

Clase teórica 7 (Videos)

Funciones bilineales (Tabla 3.1) Cap 3. 5ch

$$T(s) = K \frac{s+Z}{s+P}$$

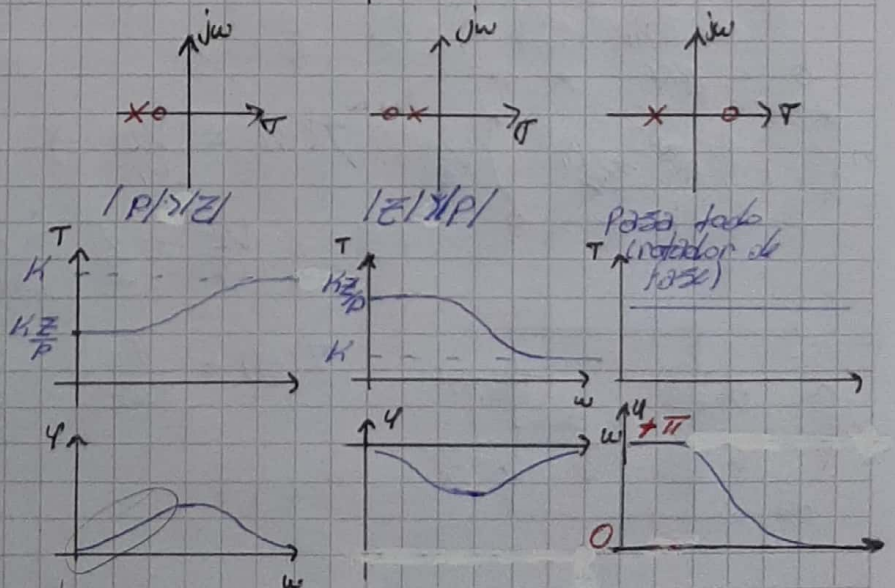
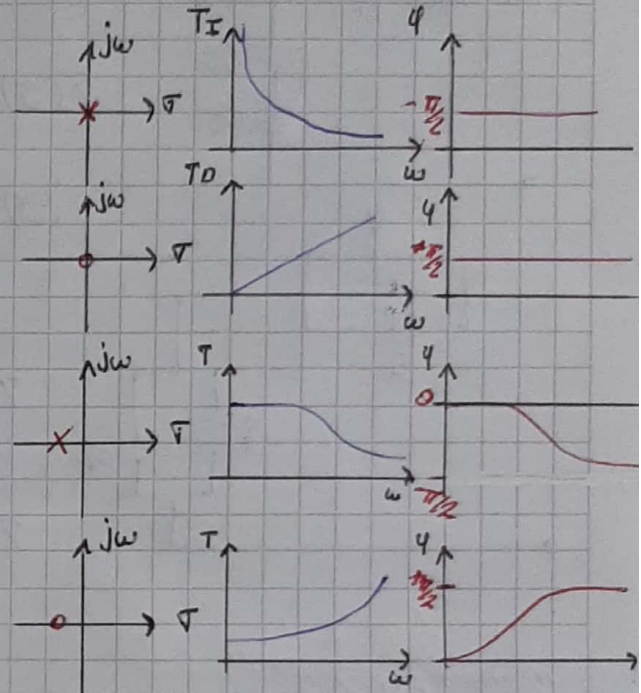
$$T_I(s) = \frac{K}{s} \quad \text{integrador}$$

$$T_D(s) = sK \quad \text{derivador (cero en el origen)}$$

$$T(s) = \frac{K}{s+P} \quad \text{integrador real}$$

$$T(s) = K(s+Z) \quad \text{derivador real}$$

$$T(s) = K \frac{(s+Z)}{(s+P)}$$



En este BW el retardo de grupo es cte

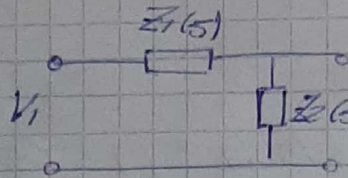
$$D = \frac{d\phi}{d\omega}$$

NOTA +2h [9:45] → [14:00]

Implementaciones PASIVAS

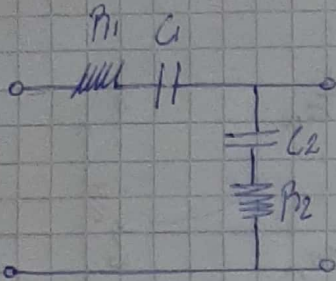
$$T(s) = K \frac{s+Z}{s+P}$$

$K < 1$



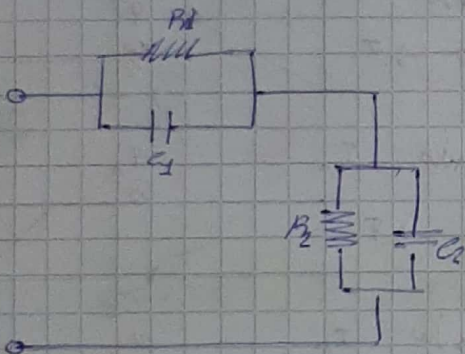
Solo con un tipo de componente
reactivo
RL o RC

$$T(s) = \frac{V_2}{V_1} = \frac{Z_2}{Z_1 + Z_2} = \frac{Y_1}{Y_1 + Y_2}$$



$$T(s) = \frac{P_2 + \sqrt{s}C_2}{P_1 + P_2 + \sqrt{s}C_1 + \sqrt{s}C_2}$$

$$T(s) = \frac{P_2}{P_1 + P_2} \frac{s + \sqrt{s}C_2 P_2}{s + \frac{C_1 + C_2}{C_1 C_2 (P_1 + P_2)}}$$

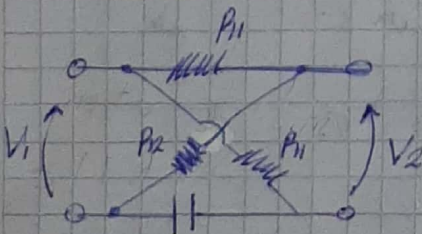


$$T(s) = \frac{G_1 + sC_1}{G_1 + G_2 + sC_1 + sC_2}$$

$$T(s) = \frac{C_1}{C_1 + C_2} \frac{s + \frac{G_1/C_1}{s + \frac{G_1 + G_2}{C_1 + C_2}}}{s + \frac{G_1 + G_2}{C_1 + C_2}}$$

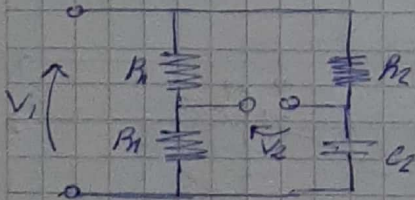
K $\frac{Z}{P}$

Circuito Lattice



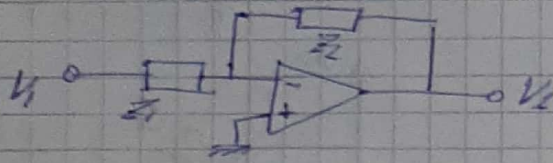
$$T(s) = \frac{V_2}{V_1} \Rightarrow V_2 = \frac{V_1}{2} - \frac{G_2}{G_2 + sC} V_1$$

$$\frac{V_2}{V_1} = \frac{1}{2} - \frac{G_2}{G_2 + sC} = \frac{sC - G_2}{2(sC + G_2)}$$

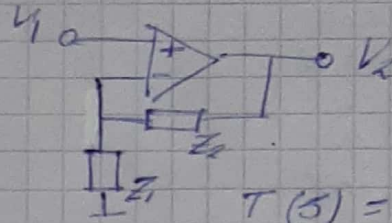


$$T(s) = \frac{1}{2} \cdot \frac{s - G_2/C_2}{s + G_2/C_2}$$

Implementaciones activas

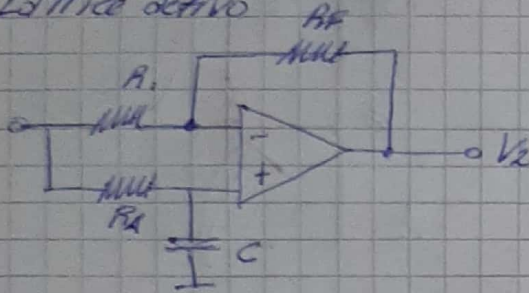


$$T(s) = -\frac{Z_2}{Z_1}$$



$$T(s) = 1 + \frac{Z_2}{Z_1}$$

Lattice activo



$$T(s) = -K \frac{s - \frac{1}{RCRA}}{s + \frac{1}{RCRA}}$$

$$K = \frac{RF}{R1}$$

Funciones Biquadráticas (Cap 5 Sch)

$$T(s) = \frac{s^2 + \frac{\omega_z}{Q_z} s + \omega_z^2}{s^2 + s \frac{\omega_0}{Q_p} + \omega_0^2} K$$

$$T_{LP} = \frac{\omega_0^2 K}{s^2 + s \frac{\omega_0}{Q} + \omega_0^2}$$

$$T_{BP} = \frac{s \frac{\omega_0}{Q} K}{s^2 + s \frac{\omega_0}{Q} + \omega_0^2}$$

$$T_{HP} = \frac{s^2 K}{s^2 + s \frac{\omega_0}{Q} + \omega_0^2}$$

$$T_{BP} = \frac{s^2 - s \frac{\omega_0}{Q} + \omega_0^2}{s^2 + s \frac{\omega_0}{Q} + \omega_0^2}$$

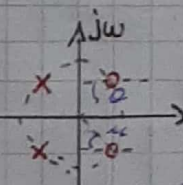
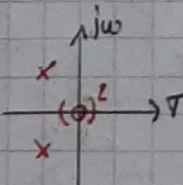
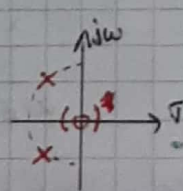
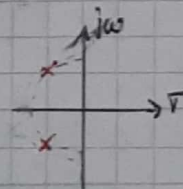
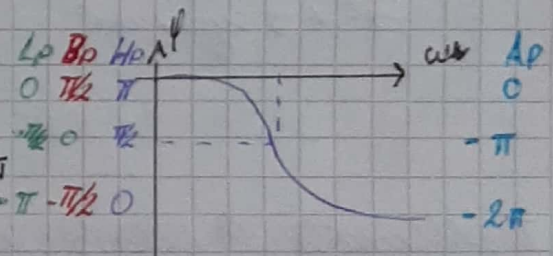


Fig 5.10 del Sch



$$\phi = \phi_z - \phi_p$$

$$\pi + \theta$$

$$\pi - \mu$$

$$\phi_z|_{\omega=0} \Rightarrow \mu = 0 \quad 2\pi$$

$$\omega \rightarrow \infty \quad \left. \begin{array}{l} \theta/\phi_0 \rightarrow -\pi/2 \\ \mu/\mu_0 \rightarrow \pi/2 \end{array} \right\} \pi$$

~~del resto todo~~

~~$H = H_z$~~ • $T_{Notch} = T_{BE} = \frac{s^2 + \omega_0^2}{s^2 + \frac{s\omega_0}{Q} + \omega_0^2}$

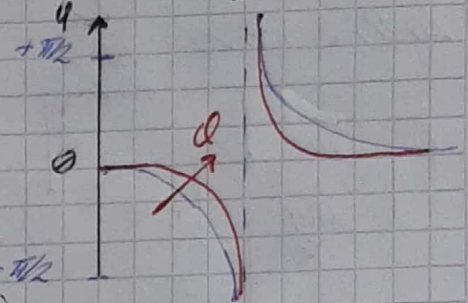
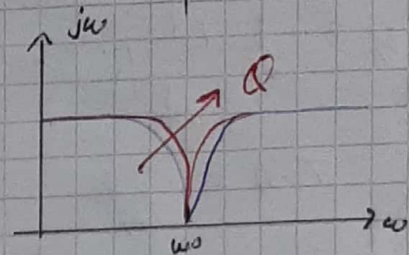
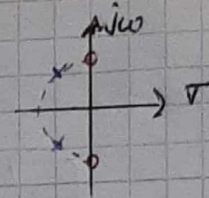
Fase

$\omega < \omega_0$
 $\phi_z = 0$

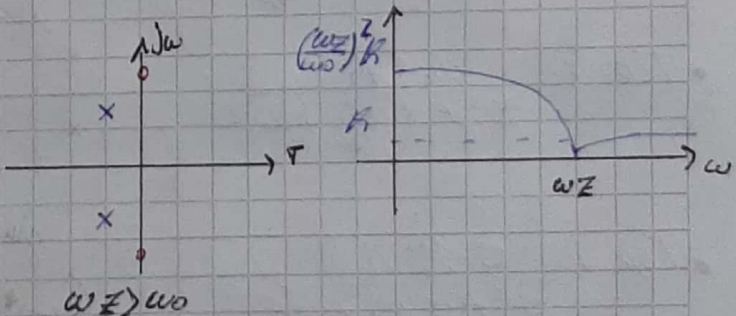
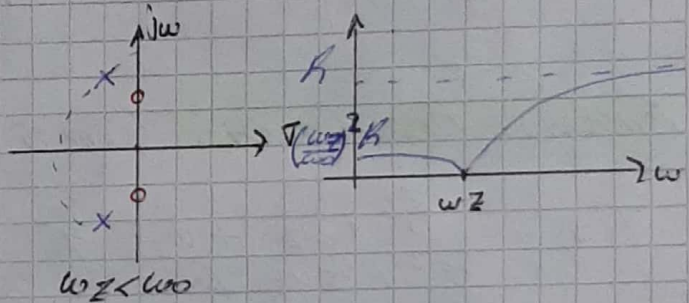
$\omega > \omega_0$
 $\phi_z = +\pi$

$\phi_z + \pi \rightarrow -\pi$

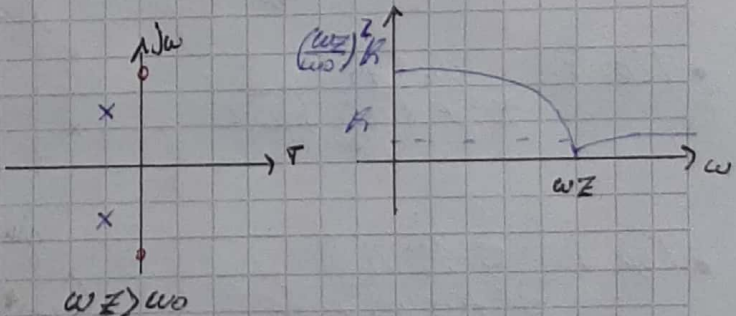
~~Fase~~



• $T_{HP} = \frac{s^2 + \omega_z^2}{s^2 + \frac{s\omega_0}{Q} + \omega_0^2}$

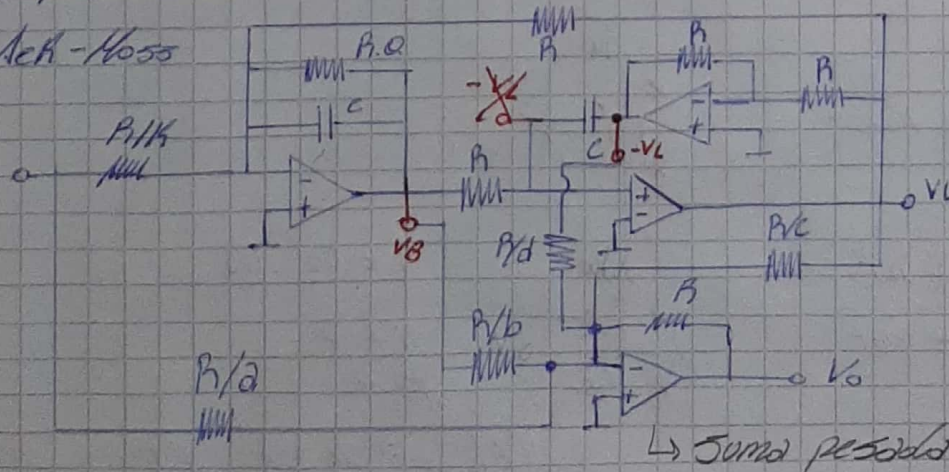


• $T_{LP} = \frac{s^2 + \omega_z^2}{s^2 + \frac{s\omega_0}{Q} + \omega_0^2}$



Implementación de circuitos por suma de transformadas

Act-Hoss



NOTA

$$\begin{cases} \frac{V_L}{V_i} = - \frac{K\omega_0^2}{s^2 + 5\frac{\omega_0}{Q} + \omega_0^2} = T_L(s) \\ \frac{V_B}{V_i} = - \frac{5KQ(\frac{\omega_0}{Q})}{s^2 + 5\frac{\omega_0}{Q} + \omega_0^2} \end{cases}$$

$$T(s) = \frac{s^2 + 5\frac{\omega_0}{Q} + \omega_0^2}{s^2 + 5\frac{\omega_0}{Q} + \omega_0^2} = \frac{V_o}{V_i}$$

Transmisorncia resonator notch

$$\begin{aligned} T_{LN}(s) &= \frac{s^2 + \omega_0^2}{s^2 + 5\frac{\omega_0}{Q} + \omega_0^2} \\ &= \frac{s^2}{\dots} + \frac{\omega_0^2}{\dots} \\ &\quad \text{HP} \quad + \quad \text{NL. H} \end{aligned}$$

$$V_o = -[aV_i + bV_B + cV_L + d(-V_L)]$$

$$T(s) = -[a + bT_{BP} + cT_{LP} + d(-T_{LP})]$$

Para formar una ecuación completa, sumo todas las salidas de cada etapa x CTES y luego le doy valores a esas CTES

$$T(s) = -\frac{a(s^2 + 5\frac{\omega_0}{Q} + \omega_0^2) + b(-5K\omega_0) + (c-d)(-K\omega_0^2)}{s^2 + 5\frac{\omega_0}{Q} + \omega_0^2}$$

$$T(s) = -\frac{s^2(a + 5\frac{\omega_0}{Q}(a - bKQ)) + \omega_0^2[a + K(d - c)]}{s^2 + 5\frac{\omega_0}{Q} + \omega_0^2}$$

$$T(s) = -\frac{s^2(a + 5\frac{\omega_0}{Q}(a - bKQ)) + \omega_0^2[a + K(d - c)]}{s^2 + 5\frac{\omega_0}{Q} + \omega_0^2}$$

Ejemplo pasa banda

$$\begin{cases} Q=0 & R/d = 0 \\ K(d-c) = 0 & \Rightarrow d=c \text{ o sea } R/d = R/c \end{cases}$$

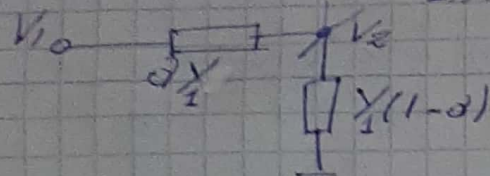
EJ. Notch

$$\begin{cases} a \neq 0 \text{ preferentemente } a=1 \\ 1 - bKQ = 0 & \Rightarrow b = \frac{1}{KQ} \end{cases}$$

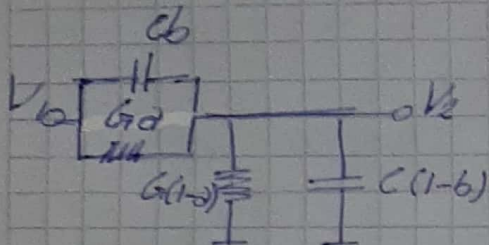
Voltage Feed Forward

Bicúadrática completa

1/1 T a component off ground



Es el mismo componente que se lo separa y se le pone una tensión en medio (V_2)



$$\frac{V_2}{V_1} = \frac{G_d a + s C b}{G_d a + s C b + G(1-a) + s C(1-b)}$$

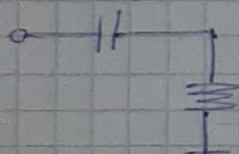
$$= \frac{G_d a + s C b}{G + s C}$$

Los polos no varían

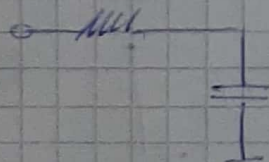
$$\frac{V_2}{V_1} = b \frac{s + \frac{G_d a}{C b}}{s + \frac{G}{C}}$$

$$0 \leq a, b \leq 1$$

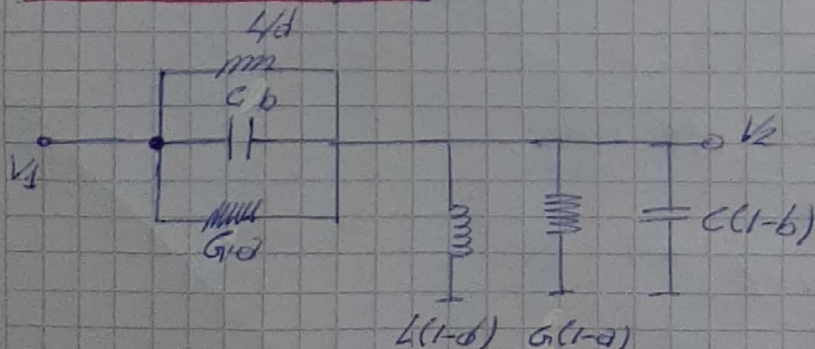
Si $a=0$ y $b=1$ Resaca alto



Si $a=1$ y $b=0$ Resaca bajo



Bicúadrática completa



$$T(s) = \frac{V_2}{V_1} = \frac{G_d a + s C b + \frac{d}{s L}}{G + s C + \frac{1}{s L}} = \frac{s L G_d a + s^2 L C b + d}{s^2 + s G L + \frac{1}{L C}}$$

NOTA

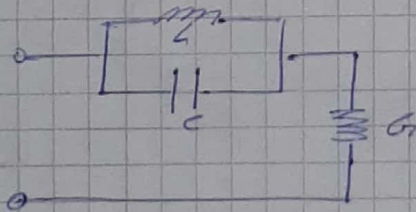
$$T(s) = \frac{dbX}{24} \frac{s^2 + s \frac{Gd}{Cb} + \frac{1}{LC} \cdot \frac{d}{b}}{s^2 + s \frac{G}{C} + \frac{1}{LC}}$$

$$\left[T(s) = b \frac{s^2 + s \frac{Gd}{Cb} + \frac{1}{LC} \frac{d}{b}}{s^2 + s \frac{G}{C} + \frac{1}{LC}} \right]$$

Ej: Notch

$$T(s) = \frac{s^2 + \omega_z^2}{s^2 + s\gamma_Q + 1}$$

$$\begin{cases} \frac{1}{LC} = 1 \\ d=0 \\ \frac{d}{b} = \omega_z^2 = 1 \end{cases} \quad \begin{array}{l} \text{0dB en la banda de paso si } b=1 \text{ y } d=1 \\ \hookrightarrow \text{pero esto sería un Notch puro sin} \\ \text{componentes levantados de notch} \end{array}$$



Ej: Notch paso bajos

$$T(s) = \frac{s^2 + \omega_z^2}{s^2 + s\gamma_Q + 1}$$

$$\frac{1}{LC} = 1$$

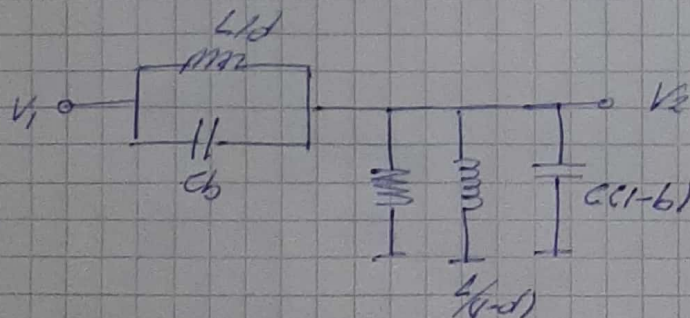
$$d=0$$

$$1 < \frac{d}{b} = \omega_z^2 \Rightarrow b < d$$

$$T(s)|_{s=0} = d$$

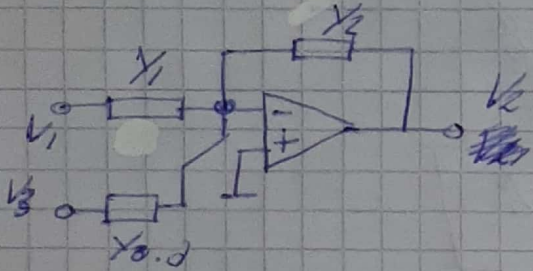
$$T(s)|_{s \rightarrow \infty} = b$$

$$\left. \begin{array}{l} d \leq 1 \\ b \leq 1 \end{array} \right\}$$



Levantar admitancias con circuitos activos

↳ se presintonan las tierras virtuales



Siendo $V_1 = V_3$

$$V_2 = - \frac{V_1 Y_1 + V_3 Y_{0.2}}{Y_2}$$

$$T(s) = - \frac{Y_1 + Y_{0.2}}{Y_2}$$