

## TA Session 06

### Part I . Summary of Week 6 lectures.

- (1) examples on Z transform.
- (2) implement Z transform using contour integration Method.

$x(z) \leftarrow Z\text{transform then}$

$$x(n) = \frac{1}{2\pi j} \oint_{\Gamma} x(z) z^{n-1} dz$$

$\Gamma \rightarrow$  closed contours in anti clockwise direction enclosing all poles of  $x(z) z^{n-1}$

$$\therefore x(n) = \sum_{i=1}^P \text{Res}_{z \rightarrow p_i} [x(z) z^{n-1}]$$

= sum of residuals evaluated at each pole of  $x(z) z^{n-1}$ .

- (3) Fourier series of continuous time

signals

→ Representation of the continuous time domain signal in the frequency or spectral domain.

→ For periodic signals.

- (i) complex exponential Representation

$x(t) \rightarrow$  periodic in  $T_0$

then

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{jk\omega_0 t} \quad ; \quad \omega_0 = 2\pi/T_0$$

$$c_k = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) e^{-jk\omega_0 t} dt$$

↓  
Fourier coefficients

$$c_0 = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) dt \leftarrow \text{avg of } x(t) \text{ over a period} \\ \downarrow \\ \text{dc component}$$

(\*) For real signals

$$c_k^* = c_{-k} \left[ \begin{array}{l} \text{conjugate} \\ \text{symmetry} \end{array} \right]$$

&  $|c_k| = |c_{-k}| \leftarrow \text{magnitude of Fourier coefficient have even symmetry}$

$\angle c_k = -\angle c_k + \text{phase of Fourier coefficient have odd symmetry.}$

## (2) Trigonometric Representation.

$$x(t) = \frac{a_0}{2} + \sum_{n=1}^{\infty} (a_n \cos(k\omega_0 t) + b_n \sin(k\omega_0 t))$$

$$a_k = \frac{2}{T_0} \int_{-T_0/2}^{T_0/2} x(t) \cos(k\omega_0 t) dt$$

$$b_k = \frac{2}{T_0} \int_{-T_0/2}^{T_0/2} x(t) \sin(k\omega_0 t) dt$$

## Relation b/w Trigonometric & Complex Exp representation

$$\begin{aligned} a_0 &= 2c_0 \\ a_k &= c_k + c_{-k} \\ b_k &= j(c_k - c_{-k}) \end{aligned}$$

$$\begin{aligned} c_0 &= a_0/2, \quad c_k = \frac{a_k - jb_k}{2} \\ c_{-k} &= \frac{a_k + jb_k}{2} \end{aligned}$$

If  $x(t)$  is real,  $a_k$  &  $b_k$  are real!

$$a_k = 2 \operatorname{Re} \{ c_k \}$$

$$b_k = -2 \operatorname{Im} \{ c_k \}$$

(\*) For even signals:

$$\text{i.e. } x(-t) = x(t)$$

$$x(t) = \frac{a_0}{2} + \sum_{k=1}^{\infty} (a_k \cos(k\omega_0 t) + b_k \sin(k\omega_0 t))$$

$$x(-t) = \frac{a_0}{2} + \sum_{k=1}^{\infty} (a_k \cos(k\omega_0 t) - b_k \sin(k\omega_0 t))$$

$$\neq x(t)$$

$\therefore b_k = 0 \quad \{ \text{as they are coefficients of } \sin(k\omega_0 t) \text{ odd signal} \}$

(x) for odd signals

$$a_k = 0$$

(3) Conditions for existence of Fourier Series

(i)  $x(t)$  is absolutely integrable

$$\text{over } T_0 \quad \int_{T_0} |x(t)| dt < \infty$$

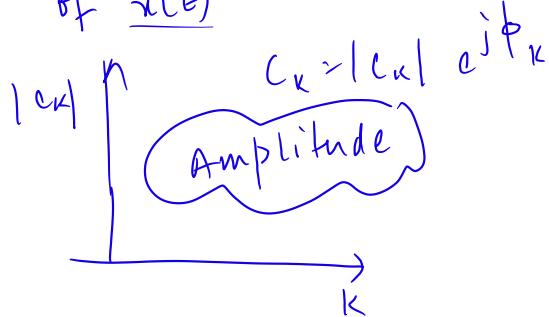
(ii)  $x(t)$  has finite # of maxima/minima in any finite interval

(iii)  $x(t)$  has finite # of discontinuities & there should be finite

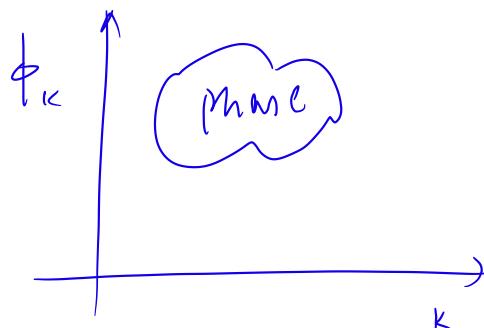
  
Dirichlet conditions:  
 $\Downarrow$

They are "sufficient" for existence of Fourier Series but not necessary

(4) Amplitude and Phase spectrum  
of  $x(t)$

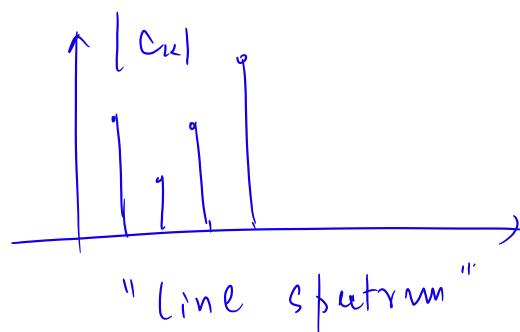


$$(w = kw_0)$$



$$(w = kw_0)$$

Discrete  $\rightarrow$



(5) Parseval's Theorem.

$$\lim_{T_0 \rightarrow \infty} \frac{1}{T_0} \int_{-T_0}^{T_0} |x(t)|^2 dt$$

$\stackrel{\text{Power of C.T. signal.}}{\sim}$

$$= \sum_{k=-\infty}^{\infty} |c_k|^2$$

Power of the signal remains equal  
in both time domain and  
frequency domain representation

$c_k$  as a fn of  $k$  is a discrete  
sequence.

$$\left\{ \dots, c_{-1}, c_0, c_1, c_2, \dots \right\} \leftarrow$$

$\hookrightarrow \sum_{i=-\infty}^{\infty} |c_i|^2 = \text{Power of discrete seq.}$

$x(t)$  is a real signal:

$$\begin{aligned} c_n^* &= c_{-n} \\ c_k &= \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) e^{-jk\omega_0 t} dt \\ c_k^* &= \left( \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) e^{-jk\omega_0 t} dt \right)^* \\ &= \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} \left( x(t) e^{-jk\omega_0 t} \right)^* dt \\ &= \frac{1}{T_0} \int_{-T_0}^{T_0} x^*(t) e^{jk\omega_0 t} dt \\ &= \frac{1}{T_0} \int_{-T_0}^{T_0} x(t) e^{jk\omega_0 t} dt \\ &\quad \text{Real signal} \\ &\approx \frac{1}{T_0} \int_{-T_0}^{T_0} x(t) e^{-j(-k)\omega_0 t} dt = c_{-k}. \\ \Rightarrow c_k^* &= c_{-k}. \quad (|c_k| = |c_k^*| = |c_{-k}|). \\ \angle c_k^* &= -\angle c_k. \\ \hookrightarrow \angle c_k &= -\angle c_k. \end{aligned}$$

## Part II Tutorials:

① Determine the inverse Z-transform

$$\text{of } X(z) = \frac{1}{1 - 0.8z^{-1} + 0.12z^{-2}}$$

- a) if ROC is  $|z| > 0.6$  { ROC in exterior part of circle}
- b) if ROC is  $|z| < 0.2$
- c) if ROC is  $0.2 < |z| < 0.6$ 
  - { ROC in interior part of circle}
  - { ROC in an annular region}

Sol<sup>n</sup> do the partial fraction expansion of  $X(z)/z$

$$\begin{aligned} X(z) &= \frac{1}{1-0.8z^{-1}+0.12z^{-2}} \\ &= \frac{1}{z^{-2}(z^2 - 0.8z + 0.12)} \\ &= \frac{z^2}{(z^2 - 0.8z + 0.12)} \\ &= \frac{z^2}{(z-0.6)(z-0.2)} \end{aligned}$$

$$\Rightarrow \frac{X(z)}{z} = \frac{z}{(z-0.6)(z-0.2)} ; P_1 = 0.6 \quad P_2 = 0.2$$

$$\frac{X(z)}{z} = \frac{\lambda_1}{z-0.6} + \frac{\lambda_2}{z-0.2} \quad \text{simple poles.}$$

$$\lambda_1 = (z-0.6) \left. \frac{X(z)}{z} \right|_{z=0.6}$$

$$= \left. \frac{z}{z-0.2} \right|_{z=0.6}$$

$$= \frac{0.6}{0.4} = 1.5$$

$$\lambda_2 = (z-0.2) \left. \frac{X(z)}{z} \right|_{z=0.2} = -0.5$$

$$\frac{X(z)}{z} = \frac{1.5}{z-0.6} - \frac{0.5}{z-0.2}$$

$$X(z) = 1.5 \frac{z}{z-0.6} - \frac{0.5z}{z-0.2}$$

(i) ROC  $|z| > 0.6 \rightarrow$  exterior point

of circle  $|z| = 0.6$

$\therefore x(n)$  should be right

handed signal

$$x(n) = 1.5 (0.6)^n u(n) - 0.5 (0.2)^n u(n)$$

(ii) ROC  $|z| < 0.2 \rightarrow$  interior point of circle  $|z|=0.2$

$\Rightarrow x(n)$  should be left sided!

$$x(n) = -1.5 \left(0.6\right)^n u(-n-1) + 0.5 \left(0.2\right)^n u(-n-1).$$

(iii) ROC  $0.2 < |z| < 0.6 \Rightarrow$  annular region.

$\Rightarrow x(n)$  should be two sided!

$$x(n) = x_1(n) + x_2(n)$$

$$\text{where } x_1(n) = \text{IZT of } \left(\frac{1.5}{z-0.6}\right)$$

$$x_1(n) = 12 + \left(\frac{-0.5}{z-0.2}\right).$$

$$\text{for } \frac{1.5}{z-0.6} ; |z| < 0.6$$

$$\Rightarrow x_1(n) = -1.5 \left(0.6\right)^n u(-n-1)$$

$$\text{for } \frac{-1.5}{z-0.2} ; |z| > 0.2$$

$$x_2(n) = -0.5 \left(0.2\right)^n u(n)$$

$$\therefore x(n) = -1.5 \left(0.6\right)^n u(-n-1) - 0.5 \left(0.2\right)^n u(n)$$

■

② Determine the complex exponential Fourier series representation for

each of the following signals:

a)  $x(t) = \cos(\omega_0 t)$

b)  $x(t) = \sin(\omega_0 t)$

c)  $x(t) = \cos(\alpha t + \pi/4)$

d)  $x(t) = \cos(\omega t) + \cos(\beta t)$

e)  $x(t) = \sin^2(\omega t)$

Solution  $x(t) = \cos(\omega_0 t)$

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{j k \omega_0 t} \quad \text{--- (1)}$$

$$= e^{j \omega_0 t} + e^{-j \omega_0 t} \quad \begin{bmatrix} \text{Euler's formula} \\ \text{--- (2)} \end{bmatrix}$$

Equate ① & ②

$$\frac{e^{jw_0 t} + e^{-jw_0 t}}{2} = \sum_{k=-\infty}^{\infty} c_k e^{jk w_0 t}$$

Comparing the coefficients for every term:

$$c_0 = \frac{1}{2}, c_1 = \frac{1}{2}, c_k = 0 \quad (k \neq 1)$$

$$\begin{aligned} & \rightarrow \frac{1}{2} e^{jw_0 t} + \frac{1}{2} e^{-jw_0 t} \\ &= \dots + c_{-3} e^{-j3w_0 t} + c_{-2} e^{-j2w_0 t} \\ & \quad + c_{-1} e^{-jw_0 t} + c_0 + c_1 e^{jw_0 t} \\ & \quad + c_2 e^{j2w_0 t} + \dots \end{aligned}$$

$$e^{jw_0 t} = 2 \quad c_0 = \frac{1}{2}, \quad c_2 = 0,$$

$$c_{-1} = -\frac{1}{2}.$$

$$\begin{aligned} (b) \quad n(t) &= \sin(w_0 t) \\ &= \frac{1}{2j} (e^{jw_0 t} - e^{-jw_0 t}) \\ &\rightarrow \text{Euler's Formula.} \end{aligned}$$

∴ similar to part (a)

$$c_1 = \frac{1}{2j}$$

$$c_{-1} = -\frac{1}{2j} \quad c_k = 0 \quad (k \neq 1)$$

$$i) \quad n(t) = \sin(2t + \pi/4)$$

$$\begin{aligned} w_0 &= 2 \\ \therefore n(t) &= \sum_{n=-\infty}^{\infty} c_n e^{jn2t} \quad \text{--- (1)} \end{aligned}$$

$$n(t) = \text{Im } [2t + \pi/4] = \frac{e^{j(2t + \pi/4)} - e^{-j(2t + \pi/4)}}{2}$$

$$\begin{aligned}
 &= \frac{1}{2} e^{j\omega t} e^{j\pi/4} + \frac{1}{2} e^{j\omega t} e^{-j\pi/4} \\
 &= \frac{1}{2} \left( \frac{1}{\sqrt{2}} + j \frac{1}{\sqrt{2}} \right) e^{j\omega t} + \frac{1}{2} \left( \frac{1}{\sqrt{2}} - j \frac{1}{\sqrt{2}} \right) e^{-j\omega t} \\
 &\quad \text{--- (3)}
 \end{aligned}$$

Comparing coefficients  $u/w(1) \& (2)$

$$\begin{aligned}
 c_0 &= \frac{1}{2} \left( \frac{1}{\sqrt{2}} + j \frac{1}{\sqrt{2}} \right) \\
 c_1 &= \frac{1}{2} \left( \frac{1}{\sqrt{2}} - j \frac{1}{\sqrt{2}} \right)
 \end{aligned}$$

$$c_k = 0, |k| \neq 1$$

$$d) x(t) = cu_1t + cu_2t$$

$$\begin{aligned}
 T_1 : \text{period of } cu_1t &= \frac{2\pi}{u_1} \\
 &= \frac{\pi}{2}
 \end{aligned}$$

$$cu_1(t+T) = cu_1(u_1t + 4T) = cu_1u_1t$$

$$u_1T = n2\pi$$

$$T = n \cdot \frac{2\pi}{u_1}, n \in \mathbb{Z}$$

$$n=1; \text{Fundamental Period} \\ = \frac{2\pi}{u_1}$$

$$\begin{aligned}
 T_2 : \text{period of } cu_2t &= \frac{2\pi}{u_2} \\
 &= \frac{1}{12} \text{ LCM of } \left( \frac{2\pi}{u_1} \times 12, \frac{2\pi}{u_2} \times 12 \right) \\
 &= \frac{1}{12} \text{ LCM of } (6\pi, 4\pi). \\
 &= \frac{1}{12} \pi \text{ LCM of } (6, 4). \\
 &= \frac{1}{12} \cdot 12\pi \\
 &= \pi
 \end{aligned}$$

$$\begin{aligned}
 \therefore \text{Fundamental Period of } x(t) &= cu_1u_1t + cu_2u_2t \\
 &= \pi
 \end{aligned}$$

$$\therefore \text{Fundamental freq} = \omega_0$$

$$\begin{aligned} &= 2\pi f_0 \\ &= 2\pi/T_0 \\ &= 2\pi/\pi \\ &= 2 \\ &= \boxed{\omega_0 = 2} \end{aligned}$$

$$x(t) = \sum_{k=-\infty}^{\infty} c_k e^{j2kt} \quad \longrightarrow \textcircled{1}$$

$$\begin{aligned} x(t) &= \cos \omega t + j \sin \omega t \\ &= \frac{e^{j\omega t} + e^{-j\omega t}}{2} + \frac{e^{j6t} + e^{-j6t}}{2} \\ &= \frac{e^{-j6t}}{2} + \frac{e^{-j4t}}{2} + \frac{e^{j4t}}{2} + \frac{e^{j6t}}{2} \end{aligned} \quad \longrightarrow \textcircled{2}$$

Comparing (1) & (2) -

$$\left\{ \begin{array}{l} c_{-3} = \frac{1}{2}, \quad c_{-2} = \frac{1}{2} \\ c_2 = \frac{1}{2}, \quad c_3 = \frac{1}{2} \end{array} \right\}$$

i)  $x(t) = \sin^2(t)$

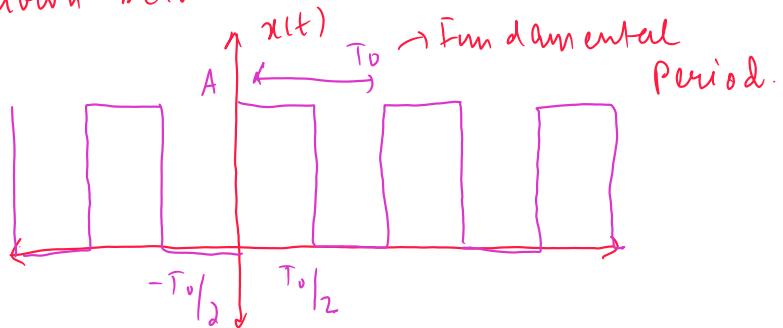
$$\sin^2(t) = \frac{1 - \cos 2t}{2} \quad \hookrightarrow \boxed{\omega_0 = 2}$$

$$\begin{aligned} x(t) &= \sum_{n=-\infty}^{\infty} c_n e^{jn2t} \\ \text{Now } x(t) &= \sin^2(t) \\ &= \left( \frac{e^{jt} - e^{-jt}}{2j} \right)^2 \\ &= \frac{e^{j2t} + e^{-j2t} - 2}{-4} \\ &= -\frac{1}{4} e^{j2t} - \frac{1}{4} e^{-j2t} + \frac{1}{2} \end{aligned}$$

$$\begin{aligned} c_0 &= -\frac{1}{4}, \quad c_1 = -\frac{1}{4} \\ c_2 &= \frac{1}{2} \end{aligned}$$

$$c_k = 0 \quad |k| \neq 0$$

③ Consider the periodic square wave as shown below:



a) Determine the complex exponential Fourier series of  $x(t)$

b) Determine the trigonometric Fourier series of  $x(t)$ -

$$\begin{aligned}
 \text{Sol}^n \quad a) \quad x(t) &= \sum_{k=-\infty}^{\infty} c_k e^{j k \omega_0 t} \\
 \omega_0 &= 2\pi/T_0 \\
 c_k &= 1/T_0 \int_0^{T_0} x(t) e^{-j k \omega_0 t} dt \\
 &= 1/T_0 \int_0^{T_0/2} A e^{-j k \omega_0 t} dt \\
 &= \frac{A}{T_0} \int_0^{T_0/2} e^{-j k \omega_0 t} dt \\
 &= \frac{A}{T_0} \left[ \frac{1}{-jk\omega_0} e^{-jk\omega_0 t} \Big|_0^{T_0/2} \right] \\
 &= \frac{A}{T_0} \left[ \frac{-1}{jk\omega_0} \left( e^{-jk\omega_0 T_0/2} - 1 \right) \right] \\
 &= \frac{A}{T_0} \left[ \frac{1 - e^{-jk\omega_0 T_0/2}}{jk\omega_0} \right]
 \end{aligned}$$

$$\omega_0 = 2\pi/T_0 \Rightarrow \omega_0 T_0 = 2\pi$$

$$= \frac{A}{T_0} \left[ \frac{1 - e^{-jk\omega_0 T_0/2}}{jk\omega_0} \right]$$

$$= \frac{A}{jk2\pi} \left[ 1 - e^{-jk\pi} \right]$$

$$= \frac{A}{jk2\pi} \left[ 1 - (-1)^k \right] \quad \begin{aligned} & \because e^{j\pi} \\ & = \cos(\pi) - j\sin(\pi) \\ & = -1 + 0 \\ & = -1 \end{aligned}$$

$\Rightarrow k \neq 0$  :-

(i)  $k$  is even :  $k = 2m ; m \in \mathbb{Z}$

$$\Rightarrow (-1)^k = 1 \Rightarrow c_k = 0$$

(ii)  $k$  is odd  $\Rightarrow k = 2m+1 ; m \in \mathbb{Z}$

$$(-1)^k = -1 \Rightarrow c_k = \frac{A}{jk\pi}$$

$$\stackrel{k=0}{=} c_k = \frac{A}{jk2\pi} \left( 1 - (-1)^k \right)$$

$\left(\frac{0}{0}\right)$  form  $\rightarrow$  <sup>NOE</sup> applicable

$$c_k = \frac{1}{T_0} \int_0^{T_0} x(t) e^{-jkw_0 t} dt$$

$$c_0 = \frac{1}{T_0} \int_0^{T_0} x(t) dt = \frac{1}{T_0} \int_0^{T_0} A dt$$

$$= \frac{1}{T_0} A \cdot T_0 / 2$$

$$= A/2$$

Summarising

$$c_0 = A/2, \quad c_{2m} = 0$$

$$c_{2m+1} = \frac{A}{j(2m+1)\pi}$$

$$x(t) = A/2 + \frac{A}{j\pi} \left\{ \sum_{m=-\infty}^{\infty} \frac{1}{2m+1} e^{j(2m+1)w_0 t} \right\}$$

where  $w_0 = 2\pi/T$

b)  $w \cdot k \cdot t$   $x(t)$  is real

$$a_0 = 2c_0 = 2 \cdot A/2 = A$$

$$a_{2m} = b_{2m} = 0 \quad ; \quad m \geq 0$$

$$a_{2m+1} = 2Rc \{ c_{2m+1} \} = 0$$

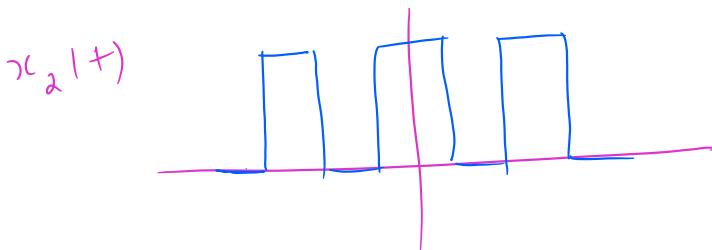
$$b_{2m+1} = -2gn \left\{ c_{2m+1} \right\} = \frac{2A}{(2m+1)\pi}$$

$$\therefore x(t) = A_0 + \sum_{m=0}^{\infty} \frac{1}{2m+1} \sin((2m+1)w_0 t)$$

$$\Rightarrow x(t) = \frac{A}{2} + \frac{2A}{\pi} \left( \sin w_0 t + \frac{1}{3} \sin 3w_0 t + \frac{1}{5} \sin 5w_0 t + \dots \right)$$

↳ 1st square pulse train π

contains odd harmonics



Square pulse train contains odd harmonics.