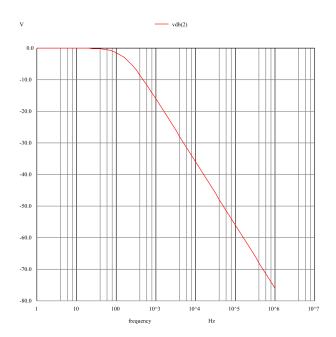


Figure 6: Waveform for RC Lowpass Filter

#### 4. RC Highpass Filter

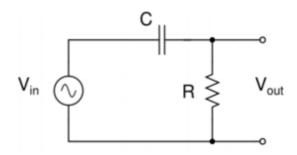


Figure 7: RC Highpass Filter Circuit

The SPICE code for simulation of a RC Highpass filter is given below.

```
Dhruv Shah 190020039 RC Highpass Filter
*Circuit Description
r1 2 0 10k
c1 1 2 0.1u
*Voltage Description
vin 1 0 dc 0 ac 1
*AC analysis for 1 Hz to 1MHz, 10 points per decade
. ac dec 10 1 1Meg
.control
run
*Magnitude dB plot for v(2) on log scale
plot vdb(2) xlog
set hcopypscolor = 1
hardcopy high_pass.ps vdb(2)
.\,\mathrm{endc}
. end
```

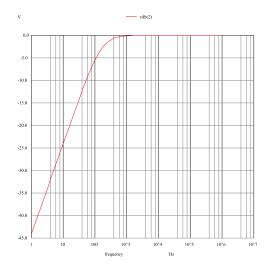


Figure 8: Waveform for RC Highpass Filter

#### 5. RC Bandpass Filter

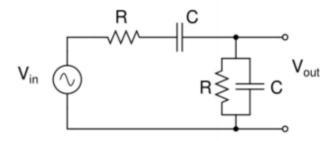


Figure 9: RC Bandpass Filter Circuit

The transfer function can be calculated using the principle of voltage divider,

$$\frac{V_{\text{out}}}{V_{\text{in}}} = \frac{sRC}{1 + 3sRC + s^2(RC)^2}$$

At  $f_0$ , the gain is maximum. Taking  $sRC = \alpha$  gives,

$$\frac{V_{\text{out}}}{V_{\text{in}}} = \frac{\alpha}{1 + 3\alpha + \alpha^2}$$
$$= \frac{1}{3 + \alpha + \frac{1}{\alpha}}$$

The function is maximised at  $\alpha=1$  with a maximum value of  $\frac{1}{5}$ . This shows that  $f_0=\frac{1}{2\pi RC}$ . To calculate -3dB points, we set the gain to  $\frac{1}{\sqrt{2}}$  times the maximum gain.

$$\frac{\alpha}{1+3\alpha+\alpha^2} = \frac{1}{5\sqrt{2}}$$

$$\alpha^2 + \left(3-5\sqrt{2}\right)\alpha + 1 = 0$$

$$\alpha = \frac{4.07 \pm \sqrt{16.57-4}}{2}$$

$$\alpha = 2.03 \pm 1.77$$

From this we get that,

$$f_0 = \frac{1}{2\pi RC}$$

$$f_l = \frac{1}{2\pi RC}(0.26)$$

$$f_h = f_h = \frac{1}{2\pi RC}(3.8)$$

The SPICE code for simulation of a RC Bandpass filter is given below.

```
Dhruv Shah 190020039 RC bandpass Filter
*Circuit Description
r1 2 0 10k
c1 2 0 0.1u
r2 1 3 10k
c2 3 2 0.1u
*Voltage Description
vin 1 0 dc 0 ac 1
*AC analysis for 1 Hz to 1MHz, 10 points per decade
. ac dec 10 1 1Meg
. control
run
*Magnitude dB plot for v(2) on log scale
plot vdb(2) xlog
```

set hcopypscolor = 1hardcopy RC\_band\_pass.ps vdb(2)

\* Measurements

meas ac peak MAX vmag(2) meas ac fpeak WHEN vmag(2)=peak let f3db = peak / sqrt(2)meas ac f1 WHEN vmag(2) = f3db RISE = 1 meas ac f2 WHEN vmag(2) = f3db FALL = 1 .endc

. end

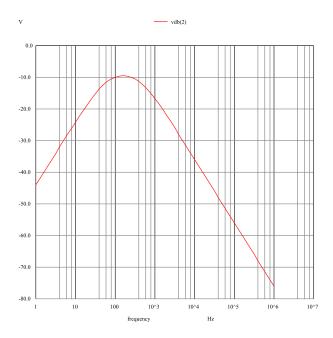


Figure 10: Waveform for RC Bandpass Filter

The comparison of simulation results with theory is summarised in the table below,

Sr no.	Quantity	Theoretical	Simulated
1	$f_0$	159.15	158.49
2	$f_l$	41.38	48.38
3	$f_h$	604.79	527.66

#### RLC Bandpass fiter

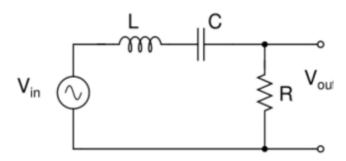


Figure 11: RLC Bandpass Filter Circuit

A similar analysis as before shows that,

$$f_0 = \frac{1}{2\pi\sqrt{LC}}$$

$$f_l = \frac{1}{2\pi} \left( -\frac{R}{2L} + \sqrt{\frac{R^2}{2L} + \frac{1}{LC}} \right)$$

$$f_h = \frac{1}{2\pi} \left( \frac{R}{2L} + \sqrt{\frac{R^2}{2L} + \frac{1}{LC}} \right)$$

The SPICE code for simulation of a RC Bandpass filter is given below.

```
Dhruv Shah 190020039 RLC Bandpass Filter
*Circuit Description
r1 2 0 1k
12 1 3 10m
c2 3 2 0.1u
*Voltage Description
vin 1 0 dc 0 ac 1
*AC analysis for 1 Hz to 1MHz, 10 points per decade
. ac dec 10 1 1Meg
. control
run
*Magnitude dB plot for v(2) on log scale
plot vdb(2) xlog
set hcopypscolor = 1
hardcopy RLC_band_pass.ps vdb(2)
*Measurements
meas ac peak MAX vmag(2)
meas ac fpeak WHEN vmag(2)=peak
let f3db = peak / sqrt(2)
meas ac f1 WHEN vmag(2) = f3db RISE = 1
meas ac f2 WHEN vmag(2) = f3db FALL = 1
.endc
. end
```

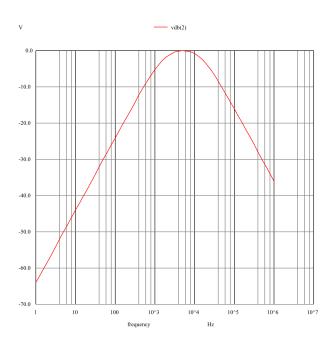


Figure 12: Waveform for RLC Bandpass Filter

The comparison of simulation results with theory is summarised in the table below,

Sr no.	Quantity	Theoretical	Simulated
1	$f_0$	5032.92	5011.87
2	$f_l$	1457.98	1468.64
3	$f_h$	17373.48	17459.91

## 3 Major Learnings from this experiment

- 1. An RC integrator circuit has good performance when  $T\gg \tau$ .
- 2. An RC differentiator circuit has good performance when  $T \ll \tau$ .

## 4 Challenges Faced

The major challenge faced in this experiment was gaining familiarity with SPICE simulation.

# 5 Questions and Clarifications

None