# Control Systems

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Abstract—This manual is an introduction to control systems based on GATE problems.Links to sample Python codes are available in the text.

Download python codes using

svn co https://github.com/gadepall/school/trunk/control/codes

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#### 1 Signal Flow Graph

2 Gain of Feedback Circuits

### 2.1 Current Amplifiers

2.1.1. For the feedback current amplifier shown in 2.1.1, Draw the Small-Signal Model eglect the Early effect in  $Q_1$  and  $Q_2$ .

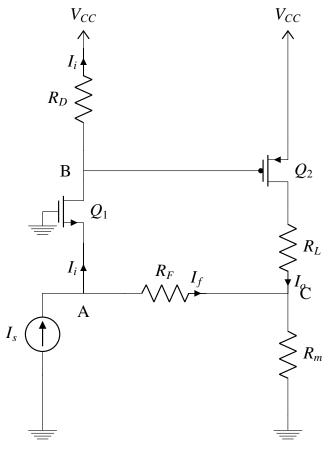


Fig. 2.1.1

**Solution:** While drawing a Small-Signal Model, we ground all constant voltage sources and open all constant current sources. All Small-Signal paramters are obtained from DC-Analysis of the circuit. Neglecting Early effect, in Small-Signal Analysis a N-MOSFET is modelled as a Current Source with value of current equal to  $g_m v_{gs}$  flowing from Drain to

Source. Whereas a P-MOSFET is modelled as a Current Source with value of current equal to  $g_m v_{sg}$  flowing from Source to Drain.

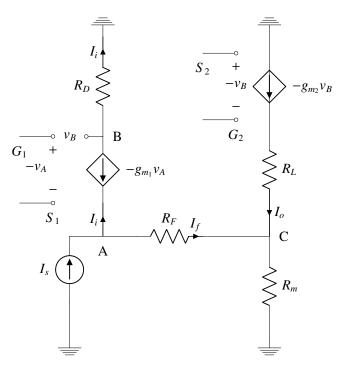


Fig. 2.1.1: Small Signal Model

2.1.2. Describe how the given circuit is a Negetive Feedback Current Amplifier.

> **Solution:** For the feedback to be negative,  $I_f$ must have the same polarity as  $I_s$ . To ascertain 2.1.5. Find the Expression for the Closed-Loop Gain that this is the case, we assume an increase in  $I_s$  and follow the change around the loop: An increase in  $I_s$  causes  $I_i$  to increase and the drain voltage of  $Q_1$  will increase. Since this voltage is applied to the gate of the p-channel device  $Q_2$ , its increase will cause  $I_o$ , the drain current of  $Q_2$ , to decrease. Thus, the voltage across  $R_M$  will decrease, which will cause  $I_f$ to increase. This is the same polarity assumed for the initial change in  $I_s$ , verifying that the feedback is indeed negative.

2.1.3. Find the Expression for the Open-Loop Gain  $G = \frac{I_o}{I}$ , from the Small-Signal Model. For simplicity, neglect the Early effect in  $Q_1$  and  $Q_2$ .

**Solution:** In Small-Signal Model,

$$v_B = I_i R_D (2.1.3.1)$$

$$v_{gs_2} = v_B = I_i R_D (2.1.3.2)$$

In Small-Signal Analysis, P-MOSFET is modelled as a current source where current flows from Source to Drain. So, the value of current flowing from Source to Drain in P-MOSFET is,

$$I_o = -g_{m_2} v_{gs_2} = -g_{m_2} I_i R_D (2.1.3.3)$$

So, the Open-Circuit Gain is

$$G = \frac{I_o}{I_i} = -g_{m_2} R_D \tag{2.1.3.4}$$

2.1.4. Find the Expression of the Feedback Factor  $H = \frac{I_f}{I_r}$ , from Small-Signal Model. For simplicity, neglect the Early effect in  $Q_1$  and  $Q_2$ .

> $I_o$  is fed to a current divider formed by  $R_M$ and  $R_F$ . Current Mixing results in a reduced input resistance, the voltage at the source node of  $Q_2$  will be close to zero. Hence the voltage at point 'A' is very small and is considered,  $v_A \simeq 0$ . So  $R_F$  and  $R_M$  are parallel and Voltage Drop across them is same.

$$(I_o + I_f)R_M \simeq -I_f R_F$$
 (2.1.4.1)

$$\frac{I_f}{I_o} \simeq -\frac{R_M}{R_F + R_M} \tag{2.1.4.2}$$

So, the Feedback Factor,

$$H \equiv \frac{I_f}{I_o} \simeq -\frac{R_M}{R_F + R_M} \tag{2.1.4.3}$$

 $T = \frac{I_o}{I_s}$ . For simplicity, neglect the Early effect in  $Q_1$  and  $Q_2$ .

#### **Solution:**

From Open-Loop Gain and Feedback Factor,

$$I_s = I_i + I_f$$
 (2.1.5.1)

$$I_s = \frac{I_o}{G} + HI_o$$
 (2.1.5.2)

$$GI_s = I_o(1 + GH)$$
 (2.1.5.3)

$$\frac{I_o}{I_s} = \frac{G}{1 + GH}$$
 (2.1.5.4)

$$\frac{I_o}{I_s} = -\frac{g_{m_2} R_D}{1 + g_{m_2} R_D / \left(1 + \frac{R_F}{R_M}\right)}$$
(2.1.5.5)

So the Block Diagram of Feedback Current Amplifier is

where 
$$G = -g_{m_2}R_D$$
 and  $H = -\frac{R_M}{R_E + R_M}$ 

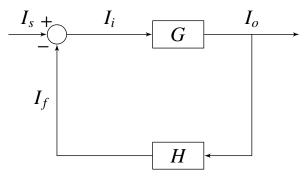


Fig. 2.1.5

So, the value of Closed-Loop Gain is

$$T = \frac{I_o}{I_s} = -\frac{g_{m_2}R_D}{1 + g_{m_2}R_D/\left(1 + \frac{R_F}{R_M}\right)}$$
 (2.1.5.6)

- 3 Bode Plot
- 4 SECOND ORDER SYSTEM
- 5 Routh Hurwitz Criterion
  - 6 STATE-SPACE MODEL
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      - 11 Oscillator
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  - 14 PID Controller