

Digital Image Processing

COSC 6380/4393

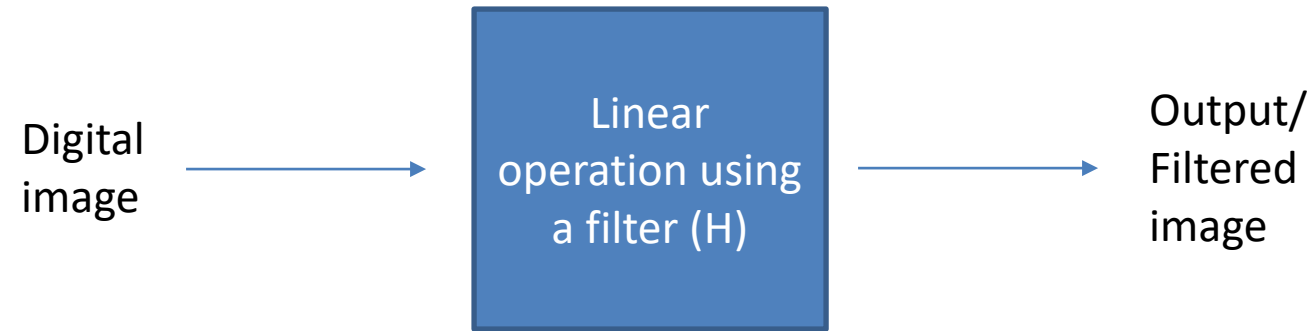
Lecture – 22

Mar. 6th, 2023

Pranav Mantini

Slides from Dr. Shishir K Shah, and Frank Liu

Linear Image Filtering (Review)



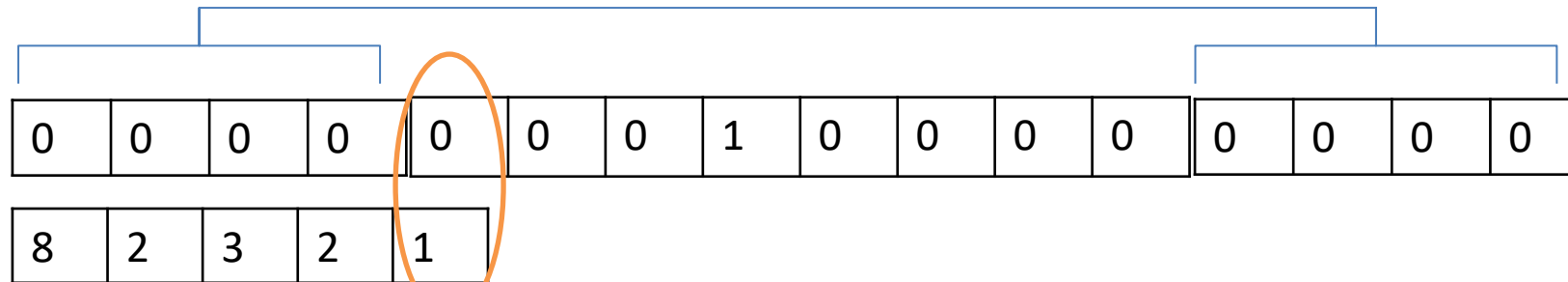
Review

$$f \begin{array}{|c|c|c|c|c|c|c|c|} \hline 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ \hline \end{array} \otimes w \begin{array}{|c|c|c|c|c|} \hline 1 & 2 & 3 & 2 & 8 \\ \hline \end{array}$$

Review

$$f \begin{array}{|c|c|c|c|c|c|c|c|} \hline 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ \hline \end{array} \otimes \begin{array}{|c|c|c|c|c|} \hline 1 & 2 & 3 & 2 & 8 \\ \hline \end{array}$$

Zero Padding



Initial Position

Review

$$f \begin{array}{|c|c|c|c|c|c|c|c|} \hline 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ \hline \end{array} \otimes w \begin{array}{|c|c|c|c|c|} \hline 1 & 2 & 3 & 2 & 8 \\ \hline \end{array}$$

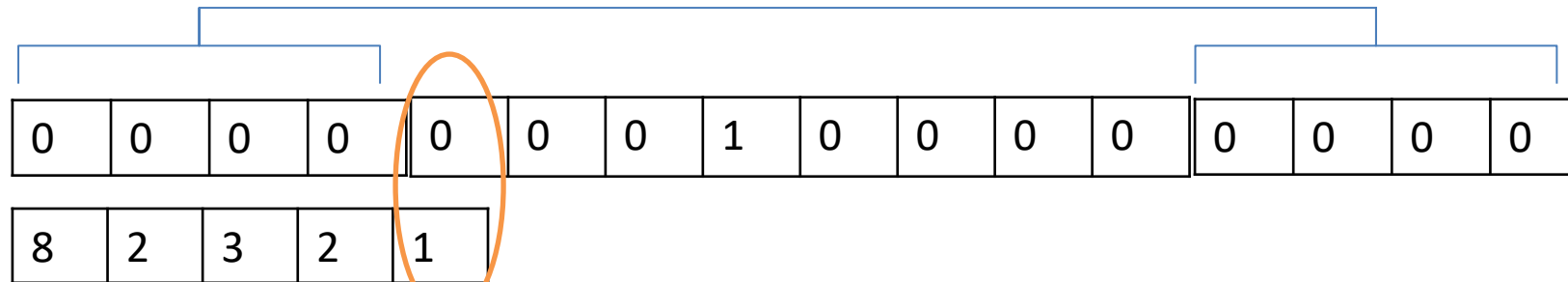
Cropped Convolution result

0	1	2	3	2	8	0	0
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Review

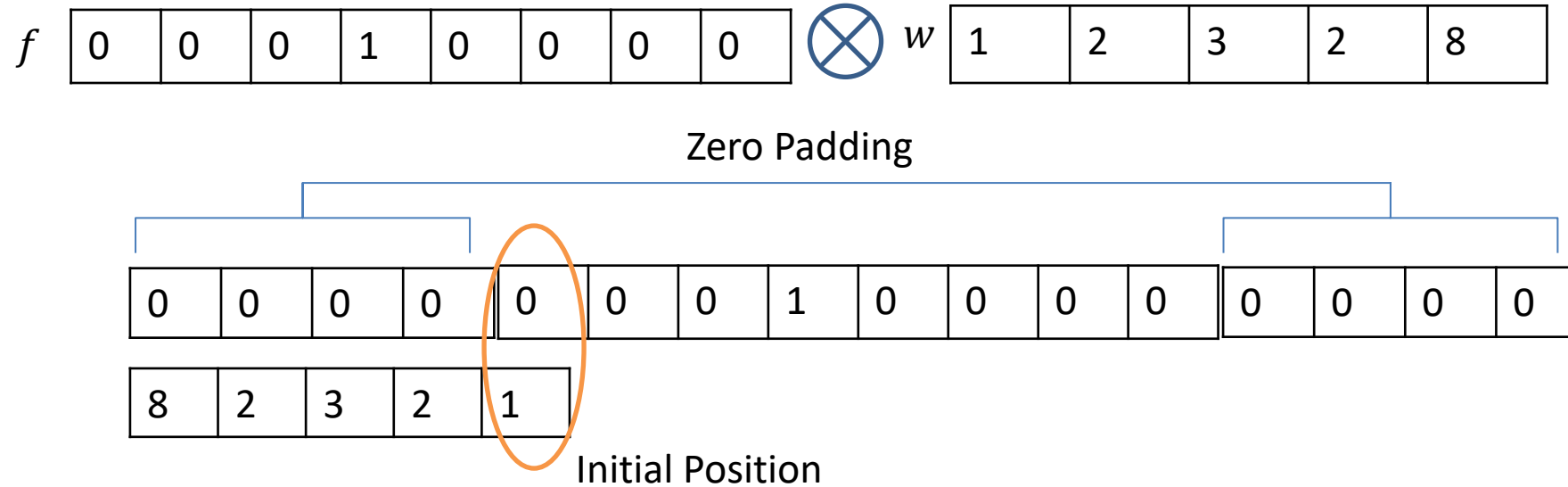
$$f \begin{array}{|c|c|c|c|c|c|c|c|} \hline 0 & 0 & 0 & 1 & 0 & 0 & 0 & 0 \\ \hline \end{array} \otimes \begin{array}{|c|c|c|c|c|} \hline 1 & 2 & 3 & 2 & 8 \\ \hline \end{array}$$

Zero Padding



Initial Position

Review



$$f(t) \otimes w(t) = \sum_{\tau=-2}^2 w(\tau) f(t - \tau)$$

Spatial Convolution Operator

The convolution of a filter $w(x, y)$ of size $m \times n$ with an image $f(x, y)$, denoted as $w(x, y) \otimes f(x, y)$

$$w(x, y) \otimes f(x, y) = \sum_{s=-a}^a \sum_{t=-b}^b w(s, t) f(x-s, y-t)$$



Linear Systems

And Linear Image Filtering

(Review)

- A process that accepts a signal or image I as input and transforms it by an act of linear convolution is a type of **linear system**
- **Example**



Some Specific Goals

- **smoothing** - remove noise from bit errors, transmission, etc
- **deblurring** - increase **sharpness** of blurred images
- **sharpening** - emphasize significant features, such as **edges**
- **combinations** of these

Review: Two Smoothing Averaging Filter Masks

$$\frac{1}{9} \times$$

1	1	1
1	1	1
1	1	1

$$\frac{1}{16} \times$$

1	2	1
2	4	2
1	2	1

a b

FIGURE 3.32 Two 3×3 smoothing (averaging) filter masks. The constant multiplier in front of each mask is equal to 1 divided by the sum of the values of its coefficients, as is required to compute an average.

Review: Laplace Operator

0	1	0	1	1	1
1	-4	1	1	-8	1
0	1	0	1	1	1
0	-1	0	-1	-1	-1
-1	4	-1	-1	8	-1
0	-1	0	-1	-1	-1

a b
c d

FIGURE 3.37

(a) Filter mask used to implement Eq. (3.6-6).

(b) Mask used to implement an extension of this equation that includes the diagonal terms.

(c) and (d) Two other implementations of the Laplacian found frequently in practice.

Review: Sharpening Spatial Filters: Laplace Operator

Image sharpening in the way of using the Laplacian:

$$g(x, y) = f(x, y) + c \left[\nabla^2 f(x, y) \right]$$

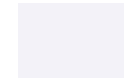
where,

$f(x, y)$ is input image,

$g(x, y)$ is sharpened images,

$c = -1$ if $\nabla^2 f(x, y)$ corresponding to Fig. 3.37(a) or (b)

and $c = 1$ if either of the other two filters is used.



Discrete Fourier Transform

Spatial Domain (x) \longrightarrow Frequency Domain (u)

Fourier Transform

$$F(u) = \int_{-\infty}^{\infty} f(x) e^{-\sqrt{-1}ux} dx$$

Discrete Fourier Transform

$$F(u) = \sum_{x=-\infty}^{\infty} f(x) e^{-\sqrt{-1}ux}$$

Frequency Domain (u) \longrightarrow Spatial Domain (x) $e^{-\sqrt{-1}x} = \cos x - \sqrt{-1}\sin x$

Inverse Fourier Transform

$$f(x) = \int_{-\infty}^{\infty} F(u) e^{\sqrt{-1}ux} du$$

Inverse Discrete Fourier Transform

$$f(x) = \sum_{u=-\infty}^{\infty} F(u) e^{\sqrt{-1}ux}$$

$$e^{\sqrt{-1}x} = \cos x + \sqrt{-1}\sin x$$

Convolution Theorem

- Let f be an image and h a filtering window
- Lets us consider the convolution

$$f(t) \otimes h(t)$$

Convolution Theorem

- Let f and h be two function
- Lets us consider the convolution

$$f(t) \otimes h(t) = \sum_{\tau=-\infty}^{\infty} f(\tau)h(t - \tau)$$

From the definition of convolution

\otimes

Convolution Theorem

- Let f and h be two function
- Lets us consider the convolution

$$f(t) \otimes h(t) = \sum_{\tau=-\infty}^{\infty} f(\tau)h(t - \tau)$$

$$F[f(t) \otimes h(t)] = \sum_{t=-\infty}^{\infty} \left[\sum_{\tau=-\infty}^{\infty} f(\tau)h(t - \tau) \right] e^{-\sqrt{-1}\mu t}$$

Computing the Fourier transform of the convolution

$$F(u) = \sum_{x=-\infty}^{\infty} f(x)e^{-\sqrt{-1}ux}$$

Convolution Theorem

- Let f and h be two function
- Lets us consider the convolution

$$\begin{aligned}
 f(t) \otimes h(t) &= \sum_{\tau=-\infty}^{\infty} f(\tau) h(t - \tau) \\
 F[f(t) \otimes h(t)] &= \sum_{t=-\infty}^{\infty} \left[\sum_{\tau=-\infty}^{\infty} f(\tau) h(t - \tau) \right] e^{-\sqrt{-1}\mu t} \\
 &= \sum_{\tau=-\infty}^{\infty} f(\tau) \left[\sum_{t=-\infty}^{\infty} h(t - \tau) e^{-\sqrt{-1}\mu t} \right]
 \end{aligned}$$

Convolution Theorem

- Let f and h be two function
- Lets us consider the convolution

$$\begin{aligned}
 f(t) \otimes h(t) &= \sum_{\tau=-\infty}^{\infty} f(\tau) h(t - \tau) \\
 F[f(t) \otimes h(t)] &= \sum_{t=-\infty}^{\infty} \left[\sum_{\tau=-\infty}^{\infty} f(\tau) h(t - \tau) \right] e^{-\sqrt{-1}\mu t} \\
 &= \sum_{\tau=-\infty}^{\infty} f(\tau) \left[\sum_{t=-\infty}^{\infty} h(t - \tau) e^{-\sqrt{-1}\mu t} \right] \\
 &= \sum_{\tau=-\infty}^{\infty} f(\tau) \left[\sum_{t=-\infty}^{\infty} h(t - \tau) e^{-\sqrt{-1}\mu(t-\tau)} \right] e^{-\sqrt{-1}\mu\tau}
 \end{aligned}$$

Convolution Theorem

- Let f and h be two function
- Lets us consider the convolution

$$\begin{aligned}
 f(t) \otimes h(t) &= \sum_{\tau=-\infty}^{\infty} f(\tau) h(t - \tau) \\
 F[f(t) \otimes h(t)] &= \sum_{t=-\infty}^{\infty} \left[\sum_{\tau=-\infty}^{\infty} f(\tau) h(t - \tau) \right] e^{-\sqrt{-1}\mu t} \\
 &= \sum_{\tau=-\infty}^{\infty} f(\tau) \left[\sum_{t=-\infty}^{\infty} h(t - \tau) e^{-\sqrt{-1}\mu t} \right] \\
 &= \sum_{\tau=-\infty}^{\infty} f(\tau) \left[\sum_{t=-\infty}^{\infty} h(t - \tau) e^{-\sqrt{-1}\mu(t-\tau)} \right] e^{-\sqrt{-1}\mu\tau} \\
 &= \sum_{\tau=-\infty}^{\infty} f(\tau) [H(\mu)] e^{-\sqrt{-1}\mu\tau} = H(\mu) \sum_{\tau=-\infty}^{\infty} f(\tau) e^{-\sqrt{-1}\mu\tau} \\
 &= H(\mu)F(\mu)
 \end{aligned}$$

Convolution Theorem

- Fourier transform pairs

$$f(t) \otimes h(t) \Leftrightarrow H(\mu)F(\mu)$$

$$f(t)h(t) \Leftrightarrow H(\mu) \otimes F(\mu)$$

The Basic Filtering in the Frequency Domain

- ▶ Modifying the Fourier transform of an image
- ▶ Computing the inverse transform to obtain the processed result

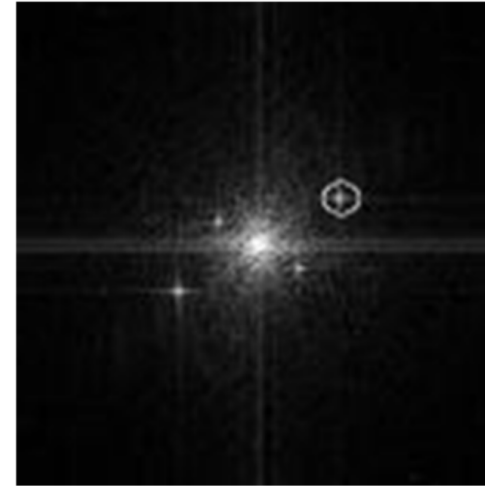
$$g(x, y) = \mathfrak{F}^{-1} \{H(u, v)F(u, v)\}$$

$F(u, v)$ is the DFT of the input image

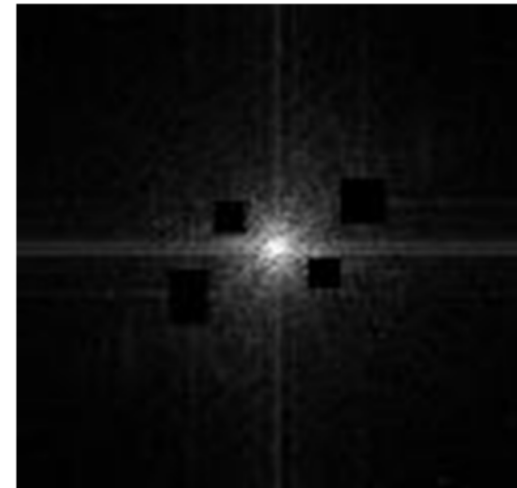
$H(u, v)$ is a filter function.

Example: Periodic Noise removal

Fourier transform

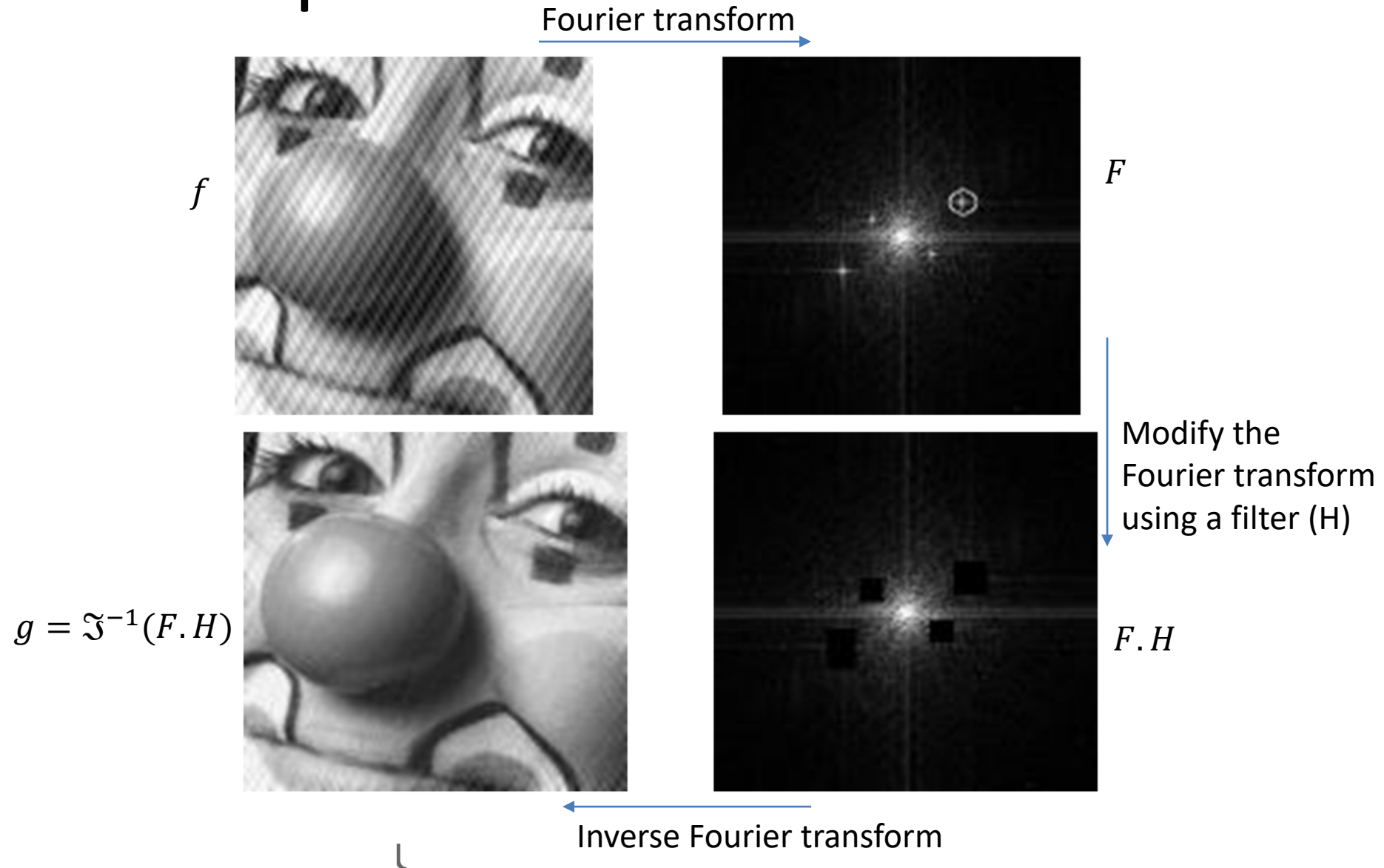


Modify the
Fourier transform
using a filter



Inverse Fourier transform

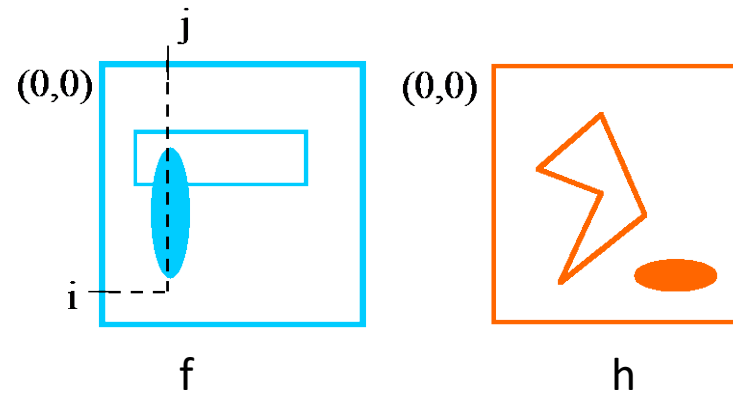
Example: Periodic Noise removal



$$\begin{aligned} g &= \mathfrak{I}^{-1}(F.H) \\ &\equiv \mathfrak{I}^{-1}(F) \otimes \mathfrak{I}^{-1}(H) \\ &= f \otimes h \end{aligned}$$

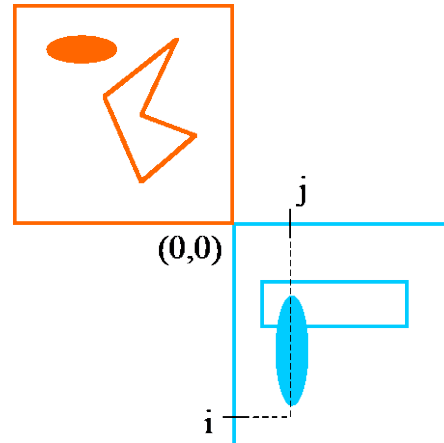
Diagrams of Convolution

- Consider the two images with image **f** and **h** and its contents **shaded** at each stage of processing shown:

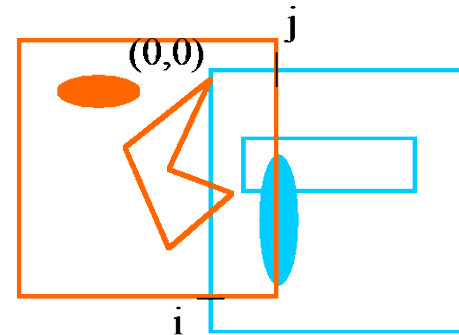


Diagrams of Convolution

- The image h is then reversed (reflected), along both axes.



- The reversed version of h is then shifted by the amount (i, j) along both axes:



$$\begin{aligned} g &= \mathfrak{I}^{-1}(F.H) \\ &\equiv \mathfrak{I}^{-1}(F) \otimes \mathfrak{I}^{-1}(H) \\ &= f \otimes h \end{aligned}$$

Review: Periodic Extension of Image

- The IDFT equation
$$I(i, j) = \frac{1}{N^2} \sum_{u=0}^{N-1} \sum_{v=0}^{N-1} \tilde{I}(u, v) W_N^{-(ui+vj)}$$

implies the **periodic extension of the image I** as well (with period N), simply by letting the arguments (i, j) take any integer value.

- Note that for any integers n, m

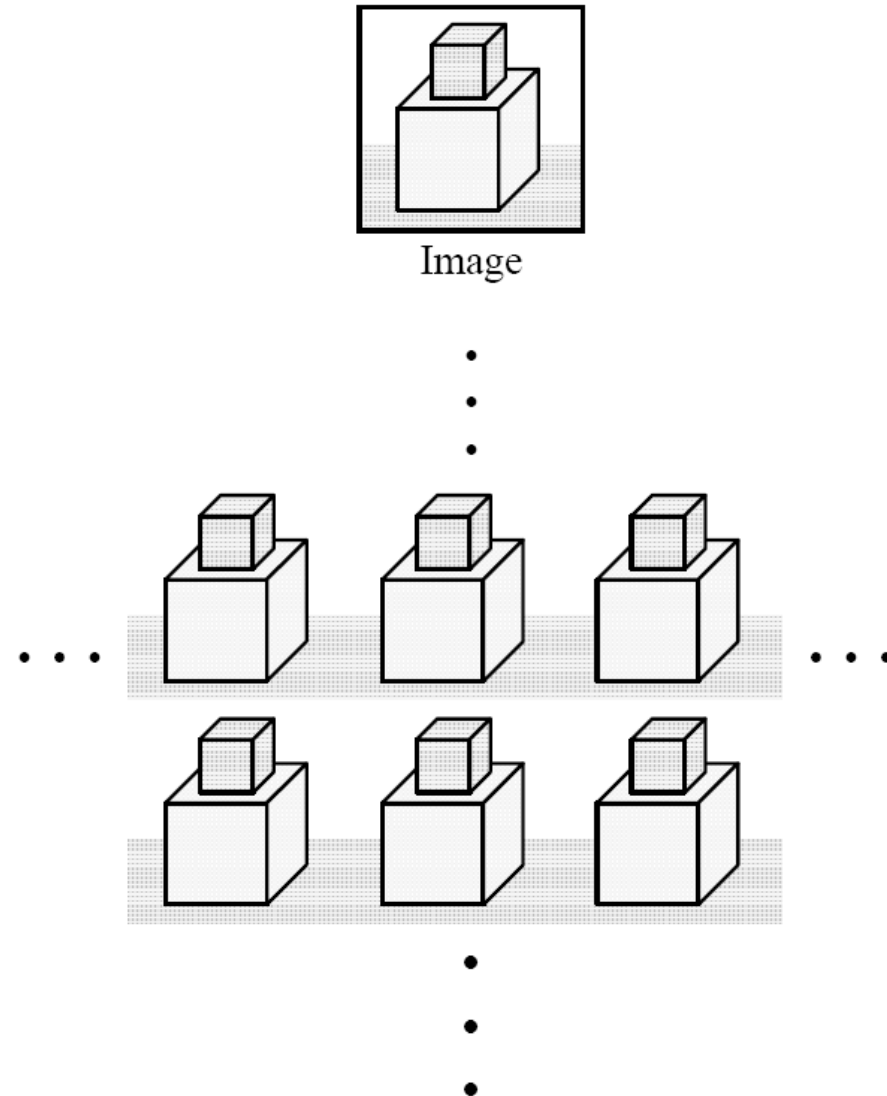
$$\begin{aligned} I(i+nN, j+mN) &= \frac{1}{N^2} \sum_{u=0}^{N-1} \sum_{v=0}^{N-1} \tilde{I}(u, v) W_N^{-[u(i+nN)+v(j+mN)]} \\ &= \frac{1}{N^2} \sum_{u=0}^{N-1} \sum_{v=0}^{N-1} \tilde{I}(u, v) W_N^{-(ui+vj)} W_N^{-N(nu+mv)} \\ &= \frac{1}{N^2} \sum_{u=0}^{N-1} \sum_{v=0}^{N-1} \tilde{I}(u, v) W_N^{-(ui+vj)} = I(i, j) \end{aligned}$$

since

$$W_N^{-N(nu+mv)} = e^{-j2\pi \frac{1}{N} \cdot N(nu+mv)} = e^{-j2\pi (nu+mv)} = 1^{(nu+mv)} = 1$$

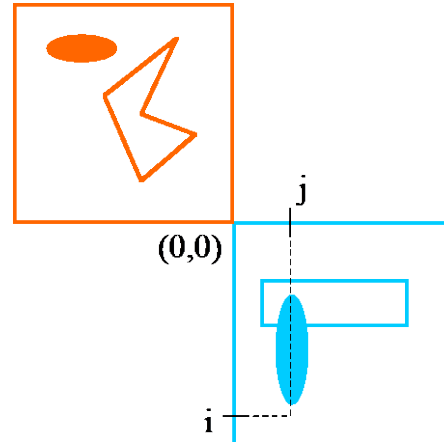
- In a sense, the DFT **implies** that the image **I** is already periodic.
- This will be extremely important when we consider **convolution**

Review: Periodic Extension of Image

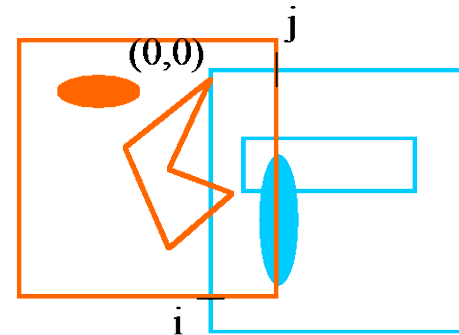


Diagrams of Convolution

- The image h is then reversed (reflected), along both axes. This requires that it be defined for negative coordinates, i.e., the periodic extension is used.



- The reversed version of h is then shifted by the amount (i, j) along both axes:



Diagrams of Convolution

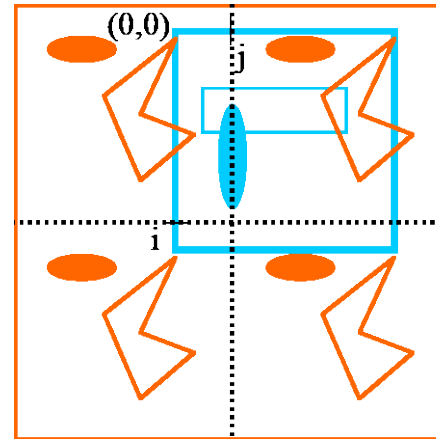
- The sum extends over $0 \leq m \leq M-1$, $0 \leq n \leq N-1$ (in blue) so some of the arguments of $h(i-m, j-n)$ fall outside the range $0, \dots, N-1$. What is computed is the summation of the product of

$$[f(m, n) ; 0 \leq m, n \leq N-1]$$

and the periodic extension of

$$[h(i-m, j-n)]$$

as shown:



Wraparound Convolution

- Wraparound convolution is a consequence of the **periodic DFT**.
- **Wraparound convolution** is an **artifact** of digital processing.

Linear Convolution by Zero Padding

- Performing linear convolution by wraparound convolution is a conceptually simple matter.
- It is accomplished by **padding** the two image arrays with **zero values**.
- **Generally**, both image arrays must be doubled in size:

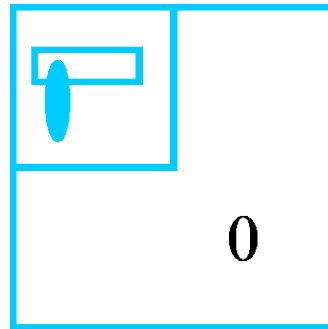


Image I_1
(zero padded)

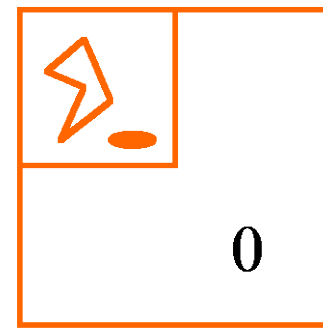
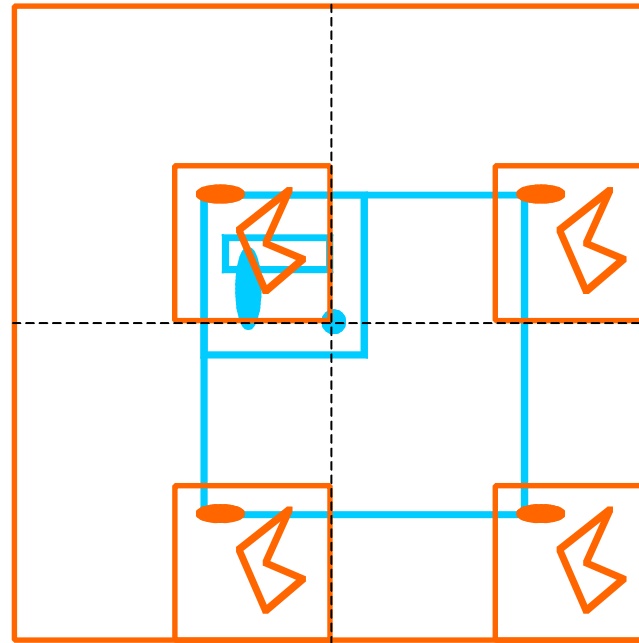


Image I_2
(zero padded)

- At the edges, no wraparound effect will occur, since the "moving" image will be weighted by zero values only outside the domain.
- This can be seen by looking at the overlaps when computing the convolution at a point (i, j) :

Linear Convolution by Zero Padding



Linear convolution by zero padding

- Remember, the summations take place only within the **blue** shaded square ($0 \leq i, j, \leq 2N-1$).
- Instead of summing over the periodic extension of the "moving image," zero values are summed with the weighted interior values.

Summary:

Steps for Filtering in the Frequency Domain

1. Given an input image $f(x,y)$ of size $M \times N$, obtain the padding parameters P and Q . Typically, $P = 2M$ and $Q = 2N$.
2. Form a padded image, $f_p(x,y)$ of size $P \times Q$ by appending the necessary number of zeros to $f(x,y)$
3. Multiply $f_p(x,y)$ by $(-1)^{x+y}$ to center its transform
4. Compute the DFT, $F(u,v)$ of the image from step 3
5. Generate a real, symmetric filter function*, $H(u,v)$, of size $P \times Q$ with center at coordinates $(P/2, Q/2)$

*generate from a given spatial filter, we pad the spatial filter, multiply the expanded array by $(-1)^{x+y}$, and compute the DFT of the result to obtain a centered $H(u,v)$.

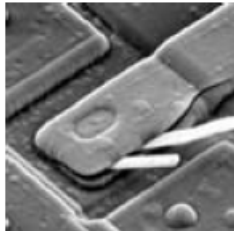
Summary:

Steps for Filtering in the Frequency Domain

6. Form the product $G(u,v) = H(u,v)F(u,v)$ using array multiplication
7. Obtain the processed image

$$g_p(x, y) = \left\{ \text{real} \left[\mathfrak{F}^{-1} \left[G(u, v) \right] \right] \right\} (-1)^{x+y}$$

8. Obtain the final processed result, $g(x,y)$, by extracting the $M \times N$ region from the top, left quadrant of $g_p(x,y)$



a	b	c
d	e	f
g	h	

FIGURE 4.36

- (a) An $M \times N$ image, f .
- (b) Padded image, f_p of size $P \times Q$.
- (c) Result of multiplying f_p by $(-1)^{x+y}$.
- (d) Spectrum of F_p .
- (e) Centered Gaussian lowpass filter, H , of size $P \times Q$.
- (f) Spectrum of the product HF_p .
- (g) g_p , the product of $(-1)^{x+y}$ and the real part of the IDFT of HF_p .
- (h) Final result, g , obtained by cropping the first M rows and N columns of g_p .

Next

- We will design filter to perform smoothing, sharpening in frequency domain.