# **Switched Capacitor Voltage Inverter with Shutdown**

#### **Features**

- Inverts Input Supply Voltage
- 25mA Output Current with a Voltage Drop of 250mV
- 0.45mA Quiescent Current at 3.3V Supply Voltage
- 99% Voltage Conversion Efficiency
- 1.8V to 5V operating range
- Require Only Two Capacitors
- Over-Temperature Protection
- Logic Controlled Shutdown
- 6-Pin SOT-23 Package

## **Applications**

- Cell Phone
- Small LCD Panels
- Portable Equipment
- Handy-Terminals, PDAs
- **■** Battery-Operated Equipment

## **General Description**

The G5930 is an unregulated charge-pump voltage inverter which may be used to generate a negative supply from positive input. Input voltages ranging from +1.8V to +5V can be inverted into a -1.8V to -5V output supply. The devices is ideal for both battery- powered and board level voltage conversion applications with a typical operating current of 0.45mA at 3.3V supply.

The G5930 can deliver 25mA output current with a voltage drop of 250mV. The parts are over -temperature protected.

Applications include cell phones, PDAs, and other portable equipment. The devices is available in a space-saving 6-pin SOT-23 Package.

## **Ordering Information**

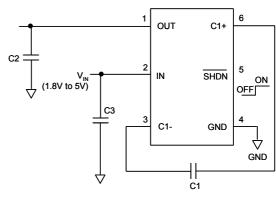
ORDER	MARKING	TEMP.	PACKAGE
NUMBER		RANGE	(Pb free)
G5930TBU	5930x	-40°C to 85°C	SOT-23-6

Note: TB: SOT23-6 U: Tape & Reel

## **Pin Configuration**

# G5930 OUT 1 6 C1+ IN 2 5 SHDN C1- 3 4 GND SOT-23-6

## **Typical Application Circuit**



C1,C2,C3: 1µF MLCC (X5R)





# **Global Mixed-mode Technology Inc.**

## **Absolute Maximum Ratings**

0.3V to +5.5V
5.5V to +0.3V
0.3V to (V <sub>IN</sub> +0.3V)
$(V_{OUT}-0.3V)$ to +0.3V
0.3V to (V <sub>IN</sub> +0.3V)
40°C to 85°C
Indefinite

Storage Temperature6	35°C to 150°C
Junction Temperature	150°C
Reflow Temperature (soldering, 10sec)	260°C
Continuous Power Dissipation (T <sub>A</sub> =70°C)	
SOT-23-6 (derate 8.7mW/°C above 70°C	696mW
ESD Rating HBM	2000V

Note: Human body model is a 100pF capacitor discharged through a  $1.5k\Omega$  resistor into each pin.

## **Electrical Characteristics**

 $(V_{IN}=+3.3V, C1=C2=C3=1\mu F, T_A=-40 \text{ to } 85^{\circ}\text{C} \text{ unless otherwise noted.}$  Typical values is at  $T_A=25^{\circ}\text{C.})$ 

PARAMETER	CONDITION	MIN	TYP	MAX	UNIT
Input Voltage, V <sub>IN</sub>	$R_{LOAD}$ =10k $\Omega$	1.8	-	5	V
Supply Current, IQ	T <sub>A</sub> =25°C		0.45	0.55	mA
Charge Pump Frequency, F <sub>SW</sub>	V <sub>IN</sub> =5V	200	270	330	KHz
Output Resistance(Note)	I <sub>LOAD</sub> =10mA		8.3		Ω
Output Displa	I <sub>LOAD</sub> =5mA		14		mV p-p
Output Ripple	I <sub>LOAD</sub> =25mA		56		mV p-p
Voltage Conversion Efficiency	No Load	99			%
Power Efficiency	I <sub>LOAD</sub> =5mA		93		%
SHDN Input Logic High		0.7xV <sub>IN</sub>			V
SHDN Input Logic Low				0.2	٧

### **Electrical Characteristics**

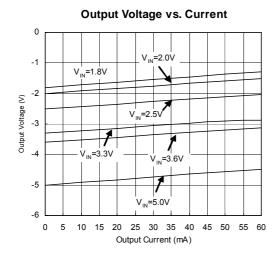
 $(V_{IN}=+5V, C1=C2=C3=1\mu F, T_A=-40 \text{ to } 85^{\circ}\text{C} \text{ unless otherwise noted.}$  Typical values is at  $T_A=25^{\circ}\text{C.})$ 

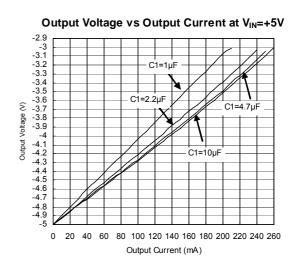
PARAMETER	CONDITION	MIN	TYP	MAX	UNIT
Supply Current, IQ	T <sub>A</sub> =25°C		0.92	1.1	mA
Charge Pump Frequency, F <sub>SW</sub>	V <sub>IN</sub> =5V		260		KHz
Output Resistance(Note)	I <sub>LOAD</sub> =10mA		8.8		Ω
Output Binala	I <sub>LOAD</sub> =5mA		25		mV p-p
Output Ripple	I <sub>LOAD</sub> =25mA		100		mV p-p
Voltage Conversion Efficiency	No Load	99			%
Power Efficiency	I <sub>LOAD</sub> =10mA		90		%
SHDN Input Logic High		$0.7xV_{\text{IN}}$			V
SHDN Input Logic Low		-	-	0.2	V

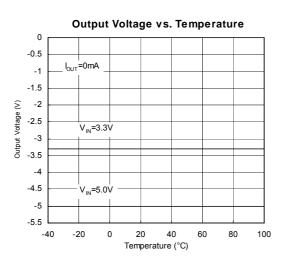
Note: Capacitor contribution (ESR component plus (1/F<sub>SW</sub>) · C) is approximately 20% of output.

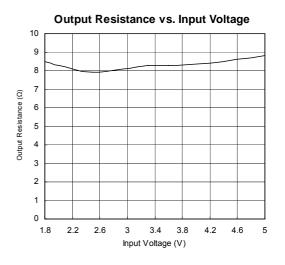
## **Typical Performance Characteristics**

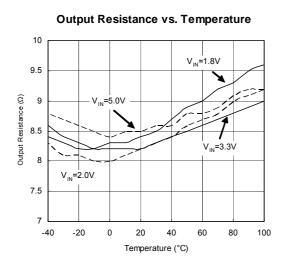
( $V_{IN}$ =3.3V,  $C_1$ = $C_2$ = $C_3$ = $1\mu$ F,  $T_A$ =25°C, unless otherwise noted.)

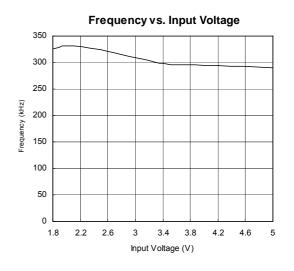




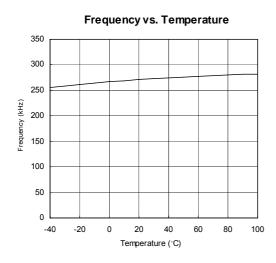


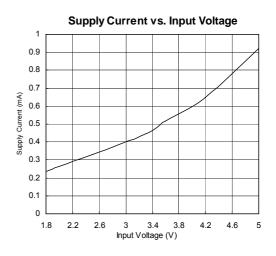


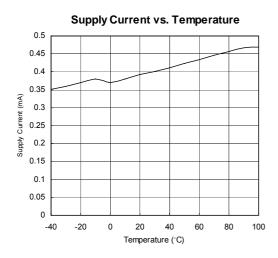


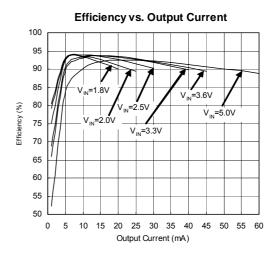


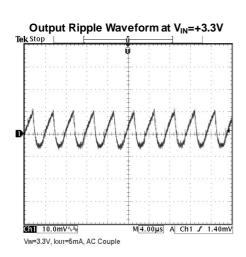
## **Typical Performance Characteristics (continued)**

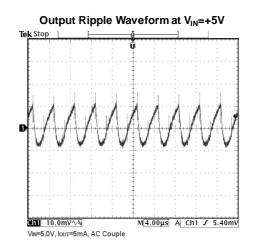




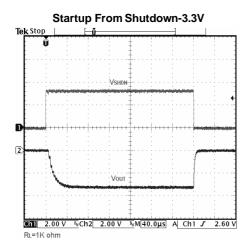


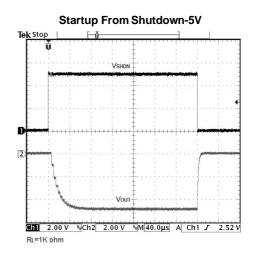




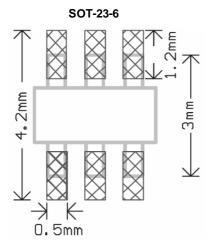


## **Typical Performance Characteristics (continued)**





## **Recommended Minimum Footprint**



## **Global Mixed-mode Technology Inc.**

## **Pin Descriptions**

PIN	NAME	FUNCTION	
1	OUT	Inverting Charge-Pump Output	
2	IN	Power-Supply Positive Voltage Input	
3	C1-	Flying Capacitor's Negative Terminal	
4	GND	Ground	
5	SHDN	Inverting Shutdown Input	
6	C1+	Flying Capacitor's Positive Terminal	

## **Detailed Description**

The G5930 contains four large switches which are switched in a sequence to inverter the input supply voltage. Energy transfer and storage are provided by external capacitors. Fig. 1 illustrates the voltage conversion scheme. When S1 and S3 are closed, C1 charges to the supply voltage  $V_{\text{IN}}$ . During this time interval, switches S2 and S4 are open. In the second time interval, S1 and S3 are open; at the same time, S2 and S4 are closed, C1 is charging C2. After a number of cycles, the voltage across C2 will be pumped to  $V_{\text{IN}}$ . Since the anode of C2 is connected to ground, the output at the cathode of C2 equals -( $V_{\text{IN}}$ ) when there are no load current.

The G5930 has a logical-controlled shutdown input. Driving  $\overline{SHDN}$  low places the G5930 in a low-power shutdown mode, and the charge pump switching halts. In the shutdown mode, supply current is smaller than 1 $\mu$ A. Besides, the OUT is activily pulled to ground.

The main application of G5930 is to generate a negative supply voltage. The range of the input supply voltage is 1.8V to 5V. The output characteristics of this

circuit can be approximated by an ideal voltage source in series with a resistance. The voltage source equals -( $V_{IN}$ ). The output resistance, Rout, is a function of the ON resistance of the internal MOSFET switches, the oscillator frequency, the capacitance and the ESR of both  $C_1$  and  $C_2$ . Since the switching current charging and discharging  $C_1$  is approximately twice as the output current, the effect of the ESR of the pumping capacitor  $C_1$  will be multiplied by four in the output resistance. The output capacitor  $C_2$  is charging and discharging at a current approximately equal to the output current, therefore, this ESR term only counts once in the output resistance. A good approximation of  $R_{out}$  is:

$$\text{Rout} \cong 2 \text{Rsw} + \frac{2}{f_{\text{OSC}} x C_1} + 4 \text{ESRc}_1 + \text{ESRc}_2$$

Where  $R_{\text{SW}}$  is the sum of the ON resistance of the internal MOSFET switches shown in Figure 1.

High capacitance, low ESR capacitors will reduce the output resistance.

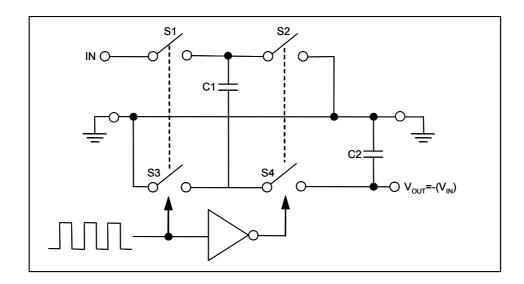


Figure 1. Ideal Voltage Inverter

## **Application Information**

**Table 1. Low-ESR Capacitor Manufacturers** 

PRODUCTION METHOD	MANUFACTURER	SERIES
Surface-Mount	AVX	X7R
Ceramic	Matsuo	X7R

#### **Capacitor Selection**

To maintain the lowest output resistance, use capacitors with low ESR (Table 1). The charge-pump output resistance is a function of C1's and C2's ESR. Therefore, minimizing the charge-pump capacitor's ESR minimizes the total output resistance.

#### Flying Capacitor (C1)

Increasing the flying capacitor's value reduces the output resistance. Above a certain point, increasing C1's capacitance has a negligible effect because the output resistance becomes dominated by the internal switch resistance and capacitor ESR.

#### **Output Capacitor (C2)**

Increasing the output capacitor's value reduces the output ripple voltage. Decreasing its ESR reduces both output resistance and ripple. Lower capacitance values can be used with light loads if higher output ripple can be tolerated. Use the following equation to calculate the peak-to-peak ripple:

$$V_{RIPPLE} = \frac{I_{L}}{f_{OSC}xC2} + 2xI_{L} + ESR_{C2}$$

The output resistance is dependent on the capacitance and ESR values of the external capacitors. The output voltage drop is the load current times the output resistance, and the power efficiency is

$$\eta = \frac{{{P_{\text{OUT}}}}}{{{P_{\text{IN}}}}} = \frac{{{I_{\text{L}}}^2}{{R_{\text{L}}}}}{{{I_{\text{L}}}^2}{R_{\text{L}}} + {I_{\text{L}}}^2}{{R_{\text{OUT}}}} + {I_{\text{O}}}({V_{\text{IN}}})$$

Where  $I_Q(V_{IN})$  is the quiescent power loss of the IC device, and  $I_L^2R_{out}$  is the conversion loss associated with the switch on-resistance, the two external capacitors and their ESRs.

#### Input Bypass Capacitor (C3)

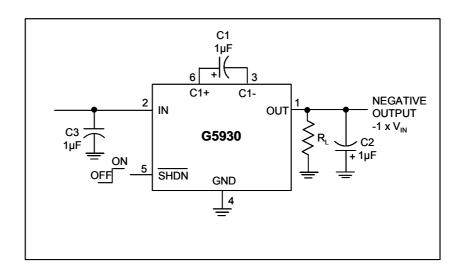
Bypass the incoming supply to reduce its AC impedance and the impact of the G5930's switching noise. A bypass capacitor with a value equal to that of C1 is recommended.

#### **Voltage Inverter**

The most common application for these devices is a charge-pump voltage inverter (Figure 2). This application requires only two external components— capacitors C1 and C2—plus a bypass capacitor, if necessary.

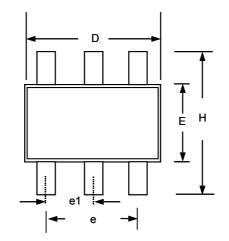
#### **Layout and Grounding**

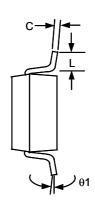
Good layout is important, primarily for good noise performance. To ensure good layout, mount all components as close together as possible, keep traces short to minimize parasitic inductance and capacitance, and use a ground plane.

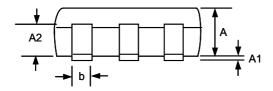


**Figure 2. Typical Application Circuit** 

## Package Information





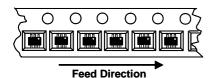


#### Note:

- 1. Package body sizes exclude mold flash protrusions or gate burrs
- 2. Tolerance ±0.1000 mm (4mil) unless otherwise specified
- 3. Coplanarity: 0.1000mm
- 4. Dimension L is measured in gage plane

SYMBOL	DIMENSIONS IN MILLIMETER			
STWIBUL	MIN	NOM	MAX	
Α	1.00	1.10	1.30	
A1	0.00		0.10	
A2	0.70	0.80	0.90	
b	0.35	0.40	0.50	
С	0.10	0.15	0.25	
D	2.70	2.90	3.10	
Е	1.40	1.60	1.80	
е		1.90(TYP)		
e1		0.95		
Н	2.60	2.80	3.00	
L	0.37			
$\theta$ 1	1°	5°	9°	

## **Taping Specification**



**SOT- 23-6 Package Orientation** 

PACKAGE	Q'TY/REEL
SOT-23-6	3,000 ea

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