

DSP HW2

Dmitrii Masnyi

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Task 1

Design a low pass FIR filter with parameters: passband $F_{\text{pass}} = 5$ MHz; stopband $F_{\text{stop}} = 6$ MHz; attenuation at least 60 dB in the stopband (out-of-band attenuation).

Let the sampling frequency be $F_s = 50$ MHz.

Determine the design with the lowest computational complexity.

Provide code.

Solution:

- **Fir1 filter**

First of all, I tried to implement fir1 filter. Because I need to guarantee out-of-band attenuation at least 60 dB I used Blackman window (I tried to implement Hamming window but its attenuation was not sufficient enough). I just ran multiple times the function `fir1()` with different parameter n and choose n that gave me results that is specified in the problem statement. Finally, I designed fir1 filter with $n = 300$, which satisfies requirements.

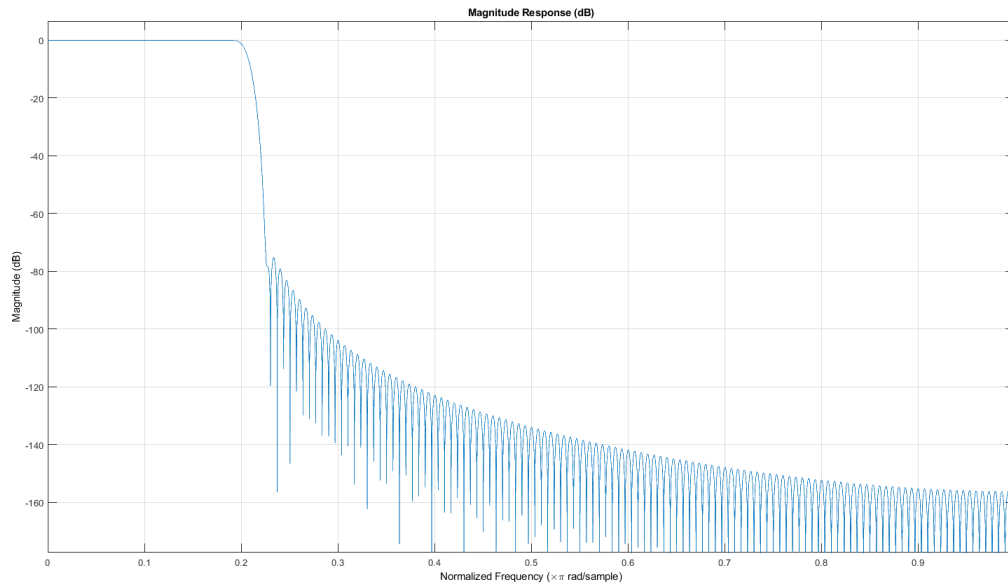


Figure 1: Magnitude response of fir1 filter with blackman window and $n = 300$.

- **Fir2 filter** Then I designed fir2 filter. I used function `fir2()` and ran it multiple with different parameter n and chose $n=210$, because fir2 with this parameter satisfies requirements.

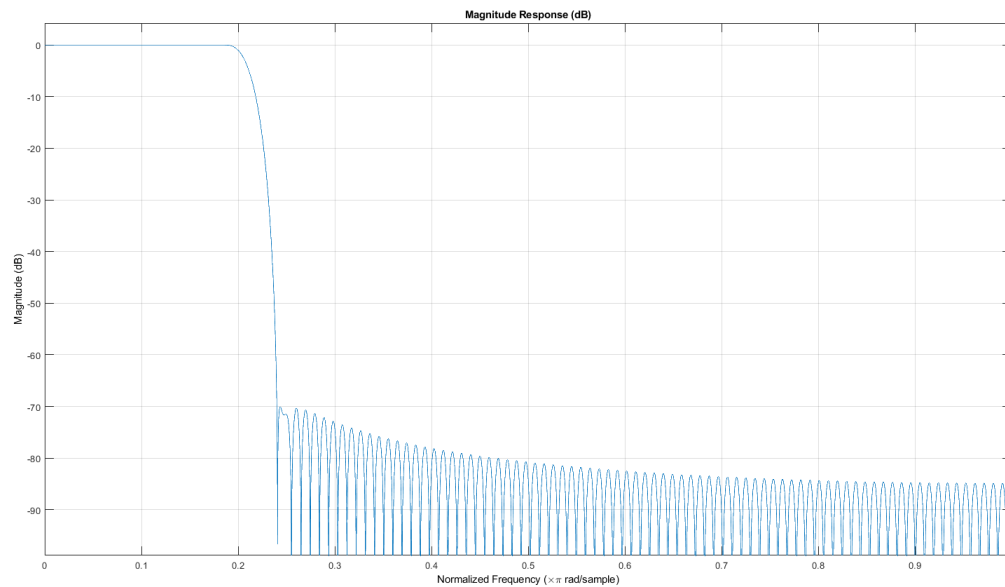


Figure 2: Magnitude response of fir2 filter with $n = 210$.

- **Firls filter**

Then I designed firls filter. I used function **firls()** and ran it multiple times with different parameter **n** and chose **n=185**, because firls with this parameter satisfies requirements.

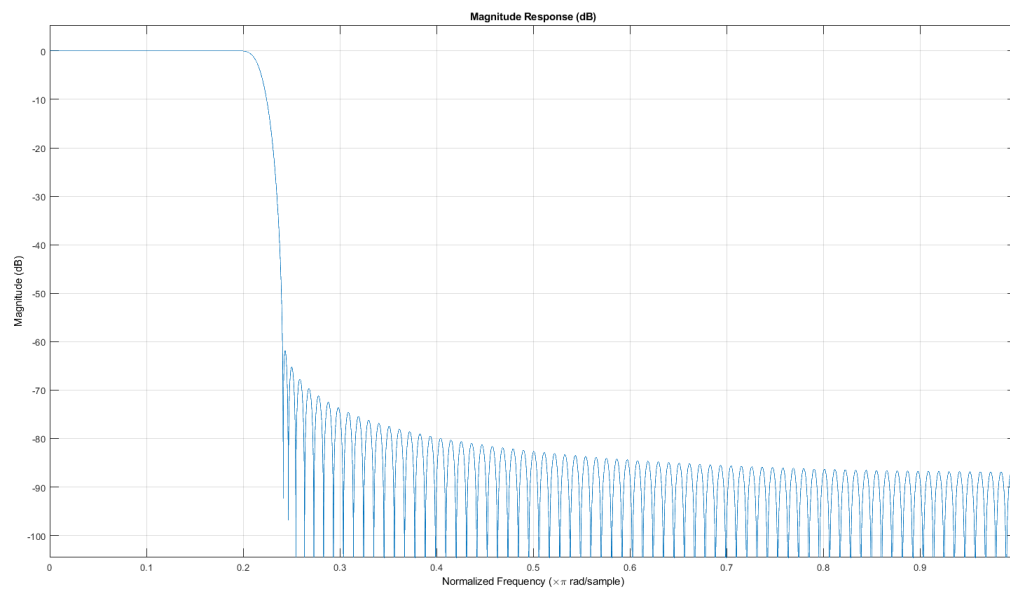


Figure 3: Magnitude response of firls filter with $n = 185$.

```

Fpass = 5e6;
Fstop = 6e6;
Fs = 50e6;
%fir1
n = 300;
delta = 0.007; % because default attenuation in wpass = -6dB, so I need to
%shift magn. resp.
f1 = fir1(n, 2 * Fpass / Fs + delta, blackman(n + 1));
%fvtool(f1);
%fir2
delta = 0.01;
freq = [0 2*Fpass/Fs (2*Fstop/Fs - delta) 1]; % subtraction delta is necessary
%for the proper filter design
mag = [1 1 0 0];
n = 210;
f2 = fir2(n, freq, mag);
%fvtool(f2);
%fir_ls
freq = [0 2*Fpass/Fs 2*Fstop/Fs 1];
mag = [1 1 0 0];
n = 185;
f3 = firls(n, freq, mag);
fvtool(f3);

```

So the filter with the lowest complexity is the firls filter, because it has the least order (**n = 185**).

Task 2

Using the impulse invariance method for analog to digital filter conversion, calculate the Chebyshev lowpass digital filter with parameters: passband 20MHz; passband ripple 0.2 dB; stopband (out-of-band) attenuation 60 dB; sampling frequency $F_s = 100$ MHz.

- Plot the impulse response for both analog and digital systems.
 - Plot the magnitude response for analog and digital systems in the frequency domain.
- Provide code.

Solution:

First of all, I need to design Chebyshev type I analog filter, then using impulse invariance approximation design the digital prototype of the analog Chebyshev type I filter.

Matlab has a lot of built-in functions for digital filter designing. For this task I will use **cheby1(n, Rp, Fpass, 's')** function, which requires these parameters:

- n - the order of the filter;
- Rp - passband ripple;
- Wp - passband;

Before using this function I need to calculate the order of the filter which I want to design. I will use a built-in Matlab function **cheb1ord(Wp,Ws,Rp,Rs)**, which returns the order of the filter n.

- Wp - passband frequency;
- Ws - stopband frequency;
- Rp - passband ripple;

- R_s - stopband attenuation;

Note that stopband frequency is not specified in this task, so I will specify it by myself. Of course, this parameter affects on the complexity on the filter: the narrower the transition band, the more complex the filter. I ran the function `cheb1ord(Wp,Ws,Rp,Rs)` with different parameter W_s and received this results:

- $W_s = 21$ MHz, $n = 25$;
- $W_s = 25$ MHz, $n = 11$;
- $W_s = 30$ MHz, $n = 8$;

Indeed, the narrower the transition band, the higher n . So, let's choose $W_s = 25$ MHz. Then $n = 11$. The next computation will be done for the parameters $W_s = 25$ MHz, $n = 11$. Then by writing simple Matlab code I received:

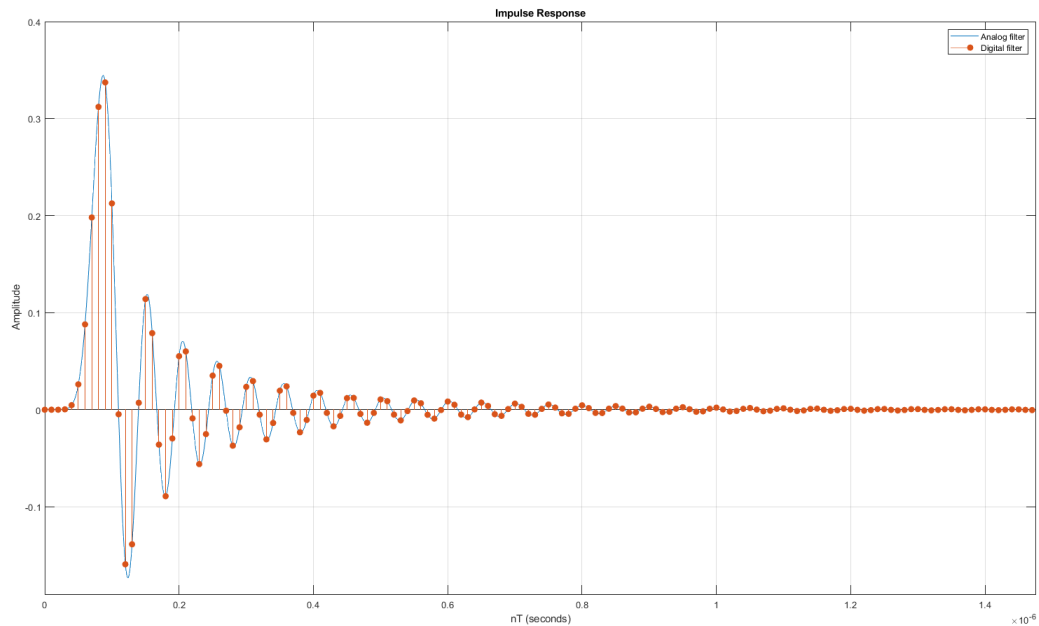


Figure 4: Impulse response of the analog and the digital filters.

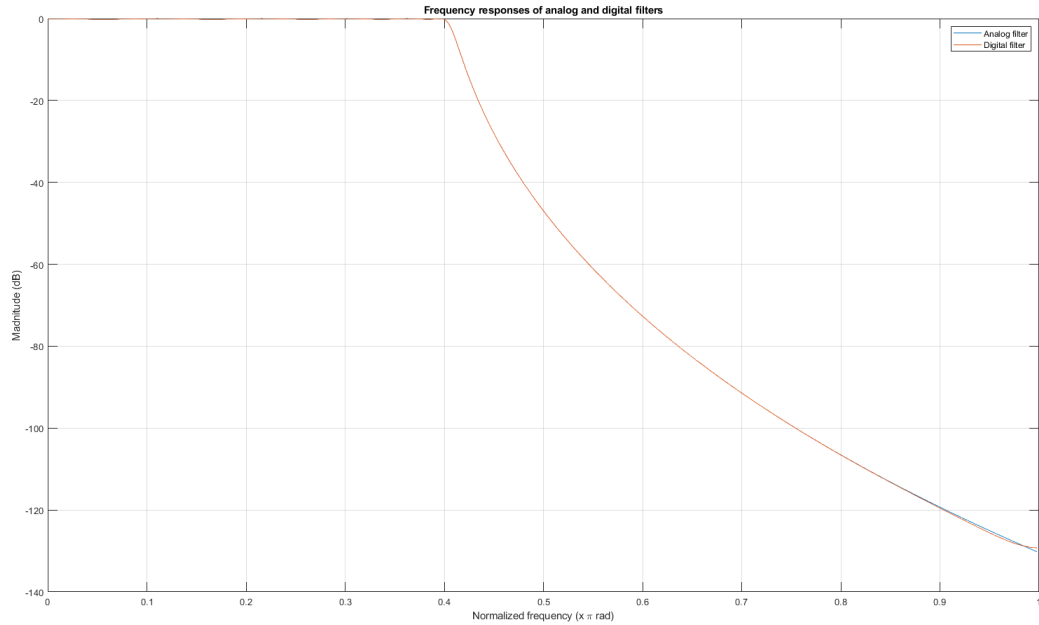


Figure 5: Frequency response of the analog and the digital filters.

```

Fpass = 20e6;
Fstop = 25e6;
Fs = 100e6;
Ripple = 0.2;
Attenuation = 60;
Ws = 2 * Fstop / Fs; % normalize the frequency
Wp = 2 * Fpass / Fs; % normalize the frequency
[n, Wp] = cheblord(Wp, Ws, Ripple, Attenuation); % compute the order of
% designing filter
[b,a] = cheby1(n, Ripple, 2 * pi * Fpass, 's'); % analog filter design B(s)/A(s)
figure(1);
[bz, az] =impinvar(b,a,Fs); % design a digital prototype of the analog
% filter B(z)/A(z)
% the following steps are similar to the seminar
[r, p] = residue(b, a); % find direct term of a Partial Fraction Expansion of the
%ratio of two polynomials
t = linspace(0, 100 / Fs, 1000);
h = real(r.'*exp(p.*t) / Fs); % analog filter impulse response
plot(t, h)
hold on;
impz(bz, az, [], Fs); % digital filter impulse invariance
legend('Analog filter ', 'Digital filter ')
grid on;
hold off;

figure(2);
[H, W] = freqz(bz, az);
[H_an] = freqs(b, a, W * Fs);

```

```

H_dig_db = 20 * log10(abs(H)); % convert magnitude to dB
H_an_db = 20 * log10(abs(H_an));
plot(W / pi, H_an_db);
hold on;
plot(W / pi, H_dig_db);
legend('Analog filter ', 'Digital filter ')
title('Frequency responses of analog and digital filters ')
ylabel('Magnitude (dB)')
xlabel('Normalized frequency (x \pi rad)')
grid on;
hold off;

```

According to the figure N!, digital filter approximates analog one with good accuracy.

Task 3

Implement a digital prototype of the analog filter with the transfer function:

$$H(s) = \frac{s + 2.5}{s^2 + 2.5s + 4}$$

using the Bilinear transformation. The sample clock frequency is $F_s = 20$ Hz.

- Determine the linear Difference Equation of the digital filter.
 - Plot impulse and frequency responses for digital and analog filters.
- Provide code.

Solution: Bilinear transformation equivalent to the substitution

$$s = \frac{2Fs(z - 1)}{(z + 1)}$$

the the transfer function of the analog filter $H(s)$

$$\begin{aligned}
 H(z) &= \frac{\frac{2Fs(z-1)}{(z+1)} + 2.5}{\left(\frac{2Fs(z-1)}{(z+1)}\right)^2 + 2.5\left(\frac{2Fs(z-1)}{(z+1)}\right) + 4} = \frac{2Fs(z-1) + 2.5(z+1)}{(z+1)\left(\frac{4Fs^2(z-1)^2}{(z+1)^2}\right) + 2.5\left(\frac{2Fs(z-1)}{(z+1)}\right) + \frac{4(z+1)^2}{(z+1)^2}} = \\
 &= \frac{(z+1)(2Fs(z-1) + 2.5(z+1))}{4Fs^2(z-1)^2 + 5Fs(z^2-1) + 4(z+1)^2} = \frac{(z+1)((2Fs+2.5)z + (2.5-2Fs))}{(4Fs^2+5Fs+4)z^2 + (8-8Fs^2)z + (4+4Fs^2-5Fs)} = \\
 &= \frac{2Fs+2.5}{4Fs^2+5Fs+4} \frac{(z+1)(z + \frac{2.5-2Fs}{2Fs+2.5})}{z^2 + \frac{8-8Fs^2}{4Fs^2+5Fs+4}z + \frac{4+4Fs^2-5Fs}{4Fs^2+5Fs+4}} = \frac{40+2.5}{1600+100+4} \frac{(z+1)(z + \frac{2.5-40}{40+2.5})}{z^2 + \frac{8-3200}{1600+100+4}z + \frac{4+1600-100}{1600+100+4}} = \\
 &= 0.0249 \frac{(z+1)(z-0.8824)}{z^2-1.8732z+0.8826} = 0.0249 \frac{z^2(1+0.1176z^{-1}+0.8824z^{-2})}{z^2(1-1.8732z^{-1}+0.8826z^{-2})} = \\
 &= 0.0249 \frac{(1+0.1176z^{-1}+0.8824z^{-2})}{(1-1.8732z^{-1}+0.8826z^{-2})} \\
 H(z) &= \frac{Y(z)}{X(z)} = 0.0249 \frac{(1+0.1176z^{-1}+0.8824z^{-2})}{(1-1.8732z^{-1}+0.8826z^{-2})} \\
 0.0249X(z)(1+0.1176z^{-1}+0.8824z^{-2}) &= Y(z)(1-1.8732z^{-1}+0.8826z^{-2}) \\
 X(z)(0.0249+0.0029z^{-1}+0.0220z^{-2}) &= Y(z)(1-1.8732z^{-1}+0.8826z^{-2})
 \end{aligned}$$

Using the time-shifting property I received the difference equation:

$$0.0249x[n] + 0.0029x[n-1] + 0.0220x[n-2] = y[n] - 1.8732y[n-1] + 0.8826y[n-2]$$

Then I just wrote a simple Matlab code, which is similar to the code on seminar and plotted impulse and frequency responses for digital and analog filters.

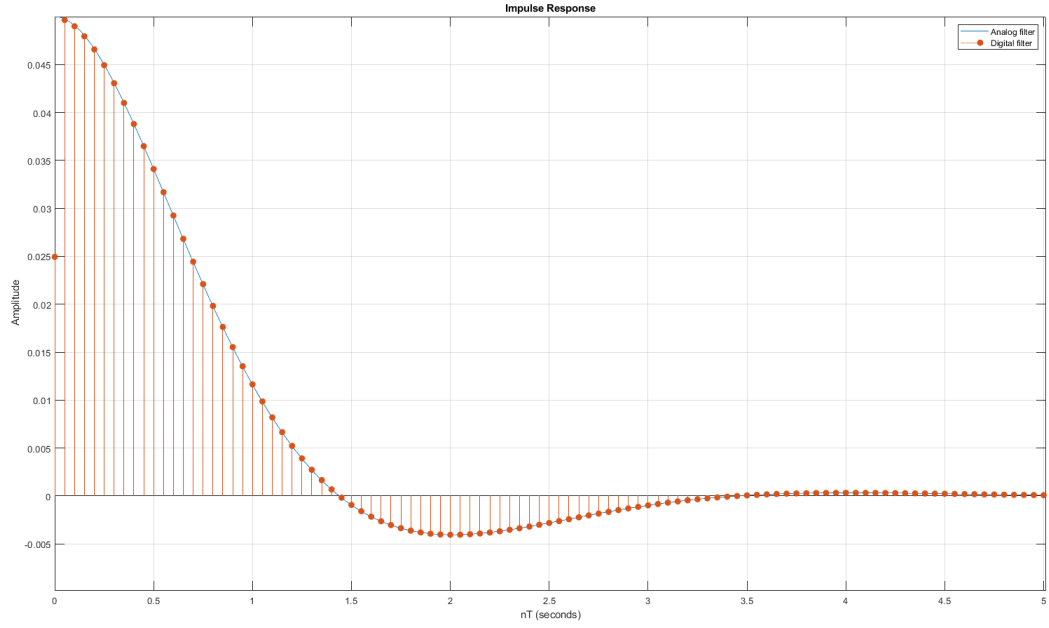


Figure 6: Impulse response of the analog and the digital filters.

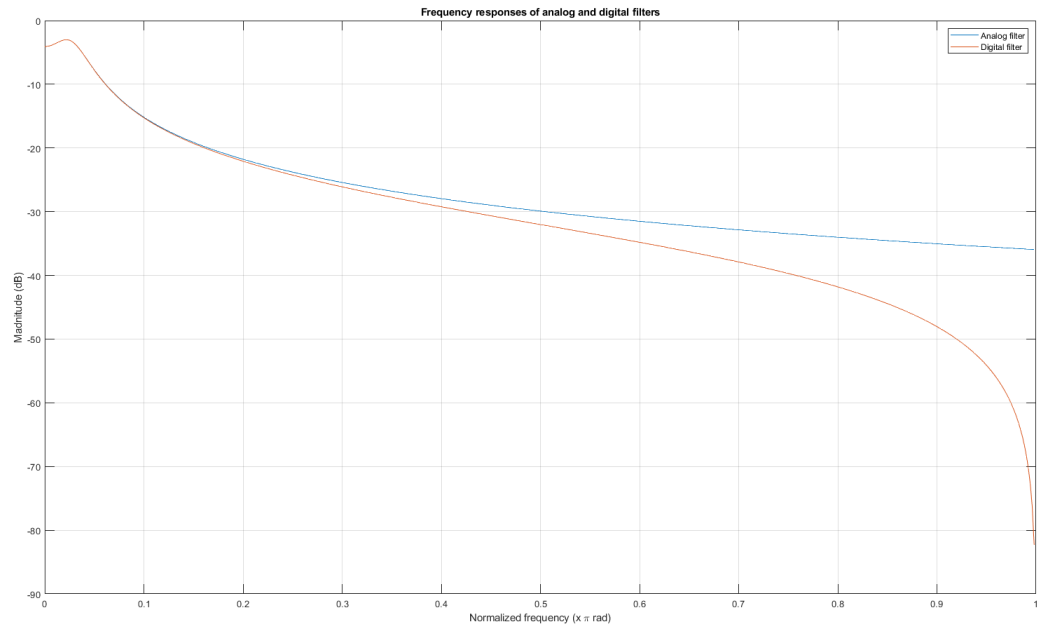


Figure 7: Frequency response of the analog and the digital filters.


```

Fs = 20;
b = [0 1 2.5];
a = [1 2.5 4];
[bz, az] = bilinear(b,a,Fs);
[r, p] = residue(b, a); % find direct term of a Partial Fraction Expansion of the
%ratio of two polynomials
t = linspace(0, 100 / Fs, 1000);
h = real(r.'*exp(p.*t) / Fs); % analog filter impulse response
figure(1);
plot(t, h);
hold on;
impz(bz, az, [], Fs);
legend('Analog filter ', 'Digital filter ')
grid on;
hold off;

figure(2);
[H, W] = freqz(bz, az);
[H_an] = freqs(b, a, W * Fs);
H_dig_db = 20 * log10(abs(H)); % convert magnitude to dB
H_an_db = 20 * log10(abs(H_an));
plot(W / pi, H_an_db);
hold on;
plot(W / pi, H_dig_db);
legend('Analog filter ', 'Digital filter ')
title('Frequency responses of analog and digital filters ')
ylabel('Magnitude (dB)')
xlabel('Normalized frequency (x \pi rad)')
grid on;

```

According to the figure with frequency response, digital prototype of the analog filter approximates the latter with good accuracy only when $f \leq 0.4Fs \rightarrow f \leq 8MHz$.

Task 4

A filter has the transfer function:

$$H(z) = 3 + 4z^{-1} + 6z^{-2} + 8z^{-3}$$

Determine the impulse response of the filter with the modified frequency response:

$$F(\omega) = H\left(\omega - \frac{\pi}{4}\right)$$

Solution:

First of all, let's move from Z-domain to frequency domain. For this substitute $z = re^{j\omega}$ to the expressions for the $H(z)$. Note: for simplicity I defined $r = 1$ so $z = e^{j\omega}$. I can do it, because r is just a constant (like a scaling coefficient for each summand).

$$H(\omega) = 3 + 4e^{-j\omega} + 6e^{-j2\omega} + 8e^{-j3\omega}$$

Then for our task substitute modified frequency: $\omega^* = \omega - \frac{\pi}{4}$

$$H(\omega) = 3 + 4e^{-j(\omega - \frac{\pi}{4})} + 6e^{-j(2\omega - \frac{\pi}{2})} + 8e^{-j(3\omega - \frac{3\pi}{4})}$$

$$H(\omega) = 3 + 4e^{\frac{j\pi}{4}}e^{-j\omega} + 6e^{\frac{j\pi}{2}}e^{-j2\omega} + 8e^{\frac{j3\pi}{4}}e^{-j3\omega}$$

So then I need to calculate inverse DTFT:

$$\begin{aligned}
h[n] &= \frac{1}{2\pi} \int_{-\pi}^{\pi} H(\omega) e^{j\omega n} d\omega \\
h[n] &= \frac{1}{2\pi} \int_{-\pi}^{\pi} (3 + 4e^{\frac{j\pi}{4}} e^{-j\omega} + 6e^{\frac{j\pi}{2}} e^{-j2\omega} + 8e^{\frac{j3\pi}{4}} e^{-j3\omega}) e^{j\omega n} d\omega \\
h[n] &= \frac{1}{2\pi} \int_{-\pi}^{\pi} (3e^j + 4e^{\frac{j\pi}{4}} e^{j\omega(n-1)} + 6e^{\frac{j\pi}{2}} e^{j\omega(n-2)} + 8e^{\frac{j3\pi}{4}} e^{j\omega(n-3)}) d\omega \\
h[n] &= \frac{1}{2\pi} \left(\frac{3}{jn} e^{j\omega n} \Big|_{-\pi}^{\pi} + 4e^{\frac{j\pi}{4}} \frac{e^{j\omega(n-1)}}{j(n-1)} \Big|_{-\pi}^{\pi} + 6e^{\frac{j\pi}{2}} \frac{e^{j\omega(n-2)}}{j(n-2)} \Big|_{-\pi}^{\pi} + 8e^{\frac{j3\pi}{4}} \frac{e^{j\omega(n-3)}}{j(n-3)} \Big|_{-\pi}^{\pi} \right) \\
h[n] &= \frac{1}{2\pi} \left(\frac{6}{n} \left(\frac{e^{j\pi n} - e^{-j\pi n}}{2j} \right) + \frac{8e^{\frac{j\pi}{4}}}{n-1} \left(\frac{e^{j\pi(n-1)} - e^{-j\pi(n-1)}}{2j} \right) + \frac{12e^{\frac{j\pi}{2}}}{n-2} \left(\frac{e^{j\pi(n-2)} - e^{-j\pi(n-2)}}{2j} \right) + \right. \\
&\quad \left. + \frac{16e^{\frac{j3\pi}{4}}}{n-3} \left(\frac{e^{j\pi(n-3)} - e^{-j\pi(n-3)}}{2j} \right) \right) \\
h[n] &= \frac{3\sin(\pi n)}{\pi n} + \frac{4e^{\frac{j\pi}{4}} \sin(\pi(n-1))}{\pi(n-1)} + \frac{6e^{\frac{j\pi}{2}} \sin(\pi(n-2))}{\pi(n-2)} + \frac{8e^{\frac{j3\pi}{4}} \sin(\pi(n-3))}{\pi(n-3)} \\
h[n] &= 3\delta[n] + 4e^{\frac{j\pi}{4}} \delta[n-1] + 6e^{\frac{j\pi}{2}} \delta[n-2] + 8e^{\frac{j3\pi}{4}} \delta[n-3]
\end{aligned}$$

$$h[n] = \begin{cases} 3 & , n = 0 \\ 4e^{\frac{j\pi}{4}} & , n = 1 \\ 6e^{\frac{j\pi}{2}} & , n = 2 \\ 8e^{\frac{j3\pi}{4}} & , n = 3 \\ 0 & , \text{otherwise} \end{cases}$$

Task 5

For a linear system with the transfer function:

$$H(z) = \frac{1z + 2}{3z^3 + 4z^2 + 5z + 6}$$

- Calculate the difference equation relating the input $x[n]$ to the output $y[n]$.
- Design block diagram realization (Direct-Form 1 and Direct-Form 2).
- Plot impulse and frequency responses.

Solution:

$$\begin{aligned}
H(z) &= \frac{Y(z)}{X(z)} = \frac{1z + 2}{3z^3 + 4z^2 + 5z + 6} = \frac{z^3(z^{-2} + 2z^{-3})}{z^3(3 + 4z^{-1} + 5z^{-2} + 6z^{-3})} \\
X(z)(z^{-2} + 2z^{-3}) &= Y(z)(3 + 4z^{-1} + 5z^{-2} + 6z^{-3})
\end{aligned}$$

Using the time-shifting property I received the difference equation:

$$\begin{aligned}
x[n-2] + 2x[n-3] &= 3y[n] + 4y[n-1] + 5y[n-2] + 6y[n-3] \\
y[n] &= \frac{1}{3}x[n-2] + \frac{2}{3}x[n-3] - \frac{4}{3}y[n-1] - \frac{5}{3}y[n-2] - 2y[n-3]
\end{aligned}$$

Then I designed block diagram realization in Direct-Form I and Direct-Form II: And then I plotted impulse and frequency responses for this linear system:

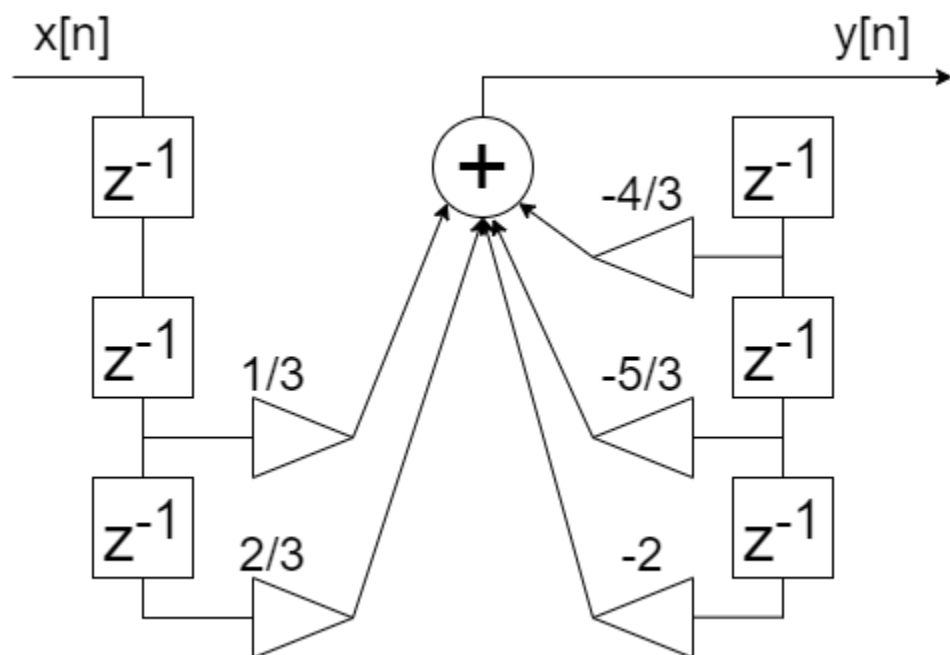


Figure 8: Direct-Form I.

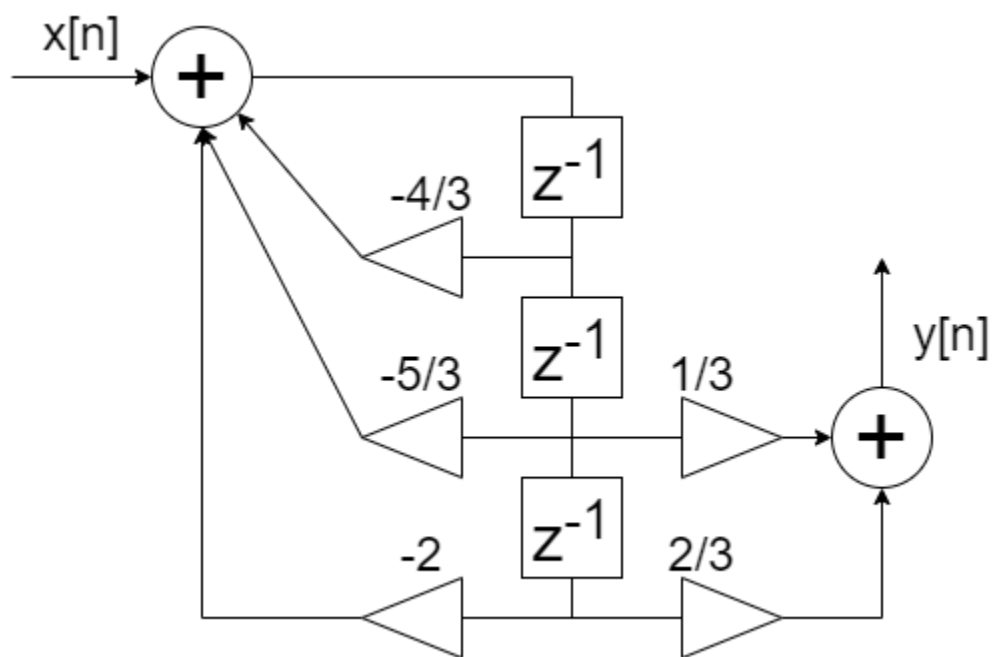


Figure 9: Direct-Form II.

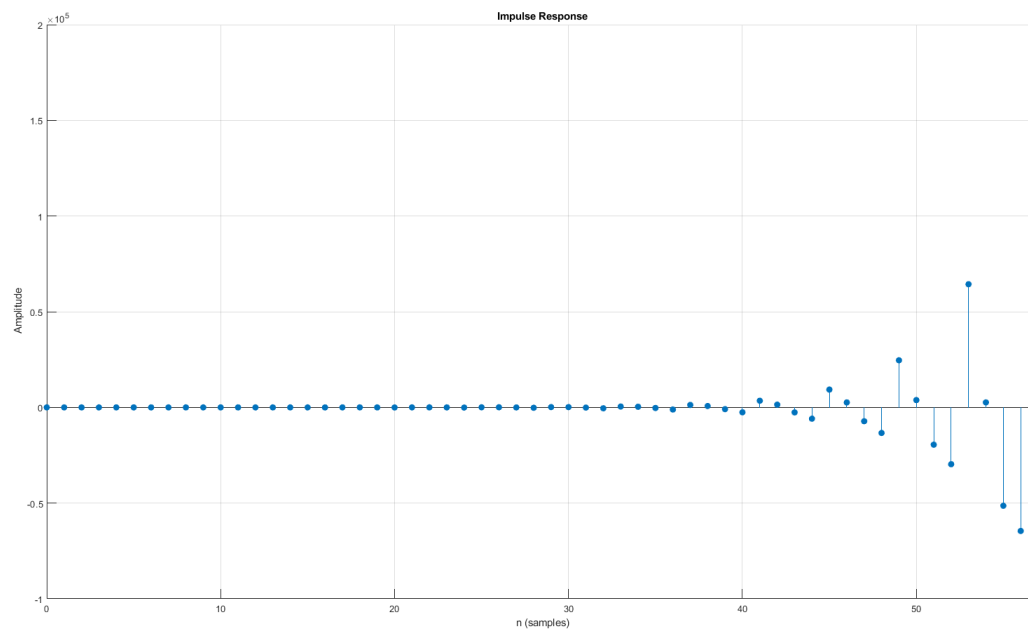


Figure 10: Impulse response of the linear system.

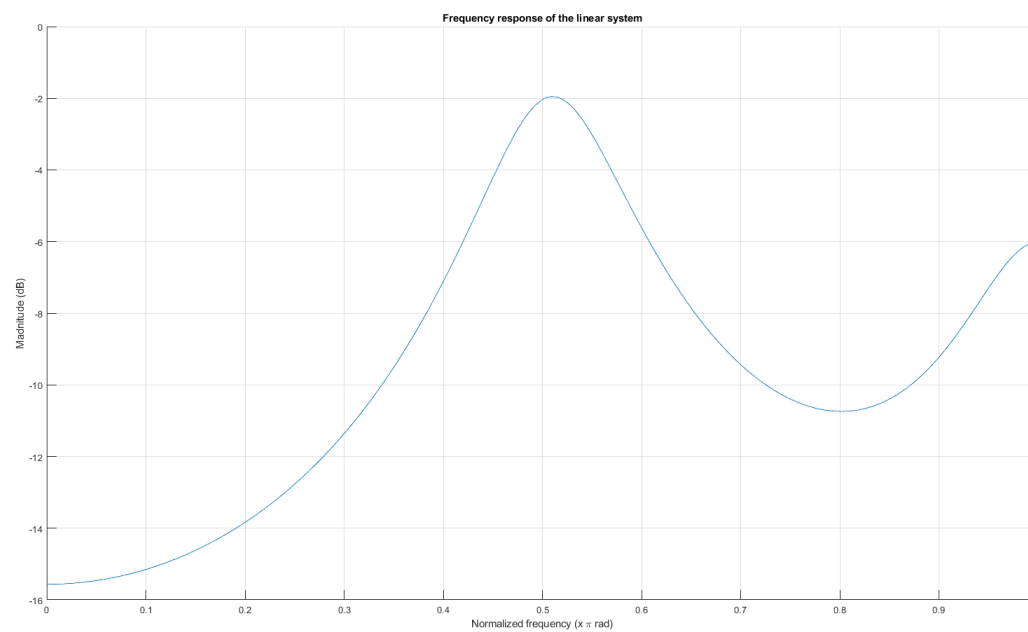


Figure 11: Frequency response of the linear system.

```

bz = [0 0 1 2];
az = [3 4 5 6];

figure(1);
hold on;
impz(bz, az);
grid on;
hold off;

figure(2);
[H, W] = freqz(bz, az);
H_dig_db = 20 * log10(abs(H)); % convert magnitude to dB
hold on;
plot(W / pi, H_dig_db);
title('Frequency response of the linear system')
ylabel('Magnitude (dB)')
xlabel('Normalized frequency (x \pi rad)')
grid on;
hold off;

```