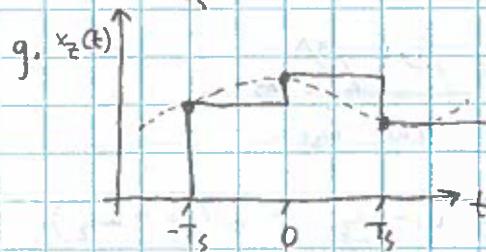
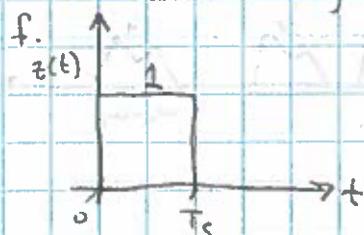


$$\text{d. } 2\omega_m \leq \frac{2\pi}{T_s}$$

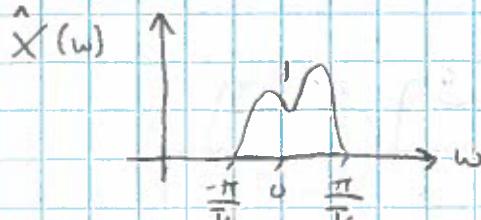
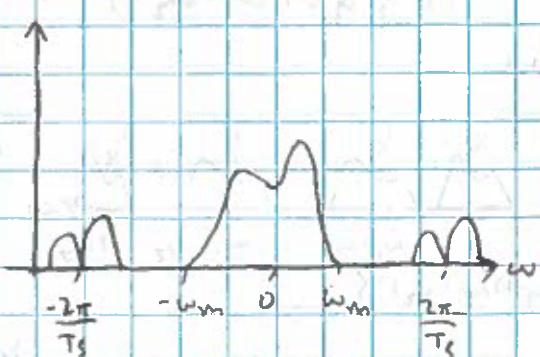
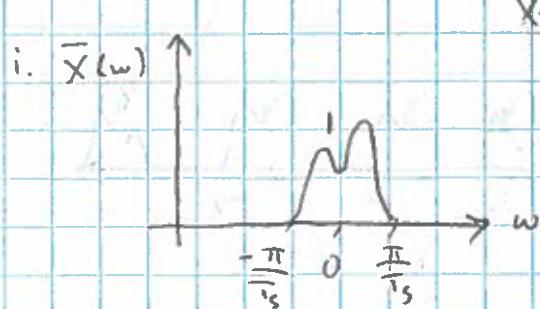
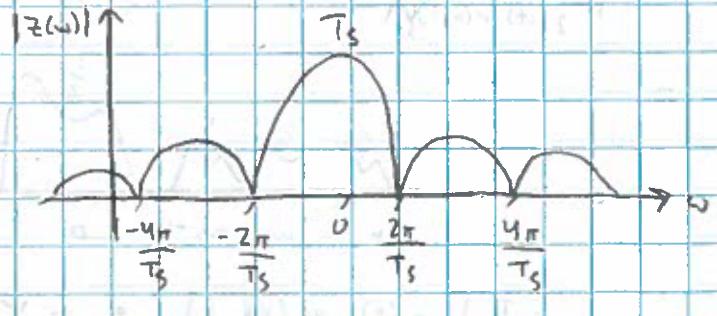
$$T_s \leq \frac{\pi}{\omega_m}$$

e. You can recover $x(t)$ from $x_p(t)$ exactly by multiplying $X_p(w)$ with a low-pass filter with cut-off frequency ω_m . In the time domain, this would be convolving with a sinc function which is the equivalent of interpolating with sincs, recovering $x(t)$ exactly.



h. $x(+ - \frac{T_s}{2}) = e^{-jw\frac{T_s}{2}} X(w)$

$$Z(w) = e^{-jw\frac{T_s}{2}} \frac{2 \sin(\frac{\pi}{2} w)}{w}$$



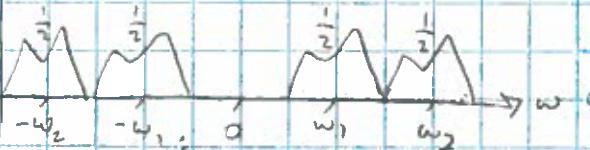
j. $\hat{X}(w)$ and $\tilde{X}(w)$ are very similar. However, the edges of $\tilde{X}(w)$ are attenuated a little due to the envelope of the sinc function which decays at a rate of $1/w$. Other than that, $\tilde{X}(w)$ and $\hat{X}(w)$ are very similar to $X(w)$.

$$k. \frac{\tilde{X}(w)}{\hat{X}(w)} = \frac{X_p(w) H(w)}{X_p(w) H(w)} = \frac{X_p(w) \tilde{F}(w)}{X_p(w)}$$

$$\left| Z\left(\frac{\pi}{T_s}\right) \right| = \frac{2 \sin\left(\frac{\pi}{2} \cdot \frac{\pi}{T_s}\right)}{\frac{\pi}{T_s}} = \frac{2}{\frac{\pi}{T_s}} = \frac{2 T_s}{\pi}$$

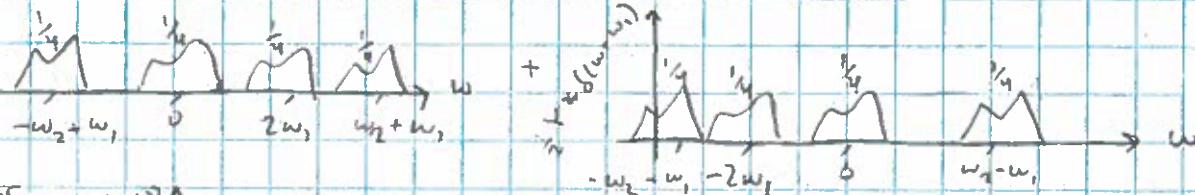
$$\frac{\tilde{X}(w)}{\hat{X}(w)} = \frac{2 T_s}{\pi}$$

2. a. $y(t)$



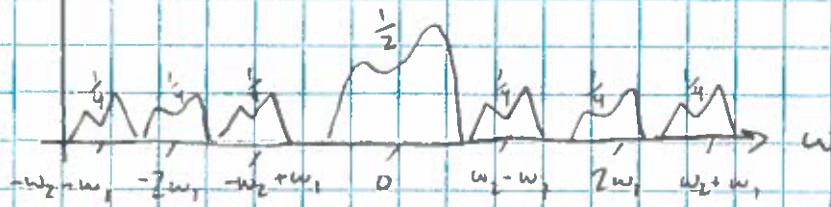
$$b. \text{FT} \{ y(t) \cos(w_1 t) \} = \frac{1}{2\pi} Y * \pi \delta(w-w_1) + \pi \delta(w+w_1)$$

$$= \frac{1}{2} Y * \delta(w-w_1) + \delta(w+w_1)$$

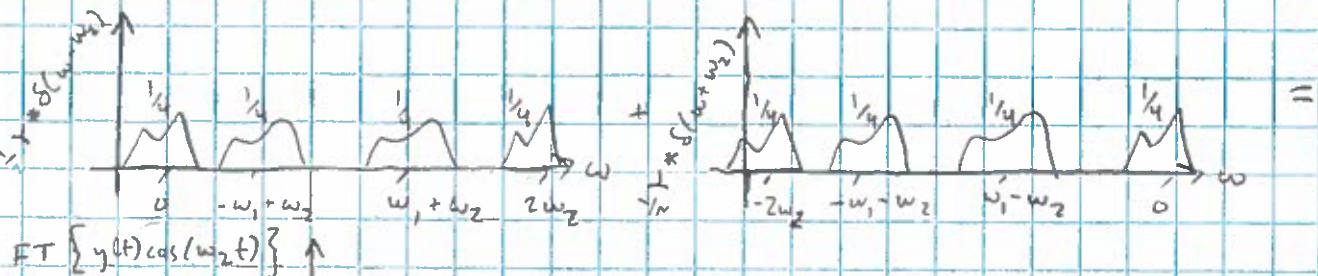


$$\text{FT} \{ y(t) \cos(w_1 t) \}$$

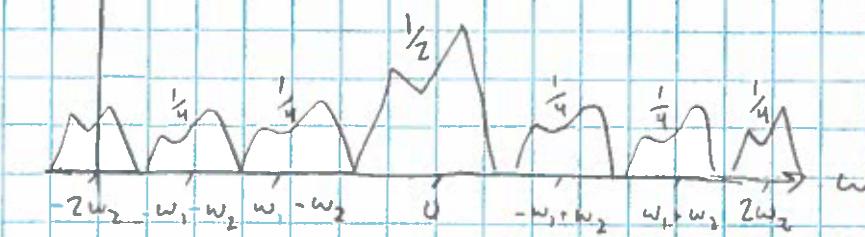
=



$$\text{FT} \{ y(t) \cos(w_2 t) \} = \frac{1}{2} Y * \delta(w-w_2) + \delta(w+w_2)$$



$$\text{FT} \{ y(t) \cos(w_2 t) \}$$



c. $y(t)$ is a received AM radio signal from two different AM transmitters. To recover $x_1(t)$, you multiply $y(t)$ by a $\cos(\omega t)$ which demodulates the frequency components at ω . Applying a low-pass filter in the frequency domain with a cut-off of ω_m will recover a scaled version of $x_1(t)$.

To recover $x_2(t)$, you do the same thing except $y(t)$ is multiplied by a $\cos(\omega_2 t)$.

$$\begin{aligned} 3. \text{ a. } V_{in}(t) &= V_R(t) + V_L(t) + V_{out}(t) = R_i(t) + V_L(t) + V_{out}(t) \\ &= RC \frac{d}{dt} V_{out}(t) + L \frac{d}{dt} C \frac{d}{dt} V_{out}(t) + V_{out}(t) \\ V_{in}(t) &= LC \frac{d^2}{dt^2} V_{out}(t) + RC \frac{d}{dt} V_{out}(t) + V_{out}(t) \end{aligned}$$

$$\begin{aligned} \text{b. } V_{in}(w) &= LC(w)^2 V_{out}(w) + RCjw V_{out}(w) + V_{out}(w) \\ &= V_{out}(w) (-LCw^2 + RCjw + 1) \\ \frac{V_{in}(w)}{V_{out}(w)} &= -LCw^2 + 1 + RCwj \end{aligned}$$

$$\frac{V_{out}(w)}{V_{in}(w)} = H(w) = \frac{1}{-LCw^2 + 1 + RCwj}$$

$$\text{c. } |H(w)| = \frac{1}{\sqrt{(1-LCw^2)^2 + (RCw)^2}}$$

d. Since we are trying to maximize $|H(w)|$, we are trying to minimize the denominator so we are looking for where the derivative of the denominator is 0.

$$0 = \frac{d}{dw} \sqrt{(-LCw^2)^2 + (RCw)^2}$$

$$\begin{aligned} 0 &= \frac{d}{dw} (1-LCw^2)^2 + (RCw)^2 \\ &= 2(1-LCw^2) \frac{d}{dw} (1-LCw^2) + 2RCw \frac{d}{dw} RCw \\ &= 2(1-LCw^2)(-LC2w) + 2RCw(RC) \\ &= -4LCw(1-LCw^2) + 2R^2C^2w \\ &= -4LCw + 4L^2C^2w^3 + 2R^2C^2w \\ &= 2Cw(-2L + 2L^2Cw^2 + R^2C) \Rightarrow w=0 \text{ is a possible solution} \\ 0 &= -2L + 2L^2Cw^2 + R^2C \end{aligned}$$

$$2L - R^2C = 2L^2Cw^2$$

$$w^2 = \frac{2L - R^2C}{2L^2C} \Rightarrow w = \pm \sqrt{\frac{2L - R^2C}{2L^2C}} \text{ are also possible solutions}$$

To check which of these solutions are minimums, we need to check the second derivative to see if it is positive for these solutions.

$$H'(w) = -4LCw + 4L^2C^2w^3 + 2R^2C^2w$$

$$H''(w) = -4LC + 12L^2C^2w^2 + 2R^2C^2$$

$$= -4LC + 12L^2C^2 + 12L^2C^2w^2$$

$$= 12C(-2L + R^2C) + 4C(2L^2Cw^2)$$

If you substitute in $w=0$ to find the sign of the second derivative, you need to know if $-2L + R^2C > 0$. From our other solutions for w , we know $2L - R^2C \geq 0$ which means $2L \geq R^2C$. As a result, $-2L + R^2C < 0$ so the second derivative is negative which means $w=0$ is a maximum and not a minimum. Thus,

$$w = \pm \sqrt{\frac{2L - R^2C}{2L^2C}} \quad \text{maximize } |H(w)|$$

