

Wk 7 TA Session

Part I Summary of Wk 7's

Lecture

1) Fourier Transform

→ Defined for continuous
aperiodic signals.

$$2) x(w) \triangleq \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

$$(or) x(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi ft} dt$$

$\omega \triangleq$ angular frequency.

→ Representation of time domain
signal in frequency domain.

3) Inverse Fourier transform

$$x(t) \triangleq \frac{1}{2\pi} \int_{-\infty}^{\infty} x(w) e^{j\omega t} dw$$

$$(or) x(t) = \int_{-\infty}^{\infty} x(f) e^{j2\pi ft} df$$

4) Spectrum

$$x(w) = |x(w)| e^{j\angle x(w)}$$

$|x(w)|$ ↑
 Magnitude spectrum.
 $\angle x(w)$ ↓
 phase spectrum.

5) For real signals

$$\begin{aligned} x^*(w) &= x(-w) \\ \Rightarrow |x(w)| &= |x(-w)| \quad (\text{even symmetry}) \\ \& L x(-w) \\ &= -\langle x(w) \rangle \quad (\text{odd symmetry}) \end{aligned}$$

6) Dirichlet conditions for

convergence :-

- 1) $x(t)$ is absolutely integrable
- 2) $x(t)$ has finite maxima/minima
- 3) $x(t)$ has finite discontinuities.

7) Laplace transform evaluated

$s = \sigma + j\omega$ gives the expression

for Fourier Transform.

$$s = \sigma + j\omega \quad \text{if } \sigma = 0 \Rightarrow s = j\omega$$

8) Properties of $\tilde{f}(t)$:

Ref: Oppenheim
Ch-4:

$$x(t) \longleftrightarrow X(j\omega) \quad y(t) \longleftrightarrow Y(j\omega)$$

Linearity $x_1(t) + b y_1(t) \longleftrightarrow a X_1(\omega) + b Y_1(\omega)$

Time shifting $x(t - t_0) \longleftrightarrow e^{-j\omega t_0} X(\omega)$

Freq shifting $e^{j\omega_0 t} x(t) \longleftrightarrow X(\omega - \omega_0)$

Scaling

Conjugation $x^*(t) \longleftrightarrow X^*(-\omega)$

Time reversal $x(-t) \longleftrightarrow X(-\omega)$

Time & Frequency scaling $x(at) \longleftrightarrow \frac{1}{|a|} X\left(\frac{\omega}{a}\right)$

↳ compression when $a > 1$
expansion, for $a < 1$

Convolution $x(t) * y(t) \longleftrightarrow X(\omega)Y(\omega)$

Multiplication $x(t)y(t) \longleftrightarrow \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\theta)Y(\omega - \theta) d\theta$

Dif in time $\frac{d x(t)}{dt} \longleftrightarrow j\omega X(\omega)$

$$\text{Integration} \quad \int_{-\infty}^{\infty} x(t) dt = \frac{1}{j\omega} X(\omega) + \pi X(0)$$

$$\text{Diff in frequency} \quad t x(t) \quad j \frac{d}{d\omega} X(\omega)$$

Conjugate symmetry for Real signals.

$$x(t) \text{ real} \Rightarrow x(w) = x^*(-w)$$

$$\operatorname{Re}(x(w)) = \operatorname{Re}(x(-w))$$

$$q_m(x(w)) = -q_m(x(-w))$$

$$|x(w)| = |x(-w)|$$

$$X(w) = -X(-w)$$

Symmetry for real & even signals

$$x(t) \text{ R & E} \Rightarrow x(w) \text{ R+E}$$

Symmetry for real & odd signals

$$x(t) \text{ R & O} \Rightarrow x(w) \text{ Purely } q_m \text{ & odd.}$$

Even & odd decomposition for real signals

$$x(t) = \operatorname{Ev}(x(t)) + \operatorname{Od}(x(t))$$

$$\operatorname{Ev}(x(w)) = \operatorname{Re}(x(w))$$

$$\operatorname{Od}(x(w)) = j q_m(x(w))$$

$x(t)$ is real.

Duality

$$x(t) \longleftrightarrow x(w)$$

$$x(t) \longleftrightarrow 2\pi x(-w)$$

Parseval Relation for aperiodic signals.

$$\int_{-\infty}^{\infty} |x(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(w)|^2 dw.$$

(9) FT of common signals

$$\sum_{n=-\infty}^{\infty} a_n e^{jkn_0 t} \quad (\rightarrow 2\pi \sum_{n=-\infty}^{\infty} a_n \delta(n - n_0))$$

$$e^{jw_0 t} \quad (\rightarrow 2\pi s(w - w_0))$$

$$\text{cos } \omega_0 t \rightarrow \pi \left[\delta(w - \omega_0) + \delta(w + \omega_0) \right]$$

$$\text{sin } \omega_0 t \rightarrow \frac{\pi}{j} \left[\delta(w - \omega_0) - \delta(w + \omega_0) \right]$$

$$1 \longleftrightarrow 2\pi \delta(w).$$

$$n(t) = \begin{cases} 1, & |t| < T \\ 0, & |t| > T/2 \end{cases} \leftrightarrow \sum_{n=-\infty}^{\infty} \frac{2 \sin(k\omega_0 T)}{jk} \delta(w - kn\omega_0)$$

$$\sum_{n=-\infty}^{\infty} \delta(t - nT) \leftrightarrow \frac{2\pi}{T} \sum_{k=-\infty}^{\infty} \delta\left(w - \frac{2\pi k}{T}\right)$$

$$x(t) = \begin{cases} 1, & |t| < T \\ 0, & |t| > T \end{cases} \leftrightarrow \frac{2 \sin(wT)}{w}$$

$$\frac{\sin \omega t}{\pi t} \leftrightarrow x(w) = \begin{cases} 1, & |w| < \omega \\ 0, & |w| > \omega \end{cases}$$

$$\delta(t) \leftrightarrow 1$$

$$u(t) \leftrightarrow \frac{1}{jw} + \pi \delta(w)$$

$$\delta(t - \tau) \leftrightarrow e^{-j\omega \tau}$$

$$e^{-at} u(t), \operatorname{Re}\{a\} > 0 \quad \frac{1}{a + jw}$$

$$t e^{-at} u(t), \operatorname{Re}\{a\} > 0 \quad \frac{1}{(a + jw)^2}$$

$$\frac{t^{n-1}}{(n-1)!} e^{-at} u(t), \operatorname{Re}\{a\} > 0 \quad \frac{1}{(a + jw)^n}$$

(10)

) Frequency response of the system.

$$x(t) \rightarrow \boxed{h(t)} \rightarrow y(t)$$

$$x(w) + h(w) = y(w)$$

$$\Rightarrow h(w) = \frac{y(w)}{x(w)}$$

Freq
responses

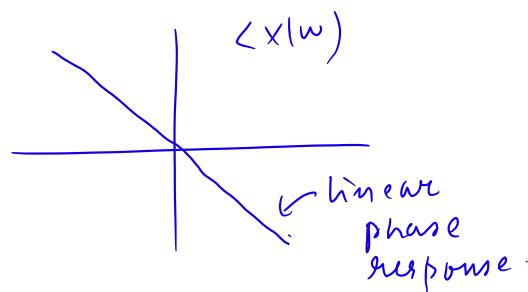
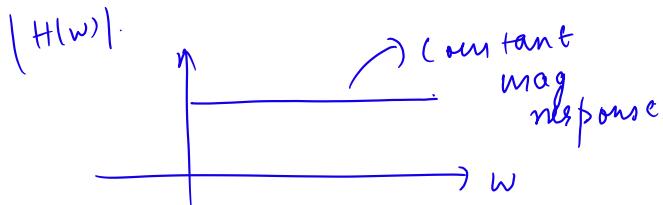
$$x(t) = e^{j\omega_0 t} \rightarrow \text{cyclic function!}$$

ii) distortionless transmission via system

$$y(t) = k x(t - t_0)$$

↳ fully scaling & time delaying

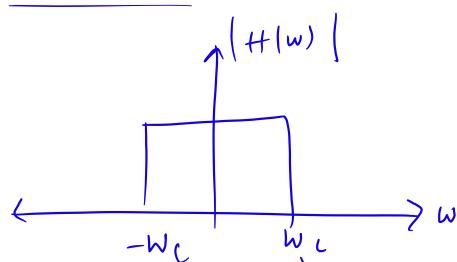
Such sys should have



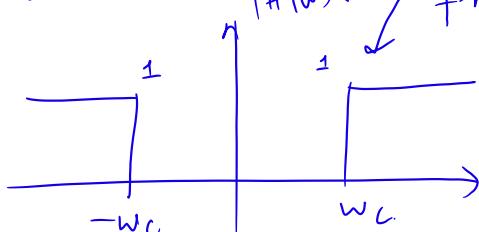
(1) ideal filters

→ allows only certain frequencies to pass.

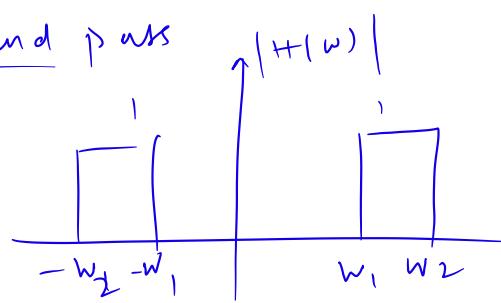
(i) Low pass:



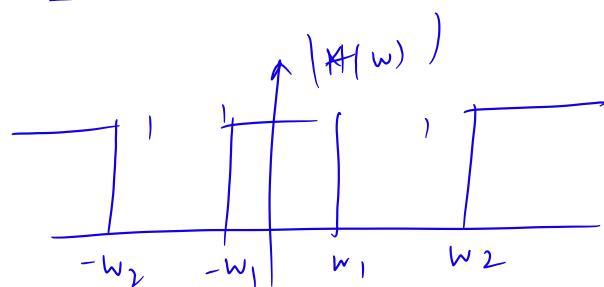
(ii) high pass:



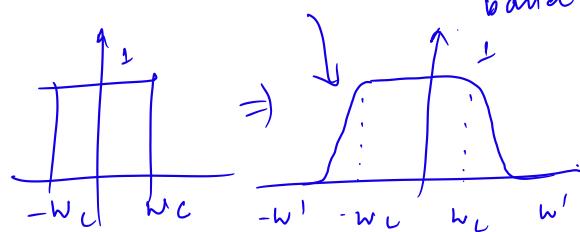
(iii) Band pass



(iv) Band stop (Notch filters)

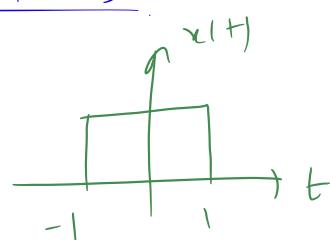


Practical filters =
Transition band



Part II tutorials

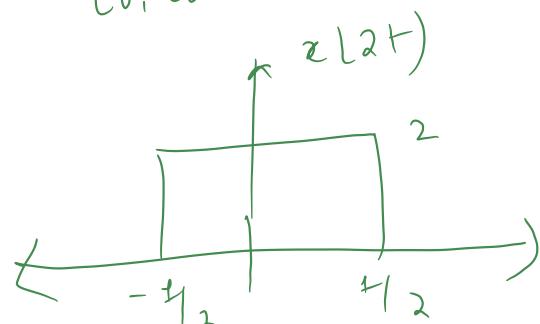
(eg) $n(t)$:



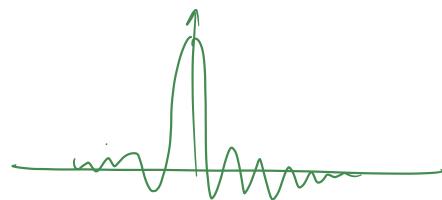
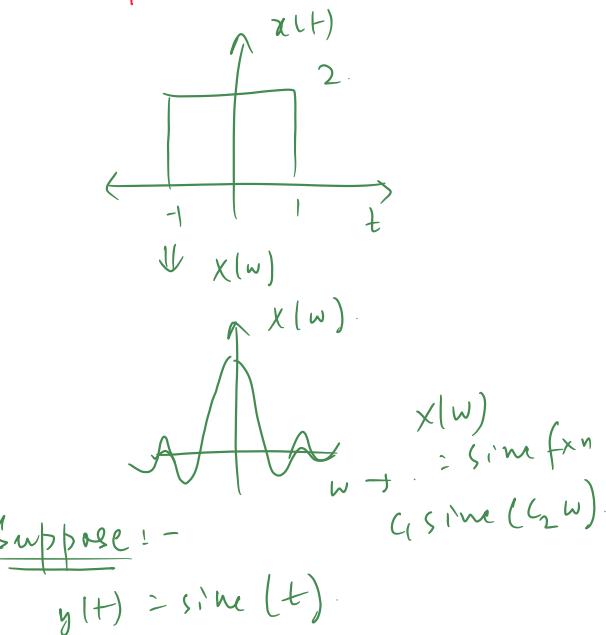
$a = 2^{-1}$
 $n(2t) :- \quad \hookrightarrow n(2t) = \begin{cases} 2, & \text{if } -1 \leq t \leq 1 \\ 0, & \text{else} \end{cases}$

$$n(2t) = \begin{cases} 2, & \text{if } -1 \leq 2t \leq 1 \\ 0, & \text{else} \end{cases}$$

$$= \begin{cases} 2, & \text{if } -\frac{1}{2} \leq t \leq \frac{1}{2} \\ 0, & \text{else} \end{cases}$$

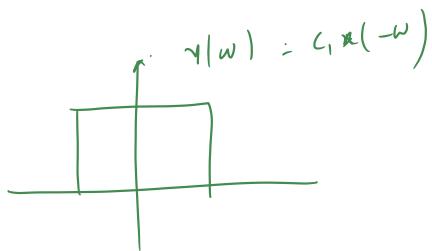


(e.g) Example of Freq. Duality

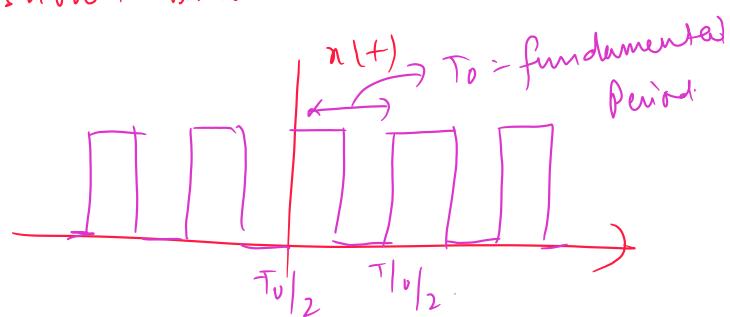


$$\downarrow Y(w)$$

$$= x(-w)$$



(3) Consider the periodic square wave shown below.



a) find complex exponential Fourier series of $x(t)$

b) find the trigonometric Fourier series of $x(t)$

$\xrightarrow{\text{SOMY}}$

a) $x(t) = \sum_{n=-\infty}^{\infty} c_n e^{jk w_0 t}$ j $w_0 = 2\pi/T_0$

$$c_k = \frac{1}{T_0} \int_0^{T_0} u(t) e^{-jk\omega_0 t} dt$$

$$= \frac{1}{T_0} \int_0^{T_0/2} u(t) e^{-jk\omega_0 t} dt + \frac{1}{T_0} \int_{T_0/2}^{T_0} u(t) e^{-jk\omega_0 t} dt$$

$$\geq \frac{1}{T_0} \int_0^{T_0/2} A e^{-jk\omega_0 t} dt$$

$$\geq \frac{A}{T_0} \int_0^{\frac{T_0}{2}} e^{-jk\omega_0 t} dt$$

$$\geq \frac{A}{T_0} \left[\frac{e^{-jk\omega_0 t}}{-jk\omega_0} \right]_0^{T_0/2}$$

$$= \frac{A}{jk\omega_0 T_0} \left\{ 1 - e^{-jk\omega_0 \frac{T_0}{2}} \right\}$$

$$\omega_0 > 2\pi/T_0 \Rightarrow \omega_0 T_0 = 2\pi$$

$$= \frac{A}{jk2\pi} \left\{ 1 - e^{-jk\pi/2} \right\}$$

$$= \frac{A}{jk2\pi} \left\{ 1 - (-1)^n \right\},$$

$k \neq 0$ (i) k is even ($k = 2m$), $m \in \mathbb{Z}$

$$\Rightarrow (-1)^k = 1 \Rightarrow c_k = 0$$

(ii) k is odd $\Rightarrow k = 2m+1$, $m \in \mathbb{Z}$

$$(-1)^k = -1 \Rightarrow c_k = \frac{A}{jk\pi}$$

$$c_k = \frac{A}{jk\pi} \left(1 - (-1)^n \right) \Big|_{n=0}$$

$(0/0)$ form \rightarrow $\frac{N_0}{jk\pi}$ applicable

$$c_n = \frac{1}{T_0} \int_0^{T_0} x(t) e^{-jk\omega_0 t} dt$$

$$c_0 = \frac{1}{T_0} \int_0^{T_0} x(t) dt = \frac{1}{T_0} \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} x(t) dt$$

$$= \frac{A}{T_0} \cdot T_0 / 2 = A/2$$

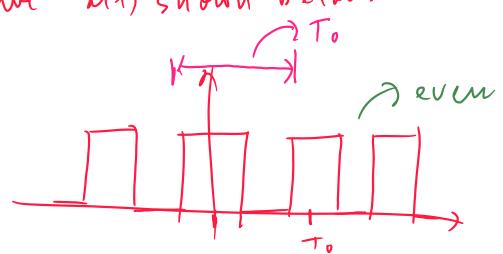
Summarising

$$\omega = \frac{\pi}{T_0} \quad c_{2m} = 0 \quad ; \quad c_{2m+1} = \frac{A}{j(2m+1)\pi}$$

$$x(t) = \frac{A}{2} + \frac{A}{j\pi} \sum_{m=-\infty}^{\infty} \frac{1}{(2m+1)} e^{j(2m+1)\omega_0 t}$$

$$\omega_p = 2\pi/T_0$$

② Consider the periodic square wave $x(t)$ shown below:-



a) Find complex and trigonometric Fourier series of $x(t)$

$$S_{FT} = \sum_{n=-\infty}^{\infty} c_n e^{jn\omega_0 t} ; \omega_0 = 2\pi/T_0$$

$$c_n = \frac{1}{T_0} \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} x(t) e^{-jn\omega_0 t} dt$$

$$= \frac{1}{T_0} \int_{-\frac{T_0}{2}}^0 0 dt + \frac{1}{T_0} \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} A e^{-jn\omega_0 t} dt$$

$$+ \frac{1}{T_0} \int_{\frac{T_0}{2}}^{\frac{T_0}{2}} 0 dt$$

$$= A/T_0 \int_{-\frac{T_0}{2}}^{\frac{T_0}{2}} e^{-jn\omega_0 t} dt$$

$$= \frac{A}{T_0} \left[\frac{e^{-jn\omega_0 t}}{-jn\omega_0} \right]_{-\frac{T_0}{2}}^{\frac{T_0}{2}}$$

$$= \frac{A}{T_0} \frac{\left(e^{-jn\omega_0 \frac{T_0}{2}} - e^{jn\omega_0 \frac{T_0}{2}} \right)}{-jn\omega_0}$$

$$\geq \frac{A}{jk\omega_0 T_0} \left[e^{jk\omega_0 t_0} - e^{-jk\omega_0 t_0} \right] \quad \text{--- (1)}$$

$$\omega_0 = 2\pi/T_0,$$

$$\omega_0 T_0 > 2\pi \quad \text{--- (2)}$$

Put (2) in (1)

$$\begin{aligned} &\geq \frac{A}{jk2\pi} \left[e^{jk2\pi/2} - e^{-jk2\pi/2} \right] \\ &\geq \frac{A}{k\pi} \left[\frac{e^{j\pi/2} - e^{-j\pi/2}}{2j} \right] \\ &= \frac{A}{k\pi} \sin(\pi/2) \end{aligned}$$

Case 1 $k \neq 0$

If k is even : $k = 2m$

$$c_k - c_{2m} = \frac{A}{2m\pi} \sin\left(\frac{2m\pi}{2}\right) = 0$$

If k is odd : $\Rightarrow k = 2m+1$

$$c_k - c_{2m+1} = \frac{A}{(2m+1)\pi} \sin\left(\frac{(2m+1)\pi}{2}\right)$$

$$= \frac{A}{(2m+1)\pi} \sin\left(\frac{2m\pi + \pi}{2}\right)$$

$$= \frac{A}{(2m+1)\pi} \sin\left(\frac{2m\pi}{2}\right)$$

$$= \frac{A}{(2m+1)\pi} (-1)^m$$

$$c_0 = \frac{1}{T_0} \int_0^{T_0} x(t) dt = \frac{A}{2}$$

Substitute for all these values.

Complex exponential Fourier series

$$x(t) = A_0 + \frac{1}{\pi} \left[\sum_{m=-\infty}^{\infty} (-1)^m / (2m+1) e^{j(2m+1)\omega_0 t} \right]$$

$$\omega_0 = 2\pi/T_0$$

$$(ii) \frac{c_0}{2} - c_0 = A(\frac{1}{2}) \Rightarrow c_0 = A$$

$$a_{2m} = 2 \operatorname{Re} \{ c_{2m} \} = 0 ; m \neq 0$$

\uparrow
 $n+1$ Real

$$a_{2m+1} = 2 \operatorname{Re} \{ c_{2m+1} \}$$

$$= \frac{(-1)^m 2A}{(2m+1)\pi}$$

$$b_{2m} = -2 \operatorname{Im} \{ c_{2m} \} = 0 \quad \left. \right\} b_n = 0$$

$$b_{2m+1} = -2 \operatorname{Im} \{ c_{2m+1} \} = 0 \quad \left. \right\} b_n \neq 0$$

$$\text{Result: } x(t) = \frac{a_0}{2} + \sum_{n=1}^{\infty} a_n \cos(n\omega_0 t) + \sum_{n=1}^{\infty} b_n \sin(n\omega_0 t) \quad \text{as } b_n = 0 \forall n$$

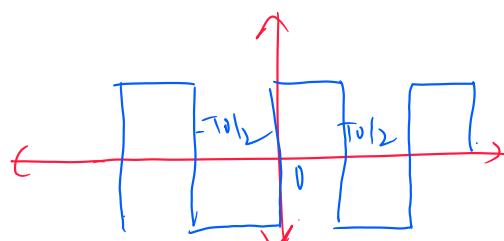
$$= A_0 + \sum_{m=1}^{\infty} a_{2m+1} \cos((2m+1)\omega_0 t)$$

$$= A_0 + \frac{2A}{\pi} \sum_{m=0}^{\infty} \frac{(-1)^m}{(2m+1)} \cos((2m+1)\omega_0 t)$$

$$= A_0 + \frac{2A}{\pi} \left\{ \cos \omega_0 t - \frac{1}{3} \cos 3\omega_0 t + \frac{1}{5} \cos 5\omega_0 t - \dots \right\}$$

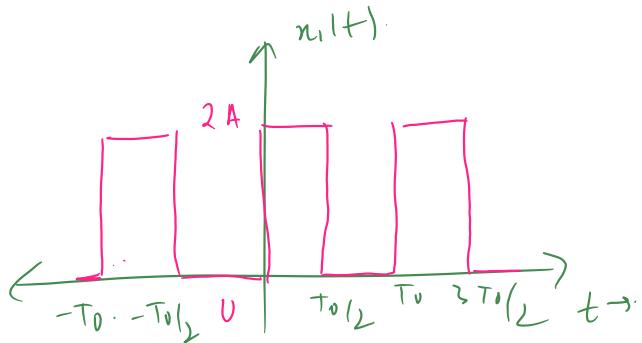
$\therefore x(t)$ contains odd harmonics.

③ Consider the periodic square wave shown below:-

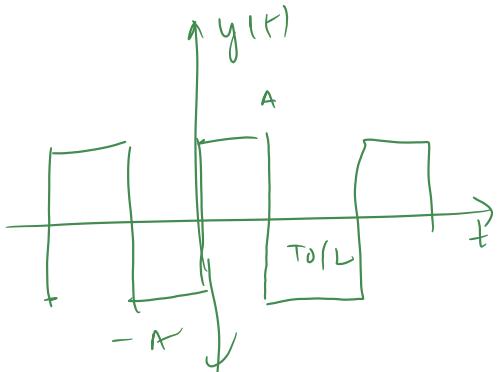


a) find complex exponential and trigonometric expansion of $x(t)$.

Sol: Consider the signal $x_1(t)$ as



$$\text{Let } y_1(t) = n_1(t) - A$$



W.L.O.G. consider $y_1(t) = x_1(t)$

$$x_1(t) = x_1(t) - A$$

\therefore From 1st question, Replace
A by $2A$.

\therefore FS expansion of $x_1(t)$

$$A + \frac{2A}{j\pi} \sum_{m=-\infty}^{\infty} \frac{1}{2m+1} e^{j(2m+1)\omega_0 t} ; \omega_0 = \frac{2\pi}{T_0}$$

\therefore Put (2) in (1)

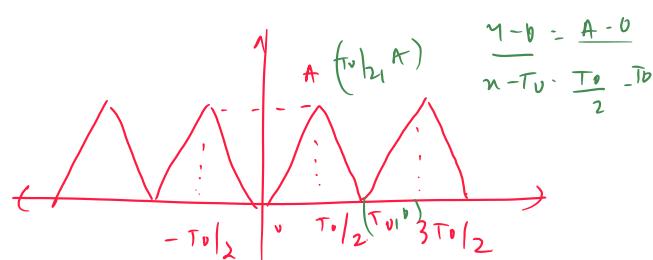
$$x_1(t) = \frac{2A}{j\pi} \sum_{m=-\infty}^{\infty} \frac{1}{2m+1} e^{j(2m+1)\omega_0 t} ; \omega_0 = \frac{2\pi}{T_0}$$

Complex FS expansion
of $x_1(t)$

Similarly, the Trigonometric FS of
 $x_1(t)$

$$x_1(t) = \frac{4A}{\pi} \sum_{m=0}^{\infty} \frac{1}{(2m+1)} \sin((2m+1)\omega_0 t) ; \omega_0 = 2\pi/T_0$$

(ii) Consider the triangular wave
 $x_1(t)$ as shown below:

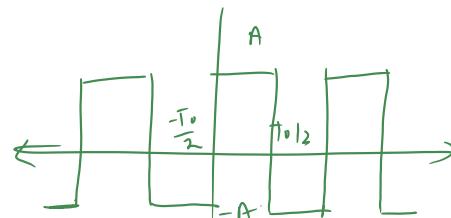


- a) Find complex exponential F.S.
 b) Find trigonometric F.S. of $x(t)$.

Soln Consider $x'(t) = \frac{d}{dt} x(t)$

$$x'(t) = \begin{cases} \frac{2A}{T_0} t & ; 0 \leq t \\ \frac{2A}{T_0} (T_0 - t) & ; \frac{T_0}{2} \leq t \leq T_0 \end{cases}$$

$$x'(t) = \begin{cases} 2A/T_0 & , 0 \leq t \leq T_0/2 \\ -2A/T_0 & , T_0/2 \leq t \leq T_0 \end{cases}$$



F.S. of $x'(t)$ from Q.3. when

$$A \rightarrow 2A/T_0 \xrightarrow{\omega} x'(t) = \frac{4A}{j\pi T_0} \sum_{m=-\infty}^{\infty} \frac{1}{2m+1} e^{-j(2m+1)\omega_0 t} ; \omega_0 = 2\pi/T_0 \quad (1)$$

Now actual F.S. of $x(t)$.

$$x(t) = \sum_{n=-\infty}^{\infty} c_n e^{jk\omega_0 t} ; \omega_0 = 2\pi/T_0$$

$$\therefore \frac{d x(t)}{dt} = \sum_{n=-\infty}^{\infty} c_n j k \omega_0 e^{jk\omega_0 t} \quad (2)$$

$$\text{But } x'(t) = \frac{d}{dt} x(t).$$

$$\therefore \text{Compare (1) \& (2)}$$

$$\text{When } k = 2m, \neq 0 \Rightarrow c_n j k \omega_0 = 0$$

$$\Rightarrow c_n = 0.$$

When $k = 2m + 1$:-

$$c_n j k \omega_0 = \frac{4A}{j\pi T_0 \pi} \cdot \frac{1}{k}$$

$$\Rightarrow C_K = \frac{-\frac{4A}{\kappa^2 T_0 w_0}}{\kappa^2 - 2\pi^2} = \frac{-\frac{4A}{\kappa^2 T_0 w_0}}{\kappa^2 - \frac{4\pi^2}{T_0^2}} = \frac{-2A}{\kappa^2 \pi^2}$$

$$w_0 = 2\pi/T_0$$

$$\Rightarrow w_0 T_0 = 2\pi$$

$$C_0 = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} a(t) dt = A/2$$

$\overbrace{\text{Collect the above values and}}$

we get (A) :-

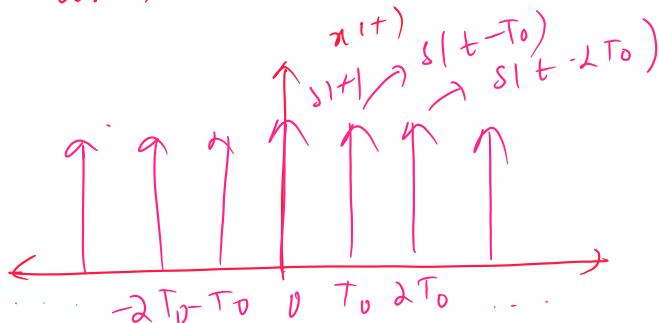
$$a(t) = \frac{A}{2} + \frac{2\pi}{\pi^2} \sum_{m=-\infty}^{\infty} \frac{1}{(2m+1)^2} e^{-j(2m+1)w_0 t}$$

Now for trigonometric F.S:

$$n(t) = \frac{A}{2} + \frac{4A}{\pi^2} \sum_{m=0}^{\infty} \frac{1}{(2m+1)^2} \times \left(\frac{e^{j(2m+1)w_0 t}}{w_0 t} \right)$$

(Q3) Consider periodic impulse

train as shown below:-



a) Find complex exponential

F.S

b) Find trigonometric exponential

F.S.

$$\stackrel{\text{Sol'n}}{=} a) n(t) = \sum_{n=-\infty}^{\infty} c_n e^{jnkw_0 t} ; w_0 = 2\pi/T_0$$

$$c_n = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} n(t) e^{-jnkw_0 t} dt$$

$$= \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} g(t) e^{-j k w_0 t} dt$$

$$= \frac{1}{T_0} e^{j 0} = \frac{1}{T_0}$$

$$\therefore \boxed{n(t) = \frac{1}{T_0} \sum_{n=-\infty}^{\infty} e^{j n k w_0 t}; w_0 = 2\pi/T_0}$$

$$(ii) x(t) = \frac{a_0}{2} + \sum_{k=1}^{\infty} a_k \cos(k w_0 t) + b_k \sin(k w_0 t)$$

$$a_k = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} (x(t)) \cos(k w_0 t) dt$$

$$= \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} g(t) \cos(k w_0 t) dt$$

$$= \frac{1}{T_0} \times 2 \int_{0}^{T_0/2} g(t) \cos(k w_0 t) dt$$

$$a_0 = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) dt = \frac{1}{T_0}$$

$$b_k = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) \sin(k w_0 t) dt$$

$$= \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} g(t) \sin(k w_0 t) dt$$

$$= \frac{1}{T_0} \sin(0) = 0$$

Collecting values for a_0, a_n, b_n in A

$$x(t) = \frac{1}{T_0} + \frac{2}{T_0} \sum_{k=1}^{\infty} \cos(k w_0 t);$$

$$w_0 = 2\pi/T_0$$

Trigonometric F-s