

## TA Session -03.

### Part I- Summary of week 3 lectures.

(1) Differential equation description

of LTI systems.

(2) i/p  $\rightarrow$  (p) relation represented as  
a differential equation

$$\sum_{m=0}^M a_m \frac{d^m y(t)}{dt^m} = \sum_{m=0}^N b_m \frac{d^m u(t)}{dt^m}$$

$$\text{So } y(t) = y_p(t) + y_h(t)$$

↑                    ↓  
Particular      homogeneous  
solution        solution.

1) Linear system - all initial conditions

$= 0$

2) Time Invariant : System has to  
be at INITIAL REST

$$\left. \begin{array}{l} (1) \text{ if } x(t) = 0 \text{ for } t \leq t_0 \\ \Rightarrow y(t) = 0 \text{ for } t \leq t_0 \end{array} \right\}$$

(2) Discrete LTI systems. (LSI)

$$x[n] \rightarrow \boxed{T(\cdot)} \rightarrow y[n]$$

Linear shift

$$s[n] \rightarrow \boxed{T(\cdot)} \rightarrow h[n]$$

Invariant

↓  
discrete time  
impulse response

$$y[n] = x[n] * h[n]$$

convolution

$$= \sum_{m=-\infty}^{\infty} x[m] h[n-m]$$

(\*) Memory (res.)

$$\text{if } h[n] = k s[n]$$

(\*) Causality

$$\text{if } h[n] = 0 \text{ for } n < 0$$

(\*) BIBO Stability

$$\text{if } \sum_{n=-\infty}^{\infty} |h[n]| < \infty \quad \begin{cases} \text{absolutely} \\ \text{summable} \end{cases}$$

$$\int |h(t)| dt < \infty$$

(\*)  $x(n) = z^n$  in the eigen function  
of discrete time LTI system.

(\*) Difference equation description:

$$\sum_{m=0}^M a_m y(n-m) = \sum_{m=0}^N b_m x(n-m)$$

$$\left[ \frac{dy^{(n)}}{dt^n} \rightarrow y(n-m) \right] \checkmark$$

### ③ Laplace Transform

↳ Represent a time domain signal in the "Laplace" Domain

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt$$

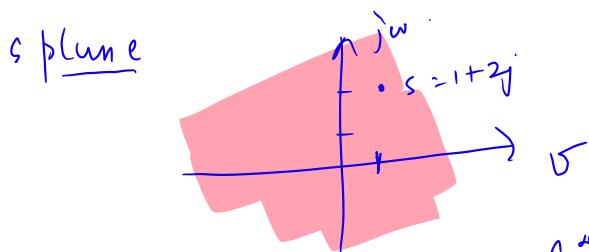
a complex number  
 $s = \sigma + j\omega$

$$x(s) \triangleq \mathcal{L}[x(t)], x(t) \xrightarrow{\mathcal{L}} x(s)$$

(\*) The transform is defined in terms of integral  $\rightarrow$  Integration should be convergent

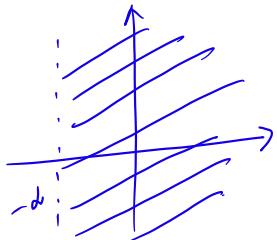
$\therefore$  LT is also characterised by its ROC

ROC: Region of s plane for which LT converges



Eg (1)  $n(t) = e^{-dt} u(t) \xrightarrow{\mathcal{L}} \int_0^{\infty} e^{-st} e^{-dt} u(t) dt = \int_0^{\infty} e^{-(s+d)t} dt$

$$x(s) = \frac{1}{s+d} : \text{Re}(s) > -d \quad t > -d$$



(\*) Two different signals can have same LT, but their ROCs can be different

(\*) Standard LTs

a)  $x(t) = \delta(t)$

$X(s) = 1$ , ROC everywhere

b)  $x(t) = u(t)$

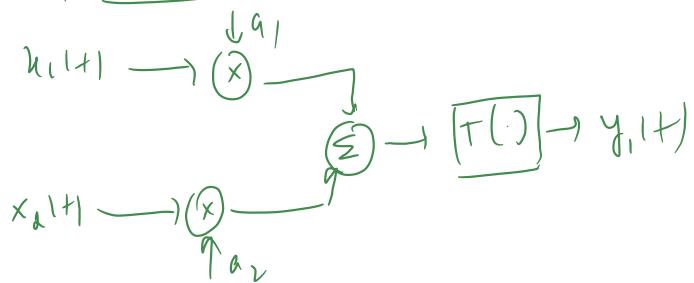
$X(s) = 1/s$ , ROC  $\text{Re}\{s\} > 0$

## Part II Tutorials

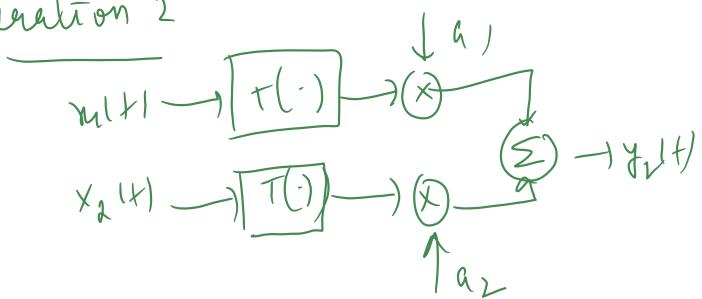
① Check whether

$y(t) = 2x(t) - 5$  is a valid op of a linear system

Operation 1

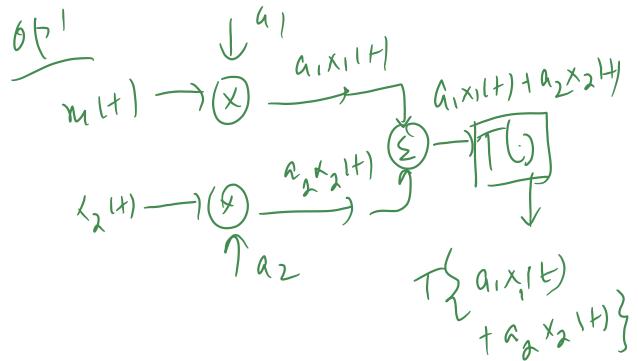


Operation 2



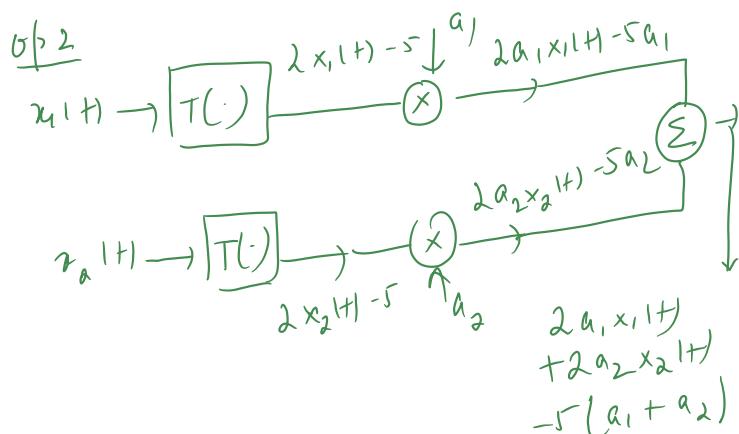
if  $y_1(t) = y_2(t)$

$T(\cdot)$  is a linear system



$$= 2 \{ a_1 x_1(t) \\ + a_2 x_2(t) \} - 5$$

$$= 2 a_1 x_1(t) \\ + 2 a_2 x_2(t) - 5$$



$y_1(t) \neq y_2(t)$   
Not linear

⑧ Perform convolution of the following signals.

$$x_1(t) = e^{-2t} u(t)$$

$$x_2(t) = e^{-5t} u(t)$$

$$\begin{aligned} \text{Soln } x_1(t) &\stackrel{\Delta}{=} x_1(t) * x_2(t) \\ &= \int_{-\infty}^{\infty} x_1(z) x_2(t-z) dz \\ &= \int_{-\infty}^{\infty} e^{-2z} u(z) e^{-5(t-z)} u(t-z) dz \\ &\quad \downarrow \quad \downarrow \\ &\quad \text{for } z \geq 0 \quad \text{for } t-z \geq 0 \\ &= \int_0^t e^{-2z} e^{-5(t-z)} dz \\ &= \int_0^t e^{-2z} e^{-5t} e^{5z} dz \end{aligned}$$

$$\begin{aligned}
 &= e^{-5t} \int_0^t e^{-2z} e^{-5z} dz \\
 &= e^{-5t} \int_0^t e^{3z} dz \\
 &= e^{-5t} \left[ \frac{e^{3z}}{3} \right] \Big|_0^t \\
 &= e^{-5t} \left( \frac{e^{3t} - 1}{3} \right) \\
 n_3(t) = & \begin{cases} \frac{e^{-2t} - e^{-5t}}{3}, & t \geq 0 \\ 0, & \text{else} \end{cases} \\
 &= e^{-5t} \left( \frac{e^{3t} - 1}{3} \right) u(t)
 \end{aligned}$$

### ③ Convolve

$$n_1(t) = \text{unit } u(t)$$

$$n_2(t) = t u(t)$$

$$\begin{aligned}
 n_3(t) &= n_1(t) * n_2(t) \\
 &= \int_{-\infty}^{\infty} u_1(z) u_2(t-z) dz \\
 &= \int_{-\infty}^{\infty} u(z) u(t-z) dz \\
 &\quad \text{for } z \geq 0 \quad \text{for } t-z \geq 0 \\
 &\quad \quad \quad t \geq z \\
 &\quad \text{or } z \leq t
 \end{aligned}$$

$$\begin{aligned}
 &= \int_0^t u(z)(t-z) dz \\
 &= t \int_0^t u(z) dz - \int_0^t z u(z) dz \\
 &= t \sin t \Big|_0^t - \int_0^t z \sin z dz \\
 &= t \{ \sin t - \cos t \} - \int_0^t z \sin z dz \\
 &= t \sin t
 \end{aligned}$$

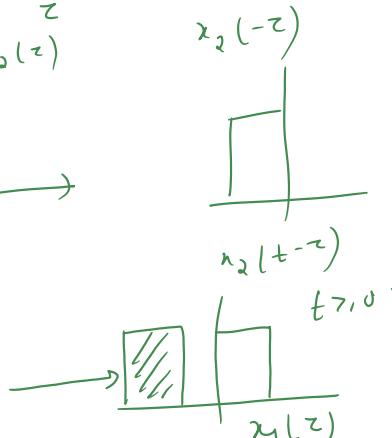
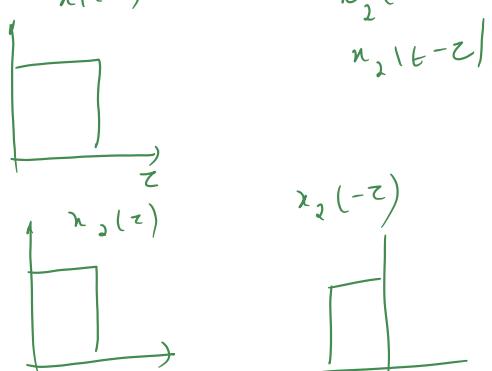
$$\begin{aligned}
 & \int_0^t z \cos z \, dz \rightarrow \text{Integrate by parts} \\
 & -z \int \cos z \, dz \Big|_0^t + \int \left( \frac{d}{dz} z \int \cos z \, dz \right) dz \\
 & = t \sin t - \int_0^t \sin z \, dz \\
 & = t \sin t + \cos z \Big|_0^t \\
 & = t \sin t + (\cos t - 1) \\
 & x_1(t) = t \sin t - \left\{ t \sin t + \cos t - 1 \right\} \\
 & = 1 - \cos t, \quad t \geq 0 \\
 \Rightarrow & \boxed{x_1(t) = (1 - \cos t) u(t)}
 \end{aligned}$$

Another interpretation

$$y(t) = \int_{-\infty}^{\infty} x_1(z) x_2(t-z) dz$$

$$z = -\omega \quad x_2(z) \rightarrow x_2(-z)$$

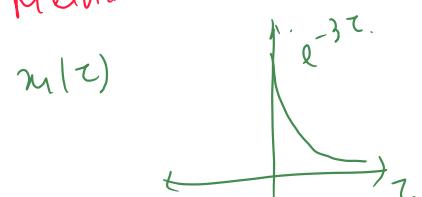
$$\begin{array}{ccc}
 x_1(z) & & x_2(-z) \\
 \downarrow & & \downarrow \\
 x_2(z) & & x_2(-z+t)
 \end{array}$$

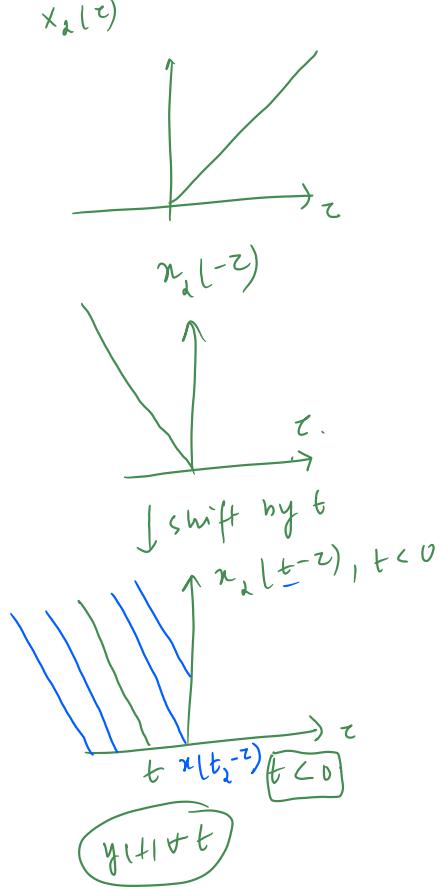


$\downarrow$   
 $y(t)$

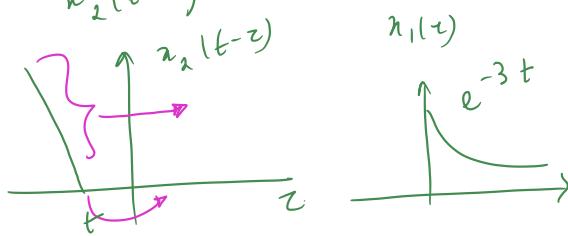
④ Convolve  $x_1(t) = e^{-3t} u(t)$ ;

$x_2(t) = t u(t)$  using graphical  
Method.



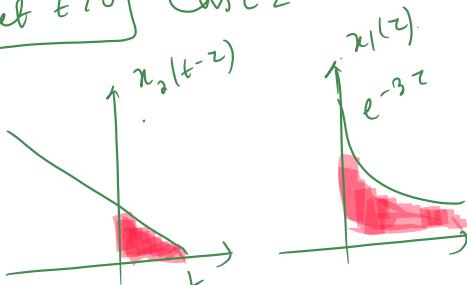


Case 1 when  $t < 0$



$$\therefore y_1(t) = \int n_1(z) x_2(t-z) dz = 0 \quad t < 0$$

Let  $t > 0$  Case 2



$$\begin{aligned} y_1(t) &= \int n_1(z) x_2(t-z) dz \\ &= \int_0^t e^{-3z} (t-z) dz \\ &= t \int_0^t e^{-3z} dz - \int_0^t e^{-3z} z dz \end{aligned}$$

$$x_3(t) = \begin{cases} t/3 + e^{-3t}/9 & \text{if } t > 0 \\ 0 & \text{if } t \leq 0 \end{cases}$$

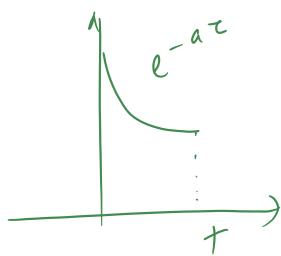
$$x_3(t) = (t/3 + e^{-3t}/9 - 1/9) u(t).$$

5 Convolve

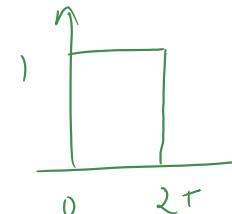
$$x_1(t) = e^{-at} ; 0 \leq t \leq T$$

$$x_2(t) = 1 ; 0 \leq t \leq 2T$$

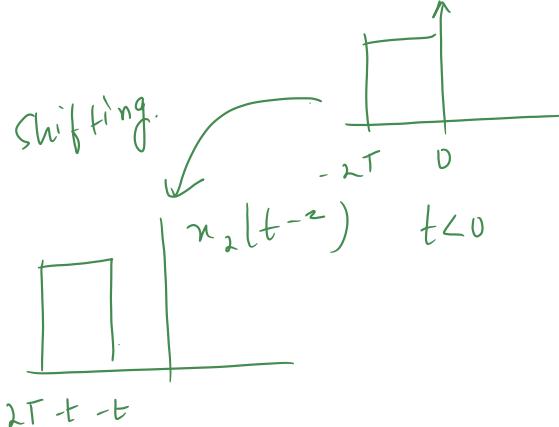
$x_1(t)$



$x_2(t)$



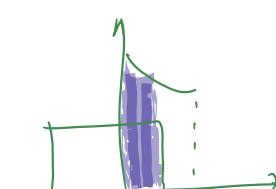
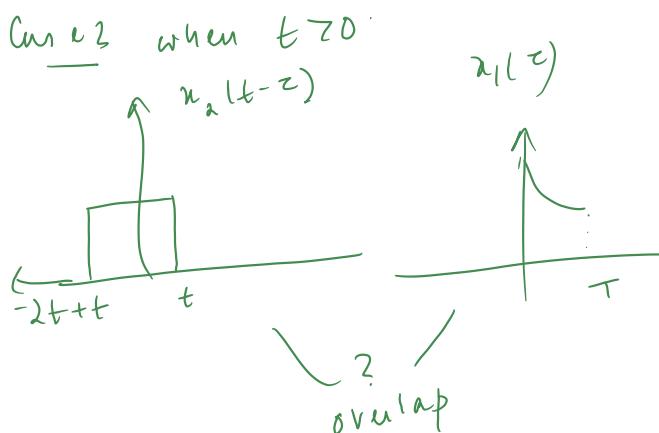
$x_2(t-z)$



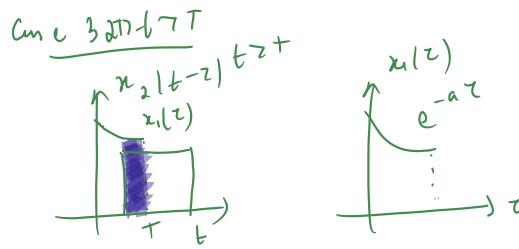
Case 1 when  $t < 0$

$y(t) = 0$ , as there is no overlap.

Case 2 when  $t > 0$



$$\begin{aligned} y(t) &= \int_0^t e^{-a\tau} \cdot 1 \, d\tau \\ &= 1/a [1 - e^{-at}] \end{aligned}$$



$$y(t) = \int_0^T x_1(t-z)x_2(z)e^{-az} dz$$

$$= \frac{1}{a} (1 - e^{-aT})$$

Case 2:  $T < t < 2T$

$$y(t) = \int_{t-2T}^t x_1(t-z)x_2(z)e^{-az} dz$$

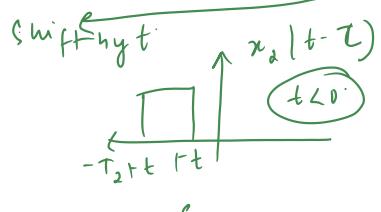
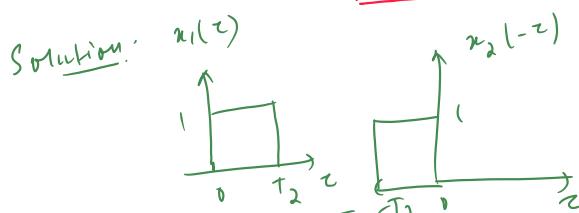
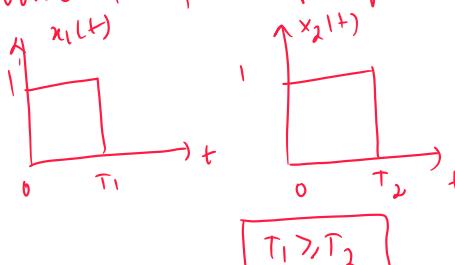
$$= \frac{1}{a} \{ e^{-a(t-2T)} - e^{-at} \}$$

Case 3:  
 $t - 2T > T$   
 $t > 3T$

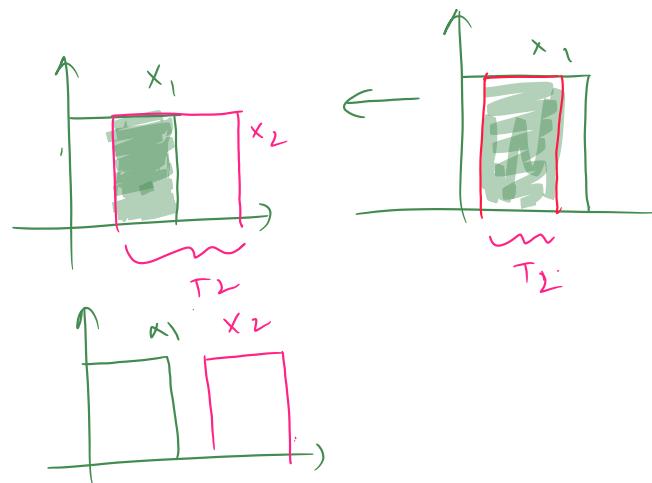
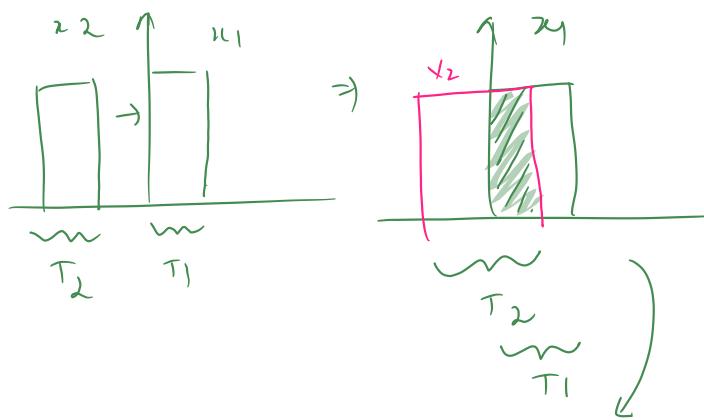


$$y(t) = \begin{cases} 0, & \text{if } t < 0 \\ \frac{1}{a} (1 - e^{-at}), & \text{if } 0 < t < T \\ \frac{1}{a} (1 - e^{-at}) - \frac{1}{a} (e^{-a(t-2T)} - e^{-at}), & \text{if } T < t < 3T \\ 0, & \text{if } t > 3T \end{cases}$$

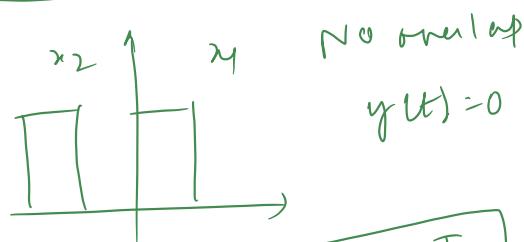
⑥ Convolve the following signals.



## Sliding 5 scenarios:-

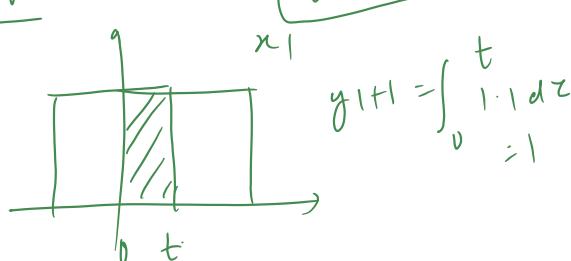


Case 01  $t < 0$



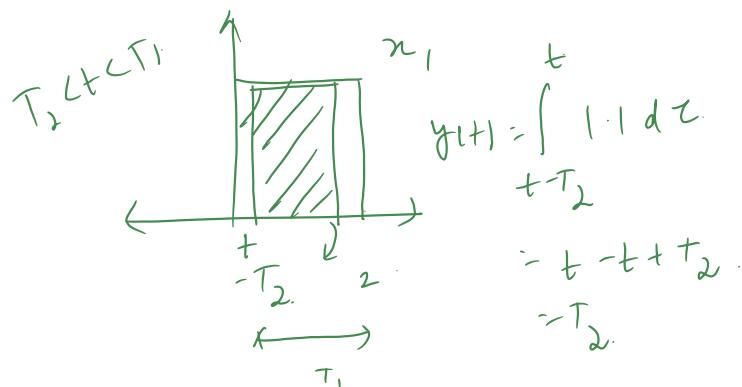
No overlap  
 $y(t) = 0$

Case 02

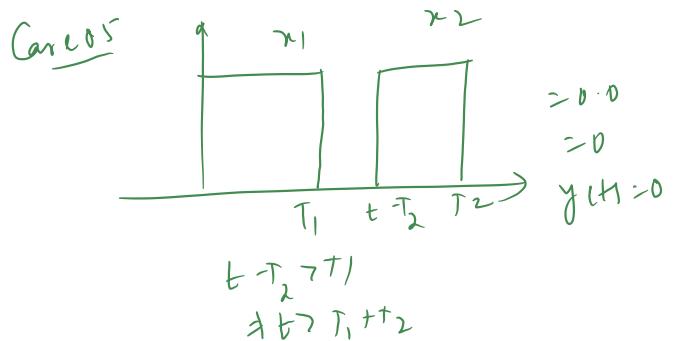
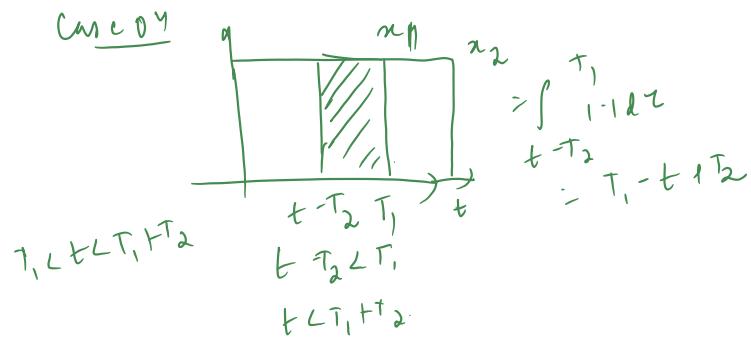


$$0 \leq t \leq T_2$$

Case -03

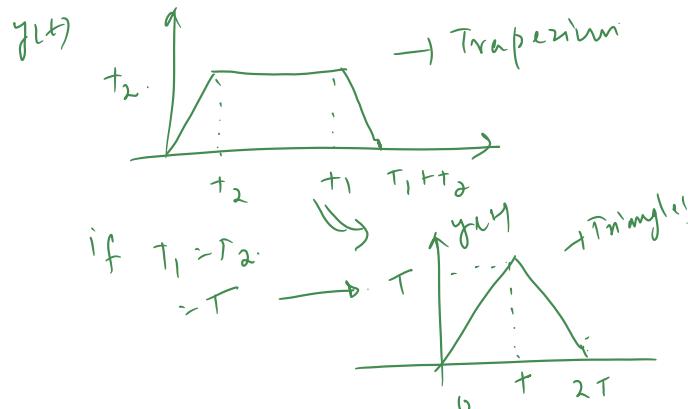


$$\begin{aligned} y_{(+)}(t) &= \int_0^t 1 \cdot 1 dz \\ &= t + T_2 \\ &= t - t + T_2 \\ &= T_2 \end{aligned}$$



Summarising

$$y(t) = \begin{cases} 0, & t < 0 \\ t, & 0 \leq t \leq T_2 \\ T_2, & T_2 \leq t \leq T_1 \\ T_1 + T_2 - t, & T_1 \leq t \leq T_1 + T_2 \\ 0, & t > T_1 + T_2 \end{cases}$$



Conclusion: convolution of 2 rectangular signals with

- a) unequal widths gives a trapezium
- b) equal widths gives a triangle.

⑦ Determine the range of values of "a" and "b" for the stability of LTI systems with impulse response

$$r(n) = \begin{cases} b^n, & n < 0 \\ a^n, & n \geq 0 \end{cases}$$

Solution: An LTI is stable if

$$\sum_{n=-\infty}^{\infty} |h(n)| < \infty$$

$$\begin{aligned} \sum_{n=-\infty}^{\infty} |h[n]| &= \sum_{n=-\infty}^{-1} |b|^n + \sum_{n=0}^{\infty} |a|^n \\ &= \sum_{n=1}^{\infty} |b|^{-n} + \sum_{n=0}^{\infty} |a|^n \\ &= \sum_{n=1}^{\infty} |b|^{-n} + |b|^{-0} - |b|^{-0} \\ &\quad + \sum_{n=0}^{\infty} |a|^n \\ &= \sum_{n=0}^{\infty} |b|^{-n} - |b|^{0} + \sum_{n=0}^{\infty} |a|^n \\ &= \sum_{n=0}^{\infty} \underbrace{(|b|^{-1})^n}_{c} - 1 + \sum_{n=0}^{\infty} |a|^n. \end{aligned}$$

$$\begin{aligned} &= \sum_{n=0}^{\infty} c^n - 1 + \sum_{n=0}^{\infty} |a|^n \quad |a| < 1 \\ \text{Since } |a| < 1 \quad \Rightarrow \quad &c = \frac{1}{|b|^{-1}} = \frac{1}{1-|b|} \quad \text{and} \quad |a| < 1 \\ &\Rightarrow |b|^{-1} \quad \downarrow \quad \text{Finite number.} \end{aligned}$$

$\therefore$  For  $B_1 B_0$  stability

$$|a| < 1 \quad \underline{\text{and}} \quad |b| > 1$$

(8) Show that

$$(i) x(t) * \delta(t) = x(t)$$

$$(ii) x(t) * \delta(t-t_0) = x(t-t_0)$$

$$(iii) x(t) * u(t) = \int_{-\infty}^t x(\tau) d\tau$$

$$(iv) x(t) * u(t-t_0) = \int_{-\infty}^{t-t_0} x(\tau) d\tau$$

$$\int_{-\infty}^{\infty} x(\tau) * \delta(t) = \int_{-\infty}^{\infty} x(\tau) \delta(t-\tau) d\tau$$

$$= x(\tau) \Big|_{t-\tau=0}$$

Sifting property

$$(ii) x(t) * \delta(t-t_0) = \delta(t-t_0) * x(t)$$

Commutativity of convolution

$$= \int_{-\infty}^{\infty} \delta(t-t_0) x(t-\tau) d\tau$$

$$= \int_{-\infty}^{\infty} x(t-\tau) \delta(t-t_0) d\tau$$

$$= x(t-t_0) \Big|_{t=t_0}$$

Sifting property

$$(iii) x(t) * u(t) = \int_{-\infty}^{\infty} x(\tau) u(t-\tau) d\tau.$$

$$u(t-\tau) = \begin{cases} 1, & \text{if } \tau < t \\ 0, & \text{if } \tau \geq t \end{cases}$$

$$iv) n(t) * u(t-t_0) = \int_{-\infty}^{\infty} n(\tau) u(t-\tau-t_0) d\tau$$

$$= \int_{-\infty}^{t-t_0} n(\tau) d\tau$$

(g) Consider a continuous time LTI

system described by

$$y(t) = \frac{1}{T} \int_{t-T/2}^{t+T/2} x(\tau) d\tau$$

a) Find and plot the impulse response  $h(t)$  of the system

b) Is the system causal?

Solution

a) Approach 1

$$\text{Put } x(z) = \delta(z)$$

$$y(t) = \frac{1}{T} \int_{t-T/2}^{t+T/2} \delta(z) dz$$

$$= \left[ 1 + u(t) \right] \Big|_{t-T/2}^{t+T/2}$$

$$= \frac{1}{T} \left[ u(t+T/2) - u(t-T/2) \right]$$

Approach 2

$$y(t) = \frac{1}{T} \int_{t-T/2}^{t+T/2} x(z) dz$$

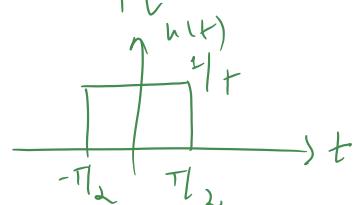
$$= \frac{1}{T} \int_{-\infty}^{t+T/2} x(z) dz - \frac{1}{T} \int_{-\infty}^{t-T/2} x(z) dz$$

$$= \frac{1}{T} \left[ x(t+T/2) + u(t+T/2) \right] - \frac{1}{T} \left[ x(t-T/2) + u(t-T/2) \right]$$

$$= \frac{1}{T} u(t) \times \left\{ u(t+T/2) - u(t-T/2) \right\}$$

impulse response

$$h(t) = \frac{1}{T} \left[ u(t+T/2) - u(t-T/2) \right]$$



b)  $h(t) \neq 0$ , for  $t < 0$

$\Rightarrow$  it's a non causal system.

⑩ Consider a C.T LTI systems

with  $i(p) - o(p)$  relation

$$y(t) = \int_{-\infty}^t e^{-s(t-s)} x(s) ds$$

a) find impulse response  $h(t)$

b) show that  $e^{st}$  is eigen func of s/s

c) find eigenvalue of s/s corresponding to  $e^{st}$  using  $h(t)$  obtained in (a).

$$\text{Solution: } h(t) = \int_{-\infty}^t e^{-(t-z)} s(z) dz$$

$$= e^{(t-z)} \Big|_{z=0} \quad t > 0$$

$$= e^{-t}, t > 0$$

$$h(t) = e^{-t} u(t)$$

$$h) x(t) = e^{st}$$

$$y(t) = x(t) * h(t)$$

$$= \int_{-\infty}^t e^{-(t-z)} e^{sz} dz$$

$$= e^{-t} \int_{-\infty}^t e^{(s+1)z} dz$$

$$= \frac{1}{s+1} e^{st} \text{ if } \operatorname{Re}\{s\} > -1.$$

$$y(t) = x(t) \Rightarrow x(t) = e^{st} \text{ eigen fcn!}$$

Eigenvalue

$$\frac{1}{s+1} = \lambda$$