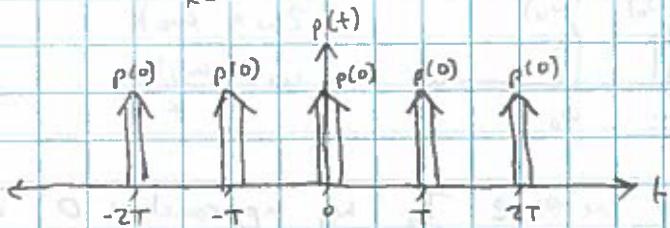


1. a. $p(t) = \sum_{k=-\infty}^{\infty} \delta(t - kT)$



b. Fourier series representation of $p(t)$: $\tilde{p}(t)$

$$\tilde{p}(t) = \sum_{k=-\infty}^{\infty} C_k e^{j \frac{2\pi}{T} kt}$$

$$C_k = \frac{1}{T} \int_{-T/2}^{T/2} p(t) e^{-j \frac{2\pi}{T} kt} dt = \frac{1}{T} \int_{-T/2}^{T/2} \sum_{n=-\infty}^{\infty} \delta(t - nT) e^{-j \frac{2\pi}{T} kt} dt$$

$$= \sum_{n=-\infty}^{\infty} \frac{1}{T} \int_{-T/2}^{T/2} \delta(t - nT) e^{-j \frac{2\pi}{T} kt} dt$$

From the plot above, we know that $p(t)$ is only non-zero at a point so from $T/2$ to $-T/2$, the integral is only evaluated at a single point, $t = 0$.

$$C_k = \sum_{n=-\infty}^{\infty} \frac{1}{T} e^{-j \frac{2\pi}{T} k(0)} = \sum_{n=-\infty}^{\infty} \frac{1}{T} = \frac{1}{T}$$

$$\tilde{p}(t) = \sum_{k=-\infty}^{\infty} \frac{1}{T} e^{j \frac{2\pi}{T} kt}$$

c. $x(t) = \sum_{k=-\infty}^{\infty} G_k e^{j \frac{2\pi}{T} kt}$

$$X(\omega) = \int_{-\infty}^{\infty} \sum_{k=-\infty}^{\infty} G_k e^{j \frac{2\pi}{T} kt} e^{-j \omega t} dt = \sum_{k=-\infty}^{\infty} G_k \int_{-\infty}^{\infty} e^{j \frac{2\pi}{T} kt - j \omega t} dt$$

$$\int_{-\infty}^{\infty} e^{j \frac{2\pi}{T} kt - j \omega t} dt = \int_{-\infty}^{\infty} e^{j + (\frac{2\pi}{T} k - \omega)t} dt \quad \text{define } \omega_0 = \frac{2\pi}{T} k - \omega$$

$$\int_{-\infty}^{\infty} e^{j t \omega_0} dt = 2\pi \delta(\omega - \omega_0) = 2\pi \delta(\omega - \frac{2\pi}{T} k + \omega) = 2\pi \delta(2\omega - \frac{2\pi}{T} k)$$

Substituting back in: $X(\omega) = \sum_{k=-\infty}^{\infty} G_k 2\pi \delta(2\omega - \frac{2\pi}{T} k)$

Now define $\omega_0 = \frac{2\pi}{T}$

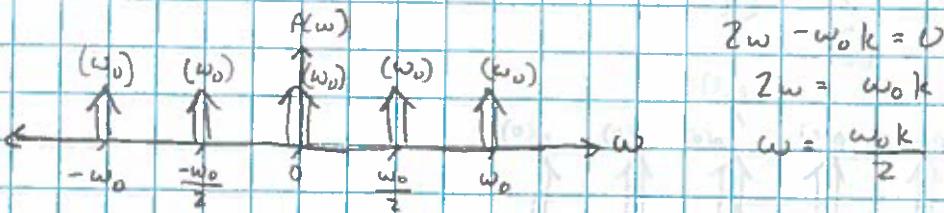
$$X(\omega) = \sum_{k=-\infty}^{\infty} G_k 2\pi \delta(2\omega - \omega_0 k)$$

$$d. C_k = \frac{1}{T}$$

$$P(w) = \sum_{k=-\infty}^{\infty} \frac{1}{T} 2\pi \delta(2w - w_0 k) = \sum_{k=-\infty}^{\infty} w_0 \delta(2w - w_0 k)$$

$$= w_0 \sum_{k=-\infty}^{\infty} \delta(2w - w_0 k)$$

e.



w_0 is $\frac{2\pi}{T}$. So if we increase T , w_0 approaches 0 which makes $P(w)$ approach 0 at all points. Decreasing T increases the spacing between the impulses and scales up the impulses in $P(w)$.

Increasing T for $p(t)$ increases the spacing between impulses. Decreasing T makes $p(t)$ approach a single impulse at $t=0$.

This behavior is kind of intuitive when transitioning from the time domain to the frequency domain as T varies.

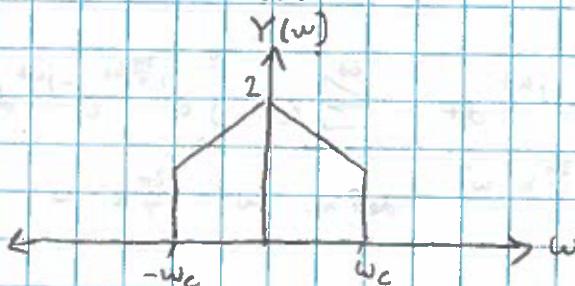
$$2. a. h(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} H(w) e^{jwt} dw \quad \int_{-\infty}^{w_c} H(w) dw = 0 \quad \int_{-\infty}^{\infty} H(w) dw = 0$$

$$\int_{-w_c}^{w_c} e^{jwt} dw = \frac{1}{jt} e^{jwt} \Big|_{-w_c}^{w_c} = \frac{1}{jt} e^{jw_c t} - \frac{1}{jt} e^{-jw_c t}$$

$$h(t) = \frac{1}{2\pi} \left(\frac{1}{jt} e^{jw_c t} - \frac{1}{jt} e^{-jw_c t} \right) = \frac{1}{\pi t} \left(\frac{1}{2j} e^{jw_c t} - \frac{1}{2j} e^{-jw_c t} \right)$$

$$h(t) = \frac{\sin(w_c t)}{\pi t}$$

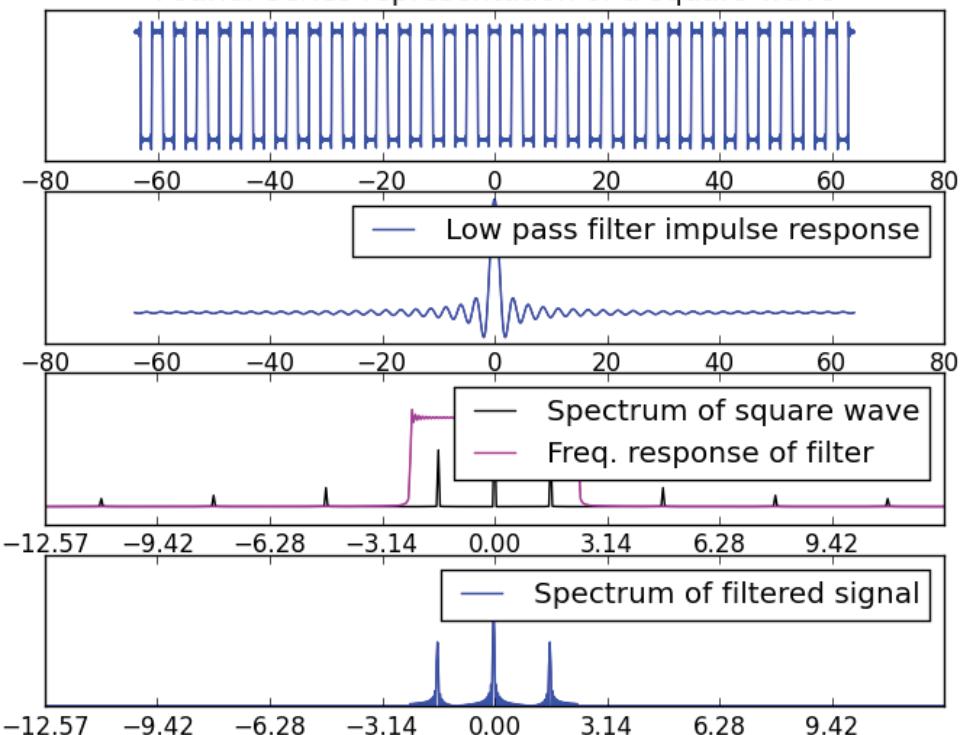
b.



c. This LTI system is known as an ideal low-pass filter with cut-off frequency w_c because it attenuates to 0 all frequencies above w_c and below $-w_c$. This is because if you multiply $H(w)$ by frequencies above w_c , they become 0.

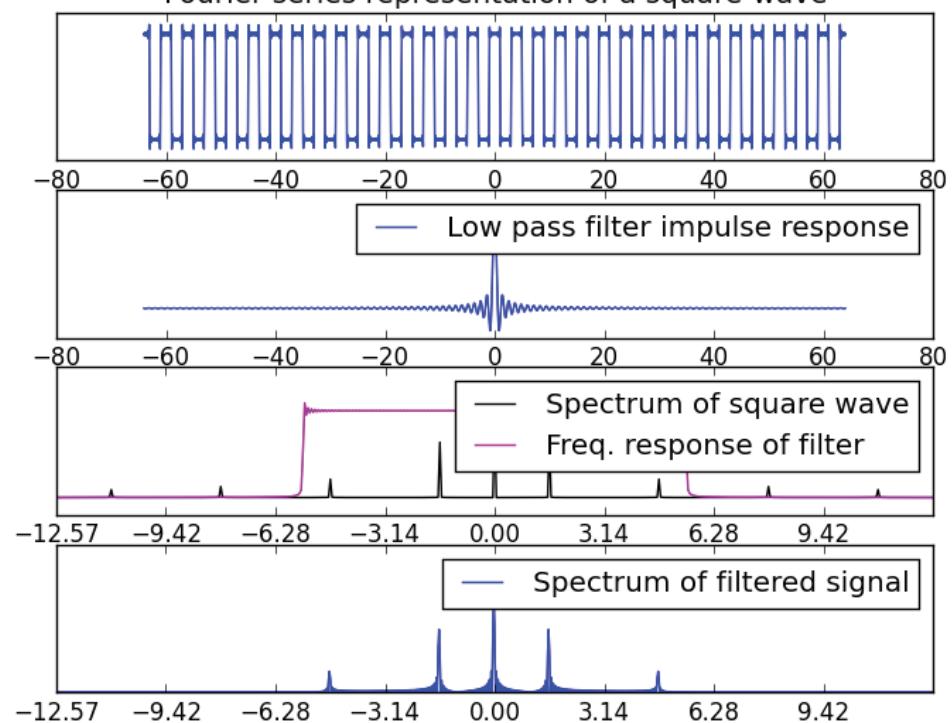
$$\omega_c = 0.75\pi$$

Fourier series representation of a square-wave

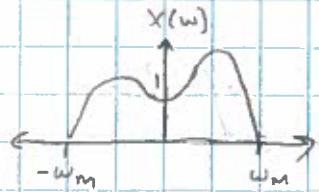


$$\omega_c = 1.75\pi$$

Fourier series representation of a square-wave



3.



$$y(t) = x(t) \cos(\omega_c t)$$

$$y(t) = x(t) h(t) \text{ where } h(t) = \cos(\omega_c t)$$

$$H(\omega) = \pi \delta(\omega - \omega_c) + \pi \delta(\omega + \omega_c)$$

$$Y(\omega) = \frac{1}{2\pi} X * H(\omega)$$

$$Y(\omega) = \frac{1}{2\pi} X * \pi \delta(\omega - \omega_c) + \pi \delta(\omega + \omega_c)$$

$$Y(\omega) = \frac{1}{2} X * \delta(\omega - \omega_c) + \frac{1}{2} X * \delta(\omega + \omega_c)$$

