

$$\& \Leftrightarrow \lambda = \frac{C}{f} = 3 \cdot 10^{12} \text{ s}^{-1}$$

Knoten: heraus  $\oplus$  Knoten zidenz- Baums: Superknoten, vorzeichen nach Baumkomk  
mind Faktor 10  $(n-1)$  rein  $\ominus$  Matrix  $\hookrightarrow [1, A_i] \cdot i = 0$   
 $\hookrightarrow \sum_i i = 0$

Graph: Verbindungsstukter

Baum: alle Knoten - keine Schleifen  
- zusammenhängend

$$A_i = 0$$

$$\begin{bmatrix} 1 & 1 \\ 1 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \\ u_3 \end{bmatrix} = \begin{bmatrix} u_1 \\ u_2 \\ u_3 \end{bmatrix}$$

$$i = \sum_i i \cdot u = A_i \cdot u$$

$$u = A_i^{-1} \cdot i$$

$$\hookrightarrow [B_b, 1] \cdot u = 0$$

Baum: Schleifen mit nur 1 Verbindungsweig

$$M = A^T \text{ (Inversz)}; A_V = -B_b^T \text{ (Baum)} \quad \text{Bild: } u = M \cdot u_M \quad \text{- aktiv: } u \cdot i < 0 \text{ (1 Punkt in II o. IV)} \\ \text{- passiv: } u \cdot i > 0 \text{ (nur in I u. III)}$$

### resistive Elemente:

implizit:  $u - u_i = 0$

parametrische:  $u = u_F(n)$

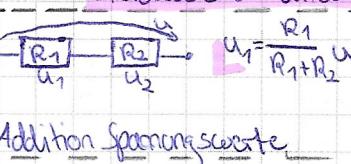
explizit:  $u = r_F \cdot i$

- Eigenschaften:
  - unipolt (Punktsymmetrie)  $\hookrightarrow$  verlustfrei:  $u \cdot i = 0$  (Kernlinie Achsen)
  - Quellenfrei: Ursprung, keine Quellen  $\hookrightarrow$  verlustbehaftet: (nur 1 nicht auf Achsen)
  - Strom/Spannungsg.: Jeder Strom/Spannungswert nur 1 Wert
  - Steig Lineare: Ursprungsgerade ( $\neq$  Nullator/Monotone)
  - Linear  $\Rightarrow$  jede Gerade
  - Stückweise linear: aus Geradenstückchen

### Dualität:

$$u \rightarrow R_d i \quad i \rightarrow \frac{u}{R_d}$$

$$R_d = \frac{R^2}{R} \quad G = \frac{1}{R^2 G}$$

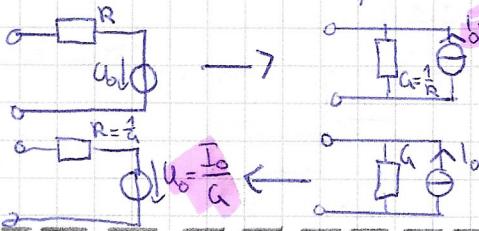


$$u = u_1 + u_2 \quad i = \frac{u_1}{R_1} = \frac{u_2}{R_2} \quad \hookrightarrow \text{Addition von Spannungen}$$

$$G = \frac{i}{u} = \frac{i_1 + i_2}{u_1 + u_2} \quad R = \frac{u_1 + u_2}{i} \quad \hookrightarrow \text{Addition von Strömen}$$

### Quellenwandlung:

Helmholtz-/Thevenin; Mayer-Norton



Ideale Diode

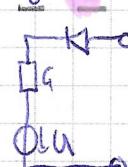
$$I: i=0, u \leq 0$$

$$II: u=0, i \geq 0$$

Konkav:  $f(u)$

$$I: i=0, u \leq u$$

$$II: i = G_u - G_u u, u \geq u$$



Konvex:

$$I: u=0, i \leq I$$

$$II: u = R_i - RI, i \geq I$$

Convex:

$$I: i=0, u \leq I$$

$$II: u = R_i + RI, i \geq I$$

$$i_D = I_S \exp\left(\frac{u_D}{U_T} - 1\right)$$

$$u_D = U_T \ln\left(\frac{i_D}{I_S} + 1\right)$$

$$u(t) = u_{AP} + \Delta u \quad i(t) = I_{AP} + \Delta i$$

### Arbeitspunktbest.:

graphisch: Schnittpunkt Quelle  $\rightarrow$  KS- / LK-Methode

& das Kernlinie  $\hookrightarrow Q^\times \rightarrow$  Spiegelung an  $u$  Achse

rechnerisch: Quelle & Last gleichsetzen

$$\text{Linearisierung: } i_F, c_{in} = \frac{\partial i_F}{\partial u_F} \Big|_{AP} (u_F - u_{AP}) + I_{AP}$$

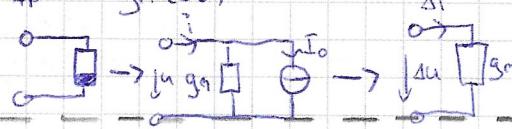
$$u_{F,lin} = \frac{\partial u_F}{\partial i_F} \Big|_{AP} \cdot (i_F - I_{AP}) + u_{AP}$$

Leitwert

Großsignal: Parallelsch. v. Widerstand & Spannungsquelle; Serienschaltung v.  $R$  und Spannungsquelle (Si zu gelassen)

Kleinsignale - Konst. Quellen zu null  
- alle Beschreibungen ein  $\Delta$

$$i - I_{AP} = \Delta i = g_m (u \Delta u)$$



Mind. Leitwert:  $G_i \geq \|G_L\|$

parametrisch

Betriebsmatrix

$$\begin{bmatrix} i_1 \\ i_2 \end{bmatrix} = \begin{bmatrix} u_1 \\ u_2 \end{bmatrix}$$

$$M u + N i = 0$$

$$\text{Bsp: } I_1 u - R_i i = 0$$

$$\rightarrow [I_1, -R_i] \begin{bmatrix} u \\ i \end{bmatrix} = 0$$

$$(\text{Bild})$$

$\rightarrow$  2 linear unabhängige  $\rightarrow$  KS- / LK- Methode  
Messungen (+1 falls Quelle) (abwesend)

1) Matrix quellenfrei bestimmen

2) Quelliektor best. (Beide steuerenden Größen zu 0)

3) Schaltung mit externen Quellen zeichnen

### Linearisierung

$$\begin{bmatrix} u_1 \\ u_2 \end{bmatrix} = Z_{AP} \begin{bmatrix} i_1 - I_1 \\ i_2 - I_2 \end{bmatrix} + \begin{bmatrix} u_1 \\ u_2 \end{bmatrix}$$

$$\begin{bmatrix} u_1 \\ u_2 \end{bmatrix} = A \begin{bmatrix} u_1 \\ u_2 \end{bmatrix}$$

$$\begin{bmatrix} u_1 \\ u_2 \end{bmatrix} = A \begin{bmatrix} u_1 \\ -u_1 \end{bmatrix}$$

## Eigenschaften

- Leistungsbilanz:  $p(t) = \bar{U}^T \bar{I} = p_1(t) + p_2(t)$
- Verlustlos:  $\bar{U}^T \bar{I} = 0 \rightarrow A = -A^T, R = -R^T$
- $\rightarrow \bar{U}^T \bar{I} + \bar{I}^T \bar{U} = 0$

- aktiv:  $\bar{U}^T \bar{I} < 0$

- passiv:  $\bar{U}^T \bar{I} \geq 0$

- Symmetrie / Umkehrbarkeit:  $A = A^T$

Vertausch d. Tore ändert nichts  
 $G = P A P^T, R = P R^T P$   
 $P = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}$

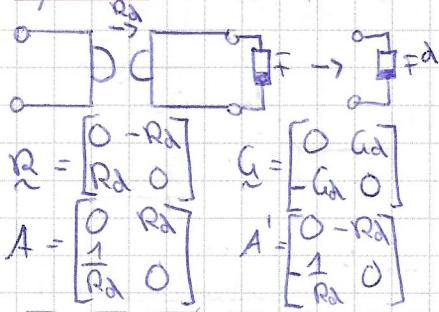
- Reziprozität (Übertragungssymmetrie)

$$R = R^T \quad G = G^T$$

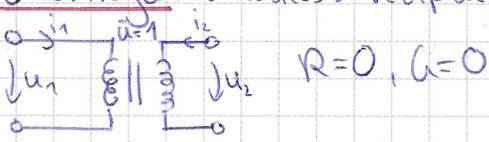
$$\frac{u_1}{i_1} = \frac{u_2}{i_2} \quad h_{12} = -h_{21} \quad h_{12}' = -h_{21}' \quad \bar{U}^T \bar{I} - \bar{I}^T \bar{U} = 0$$

$$\det(A) = 1 \quad \det(A') = 1$$

Gyator: verlustlos, nicht reziprok



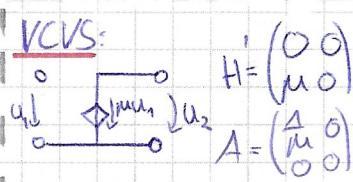
Übertrager: verlustlos & reziprok



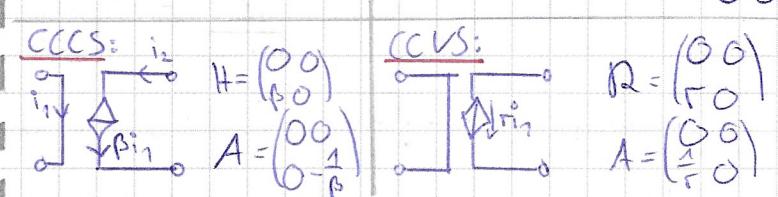
$$H = \begin{bmatrix} 0 & u_1 \\ -u_2 & 0 \end{bmatrix} \quad H' = \begin{bmatrix} 0 & -\frac{1}{u} \\ 1 & 0 \end{bmatrix} \quad A = \begin{bmatrix} u_1 & 0 \\ 0 & 1/u \end{bmatrix} \quad A' = \begin{bmatrix} 1 & 0 \\ 0 & u \end{bmatrix}$$

$$R_{\text{üb}} = \dot{u}^2 R \quad (\text{an Tor 2 Rangeschlossen})$$

## VCVS:

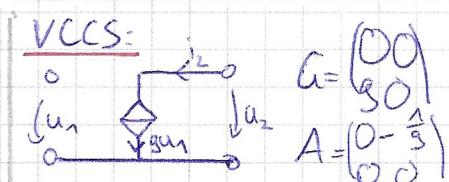


$$A = \begin{bmatrix} 1 & 0 \\ \frac{1}{\mu} & 0 \end{bmatrix}$$

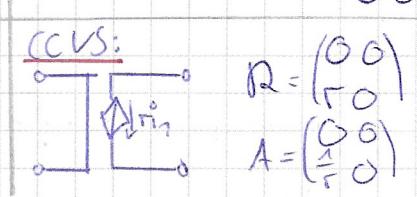


$$A = \begin{bmatrix} 0 & 1 \\ 0 & -\frac{1}{\mu} \end{bmatrix}$$

## VCCS:



$$A = \begin{bmatrix} 0 & 0 \\ 0 & -\frac{1}{\mu} \end{bmatrix}$$



$$A = \begin{bmatrix} 0 & 0 \\ \frac{1}{\beta} & 0 \end{bmatrix}$$

## Nullor

$$M = \begin{bmatrix} 10 \\ 00 \end{bmatrix} \quad N = \begin{bmatrix} 00 \\ 10 \end{bmatrix} \quad A = \begin{bmatrix} 00 \\ 00 \end{bmatrix}$$

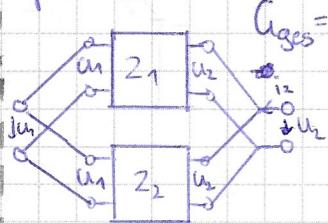
## MK:

$$H = \begin{bmatrix} 0 & -k \\ k & 0 \end{bmatrix}$$

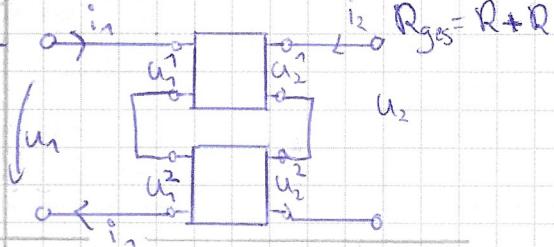
$$H' = \begin{bmatrix} 0 & -\frac{1}{k} \\ \frac{1}{k} & 0 \end{bmatrix} \quad A = \begin{bmatrix} -k & 0 \\ 0 & 1 \end{bmatrix} \quad A' = \begin{bmatrix} -\frac{1}{k} & 0 \\ 0 & k \end{bmatrix}$$

## Verschaltung

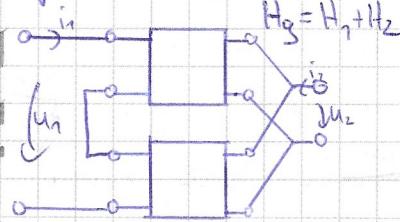
- parallel:



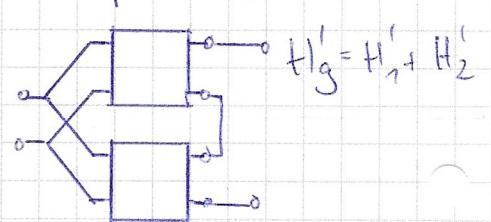
- seriell:



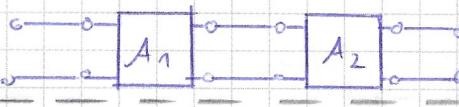
## serie-parallell parallel-seriell



## -serie-parallell-parallel-seriell



## Kettenschluss:



$$A_{\text{ges}} = A_1 \cdot A_2$$

$$A'_{\text{ges}} = A_2 \cdot A_1$$

## Aeg.

R	G	H	H'	A	A'
---	---	---	----	---	----

$$R \quad \begin{bmatrix} r_{11} & r_{12} \\ r_{21} & r_{22} \end{bmatrix} \quad \frac{1}{\det G} \begin{bmatrix} g_{22} & -g_{12} \\ -g_{21} & g_{11} \end{bmatrix} \quad \frac{1}{h_{22}} \begin{bmatrix} \det H & h_{12} \\ -h_{21} & 1 \end{bmatrix} \quad \frac{1}{h'_{11}} \begin{bmatrix} 1 & -h'_{12} \\ h'_{21} & \det H' \end{bmatrix} \quad \frac{1}{a_{21}} \begin{bmatrix} a_{11} & \det A \\ 1 & a_{22} \end{bmatrix} \quad \frac{1}{a'_{21}} \begin{bmatrix} a'_{22} & 1 \\ \det A' & a'_{11} \end{bmatrix}$$

$$G \quad \frac{1}{\det R} \begin{bmatrix} r_{22} & -r_{12} \\ -r_{21} & r_{11} \end{bmatrix} \quad \begin{bmatrix} g_{11} & g_{12} \\ g_{21} & g_{22} \end{bmatrix} \quad \frac{1}{h_{11}} \begin{bmatrix} 1 & -h_{12} \\ h_{21} & \det H \end{bmatrix} \quad \frac{1}{h'_{22}} \begin{bmatrix} \det H' & h'_{12} \\ -h'_{21} & 1 \end{bmatrix} \quad \frac{1}{a_{12}} \begin{bmatrix} a_{22} & -\det A \\ -1 & a_{11} \end{bmatrix} \quad \frac{1}{a'_{12}} \begin{bmatrix} a'_{11} & -1 \\ -\det A' & a'_{22} \end{bmatrix}$$

$$H \quad \frac{1}{r_{22}} \begin{bmatrix} \det R & r_{12} \\ -r_{21} & 1 \end{bmatrix} \quad \frac{1}{g_{11}} \begin{bmatrix} 1 & -g_{12} \\ g_{21} & \det G \end{bmatrix} \quad \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} \quad \frac{1}{\det H'} \begin{bmatrix} h'_{22} & -h'_{12} \\ -h'_{21} & h'_{11} \end{bmatrix} \quad \frac{1}{a_{22}} \begin{bmatrix} a_{12} & \det A \\ -1 & a_{21} \end{bmatrix} \quad \frac{1}{a'_{11}} \begin{bmatrix} a'_{12} & 1 \\ -\det A' & a'_{21} \end{bmatrix}$$

$$H' \quad \frac{1}{r_{11}} \begin{bmatrix} 1 & -r_{12} \\ r_{21} & \det R \end{bmatrix} \quad \frac{1}{g_{22}} \begin{bmatrix} \det G & g_{12} \\ -g_{21} & 1 \end{bmatrix} \quad \frac{1}{\det H} \begin{bmatrix} h_{22} & -h_{12} \\ -h_{21} & h_{11} \end{bmatrix} \quad \begin{bmatrix} h'_{11} & h'_{12} \\ h'_{21} & h'_{22} \end{bmatrix} \quad \frac{1}{a_{11}} \begin{bmatrix} a_{21} & -\det A \\ 1 & a_{12} \end{bmatrix} \quad \frac{1}{a'_{22}} \begin{bmatrix} a'_{21} & -1 \\ \det A' & a'_{12} \end{bmatrix}$$

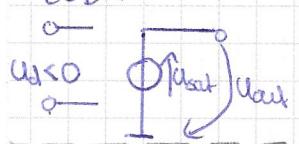
$$A \quad \frac{1}{r_{21}} \begin{bmatrix} r_{11} & \det R \\ 1 & r_{22} \end{bmatrix} \quad \frac{1}{g_{21}} \begin{bmatrix} -g_{22} & -1 \\ -\det G & g_{11} \end{bmatrix} \quad \frac{1}{h_{21}} \begin{bmatrix} -\det H & -h_{11} \\ -h_{22} & -1 \end{bmatrix} \quad \begin{bmatrix} 1 & h'_{22} \\ h'_{11} & \det H' \end{bmatrix} \quad \begin{bmatrix} a_{11} & a_{12} \\ a_{21} & a_{22} \end{bmatrix} \quad \frac{1}{\det A'} \begin{bmatrix} a'_{22} & a'_{12} \\ a'_{21} & a'_{11} \end{bmatrix}$$

$$A' \quad \frac{1}{r_{12}} \begin{bmatrix} r_{22} & \det R \\ 1 & r_{11} \end{bmatrix} \quad \frac{1}{g_{12}} \begin{bmatrix} -g_{11} & -1 \\ -\det G & g_{22} \end{bmatrix} \quad \frac{1}{h_{12}} \begin{bmatrix} 1 & h_{11} \\ h_{22} & \det H \end{bmatrix} \quad \begin{bmatrix} -\det H' & -h'_{22} \\ -h'_{11} & -1 \end{bmatrix} \quad \frac{1}{\det A} \begin{bmatrix} a_{22} & a_{12} \\ a_{21} & a_{11} \end{bmatrix} \quad \begin{bmatrix} a'_{11} & a'_{12} \\ a'_{21} & a'_{22} \end{bmatrix}$$

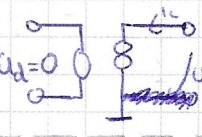
Gesucht

OP-Amps:

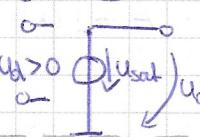
ESB I:



ESB II



ESB III



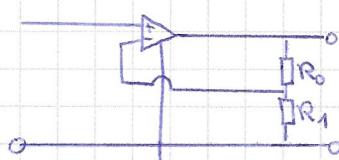
I

Usat III

Gleichungen:

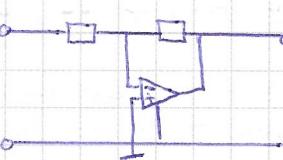
$$\begin{aligned} \text{I: } & U_d < 0, U_{out} = -U_{sat} \\ \text{II: } & U_d = 0, |U_{out}| \leq |U_{sat}| \\ \text{III: } & U_d > 0, U_{out} = U_{sat} \end{aligned}$$

Nicht invertierender Verstärker



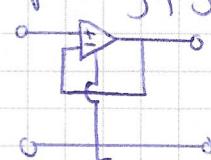
$$U_{out} = \left(1 + \frac{R_2}{R_1}\right) U_{in}$$

Invertierender Verstärker



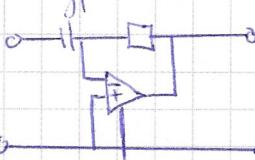
$$U_{out} = -\frac{R_2}{R_1} U_{in}$$

Spannungsfolger



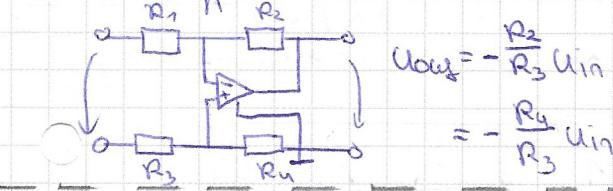
$$U_{out} = U_{in}$$

Differenzierer



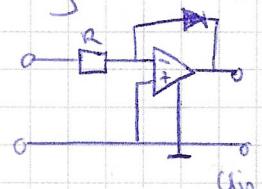
$$U_{out} = -RC \cdot U_{in}$$

Potenzialdifferenzverstärker



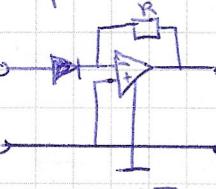
$$\begin{aligned} U_{out} &= -\frac{R_2}{R_3} U_{in} \\ &= -\frac{R_4}{R_3} U_{in} \end{aligned}$$

Logarithmierer



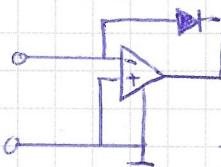
$$U_{out} = -U_T \ln \frac{U_{in}}{R \cdot I_S}$$

exponierer



$$U_{out} = -RI_S \exp\left(\frac{U_{in}}{U_T}\right)$$

Ideale Diode

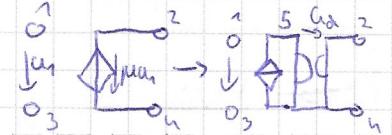


log-Mehrtore

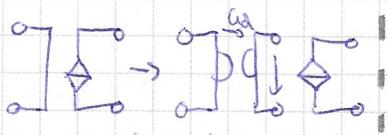
$$\begin{bmatrix} i_1 \\ i_2 \\ \vdots \\ i_p \end{bmatrix} = \begin{bmatrix} 0 & \frac{1}{u_1} & \dots & \frac{1}{u_p} \\ \frac{1}{u_1} & 0 & \ddots & 0 \\ \vdots & \ddots & \ddots & 0 \\ 0 & 0 & \dots & 0 \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \\ \vdots \\ u_p \end{bmatrix}$$

$$Q = \begin{bmatrix} 0 & R - R \\ -R & 0 & R \\ R & -R & 0 \end{bmatrix}$$

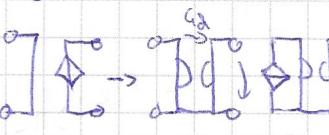
VCVS  $\rightarrow$  VCCS



CCCS  $\rightarrow$  VCCS



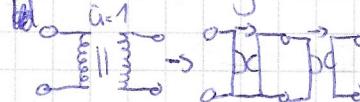
CCVS  $\rightarrow$  VCCS



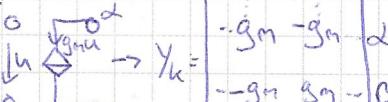
$$u = \frac{i_1}{G_d}$$

$$\begin{aligned} i_2 &= G_d u = \frac{1}{G_d} u \\ u_2 &= -G_d \frac{1}{G_d} u = u \end{aligned}$$

Idealer Absteiger



VCCS



2) Knoten nummerieren

3) Aufstellen d. Knotenleitwertsmatrix

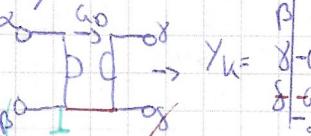
$$Y_K = \begin{bmatrix} G & -G & 0 \\ -G & G & 0 \\ 0 & 0 & \ddots \end{bmatrix}$$

5) Streichen

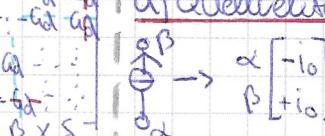
o o o Zeile add + streichen

o o o Spalte add + streichen

Gyrorator:



$\alpha$  Quellenleitor



$$\alpha \begin{bmatrix} -10 \\ 10 \end{bmatrix}$$

$$\beta \begin{bmatrix} 10 \\ 10 \end{bmatrix}$$

## Tableaugleichungen

- o KVL:  $\sum u = 0$
- o KCH:  $\sum i = 0$
- o Bauteile:  $\sum u + \sum i = 0$

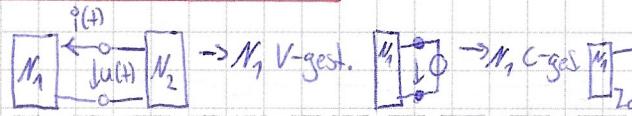
$$\left\{ \begin{array}{l} \text{KVL: } \sum u = 0 \\ \text{KCH: } \sum i = 0 \\ \text{Bauteile: } \sum u + \sum i = 0 \end{array} \right. \rightarrow \left[ \begin{array}{c|c|c} B & 0 & 0 \\ 0 & A & 0 \\ M & N & 0 \end{array} \right] \begin{bmatrix} u \\ i \\ e \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}$$

## Knotenspannungsanalyse

$$\left[ \begin{array}{c|c|c} -A^T & 1 & 0 \\ 0 & 0 & A \\ 0 & M & N \end{array} \right] \begin{bmatrix} u \\ i \\ e \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}$$

$$\left[ \begin{array}{c|c|c} B & 0 & 0 \\ 0 & 1 & -B^T \\ M & N & 0 \end{array} \right] \begin{bmatrix} u \\ i \\ e \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}$$

## Substitutionstheorem



(KVL in Bildraum) (KCH in Bildraum)

## Zweipolersatzschaltung

### Bestimmung d. Bauteilwerte

$$1) \text{ über Kurzschluss / } \text{dc} \rightarrow R = -\frac{u_0}{i_0}, \quad G = -\frac{i_0}{u_0}$$

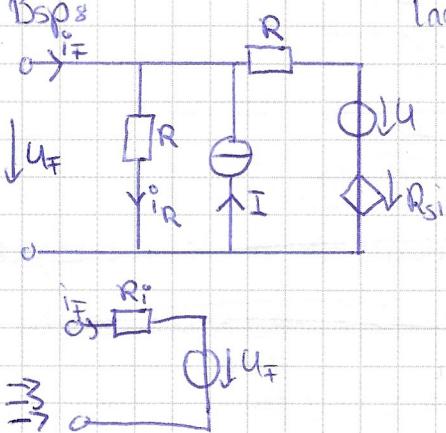
$$2) U_0 = I_0 / I_{0S} - US$$

→ alle unabh. Quellen zu null und  $u$  in abh. v.  $i$  berechnen

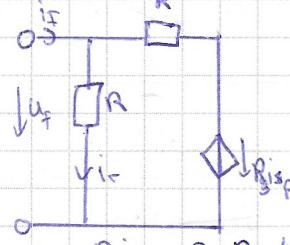
$$G_i / I_i$$

$$I_d / I_{dS}$$

## Bsp:

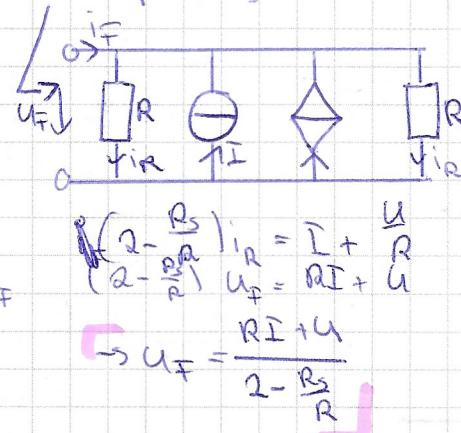


Innenwiderstand



$$R_i = \frac{U_F - U_RL}{I_F} = \frac{U_F - U_RL}{U_F / R} = \frac{R}{2 - R_s/R} U_F$$

$I_d$ -Spannung (+ Quellwandlung)



$$(2 - \frac{R_s}{R}) I_d = I + \frac{U}{R}$$

$$I_d = \frac{RI + U}{2 - \frac{R_s}{R}}$$

## Realtanzen

$$\text{ladung } q(t) = \int i(t) dt = q(t_0) + \int_{t_0}^t i(t) dt$$

$$i(t) = \frac{dq(t)}{dt}$$

$$u(t) = \frac{dq(t)}{dt}$$

Einheit

$$\Phi = L i \quad Q = C u \quad \frac{Q}{u} = 1 \text{ Farad: F}$$

$$\text{fluss } \Phi(t) = \int u(t) dt = \Phi(t_0) + \int_{t_0}^t u(t) dt$$

Eigenschaften: - Energie:  $W(t_0, t_1) = \int_{t_0}^{t_1} p(t) dt$  Dualität

- verlustfrei: Kondensator hat keine Schleifen

$$W_C = \int_{q_1}^{q_2} u(q) dq \quad W_L = \int_{\Phi_1}^{\Phi_2} i(\Phi) d\Phi$$

linear:

$$W_C = \frac{1}{2} C (u_2^2 - u_1^2) \quad W_L = \frac{1}{2} (i_2^2 - i_1^2)$$

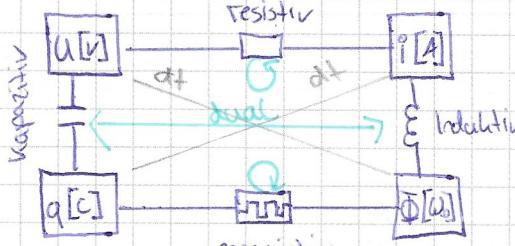
$$E_C = \frac{1}{2} C u^2$$

$$E_L = \frac{1}{2} L i^2$$

$$\Phi \rightarrow R_d q \quad q \rightarrow \frac{\Phi}{R_d}$$

$$Kapaz.: \text{Parallels } C_1 + C_2 \quad \text{Serie } C_1 || C_2$$

$$Indukt.: \text{Parallels } L_1 || L_2 \quad \text{Serie } L_1 + L_2$$



OKS o Ld o Nuller o Nullaten

→ alle  $u$  Gruppen

nicht linear Kapazität Induktivität

$$q = c(u)$$

$$\Phi = l(i)$$

$$u = x(q)$$

$$i = n(\Phi)$$

Memistor

$$q = n(\Phi)$$

$$\Phi = \mu(a)$$

## Relaxationspunkte

= Ruhepunkte (gesp. Energie minimal)

1) Schnittpunkte Achsen

- Knickepunkte
- Extremstellen

2) Energie aufnehmen, wenn bewegt von einem Relax. Punkt zu einem anderen

$$\omega = \text{Sudg} = \frac{1}{2} d\phi \geq 0$$

$\rightarrow \text{Produkt muss } > 0$

## Zeiger Darstellung:

$$a(t) = \hat{A} \cos(\omega t + \varphi) \quad \hat{A} = A e^{j\varphi}$$

$$= \operatorname{Re}\left\{\hat{A} e^{j(\omega t + \varphi)}\right\}$$

## Komplexe Leistung:

$$P = \frac{1}{2} UI^* = \frac{1}{2} \hat{U} \hat{I} e^{j\theta_u - j\theta_i} \quad (\text{Bei Zeigern})$$

$$P = P_w + j P_B$$

kompl. Leistung  
Scheinleistung:

$$= \operatorname{Re}\left\{\frac{1}{2} U I^*\right\} \quad \text{Wirkleistung}$$

$$= \operatorname{Im}\left\{\frac{1}{2} U I^*\right\} \quad \text{Blindleistung}$$

$S = |P| = \sqrt{P_w^2 + P_B^2}$

- nur an resistiven Elementen
- Mittelwert der Momentanleistung
- physikalisch: Leistung die zur period.

- Newton Raphson:

$$\begin{bmatrix} 0 & G \\ 0 & A \\ M & K \end{bmatrix} \begin{bmatrix} u \\ i \\ \dot{u} \end{bmatrix} = \begin{bmatrix} G \\ G \\ e \end{bmatrix} \quad \text{Grenzen Teil}$$

1. Wähle Initialisierung  $\begin{bmatrix} u \\ i \end{bmatrix}$  mit  $f(u, i) = 0$

2. Im  $n+1$ -ten Schritt ≈ linearisiere  $f(u, i) = 0$  in  $n$ -ter Kandidaten für den AP  $\begin{bmatrix} u \\ i \end{bmatrix}$

3. Löse das neue GLS  $\Rightarrow \begin{bmatrix} \bar{u}^{(n+1)} \\ \bar{i}^{(n+1)} \end{bmatrix}$

4. finde den neuen Kandidaten für AP in der Nähe  $(\bar{u}^{(n+1)}, \bar{i}^{(n+1)})$  mit  $f(\bar{u}^{(n+1)}, \bar{i}^{(n+1)}) = 0$

Abruchkriterium:  $\|\begin{bmatrix} \bar{u}^{(n+1)} \\ \bar{i}^{(n+1)} \end{bmatrix} - \begin{bmatrix} u^n \\ i^n \end{bmatrix}\| \leq \epsilon$  (Toleranz, Genauigkeit)

## Komplexe Wechselstromrechnung

### 1) Voraussetzungen

- lineares, zeitinvariantes System

- sinusförmige Wechselsp. o. Wechselstrom

→ alle Signal d. Schaltung sinusförmig mit gl. Frequenz

$$I = I \sin(\omega t) \approx \hat{I} \cos(\omega t - \frac{\pi}{2})$$

$$U = \hat{U} \sin(\omega t) \approx \hat{U} \cos(\omega t - \frac{\pi}{2})$$

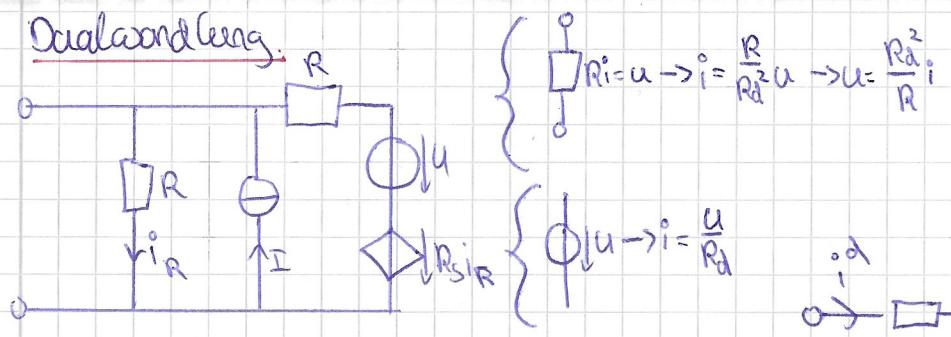
## Rechenregeln:

$$\frac{\partial}{\partial t} \leftrightarrow j\omega \quad u(t) = L i(t) \approx j\omega L I = U$$

$$\ln C: a+jb = r e^{j\varphi} \quad z_1 + z_2 = (R_1 + R_2) + j(L_1 + L_2)$$

$$a-jb = r e^{-j\varphi}$$

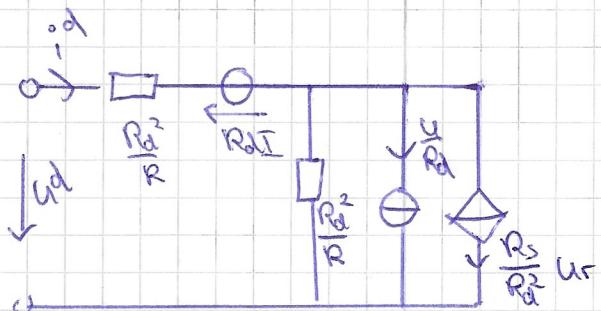
## Dualwandlung



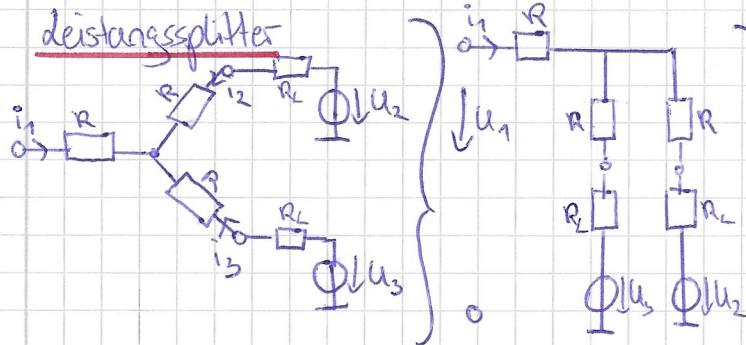
$$\left\{ \begin{array}{l} \text{---} \\ \text{---} \end{array} \right\} \frac{U}{I} \rightarrow u = R_d I$$

$$\left\{ \begin{array}{l} \text{---} \\ \text{---} \end{array} \right\} \frac{U}{I} \rightarrow i = \frac{R_s}{R_d^2} u_R$$

=>



## Leistungs splitter



Leerlaufsp. mit Superpositionsprinzip

$$U_{01} = \frac{U_3}{2}$$

$$\frac{U_3}{2} \parallel R_L \parallel R + R_L \rightarrow U_{02} = \frac{U_2}{2}$$

$$U_1 = \frac{U_3 + U_2}{2}$$

$$R_P = \frac{3R + R_L}{2}$$

$i_2$  beliebig  $\rightarrow$  Existiert eine  
Inverse Vektorbeschreibung?

Nen, da  $i_2$  nicht von den anderen  
Größen abhängt und sonst steuert.

$$\begin{bmatrix} 1 & -2 \\ 0 & 1 \end{bmatrix} u + \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} i = 0 \Rightarrow \text{In inverse  
hybridbesch.}$$

$$H \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} u + \begin{bmatrix} 1 & -2 \\ 0 & 1 \end{bmatrix} i = 0$$

$$H_A^{-1} = -\begin{bmatrix} 1 & -2 \\ 0 & 1 \end{bmatrix}^{-1} \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} = 0 - 1 \Omega$$

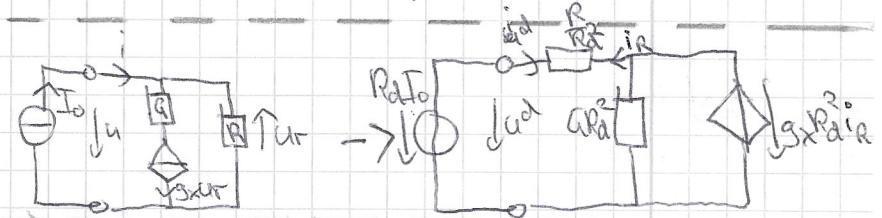
$$i = \ln\left(\frac{u}{u_0} + 1\right) \text{ mA} \quad \text{für Kleinstagssollwert } u_0$$

Messen Sie g<sub>ks</sub> = 0,2 mS

$$\rightarrow \frac{di}{du_{AP}} = \frac{1mS}{1 + \frac{u_{AP}}{u_0}} = g_{KS}$$

$$\rightarrow u_{AP} = \frac{1mS - g_{KS}}{g_{KS}} \cdot 1V = 0V$$

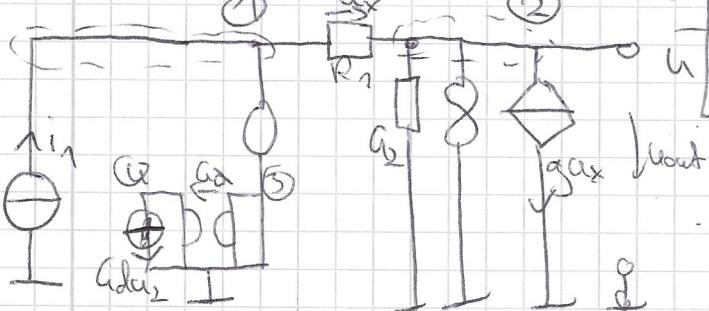
$$\rightarrow i_{AP} = \ln(5) \text{ mA}$$



G<sub>B</sub> erg. leiten Sie aus Bedingung  
d. Reciprozität eine Bed. für G<sub>B</sub> her

$$U^T I - I^T U = 0 \rightarrow U^T G_B U - U^T G_B^T I = 0$$

$$\rightarrow G_B = G_B^T \rightarrow G_B = G_2$$



$$\begin{aligned} & \text{1} \quad 2 \quad 3 \quad 4 \\ & \text{1} \quad \left[ \begin{array}{cccc} \frac{1}{R_1} & -\frac{1}{R_1} & 0 & 0 \\ \frac{1}{R_1+G_2} & \frac{1}{R_1+G_2} & 0 & 0 \end{array} \right] \quad \left[ \begin{array}{c} u_{k1} \\ u_{k2} \\ u_{k3} \\ u_{k4} \end{array} \right] = \left[ \begin{array}{c} i_1 \\ -i_2 \end{array} \right] \\ & \text{2} \quad \left[ \begin{array}{cccc} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{array} \right] \quad \left[ \begin{array}{c} u_{k1} \\ u_{k2} \\ u_{k3} \\ u_{k4} \end{array} \right] = \left[ \begin{array}{c} 0 \\ 0 \end{array} \right] \end{aligned}$$

$$u_x = u_{k1} - u_{k2}$$