

# ENGR 2420: Lab 2 Prelab Solution

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## Prelab

### 1. Diode-Connected Transistors.

- (a) If we assume that  $I_s$  is no more than 10 fA, then, in order for the current through the diode-connected transistor to be greater than 1 nA, the first term in the parentheses must be at least  $10^5$ , which means that neglecting the second term in the parentheses introduces an error of less than 10 ppm, or equivalently, 0.001 %, which is an excellent approximation.
- (b) If we were to change the current by one  $e$ -fold, the voltage across the diode-connected transistor would change by  $U_T$ , which is approximately 25 mV. If we were to change the current by one decade, the voltage across the diode-connected transistor would change by  $U_T \log 10 \approx 25 \text{ mV} \times 2.303 \approx 60 \text{ mV}$ .
- (c) To find the incremental diode resistance, we can invert the current–voltage characteristic of the forward-biased diode-connected transistor, thereby obtaining

$$V = U_T \log \frac{I}{I_s},$$

and differentiate it with respect to  $I$ , to obtain an expression for  $r_d$  as

$$r_d = \frac{\partial V}{\partial I} = \frac{U_T}{I/I_s} \cdot \frac{1}{I_s} = \frac{U_T}{I}.$$

- (d) No, the situation would not change if we were to force the voltage and measure the current instead of forcing the current and measuring the voltage. The voltage and current pairs that develop should be consistent between these two alternative experiments.
- (e) We can transform the exponential current–voltage characteristic of the forward-biased diode-connected transistor into a linear one by taking the log of both sides of the equation, to obtain

$$\begin{aligned} I &= I_s e^{V/U_T} \\ \log I &= \log (I_s e^{V/U_T}) \\ \log I &= \log I_s + \log e^{V/U_T} \\ \underbrace{\log I}_y &= \underbrace{\log I_s}_b + \underbrace{\frac{1}{U_T}}_m \cdot \underbrace{V}_x \end{aligned}$$

We can use linear regression on the log of the current and the voltage to find a best-fit line to the data. From the slope of this line, we can extract a value of  $U_T$  as  $1/m$ , and from the intercept, we can extract a value of  $I_s$  as  $e^b$ .

## 2. Characteristics of a Resistor and a Diode in Series.

- (a) With a current,  $I$ , flowing through the diode-connected transistor, the voltage drop across it will be given by  $V = U_T \log(I/I_s)$ . With a current,  $I$ , flowing through the resistor, the voltage drop across it will be given by  $V_r = IR$ . The voltage drop across the input current source is given by the sum of these two voltage drops,  $V_{in} = IR + U_T \log(I/I_s)$ .
- (b) If we were to increase the current  $I$  by a small amount,  $\delta I$ , the voltage across the diode-connected transistor would increase by  $\delta V = r_d \cdot \delta I$ , the voltage across the resistor would increase by  $\delta V_r = R \cdot \delta I$ , and the input voltage would increase by  $\delta V_{in} = \delta V_r + \delta V = (r_d + R) \delta I$ .
- (c) By equating  $\delta V_r$  and  $\delta V$ , we find that

$$\begin{aligned}\delta V_r &= \delta V \\ R \cdot \delta I &= r_d \cdot \delta I \\ R &= r_d \\ R &= \frac{U_T}{I_{on}},\end{aligned}$$

which implies that  $I_{on} = U_T/R$ .

- (d) We can obtain an expression for  $V_{on}$  as a function of  $I_{on}$ ,  $I_s$ , and  $U_T$  by inverting the diode-connected transistor's current-voltage characteristic. Doing so, we find that  $V_{on} = U_T \log(I_{on}/I_s)$ .
- (e) We can calculate the fraction of  $\delta V_{in}$  that appears across the diode-connected transistor directly as

$$\frac{\delta V}{\delta V_{in}} = \frac{r_d \cdot \delta I}{(r_d + R) \delta I} = \frac{r_d}{r_d + R} = \frac{\frac{U_T}{I}}{\frac{U_T}{I} + R} = \frac{\frac{U_T}{R}}{\frac{U_T}{R} + I} = \frac{I_{on}}{I_{on} + I}.$$

The rest of  $\delta V_{in}$  drops across the resistor, so we can obtain  $\delta V_r/\delta V_{in}$  as

$$\frac{\delta V_r}{\delta V_{in}} = 1 - \frac{\delta V}{\delta V_{in}} = 1 - \frac{I_{on}}{I_{on} + I} = \frac{I}{I_{on} + I}.$$

- (f) If  $I \ll I_{on}$ , then  $\delta V/\delta V_{in} \approx 1$  and  $\delta V_r/\delta V_{in} \approx 0$ . Consequently, practically all of  $\delta V_{in}$  results from  $\delta V$ . Because  $V$  changes logarithmically with  $I$ ,  $V_{in}$  should also increase logarithmically with  $I$  in this regime. Also, because  $I$  is so small in this regime,  $V_r \approx 0$  and  $V_{in} \approx V$ . On the other hand, if  $I \gg I_{on}$ , then  $\delta V/\delta V_{in} \approx 0$  and  $\delta V_r/\delta V_{in} \approx 1$ . Thus, in this regime, practically all of  $\delta V_{in}$  results from  $\delta V_r$ . Because  $V_r$  changes linearly with  $I$ , in this regime,  $V_{in}$  should also increase linearly with  $I$ . Moreover, because  $\delta V \approx 0$  in this regime, we have that  $V$  essentially does not change (i.e., remains constant) with increases in  $I$ .

- (g) As with the diode-connected transistor by itself, the situation that we have just described would not change if we were to force the input voltage and measure the current instead of forcing the current and measuring the input voltage. The input voltage and current pairs that develop should be consistent between these two alternative experiments. Because  $V_{\text{in}} = V_r + V$  and  $V_r > 0$ , it follows directly that  $V < V_{\text{in}}$ , so that if  $V_{\text{in}} < V_{\text{on}}$ , then  $V < V_{\text{on}}$ . Given the way that we have defined  $I_{\text{on}}$  and  $V_{\text{on}}$ , we have that  $I < I_{\text{on}}$  when  $V_{\text{in}} < V_{\text{on}}$ . Given the steepness of the current–voltage characteristic of the diode-connected transistor, if  $V_{\text{in}} < V_{\text{on}}$  by more than a few  $U_T$ , then  $I \ll I_{\text{on}}$ , and the situation will be the first one described in part f. In this case, we should expect  $I$  to increase exponentially with  $V_{\text{in}}$ . If  $V_{\text{in}} > V_{\text{on}}$  by more than a few  $U_T$ , we should expect to be in the second situation described in part f. Thus, we should expect that  $I$  will increase linearly with  $V_{\text{in}}$ .