

# ECSE 493 — Controls&Robotics Lab

## Lab 1 Report

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### 1

(a) The equation of the DC motor given in the description is:

$$J_m \ddot{\theta} + \left( b + \frac{K_t K_m}{R_a} \right) \dot{\theta} = \frac{K_t}{R_a} v_a$$

Substituting the numbers provided in the equation above, we get:

$$0.01 \ddot{\theta} + \left( 0.001 + \frac{0.02 * 0.02}{10} \right) \dot{\theta} = \frac{0.02}{10} v_a \quad (1)$$

$$0.01 \ddot{\theta} + 0.00104 \dot{\theta} = 0.002 v_a \quad (2)$$

We then apply the Laplace Transform and get the following:

$$0.01 s^2 \Theta + 0.00104 s \Theta = 0.002 V \quad (3)$$

$$\frac{\Theta}{V} = \frac{1}{s} \frac{0.002}{0.01 s + 0.00104} \quad (4)$$

$$\frac{\dot{\Theta}}{V} = \frac{0.002}{0.01 s + 0.00104} \quad (5)$$

(b) Next we input this equation in Matlab to plot the steady-state response:

```
1 num = 0.002;  
2 den = [0.01 0.00104];  
3 T = tf(num,den)  
4 step(T)
```

Which yields the plot in Figure 1

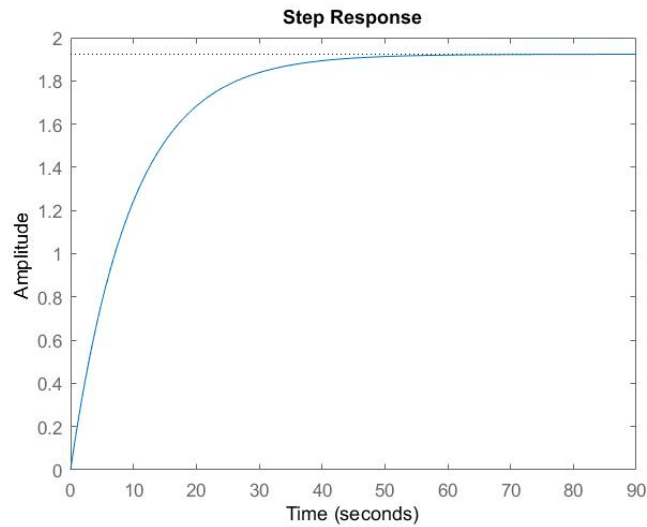


Figure 1: Step response of motor shaft speed

With this plot, we can estimate the steady-state speed of the motor to be about 1.95 rad/sec.

(c) If we assume our estimate in part (b), then 99% of the final speed would be 1.88 rad/sec, which I estimate arrives at around 45 seconds

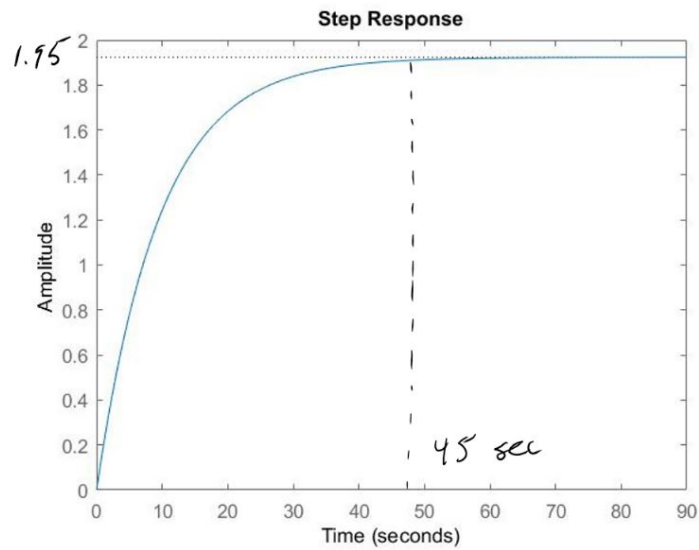


Figure 2: Steady-state and rise-time estimation

(d) The final value theorem goes as follow

$$\lim_{t \rightarrow \infty} f(t) = \lim_{s \rightarrow 0} sF(s) \quad (6)$$

$$\lim_{s \rightarrow 0} s \frac{1}{s} \frac{0.002}{0.01s + 0.00104} = \frac{0.002}{0.00104} = 1.923 \quad (7)$$

This is 0.27 rad/sec error, or a 14% overshoot, so my estimate was not that great.

(e) As we derived earlier, the transfer function between shaft angle and voltage is

$$\frac{\Theta}{V} = \frac{1}{s} \frac{0.002}{0.01s + 0.00104}$$

(f) If we add feedback to our system, we get the following block diagram, where  $G(s)$  is our plant, in this case, the motor. This system can be modeled by the following transfer function:



Figure 3: Motor with feedback system

$$H(s) = \frac{G(s)}{1 + G(s)}$$

Next we want to add a gain of  $K$  to the forward path of the system, this yields the following diagram and is described by the following transfer function:



Figure 4: Motor with feedback system and gain on forward path

$$H(s) = \frac{kG(s)}{1 + kG(s)}$$

(g)  $k$  is gain, unit-less, or V/V

- (h) From the system diagrams in figure 4, we can clearly see the input to the plant  $G(s)$  is  $k * e(s)$ . We can a value for this expression like so. First we find an expression for  $e(s)$  and one for  $\Theta(s)$ :

$$e(s) = V(s) - \Theta(s) \quad (8)$$

$$\Theta(s) = e(s) * k * G(s) \quad (9)$$

We isolate  $\Theta(s)$  in both expression, equate the equations and isolate  $k * e(s)$

$$\Theta(s) = V(s) - e(s) \quad (10)$$

$$V(s) - e(s) = e(s) * k * G(s) \quad (11)$$

$$e(s) (k * G(s) + 1) = V(s) \quad (12)$$

$$e(s) = \frac{V(s)}{k * G(s) + 1} \quad (13)$$

$$k * e(s) = \frac{k * V(s)}{k * G(s) + 1} \quad (14)$$

- (i) Next, we will replace out plant equation in the transfer function of the feedback system

$$H(s) = \frac{kG(s)}{1 + kG(s)}$$

$$\text{where } G(s) = \frac{1}{s} \frac{0.002}{0.01s + 0.00104}$$

$$H(s) = \frac{k \left( \frac{0.002}{0.01s^2 + 0.00104s} \right)}{1 + k \left( \frac{0.002}{0.01s^2 + 0.00104s} \right)} \quad (15)$$

$$H(s) = \frac{\frac{k0.002}{0.01s^2 + 0.00104s}}{\frac{0.002k + 0.01s^2 + 0.00104s}{0.01s^2 + 0.00104s}} \quad (16)$$

$$H(s) = \frac{k0.002}{0.01s^2 + 0.00104s} \frac{0.01s^2 + 0.00104s}{0.002k + 0.01s^2 + 0.00104s} \quad (17)$$

$$H(s) = \frac{k0.002}{0.002k + 0.01s^2 + 0.00104s} \quad (18)$$

- (j) We can use Matlab and plot this transfer function for different values of  $k$  : We find that a maximum

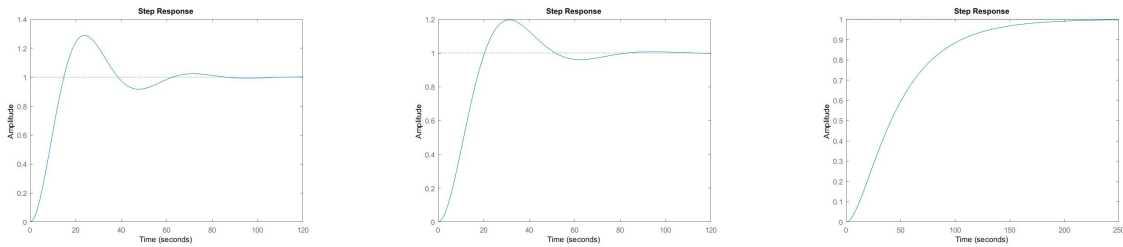
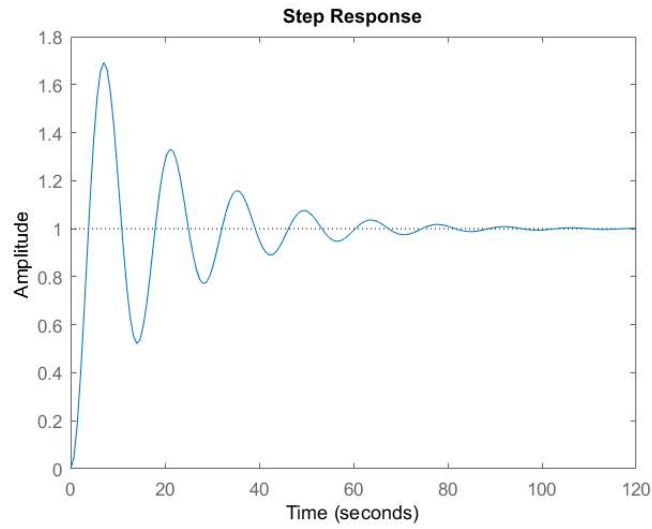


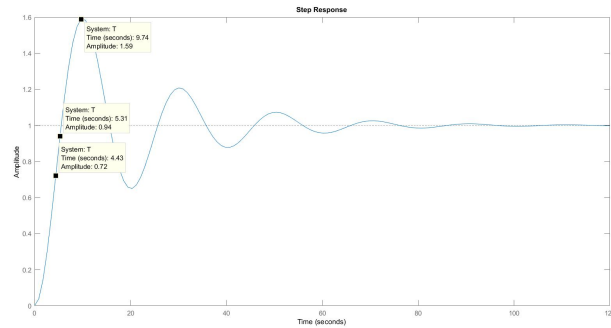
Figure 5: Servo motor response with different gains

value of 0.064 is acceptable to not overshoot past 20%.

- (k) In order to reduce our rise-time, we will increase our gain  $k$ . If we increase it to a value of 1, our rise-time goes down under 4 seconds, and lead to the following step response:

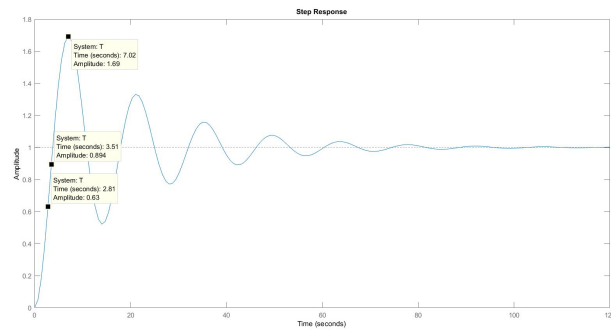


- (l) With a value of 0.5 for  $k$ , we get the following response: For this diagram, we can approximate a



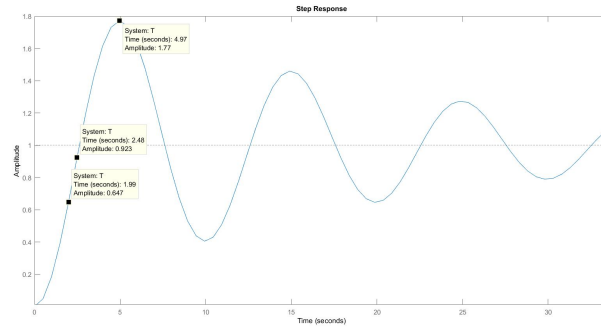
rise-time of 4.7 seconds and an overshoot of 59%.

With a value of 1 for  $k$ , we get the following response: For this diagram, we can approximate a rise-time



of 3.2 seconds and an overshoot of 69%.

With a value of 2 for  $k$ , we get the following response: For this diagram, we can approximate a rise-time



of 2.4 seconds and an overshoot of 77%.

- (m) With this experiment, we see that the parameter  $k$  influences the rise-time of the system, more specifically, how aggressive the corrections are. We can see that a large  $k$  value means a very fast acting system, but the system has a high inertia, so it is not able to stop when it approaches its target. A large gain leads to fast rise-time, large oscillations, and long settling time.

## 2

- (a) In this code  $p$  and  $q$  are the coefficients of the numerator and denominator of a transfer function, specifically

$$\frac{s + 1}{s^3 + 5s^2 + 6s}$$

- (b) The root locus diagram show how the poles of a system will change as some parameters in the transfer function change from 0 to  $\infty$ .