

# VHDL Tutorial: Learn by Example

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If we hear, we forget; if we see, we remember; if we do, we understand.

-- Proverb

[ESD book](#) | [Dalton Project](#) | [VHDL Reference](#) | [Synopsys Tutorial](#) | [ActiveHDL Tutorial](#) | [Xilinx Tutorial](#)

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### Foreword (by Frank Vahid)

<> HDL (Hardware Description Language) based design has established itself as the modern approach to design of digital systems, with VHDL (VHSIC Hardware Description Language) and Verilog HDL being the two dominant HDLs. Numerous universities thus introduce their students to VHDL (or Verilog). The problem is that VHDL is complex due to its generality. Introducing students to the language first, and then showing them how to design digital systems with the language, tends to confuse students. The language issues tend to distract them from the understanding of digital components. And the synthesis subset issues of the language add to the confusion.

We developed the following tutorial based on the philosophy that the beginning student need not understand the details of VHDL -- instead, they should be able to modify examples to build the desired basic circuits. Thus, they learn the importance of HDL-based digital design, without having to learn the complexities of HDLs. Those complexities can be reserved for a second, more advanced course. The examples are mostly from the textbook [Embedded System Design](#) by Frank Vahid and Tony Givargis. They start from basic gates and work their way up to a simple microprocessor. Most of the examples have been simulated by [Aldec](#) ActiveHDL Simulator and [Synopsys](#) Design Analyzer, as well as

synthesized with Synopsys Design Compiler . Several sequential design examples have been successfully tested on [Xilinx](#) Foundation Software and FPGA/CPLD board.



## **Basic Logic Gates** (ESD Chapter 2: Figure 2.3)

Every VHDL design description consists of at least one *entity* / *architecture* pair, or one entity with multiple architectures. The entity section of the HDL design is used to declare the *I/O ports* of the circuit, while the description code resides within architecture portion. Standardized design libraries are typically used and are included prior to the entity declaration. This is accomplished by including the code "library ieee;" and "use ieee.std\_logic\_1164.all;".

<b>Driver</b>	<a href="#">Behavior Code</a>	<a href="#">Behavior Simulation</a>	<b>Inverter</b>	<a href="#">Behavior Code</a>	<a href="#">Behavior Simulation</a>
<b>OR gate</b>	<a href="#">Behavior Code</a>	<a href="#">Behavior Simulation</a>	<b>NOR gate</b>	<a href="#">Behavior Code</a>	<a href="#">Behavior Simulation</a>
<b>AND gate</b>	<a href="#">Behavior Code</a>	<a href="#">Behavior Simulation</a>	<b>NAND gate</b>	<a href="#">Behavior Code</a>	<a href="#">Behavior Simulation</a>
<b>XOR gate</b>	<a href="#">Behavior Code</a>	<a href="#">Behavior Simulation</a>	<b>XNOR gate</b>	<a href="#">Behavior Code</a>	<a href="#">Behavior Simulation</a>



## **Combinational Logic Design** (ESD Chapter 2: Figure 2.4)

We use *port map statement* to achieve the *structural model* (components instantiations). The following example shows how to write the program to incorporate multiple components in the design of a more complex circuit. In order to simulate the design, a simple *test bench* code must be written to apply a sequence of inputs (Stimulators) to the circuit being tested (*UUT*). The output of the test bench and UUT interaction can be observed in the simulation waveform window.

<b>Combinational Logic</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>
<b>Tri-State Driver</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>

### **Discussion I: Signal vs. Variable:**

*Signals* are used to connect the design components and must carry the information between current statements of the design. On the other hand, *variables* are used within process to compute certain values. The following example shows their difference:

<b>Signal/Variable Example</b>	<a href="#">Behavior Code</a>	<a href="#">Behavior Simulation</a>
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## **Typical Combinational Components** (ESD Chapter 2: Figure 2.5)

The following behavior style codes demonstrate the concurrent and sequential capabilities of VHDL. The *concurrent statements* are written within the body of an architecture. They include *concurrent signal assignment*, *concurrent process* and *component instantiations (port map statement)*. *Sequential statements* are written within a *process* statement, *function* or *procedure*. Sequential statement include *case statement*, *if-then-else statement* and *loop statement*.

<b>Multiplexor</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>
<b>Decoder</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>
<b>Adder</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>
<b>Comparator</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>
<b>ALU</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>
<b>Multiplier</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>

## ▲ **Latch & Flip-Flops** (ESD Chapter 2.3)

Besides from the circuit input and output signals, there are normally two other important signals, *reset* and *clock*, in the sequential circuit. The reset signal is either *active-high* or *active-low* status and the circuit status transition can occur at either clock *rising-edge* or *falling-edge*. Flip-Flop is a basic component of the sequential circuits.

<b>Simple Latch</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Gate-level Implementation</a>	<a href="#">Gate-level Simulation</a>
<b>D Flip-Flop</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Gate-level Implementation</a>	<a href="#">Gate-level Simulation</a>
<b>JK Flip-Flop</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Gate-level Implementation</a>	<a href="#">Gate-level Simulation</a>

## ▲ **Typical Sequential Components** (ESD Chapter 2: Figure 2.6)

Typical sequential components consist of registers, shifters and counters. The concept of *generics* is often used to parameterize these components. Parameterized components make it possible to construct *standardized libraries* of shared models. In the behavioral description, the output transitions are generally set at the clock rising-edge. This is accomplished with the combination of the VHDL *conditional statements* (clock'event and clock='1'). During the testbench running, the expected output of the circuit is compared with the results of simulation to verify the circuit design.

<b>Register</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Gate-level Implementation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Structural Simulation</a>
<b>Shift Register</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Gate-level Implementation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Structural Simulation</a>
<b>Counter</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation</a>	<a href="#">Gate-level Implementation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Structural Simulation</a>

## **Sequential Logic Design** **(ESD Chapter 2: Figure 2.7)**

The most important description model presented here may be the *Finite State Machine (FSM)*. A general model of a FSM consists of both the combinational Logic and sequential components such as state registers, which record the states of circuit and are updated synchronously on the rising edge of the clock signal. The output function computes the various outputs according to different states. Another type of sequential model is the memory module, which usually takes a long time to be synthesized due to the number of design cells.

<b>FSM Model</b>	Behavior <a href="#">Code</a>	Test <a href="#">Bench</a>	Behavior <a href="#">Simulation</a>	Gate-level <a href="#">Implementation</a>	Synthesis <a href="#">Schematic</a>	Gate-level <a href="#">Simulation</a>
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### ○ Memories (ESD Chapter 5)

<b>RAM Module</b>	Behavior <a href="#">Code</a>	Test <a href="#">Bench</a>	Behavior <a href="#">Simulation</a>	Gate-level <a href="#">Implementation</a>	Synthesis <a href="#">Schematic</a>	Gate-level <a href="#">Simulation</a>
<b>ROM Module</b>	Behavior <a href="#">Code</a>	Test <a href="#">Bench</a>	Behavior <a href="#">Simulation</a>	Gate-level <a href="#">Implementation</a>	Synthesis <a href="#">Schematic</a>	Gate-level <a href="#">Simulation</a>

### ● **Discussion II: Behavior vs. RTL Synthesis** ([Y Chart](#))

*RTL* stands for *Register-Transfer Level*. It is an essential part of *top-down* digital design process. [Logic synthesis](#) offers an automated route from an RTL design to a *Gate-Level* design. In RTL design a circuit is described as a set of registers and a set of transfer functions describing the flow of data between the registers, (ie. *FSM + Datapath*). As an important part of a complex design, this division is the main objective of the hardware designer using synthesis. The Synopsys Synthesis Example illustrates that the RTL synthesis is more efficient than the behavior synthesis, although the simulation of previous one requires a few clock cycles.

<b>GCD Calculator</b>	Behavior <a href="#">Code</a>	RTL <a href="#">Code</a> (FSM+D)	<a href="#">Comparison</a>
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Following section illustrates the *RTL (FSM+Datapath)* method further using several design examples.

## **Custom Single-Purpose Processor Design** **(ESD Chapter 2, Chapter 4)**

The first three examples illustrate the difference between *RTL FSMD model (Finite State Machine with Datapath buildin)* and *RTL FSM + DataPath model*. From view of RT level design, each digital design consists of a *Control Unit* (FSM) and a *Datapath*. The datapath consists of storage units such as registers and memories, and combinational units such as ALUs, adders, multipliers, shifters, and comparators. The datapath takes the operands from storage units, performs the computation in the combinatorial units, and returns the results to the storage units during each state. This process typically takes one or two clock cycles.

*Data-flow (looks more like an Algorithm) modeling* is presented in the fourth example. The FIR digital filter algorithm is simulated and synthesized using VHDL. A comparison of the coding styles between the *RTL modeling* and *Algorithm level*

*modeling* highlights the different techniques.

○ **GCD Calculator (ESD Chapter2: Figure 2.9-2.11)**

<b>FSMD Modeling</b>	<a href="#">RTL Code</a>	<a href="#">Test Bench</a>	<a href="#">RTL Code Simulation</a>	<a href="#">Gate-level Implementation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>
<b>FSM + Datapath Modeling</b>	<a href="#">RTL Code</a>	<a href="#">Test Bench</a>	<a href="#">RTL Code Simulation</a>	<a href="#">Gate-level Implementation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>

○ **Simple Bridge (ESD Chapter 2: Figure 2.13-2.14)**

<b>FSMD Modeling</b>	<a href="#">RTL Code</a>	<a href="#">Test Bench</a>	<a href="#">RTL Code Simulation</a>	<a href="#">Gate-level Implementation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>
<b>FSM + Datapath Modeling</b>	<a href="#">RTL Code</a>	<a href="#">Test Bench</a>	<a href="#">RTL Code Simulation</a>	<a href="#">Gate-level Implementation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>

○ **ISA Bus Interface (ESD Chapter 4, Chapter 6)**

<b>FSM + Datapath Modeling</b>	<a href="#">RTL Code</a>	<a href="#">Test Bench</a>	<a href="#">RTL Code Simulation</a>	<a href="#">Gate-level Implementation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>
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○ **FIR Digital Filter (DSP Example)**

<b>Data-Flow Modeling</b>	<a href="#">Behavior Code</a>	<a href="#">Test Bench</a>	<a href="#">Behavior Simulation(1,2)</a>	<a href="#">Gate-level Implementation</a>	<a href="#">Synthesis Schematic</a>	<a href="#">Gate-level Simulation</a>
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### ● **Discussion III: Synopsys Power Analysis**

Synopsys tools can be used to perform Power Analysis for all the VHDL designs. Generally, the better design has smaller power consumption. On the other hand, improve the power always means sacrificing other design metrics such as performance, area size or NRE cost. Therefore, a designer need to balance these metrics to find the best implementation for the given application and constraints. Please check out the [power analysis](#) results of Adder, Counter, ISA controller, Bridge controller and FIR Filter. As we expected, FIR digital filter has the biggest power consumption because it has a more complex circuit doing DSP computation. Synopsys power analysis tutorial can be found [here](#).

### ● **Discussion IV: Synthesis with Timing Constraints**

When we design and simulate the high-level (either behavior or RTL) code, we only care about design *functionality*. However, in VHDL synthesis, the *timing* and the *functionality* of a design must always be considered together. Therefore, once the design has been synthesized, the second goal of simulation is to quickly verify that the gate-level implementation meets timing requirements. We use this idea (coding -> simulation -> synthesis -> simulation) to test all of the examples in this tutorial.

Another common way is to apply the timing constrains on the design during synthesis. then the timing report is checked to see if the slack, which is the required delay minus the actual delay, is MET or VIOLATED. If VIOLATED, we should go back to the VHDL code and re-write it to improve timing. The whole design will be compiled and tested again.

<b>Counter</b>	<a href="#">Behavior Code</a>	<a href="#">Sythesis Script File</a>	<a href="#">Timing Report</a>
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## ● Discussion V: Relationship between Area and Timing

During Synopsys synthesis, ordinary combinational logic will go through several of what are known as mapping optimizations. In a normal optimization, the synthesis tool will optimize in relation to the set constraints. It is usual to talk about moving along a "banana curve" on the area and time axes. This means that the tougher the timing constraints, the larger the design will be, and vice versa. The results from two different synthesis constraints applied on the same design are shown below.

<b>FIR filter</b>	Sample Synthesis <a href="#">Script</a>	Comparison <a href="#">Table</a>	<a href="#">Banana Curve</a>
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## **General-Purpose Processor Design** (ESD Book Chapter 3, Figure 3.15)

As indicated in the previous part, an Application Specific Integrated Circuit (ASIC) is specified with behavior descriptions which are presented in the form of particular algorithm or flowchart. A general purpose processor, on the other hand, is specified completely by its *instruction set* (IS). A sequence of instructions is required for the computation of a mathematical expression or any other similar computational task. To illustrate the whole procedure, a simple *Pseudo-Microprocessor* model is used which contains seven instructions (ESD book figure 3.7). The RT level design method from previous examples is used again to construct this microprocessor. The CPU will *fetch*, *decode*, and *execute* each instruction in order to get the final result.

For test purposes, a short program (sequential instructions) is loaded into the memory. After execution, this program will obtain 10 Fabonacci Numbers, and store the results into specific memory address. The design was implemented using Active-HDL and Synopsys Design Compiler. (Please note that PC.vhd need a little modify to get correct synthesis result. Just a practice for the reader.)

<b>Top Level Structural Code</b> <a href="#">microprocessor.vhd</a>	<b>Design Hierarchy</b> <a href="#">cpuhierarchy.jpg</a>	<b>Test Bench of CPU Design</b> <a href="#">TB_mp.vhd</a>
<b>Control Unit</b> <a href="#">ctrl_unit.vhd</a>	<b>Block Diagram</b> <a href="#">cpublock.jpg</a>	<b>Data Path</b> <a href="#">datapath.vhd</a>
<a href="#">controller.vhd</a> <a href="#">PC.vhd</a> <a href="#">IR.vhd</a> <a href="#">bigmux.vhd</a>	<b>Memory</b> <a href="#">memory.vhd</a>	<a href="#">smallmux.vhd</a> <a href="#">reg_file.vhd</a> <a href="#">alu.vhd</a> <a href="#">obuf.vhd</a>
<b>Synopsys Sythesis Script Files</b> <a href="#">cpusyn.scr</a> <a href="#">syn_ctrl_unit.inc</a> , <a href="#">syn_PC.inc</a> , <a href="#">syn_IR.inc</a> , <a href="#">syn_datapath.inc</a> , <a href="#">syn_reg_file.inc</a> <a href="#">syn_alu.inc</a> , <a href="#">syn_memory.inc</a> <a href="#">syn_controller.inc</a> , <a href="#">syn_obuf.inc</a> , <a href="#">syn_bigmux.inc</a> , <a href="#">syn_smallmux.inc</a>		
<b>Behavior Simulation Script File</b> <a href="#">cpusim.scr</a>	<b>Simulation Waveform</b> <a href="#">cpusim1.jpg</a> <a href="#">cpusim2.jpg</a>	<b>Gate-Level Simulation Script File</b> <a href="#">cpugatesim.scr</a>

## ● Discussion V: VHDL vs. Verilog

There are now two industry standard hardware description languages, VHDL and Verilog. It is important that a designer knows both of them although we are using only VHDL in class. Verilog is easier to understand and use. For several years it has been the language of choice for industrial applications that required both simulation and synthesis. It lacks, however, constructs needed for system level specifications. VHDL is more complex, thus difficult to learn and use. However it offers a lot more flexibility of

the coding styles and is suitable for handling very complex designs. Here is a [great article](#) to explain their difference and tradeoffs.

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## **Appendix: Modeling a real industry chip - HD 6402** **(ESD Chapter 4)**

### I. [Specification](#) of HD 6402

#### II. Behavior Modeling of UART Transmitter

(1) Behavior [Code](#) (2) Gate-level [design](#) (3) Test Benches - [1](#), [2](#), [3](#) (4) Synopsys

##### Simulation

Case#1: one 8-bit word, 1 start, 2 stops, and even parity, or Data=11000101, Control Word=11011. ( Gate-level [Simulation](#) )

Case#2: three 5-bit words, 1 start, 1 stop, and no parity, or Data=11010 & 00101 & 10001, Control Word=00100. ( Gate-level [Simulation](#) )

Case#3: two 6-bit words, 1 start, 2 stops, and odd parity, or Data=110010 & 101101, Control Word=01000. ( Gate-level [Simulation](#) )

#### III. Behavior Modeling of UART Receiver

(1) Behavior [Code](#) (2) Gate-level [design](#) (3) Test Benches - [1](#), [2](#), [3](#) (4) Synopsys

##### Simulation

Case#1: two 6-bit words, 1 start, 2 stops, and even parity, (Data=111001 & 100101, Control Word=01101). ( Gate-level Design [Simulation](#) )

Case#2: one 8-bit words, 1 start, 1 stop, and odd parity, (Data=10111001, Control Word=11000). ( Gate-level Design [Simulation](#) )

Case#3: three 5-bit words, 1 start, 1 stop, and no parity, (Data=01001 & 01110 & 00100, Control Word=00010. ( Gate-level Design [Simulation](#) )

#### IV. Structural Modeling of HD-6402

(1) Behavior [Code](#) (2) Gate-level [design](#) (3) Test [Bench](#) (4) Synopsys [Simulation](#)

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