

# Wavelet transforms for unbounded signals

part 2: towards wavelets from filters

Applying our wavelet transforms to  
unbounded signals

$$l(z/n) \rightsquigarrow l_o(z)$$

Haar

$$x \rightarrow \begin{bmatrix} x_{\text{even}} \\ x_{\text{odd}} \end{bmatrix} \rightsquigarrow \frac{1}{2} \begin{bmatrix} I & I \\ I & -I \end{bmatrix} \begin{bmatrix} x_{\text{even}} \\ x_{\text{odd}} \end{bmatrix} = \begin{bmatrix} s \\ d \end{bmatrix}$$

$$\text{CDF}(2,2) \quad x \mapsto \begin{bmatrix} x_{\text{even}} \\ x_{\text{odd}} \end{bmatrix} \rightsquigarrow \text{DUP} \begin{bmatrix} x_{\text{even}} \\ x_{\text{odd}} \end{bmatrix} = \begin{bmatrix} s \\ d \end{bmatrix}$$

$$\frac{1}{4\sqrt{2}} \begin{bmatrix} -s^{-1} + 6I - s & 2I + 2s \\ -2s^{-1} - 2I & 4I \end{bmatrix}$$

$\boxed{l_0(z)}$  case

Def  $x_{2\downarrow}[k] = x[2k]$

and  $\boxed{2\downarrow} : l_0(z) \rightarrow l_0(z)$

$$x \longmapsto x_{2\downarrow}$$

Def  $(Sx)[k] = x[k-1]$

$x_0 = \boxed{2\downarrow} x$

" $x_{\text{even}}$ "

$x_1 = \boxed{2\downarrow} S^{-1} x$

" $x_{\text{odd}}$ "

$$x_0[k] = \boxed{[2\downarrow]} x[k] = x[2k]$$

$$\begin{aligned} x_1[k] &= \boxed{[2\downarrow]} S^{-1}x[k] = (S^{-1}x)[2k] \\ &= x[2k+1] \end{aligned}$$

Use these like  $x_{\text{even}} \uparrow \downarrow x_{\text{odd}}$ .

Standard Wavelet Transforms on  
 $\lambda_0(z)$

Haar :  $x \sim \begin{bmatrix} x_0 \\ x_1 \end{bmatrix} \sim \frac{1}{2} \begin{bmatrix} I & I \\ -I & I \end{bmatrix} \begin{bmatrix} x_0 \\ x_1 \end{bmatrix} = \begin{bmatrix} s \\ d \end{bmatrix}$

$s_{q1} \quad s = \frac{1}{2}(x_0 + x_1) \quad d = \frac{1}{2}(x_1 - x_0)$

$$\begin{aligned} s[k] &= \frac{1}{2}(x_0[k] + x_1[k]) \\ &= \frac{1}{2}(x[2k] + x[2k+1]) \end{aligned}$$

$$d[k] = \frac{1}{2} \left( x[2k+1] - x[2k] \right)$$

Standard Wavelet Transforms on  
 $\lambda_0(z)$

$$\text{CDF}(z, 2) : \mathbf{x} \rightsquigarrow \begin{bmatrix} x_0 \\ x_1 \end{bmatrix} \rightsquigarrow \frac{1}{4\sqrt{2}} \begin{bmatrix} -S^{-1} + 6I - S & 2I + 2S \\ -2S^{-1} - 2I & 4I \end{bmatrix} \begin{bmatrix} x_0 \\ x_1 \end{bmatrix}$$

$$= \begin{bmatrix} s \\ a \end{bmatrix}$$

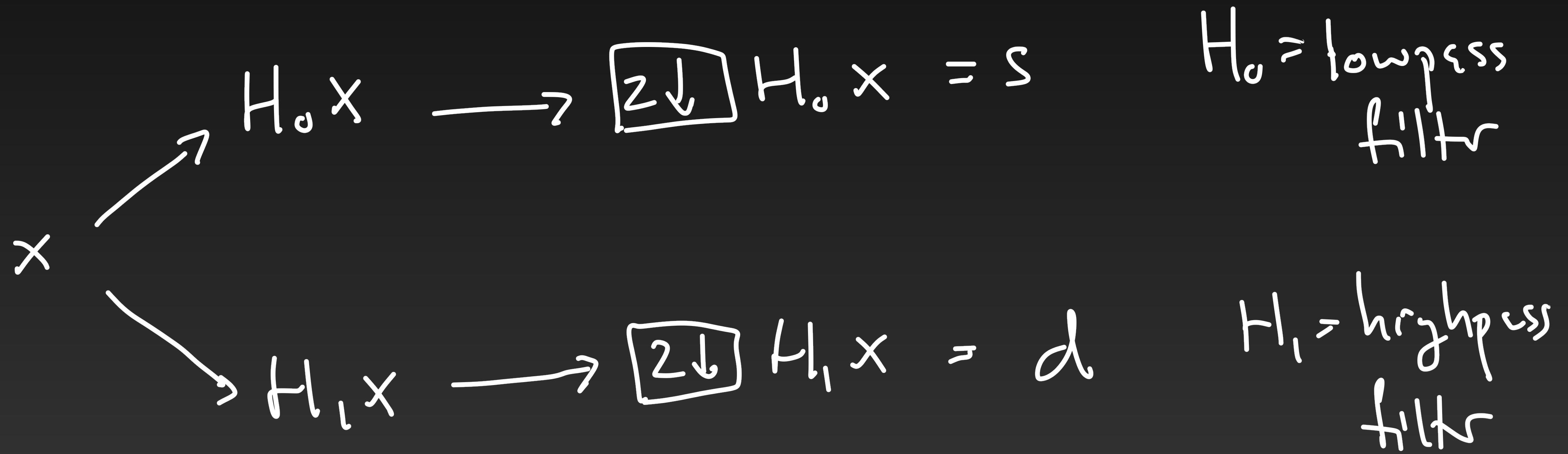
$$\rightsquigarrow s[k] = -S^{-1}x_0[k] + 6x_0[k] - Sx_0[k] + 2x_1[k] + 2Sx_1[k]$$

$$= -x[2k+2] + 6x[2k] - x[2k-2] + 2x[2k+1] + 2x[2k-1]$$

...

But why?

Alternate construction of wavelets



Filters (FIR = finite impulse response)

Described by convolutions

$$x \rightarrow h * x$$

Def  $h * x = \sum_n \left( \sum_{j+k=n} h[j] x[k] \right) e_n$

Alternatively Def  $e_j * e_k = e_{j+k}$  : foil (distribute)

$$(e_{-1} + 3e_2 + e_4) * (e_1 - e_3)$$

$$\begin{aligned}
 &= \underbrace{e_{-1} * e_1}_{e_0} - \underbrace{e_{-1} * e_3}_{e_2} + \underbrace{3e_2 * e_1}_{e_3} - \underbrace{3e_2 * e_3}_{e_5} + \underbrace{e_4 * e_1}_{e_6} \\
 &\quad - \underbrace{e_4 * e_3}_{e_7}
 \end{aligned}$$

$$= e_0 - e_2 + 3e_3 - 3e_5 + e_5 - e_7$$

$$= e_0 - e_2 + 3e_3 - 2e_5 - e_7$$

Ex:

$$h_0 = e_{-1} + e_0$$

$$h_0 * x = x + Sx$$

$$h_1 = e_{-1} - e_0$$

$$h_1 * x = -x + Sx$$

$$S = \boxed{2\downarrow} (h_0 * x)[k] = x[2k] + x[2k+1]$$

$$d = \boxed{2\downarrow} (h_1 * x)[k] = -x[2k] + x[2k+1]$$

If we have a filter  $l_0(z) \rightarrow l_0(z)$ , given by  $h$

$$X(z) H(z) = z\text{-transform of } X * h.$$

$\Rightarrow$  Frequencies are affected by  $h$  in proportion  
to their presence in  $h$ .

$$\text{Haar: } h_0 = e_{-1} + e_0 \sim H_0(z) = z^{-1} + 1$$

$$h_1 = e_{-1} - e_0 \sim H_1(z) = z^{-1} - 1$$

Power spectral densities:

$$P_0(\omega) = |H_0(e^{i\omega})|^2 = (e^{-i\omega} + 1)(e^{i\omega} + 1) = 2 + 2 \cos \omega$$

$$P_1(\omega) = 2 - 2 \cos \omega$$

$\omega = 0$  low  
 $\omega = \pi$  high

Hear inconvenience

$H_0$  cuts off highest freq. well

cuts off high-ish freq so-so

$H_1$  --- similar.

Quest: Find filters which do this better &  
which are reversible!