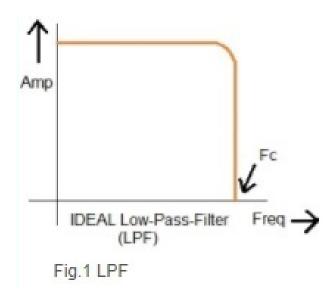
FIR Filter Design by Window Function

Basic Types of Filters

An ideal Lowpass filter allows to pass all frequency components of a signal below a designated cutoff frequency ,Fc, and rejects all frequency components of a signal above Fc.



An ideal Highpass filter allows to pass all frequency components of a signal above a designated cutoff frequency, Fc, and rejects all frequency components below Fc.

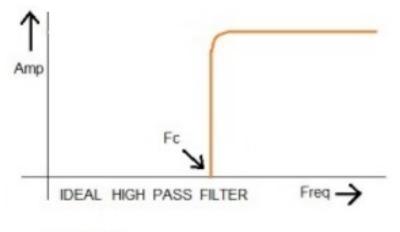
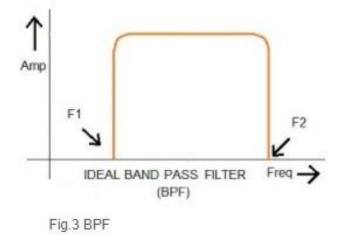


Fig.2 HPF

Basic Types of Filters

An ideal Bandpass filter allows to pass all frequency components of a signal within a certain range and rejects all frequency components outside of that range.



An ideal Bandstop filter rejects all frequency components of a signal within a certain range and allows to pass all frequency components outside of that range.

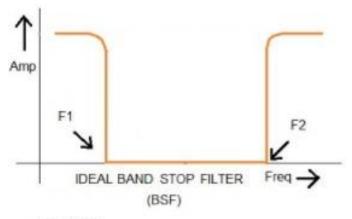


Fig.4 BSF

FIR Filter

- FIR (Finite Impulse Response) filter is a filter whose impulse response is of finite duration.
- For causal system,

$$h[n]=0$$
 for $n<0$

Summary of ideal impulse responses for standard FIR filters.

Filter Type	Ideal Impulse Response $h(n)$ (noncausal FIR coefficients)			
Lowpass:	$h(n) = \begin{cases} \frac{\Omega_c}{\pi} & n = 0\\ \frac{\sin{(\Omega_c n)}}{n\pi} & \text{for } n \neq 0 & -M \leq n \leq M \end{cases}$			
Highpass:	$h(n) = \begin{cases} \frac{\pi - \Omega_c}{\pi} & n = 0\\ -\frac{\sin(\Omega_c n)}{n\pi} & \text{for } n \neq 0 & -M \leq n \leq M \end{cases}$			
Bandpass:	$h(n) = \begin{cases} \frac{\Omega_H - \Omega_L}{\pi} & n = 0\\ \frac{\sin(\Omega_H n)}{n\pi} - \frac{\sin(\Omega_L n)}{n\pi} & \text{for } n \neq 0 & -M \leq n \leq M \end{cases}$			
Bandstop:	$h(n) = \begin{cases} \frac{\pi - \Omega_H + \Omega_L}{\pi} & n = 0\\ -\frac{\sin(\Omega_H n)}{n\pi} + \frac{\sin(\Omega_L n)}{n\pi} & \text{for } n \neq 0 & -M \leq n \leq M \end{cases}$			

Causal FIR filter coefficients: shifting h(n) to the right by M samples. Transfer function:

$$H(z) = b_0 + b_1 z^{-1} + b_2 z^{-2} + \cdots + b_{2M} z^{-2M}$$

where $b_n = h(n - M), n = 0, 1, \cdots, 2M$

Consider the Ideal Impulse Response for Standard Lowpass FIR Filter

• Lowpass:

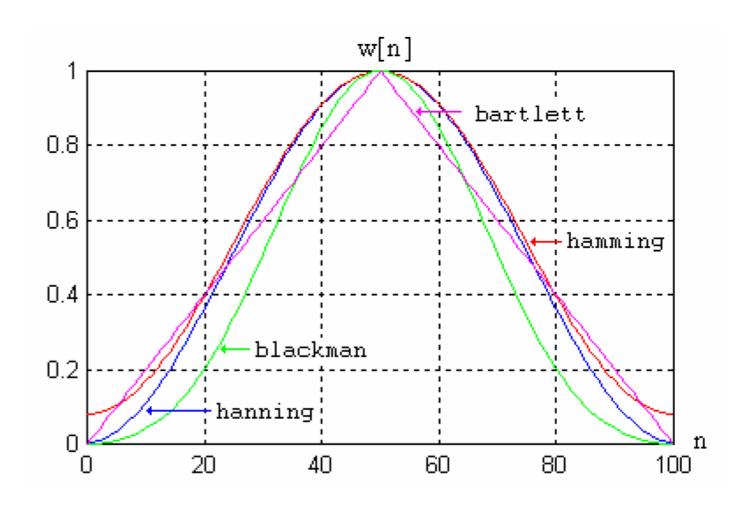
$$h[n] = egin{cases} rac{\Omega_c}{\pi} & for \ n=0 \ rac{\sin(\Omega_c n)}{n\pi} & for \ n \neq 0 \end{cases}$$
 -M \leq n \leq M

$$\Omega_C = \frac{2\pi F_C}{F_S} = \text{cutoff frequency in rad/sample}$$

$$Tap = 2M + 1 = \text{filter coefficient/delay pair (must be odd)}$$

$$FilterOrder = Tap - 1$$

Common FIR Filter Window Functions,w[n]



Calculation of FIR Filter Window Function, w[n]

Window Type	Window Function $w(n)$, $-M \le n \le M$
Rectangular	1
Hanning	$0.5 + 0.5\cos\left(\frac{\pi n}{M}\right)$
Hamming	$0.54 + 0.46 \cos(\frac{\pi n}{M})$
Blackman	$0.42 + 0.5\cos\left(\frac{n\pi}{M}\right) + 0.08\cos\left(\frac{2n\pi}{M}\right)$

$$Tap = 2M + 1$$
 $FilterOrder = Tap - 1$

Steps in the design of FIR Filter

- Convert the cutoff frequency, if given in Hz, to radian/sample
- II. Calculate M
- III. Calculate the impulse response, h[n]
- IV. Calculate the window function, w[n]
- V. Write the windowed impulse response, (product of III and IV) $h_W[n]=h[n]w[n]$
- VI. Delay the $h_W[n]$ by M samples. The result is called the filter coefficient b_n

Example #1

- a) Design a 3-tap FIR lowpass filter with cutoff frequency of 800Hz and a sampling rate of 8000Hz using the Hamming window function.
- b) Determine the transfer function and difference equation of the designed FIR system

Solution to Example #1

- a) Design a 3-tap FIR lowpass filter with cutoff frequency of 800Hz and a sampling rate of 8000Hz using the Hamming window function.
- b) Determine the transfer function and difference equation of the designed FIR system
- I. Convert the cutoff frequency, if given in Hz, to radian/sample
- II. Calculate M
- III. Calculate the impulse response, h[n]
- IV. Calculate the window function, w[n]
- V. Write the windowed impulse response, $h_w[n]=h[n]$ w[n]
- VI. Delay the $h_W[n]$ by M samples. The result is called the filter coefficient b_K

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SOLUTION:

$$\Omega_C = \frac{2\pi F_C}{F_S} = \frac{2\pi (800 Hz)}{8000 samples / sec} = 0.2\pi \frac{rad}{sample}$$

$$Tap = 2M + 1$$

 $M = (Tap - 1)/2 = (3-1)/2 = 1$
 $FilterOrder = Tap - 1 = 3 - 1 = 2$

$$h[0] = \frac{0.211}{\pi} = 0.2$$

$$h[1] = \frac{\sin(0.2\pi(1))}{(1)(\pi)}$$

$$h[-1] = \frac{\sin(0.2\pi(-1))}{\sin(0.2\pi(-1))}$$

$$h[n] = \begin{cases} \frac{\Omega_c}{\pi} & \text{for } n = 0 \\ \frac{\sin(\Omega_c n)}{n\pi} & \text{for } n \neq 0 \end{cases} -M \le n \le M$$

$$h[n] = \{0.187, 0.2, 0.187\}$$

Hamming
$$w[n] = 0.54 + 0.46 \cos(\frac{\pi n}{M})$$
 $w[n] = \{0.08, 1, 0.08\}$

$$h_W[n] = w[n] h[n] = \{0.01496, 0.2, 0.01496\}$$

$$b_n = h_W[n-1] = \{0.01496, 0.2, 0.01496\}$$

These are the filter coefficients

Continuation of Solution to Example#1

$$b_n = \{0.01496, 0.2, 0.01496\}$$

b) Determine the transfer function and difference equation of the designed FIR system

$$H(z) = 0.01496 + 0.2z^{-1} + 0.01496z^{-2}$$
 Transfer Function

Since $H(z) = \frac{\gamma(z)}{\chi(z)}$

Then, $\gamma(z) = H(z)\chi(z) = 0.01496\chi(z) + 0.2z^{-1}\chi(z) + 0.01496z^{-2}\chi(z)$

Therefore, the difference equation of the designed FIR Low Pross Filter is:

 $y(n) = 0.01496 \times [n] + 0.2 \times [n-1] + 0.01496 \times [n-2]$

Matlab Implementation of Example #1 using FIR1

```
Based on our computation, the filter coefficients for Problem #1 are:
b_n = \{0.01496, 0.200, 0.01496\}
Implementing this using fir1 command in matlab,
>> h=fir1(2,0.2,'low')
h =
  0.0651 0.8698 0.0651
note: fir1(order, cutoff, filter type);
order=tap-1; the pi on cutoff is already included in the fir1 function
```

Notice that the values are very much different. But if we will use a higher order... Say, 24 (tap is 25)

Matlab Implementation of Example #1 using FIR1.

• Using the formula, the filter coefficients for 25-tap are...

• Using MATLAB fir1 function...

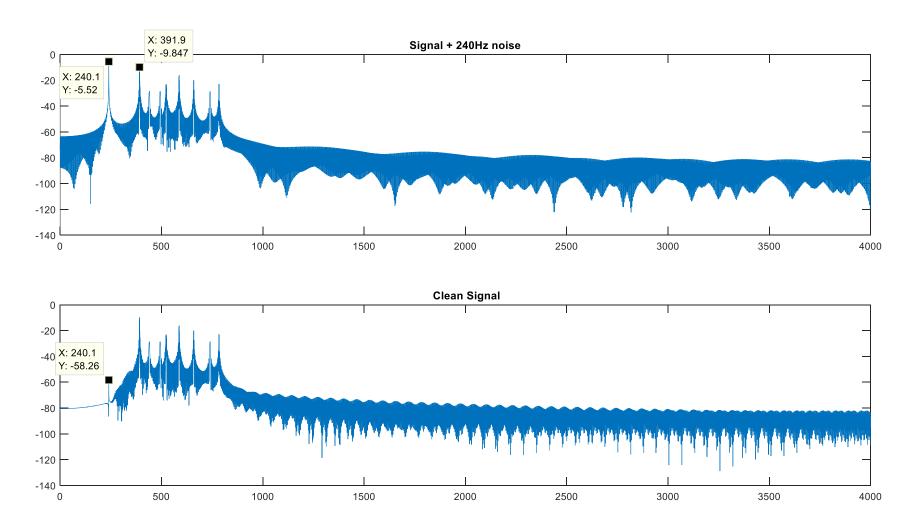
```
>> h=fir1(24,0.2,'low')
h =
        0.0016 -0.0000 -0.0044 -0.0117 -0.0181 -0.0168
0.0020
                                                         0.0000
                0.1415
                        0.1835
0.0359
        0.0870
                                  0.1992
                                         0.1835 0.1415
                                                          0.0870
0.0359
        0.0000 -0.0168 -0.0181 -0.0117
-0.0044 -0.0000 0.0016
                        0.0020
```

From this we can conclude that <fir1> can give us accurate results when using more taps.

Matlab Implementation of Example #1 using FIR1.

```
clc;
clear all;
close all;
[x1,Fs]=audioread('menuethum.wav');
%Create HPF, Order 250, Length 251, Wn=2*300/8000
%300Hz is the cutoff and 8000Hz is the Fs. pi is internal to <fir1>
h=fir1(250,0.075,'high');
y=filter(h,1,x1); % num=h and den=1
%y=conv(h,x1); %you can also use conv instead of filter
sound(x1,Fs);
pause(7)
sound(y,Fs);
pause(7)
%PSD Plots
figure(1), subplot(2,1,1);
[Px,Fx] = periodogram(x1,[],length(x1),Fs); %Original signal with 240Hz noise
plot(Fx,10*log10(Px));
title('Signal + 240Hz noise');
subplot(2,1,2);
[Py,Fy] = periodogram(y,[],length(y),Fs); %Filtered Signal
plot(Fy,10*log10(Py));
title('Clean Signal')
[H,W]=freqz(h);
dB=mag2db(abs(H));
figure(2),plot(W/pi,dB);
```

Matlab Implementation of Example #1 using FIR1.



Example #2

Given below is the calculated filter impulse response

- Determine the windowed coefficient $h_w[n]$ (or windowed impulse response). Use Hamming window function.
- Determine the FIR filter coefficients.

SOLUTION:

$$Tap = 7 = 2M+1$$
 $M = \frac{7-1}{2} = 3$

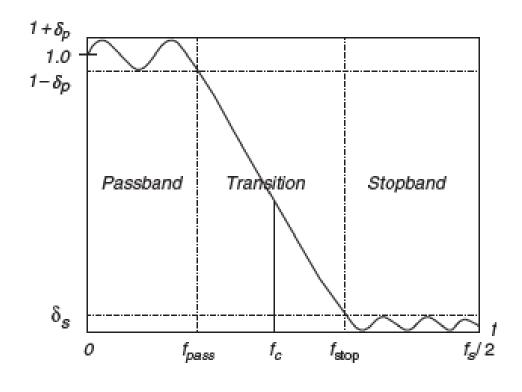
Tap =
$$7 = 2M+1$$
 a) $w[n] = 0.54 + 0.44 col(\frac{\pi n}{M}) -3 \le n \le 3$
 $M = \frac{7-1}{2} = 3$ substitute the values of n
 $w[n] = [0.08, 0.31, 0.77, 1, 0.77, 0.31, 0.08]$
 $w[n] = h[n] \cdot w[n]$
 $w[n] = [0.006, 0.04934, 0.17331, 0.25, 0.17331, 0.04934, 0.006]$
by $= hw[n-3] = [0.006, 0.04934, 0.17331, 0.25, 0.17331, 0.04934, 0.006]$

FIR filter length, N, estimation using Window Functions

$$\Delta f = \frac{|f_{stop} - f_{pass}|}{F_S}$$

Δf=normalized transition width

$$\delta_p \ dB = 20 \cdot \log_{10} (1 + \delta_p)$$
$$\delta_s \ dB = -20 \log_{10} (\delta_s)$$



FIR filter length estimation using window functions (normalized transition width $\Delta f = |f_{stop} - f_{pas\,s}|/f_s)$.

Window Type	Window Function $w(n)$, $-M \le n \le M$	Window Length, N	Passband Ripple (dB)	Stopband Attenuation (dB)
Rectangular	1	$N = 0.9/\Delta f$	0.7416	21
Hanning	$0.5 + 0.5 \cos(\frac{\pi n}{M})$	$N = 3.1/\Delta f$	0.0546	44
Hamming	$0.54 + 0.46\cos(\frac{\pi n}{M})$	$N = 3.3/\Delta f$	0.0194	53
Blackman	$0.42 + 0.5\cos\left(\frac{n\pi}{M}\right) + 0.08\cos\left(\frac{2n\pi}{M}\right)$	$N = 5.5/\Delta f$	0.0017	74

Example:

A lowpass filter using rectangular window has the following specifications:

Passband: 0-1850Hz

Stopband: 2150-4000Hz

Sampling Rate: 8000Hz

Required:

a) Estimate the filter length. (Taps in the filter length and is always odd)

b)Calculate the cutoff freq.

Solution:

$$\Delta f = \frac{1}{5} + \frac{1}{5} + \frac{1}{5} \frac{1}{5}$$

From the table on the previous slide,

N = 0.9/As = 0.9/0.0375 = 24 we 25 Taps

Passband Transition Stopband $f_s/2$ t_{pass} t_{stop} 1850Hz 2150Hz 4000Hz This is the cuting frequency

Assignment:

For #s 1-2:Estimate the filter length and calculate the cutoff freq in Hz.

1) A highpass filter using Hamming Window has the following specifications:

Stopband: 0-1500Hz Passband: 2500-4000Hz Sampling Rate: 8000Hz

2) A Bandpass filter using Hamming window has the following specifications:

Lower Stopband: 0-500Hz Upper Stopband: 3500-4000Hz

Passband: 1600-2300Hz Sampling Rate: 8000Hz

3) Design a 5-tap FIR band reject (also called band stop) filter with lower cutoff frequency of 2kHz and upper cutoff frequency of 2400Hz, and a sampling rate of 8000Hz using the Hamming window method.

Determine the transfer function and difference equation.

Answers to Assignment:

- 1) $\Delta f=0.125$; N= 26.4 use 27 Taps; $f_c=2kHz$
- 2) Use 25 Taps (solution is show below); f_{CLower} = 1050Hz

$$\Delta f_1 = \frac{|500 - 1600|}{8000} = 0.1375$$
 $N_1 = \frac{3.3}{0.1375} = 24$

$$\Delta f_2 = \frac{|3500 - 2300|}{8000} = 0.15$$
 $N_1 = \frac{3.3}{0.1375} = 24$

$$N_2 = \frac{3.3}{0.15} = 22$$
 $N_2 = \frac{3.3}{0.15} = 22$

3) $b_n = \{0.00748, 0.00841, 0.9, 0.00841, 0.00748\}$ $H(z)=0.00748+0.00841 z^{-1} +0.9 z^{-2} +0.00841 z^{-3} +0.00748z^{-4}$ y(n)=0.00748 x[n]+0.00841 x[n-1] +0.9 x[n-2] +0.00841 x[n-3] +0.00748x[n-4]