A Survey of Low-Power Transceivers and Their Applications

Johannes Blanckenstein, Jirka Klaue, and Holger Karl

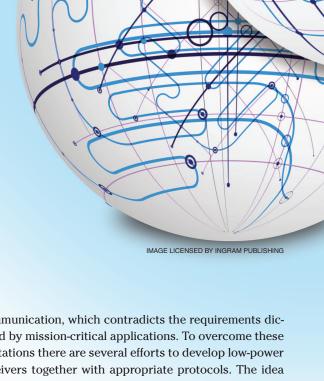
Abstract

In wireless sensor networks (WSNs) energy efficiency and communication reliability are often conflicting requirements. Additionally, some application areas such as industrial automation or infrastructure monitoring impose strict latency bounds. Low-power receivers (< 1 mW power consumption) together with adapted MAC protocols have the potential to meet these diverse requirements. We present an overview of state-of-the-art low-power receivers and relate their characteristics to requirements for different application areas. We compare low-power receivers to duty-cycled transceivers and present applications depending on them. For this, we use power consumption, sensitivity, and data rate as key performance figures for low-power receivers. Based on the characteristics of the applications we derive guidelines for using low-power receivers instead of duty-cycled transceivers.

I. Introduction

ndustrial process automation and infrastructure monitoring require dependable wireless sensor networks. These application areas demand not only energy efficiency but also reliable and timely data communication. Several Medium Access Control (MAC) protocols are developed to specifically fulfill these demands. A thorough review about such mission-critical protocols and their limitations is given in [1].

A common way to reduce the energy consumption is to turn off the transceiver periodically. Such a duty cycling can lead to increased delay and a less reliable communication, which contradicts the requirements dictated by mission-critical applications. To overcome these limitations there are several efforts to develop low-power receivers together with appropriate protocols. The idea is to have receivers with such a low power consumption that it becomes feasible to keep them turned on all the time. Depending on the target application such a receiver can fulfill a variety of functions; among others, it can be used as a "classical" receiver or as a wake-up receiver (WUR). As the name indicates, a WUR is used to wake up



Digital Object Identifier 10.1109/MCAS.2015.2450634 Date of publication: 19 August 2015 Johannes Blanckenstein and Jirka Klaue are with Airbus Group Innovations, Dept. TX4CW, 81663 Munich, Germany (e-mail: johannes.blanckenstein@airbus.com). Holger Karl is with University of Paderborn, Germany (e-mail: holger.karl@upd.de).

an otherwise powered-down node to initiate appropriate tasks. A node will thus be woken only on demand and not periodically, which can decrease communication delay and can enable more time-critical protocols.

The paper is structured as follows: Sec-

The paper is structured as follows: Section II gives an overview about different requirements for low-power receivers, which are depending on the application area. In Section III, state-of-the-art low-power receivers are presented. They are compared to each other and their benefits and limitations are highlighted. In Section IV, wake-up receivers are compared to duty-cycled

transceivers and their best field of application is defined. How to use wake-up receivers in specific applications is presented in Section V, and Section VI concludes this paper.

II. Requirements

Depending on the application area, different requirements are demanded from low-power receivers. The requirements are mostly influenced by the communication range, the channel characteristics, the message flow, and the available energy. All these requirements have a direct impact on the key performance figures for low-power receivers. Communication range and channel characteristics directly influence the necessary

THIRD QUARTER 2015

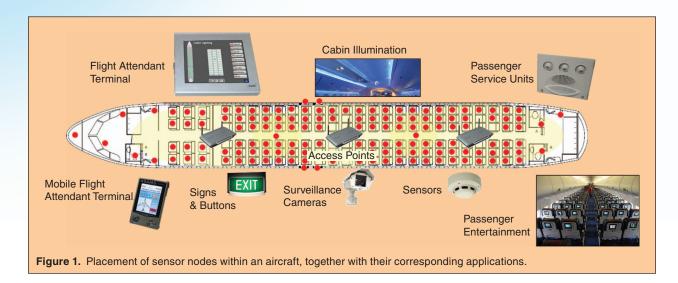
receiver *sensitivity*, and the message flow influences the necessary *data rate*. To improve either one of them, more energy has to be spent. Hence, the available energy is the limiting factor for all receiver designs; typically a very low *power consumption* is essential.

The requirements for four representative scenarios will be inspected: the aeronautical case, the wireless body area network (WBAN), smart metering, and industrial applications.

A. Aeronautical Use Case

One challenge for wireless sensor networks in the aeronautical case is the high density of wireless sensor nodes [2]. The quantity of sensor nodes within an aircraft can easily reach 1000 or more. Figure 1 shows possible locations for sensor nodes within an aircraft, together with corresponding applications.

It is challenging to consolidate all these applications and their sensor nodes within one common network. With that many nodes, a typical CSMA protocol ceases to be feasible and a more sophisticated protocol has to be used. In [2] a possible solution is presented; it uses a TDMA protocol with special reliability features. However, this protocol is developed for periodic applications and, therefore, in its present implementation not very well suited for event-triggered applications. For such cases, currently, duty-cycled transceivers are used to wake up the nodes at pre-determined points in time and then to start the protocol. For many applications, such a wake-up is only necessary once a day and therefore a very long duty cycle is used and a very low energy consumption is possible. Due to a very low duty cycle the introduced latency of the system is very high (> 10 s)while the responsiveness of the system has to be in the order of 200 ms, such a low duty cycle is not feasible.



7

To make the system autonomous, the power supply for the nodes is provided by an energy harvester [3] in combination with a power management board [4]. Such an energy harvester can provide up to 23J per flight. With two flights a day and an efficiency of 54% of the power management board [4], only 288 μ W are available on average. When the nodes are woken up, they have to start a measurement and transmit these values to a common sink. Measuring can consume a lot of energy, therefore, the available average power is further reduced to around 100 μ W. Additionally, channel measurements [2, 5–8] suggest a minimum required sensitivity of -80 dBm to support a communication range of 10 m.

Hence, the requirements for a low-power receiver in the aeronautical case are an average power consumption below $100\,\mu\text{W}$, a sensitivity better than $-80\,\text{dBm}$, and a delay below $200\,\text{ms}$.

B. Wireless Body Area Network (WBAN)

A WBAN consists of devices that are all located in close vicinity of the human body. Typical applications for WBANs are in health care or in the multimedia sector. Latre' et al. [9] give an overview of WBANs in general and the requirements resulting from their applications. As stated there, a WBAN will consist of around 20–50 sensor nodes, which are used to monitor bodily functions like heartbeat or body temperature, as well as to connect multimedia devices like microphones or headmounted displays. Those devices can have very different needs for data rates in the region of only a few kb/s; to several hundreds of kb/s; in sum the demands can reach a few Mb/s.

The advantage of low-power receivers in WBANs can be twofold: either a low-power receiver is used for low data rate applications and will replace a "normal" receiver, or it is used as a wake-up receiver to initiate a possible high data rate communication. Either way, the power consumption needs to be very limited. For example, some health care devices are implemented inside the human body, like insulin reservoirs or pacemakers. It is clear that with such devices a regular battery change is no option. Additionally, because the devices are implemented inside the body, there is a strict size constraint for the nodes, e.g., it is not possible to use large batteries. Therefore, the power consumption has to be as low as possible.

The short communication distance of about 1 m that the devices have to cover can help to keep the power consumption low. In [10] a large measurement campaign was conducted to model the behavior of WBAN channels. A stochastic channel model was derived from it where each link is considered separately. The highest path loss was found around 50 dB, with all radio links

on the front side of the body. With a path loss in the region of $40-50~\mathrm{dB}$ the sensitivity requirements for low-power receivers are lessened; a sensitivity of $-40~\mathrm{dBm}$ would be enough to receive a signal transmitted with 0 dBm. Without the need to have a high sensitivity even more energy-efficient designs can be implemented.

C. Smart Metering

Another field of application for low-power receivers is in smart metering and home automation. In both cases, radio communication takes place inside a building and the requirements for both are similar. The maximum communication distance is typically about 15 m, but it has to be differentiated if the communication takes place on the same floor or across multiple floors. At a distance of 15 m for both cases the path loss is in the same order of magnitude but at shorter distances the multiple-floor scenario has a larger path loss than the same-floor scenario. In [11] a channel is characterized for both cases. A path loss of around 100 dB was measured at a distance of 15 m. At 10 m a path loss of 100 dB was measured for the multiple-floor scenario and a path loss of 80 dB for the same-floor scenario. A similar behavior is described in [12] for the same-floor scenario.

Therefore, a low-power receiver for these applications has to have a high sensitivity. With a maximum allowed effective radiated power (ERP) of 25 dBm in the European 868 MHz band, a receiver should have at least a sensitivity of -75 dBm to be able to receive packets at a distance of 15 m. But commercially available transceivers normally have an output power around 0 dBm. To generate an output power of 25 dBm typically a discrete power amplifier has to be used. Assuming a high efficiency of 50%, additional $28 \, \mathrm{dBm} (= 631 \, \mathrm{mW})$ has to be spent to reach these high output power levels. Since a very high output power is rarely possible for low-power devices, a higher sensitivity should be used.

The required data rates for home automation/smart metering applications are moderate but the acceptable delay for a communication might be more demanding. If human interaction is involved, a latency larger than 0.5 s might not be tolerable. The allowed maximum energy consumption for the smart metering/home automation area depends heavily on the specific application. For some devices, like heat or electric meters, the battery has to provide enough energy for around 10 years, and because of that, large and expensive batteries have to be provided. Reducing the energy consumption of the communication system can reduce the overall costs immensely.

D. Industrial Applications

Control loops have the strictest requirements in the industrial applications field. Not only are they loss-intolerant

IEEE CIRCUITS AND SYSTEMS MAGAZINE THIRD QUARTER 2015

but also delay-intolerant. For control loops, a maximum allowed delay in the order of 25 ms is common.

The wireless channel for industrial applications is similar to the one for the smart metering/home automation applications. The big difference is the target communication distance of 100 m. In [13] a characterization for the industrial indoor channel is done. There, three types of topographies are distinguished: line-of-sight (LOS), obstructed line-of-sight (OBS) with light surrounding clutter, and obstructed line-of-sight with heavy surrounding clutter. For these three topologies channel measurements were done in three different bands, at 900 MHz, at 2.4 GHz, and at 5.2 GHz. From these measurements 9 different channel characteristics were derived, three for each frequency band and three for each topology, respectively.

In [13], it was argued that the log-distance path loss model is applicable for each characteristic and the corresponding path-loss exponents n and the path loss values $PL(d_0)$ at a reference distance $d_0=15\,\mathrm{m}$ were derived. When comparing the path loss at a distance of 100 m for the three topologies at 900 MHz, it was verified that the path loss behaved as assumed. The smallest path loss was observed in the LOS case with 98 dB, followed by the OBS with light surrounding clutter with 100 dB. The biggest path loss was observed in the case of OBS with heavy surrounding clutter with 110 dB. For the 2.4 GHz band the behavior was similar with 100 dB, 101 dB, and 112 dB, respectively. Assuming again a maximum transmit power of 25 dBm, a minimum sensitivity of $-85\,\mathrm{dBm}$ is necessary.

In industrial applications bigger devices and therefore bigger batteries might be possible, but the total cost of the system has to be kept as small as possible. A receiver that fulfills the sensitivity and delay requirements while fulfilling the strict energy consumption demands can help to lower the overall system and operation cost.

E. Summary

Depending on the application scenarios the requirements for low-power receivers differ. For some scenarios a very high sensitivity is necessary whereas in other scenarios a high sensitivity is not essential. As will be shown in Section III, power consumption, sensitivity, and data rate are competing features and can be traded against each other; depending on the scenario these values have to be carefully chosen.

In Table 1 estimates for the system requirements for low-power receivers are given. These values are reference points for the system development and shall help to categorize low-power receivers; they are not meant to be strict, fixed limits for the system requirements. As can be seen, the highest requirement for sensitivity is given in the industrial case, closely followed by the aeronautical and smart metering applications. Only WBAN applications have a moderate sensitivity requirement owing to the short communication distance. The delay and data rate requirements are closely coupled; a very low data rate leads to a high communication delay which can be challenging. In some receiver concepts a data rate of only several hundred bits per seconds is possible; for instance, if 100 bytes have to be transmitted at a data rate of 1 kb/s this would lead to a minimum delay of 800 ms. Hence, the data rate of the low-power receivers have a direct impact on the possible minimum communication delay. The industrial case has the most stringent delay requirements. Regarding the data rate, WBAN applications have the highest demand because of possible multimedia applications and also the strictest requirements for power consumption.

Generally, industrial applications have the highest sensitivity and delay demands while still requiring a very low power consumption. Aeronautical applications have slightly lesser sensitivity demands but at the same time stricter power consumption demands. Smart metering applications have lessened overall system requirements. Lastly, for WBAN applications a very low power consumption is clearly the most important factor, which can be met by providing only a moderate sensitivity.

III. Low-Power Receiver

Several concepts for low-power receivers are available [14–58], differing in complexity and performance.

Table 1. Requirements for low-power receivers.									
	Communication Distance	Sensitivity	Power Consumption	Data Rate	Delay	Payload Size			
Aeronautical case	10 m	-80 dBm	100 μW	250 kb/s	200 ms	10 to 100 byte			
WBAN	1 m	$-40\mathrm{dBm}$	10 μW	1000 kb/s	1000 ms	5 to 4000 byte			
Smart metering	15 m	$-75\mathrm{dBm}$	100 μW	10 kb/s	500 ms	10 to 100 byte			
Industrial	100 m	$-85\mathrm{dBm}$	$200\mu W$	250 kb/s	25 ms	10 to 200 byte			

Generally, industrial applications have the highest sensitivity and delay demands while still requiring a very low power consumption. Aeronautical applications have slightly lesser sensitivity demands but at the same time stricter power consumption demands.

Smart metering applications have lessened overall system requirements.

A. Technology

The technologies used for low-power receivers vary over a wide range; from simple passive energy detection to envelope detection with several correlation stages. Each technology has its own advantages and disadvantages. Depending on the requirements for the low-power receivers, the best technology has to be selected carefully. Some are better for high data rates, others for high sensitivity or very low power consumption.

1) Modulation

The first distinctive feature is the chosen modulation format. Most of the concepts use either amplitude modulation in its binary form On-Off-Keying (OOK), or they use Frequency-Shift-Keying (FSK). Only four of the presented concepts differ here. The concept presented by Le-Huy et al. [50] uses binary Pulse-Width-Modulation (PWM), while the concept by Marinkovic et al. [28] uses a multi-stage approach beginning with an OOK modulated signal followed by a PWM symbol. The other concepts use Phase-Shift-Keying (PSK); Yan et al. [57] use the binary form BPSK and Lee et al. [51] use the quaternary form Quadrature-Phase-Shift-Keying (QPSK), which could also be implemented as two parallel BPSK demodulators.

The advantage of FSK (and also PSK) over OOK is that it is less susceptible to noise and fading. Additionally,

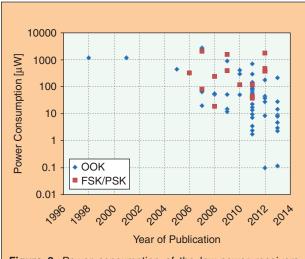


Figure 2. Power consumption of the low-power receivers using the OOK [14–47] and FSK/PSK [48–58] concepts.

because of the constant power level of the carrier, the amplifier design is simpler with FSK and there is no need for an adaptable threshold to decide which symbol was received. OOK has the advantage of overall implementation simplicity which can be transferred into energy efficiency. For instance, just a simple envelope detector can be used to detect if the carrier frequency is turned on. Such a detector can easily be implemented by using a diode and a resistor-capacitor oscillator circuit as a bandpass filter. All the concepts that have a power consumption below $10\,\mu\text{W}$ are using OOK. This can be seen in Figure 2.

2) Implementation

The second distinctive feature is the implementation of the concept. Several variants are used: super-regenerative receivers, superheterodyne receivers, injection-locked local oscillators (LO), envelope detectors with or without an intermediate frequency (IF), and in two cases envelope detectors followed by purely analog correlation stages.

The concepts in [14, 15, 20, 27, 32, 36, 43, 44, 47, 48] are using a super-regenerative receiver design. They all have a data rate above 100 kb/s with a sensitivity varying over a broad range. Two designs [22, 35] are using envelope detectors followed by purely analog non-coherent correlators and, thus, are able to reach very low power consumption values. Five concepts are using duty cycling [18, 25, 29, 30, 46], where [18, 29, 30, 46] are duty cycling even within a bit having an ON time in the range of only 100 ns. This reduction in sampling time decreases the power consumption to a level of several µW at a low data rate ($< 1 \,\mathrm{kb/s}$) without increasing the latency due to duty cycling. The concepts in [45, 53, 57, 58] are using injection-locked oscillators for either creating a power efficient "virtual LO" or creating a power efficient FSK/PSK demodulator. The concepts in [18, 21, 31, 33,] are using an uncertain IF architecture where the LO can vary in the range of 100 MHz from its desired frequency and therefore can be implemented more power-efficiently. In [25, 41, 52, 57] no IF is used, the symbols are demodulated directly which again can simplify the receiver design immensely. A very power-efficient design is possible but at the cost of a low sensitivity and the need of an external surface acoustic wave (SAW) filter.

Regardless of the particular specifications of the designs, the commonly important characteristics of all

low-power receivers are *power consumption*, *data rate*, and *sensitivity*. None of these technologies can clearly be identified to outperform the others in all of these characteristics. All parts of a receiver strongly influence its performance; the interaction between them accounts for the overall power consumption. Therefore, all the low-power receivers will be compared with each other regardless of the specific technology.

B. Comparison

1) Sensitivity vs. Power Consumption

Firstly, the power consumption of the low-power receivers is related to their sensitivity. If possible, the sensitivity is measured at a bit error rate (BER) of 10^{-3} ; due to correlation stages, for some concepts the sensitivity can only be measured at packet error rate (PER) levels because their functional principle makes no information available at BER level. This has to be kept in mind when comparing the sensitivity levels. The supply voltages vary in the range of 0.4V-3V. Some concepts use multiple supply voltage levels, meaning the actual power consumption might be higher due to the necessity of voltage converters when implementing the concepts on a communication platform.

The behavior of some concepts is adjustable, therefore, these receivers will have several connected data points in the figures. The data points of some concepts—[29, 30] in Figure 3 and [24, 53] in Figure 4—describe a vertical line, suggesting that one receiver setting outperforms all the others. However, this advantage is only valid for the examined parameters. As can be seen in the corresponding figure, there is no receiver setting

which outperforms all the others. Nevertheless all data points are drawn to indicate that several receiver settings are feasible, even if they draw a vertical line.

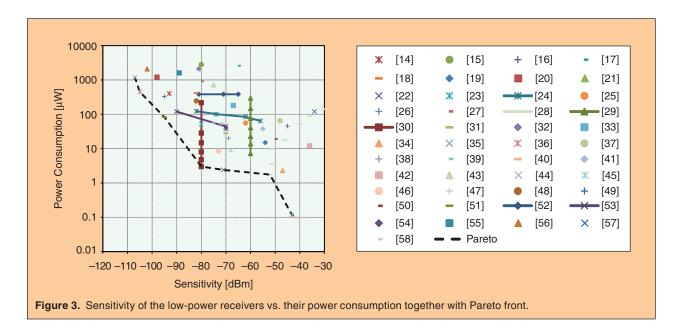
As can be seen in Figure 3, it is very hard to reach a power consumption smaller than $10\,\mu\text{W}$. Only nine concepts [22, 28–30, 34, 35, 40, 45, 46] are below that level, where [35, 40] reach an outstanding power consumption around $100\,\text{nW}$. All those concepts can meet the moderate sensitivity demands for the WBAN applications.

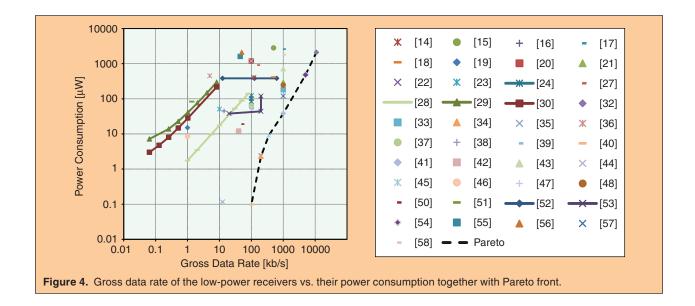
2) Data Rate vs. Power Consumption

Secondly, the data rate of the receivers is related to their power consumption. Depending on the concepts, different coding techniques are used. Some designs [19, 34, 42] use Manchester coding to be able to recover the clock signal from the data. With Manchester coding the net data rate is only half of the gross data rate. Other designs [16, 22, 28–30, 34, 35, 37, 43, 50] use spreading techniques to increase the sensitivity or to distinguish between codes. Here, the net data rate is in the order of the spreading factor smaller than the gross data rate. In the figures, gross data rates before spreading, correlation, coding, etc. are shown.

As can be seen, it is hard to reach high data rates while keeping a low power consumption. Only ten designs [17, 21, 32, 33, 41, 43, 48, 54, 57, 58] reach a data rate of at least $1\,\text{Mb/s}$. From those only three [33, 41, 57] have a power consumption smaller than $200\,\mu\text{W}$.

Regarding the requirements in Table 1, it can be seen that for the aeronautical and the WBAN case no low-power receiver concept fulfills the power consumption, sensitivity, and data rate requirements. For the industrial case only [53] and for the smart metering case only





[23, 47] fulfill all three requirements at the same time. The characteristics of these concepts would be sufficient to fulfill the sensitivity and data rate demands of those applications while reducing the power consumption dramatically.

For the aeronautical and for the WBAN scenarios currently no low-power receiver concept fulfills all the requirements to replace the main receiver of the system. However, a low-power receiver implemented as a wake-up receiver can have advantages over duty-cycling a main transceiver: either the overall power consumption and the communication delay can be reduced or at least the communication delay can be reduced while maintaining the same power consumption a duty-cycled transceiver would generate. The following section addresses this aspect more thoroughly.

IV. Wake-Up Receiver vs. Duty-Cycled Receiver

A common way to reduce the power consumption of a wireless node is to turn off all devices that are currently not used and to duty-cycle its transceiver. With a decreasing duty cycle the power consumption in the receiver is decreased, but, at the same time, the mean communication delay is increased.

In Figure 5 the periodic behavior of a duty-cycled receiver is depicted. The duty cycle follows $(T_{\rm on}/T_p)\cdot 100\%$, where $T_{\rm on}$ is the time the transceiver is turned on and T_p is the duration of the period. In this case the receiver has a duty cycle of 25%. The receiver is only able to receive a packet correctly if the transmit duration $T_d \leq T_{\rm on}$ and the packet is transmitted during the ON time of the receiver, which means that only the third packet in Figure 5 can be received correctly. It is only possible to communicate with a duty-cycled node

12

in certain time frames; the period P equals the maximum additional communication delay before this frame arrives. Thus, there is a trade-off between power consumption and communication delay.

Additionally, when duty cycling, an overall more complex communication protocol is needed. To keep the communication delay as low as possible, the transmitting node either has to know beforehand at which time its communication partner is listening or it has to transmit the packet several times until it is certain that a listening slot of the receiving node is met. In the first case, the synchronous schedule, the packet has to be sent only once, but the clocks of all nodes have to be kept synchronous. In order to do this, additional communication has to be done which complicates the communication protocol and might result in higher power consumption for the transmitting node. In the second case, the asynchronous schedule, clocks do not have to be kept synchronous. In order to keep a low communication delay the packets have to be transmitted several times, which also increases the power consumption for the transmitting node. Hence, duty cycling not only increases the communication delay but might increase the power consumption for the transmitting node as well.

Following formula can be used to estimate the mean power consumption for a duty-cycled receiver.

$$P_{\text{mean}} = \frac{P_{\text{switch}} \cdot T_{\text{switch}} + P_{\text{on}} \cdot T_{\text{on}} + P_{\text{off}} \cdot T_{\text{off}}}{T_{p}} \tag{1}$$

When calculating the power consumption, the switching time $T_{\rm switch}$ from the OFF state to the ON state is not negligible. According to Equation (1) the mean power consumption for two commercially available state-of-the-art transceivers is calculated for varying duty cycles.

IEEE CIRCUITS AND SYSTEMS MAGAZINE THIRD QUARTER 2015

The first transceiver considered is the Atmel AT86RF212 [59], which implements IEEE802.15.4 for the 700/800/900 MHz band. The second transceiver considered is the Nordic Semiconductor nRF51822 [60], which implements Bluetooth LE (2.4 GHz band). The second device is a System-on-Chip, which comprises a transceiver and a microcontroller but only the transceiver figures are taken into account. Both devices implement very different physical layers but both are optimized for low-power communication. For IEEE802.15.4 the 20 kb/s and 40 kb/s BPSK modulation as well as the 250 kb/s

OFF Receiver ON ON ON Time Ton-Period T_p Unsuccsessful Unsuccsessful Succsessful Transmitter Transmission Transmission Transmission Transmit Duration T_d Figure 5. A 25% duty-cycled receiver, together with a transmitter that is sending three

packets until a successful transmission occurs.

O-OPSK modulation is considered. For the nRF51822 the 1 Mb/s Bluetooth LE as well as the proprietary 250 kb/s and 2 Mb/s GFSK modulation is considered.

In Table 2 the characteristics for the different operation modes are shown. Regardless of the underlying physical layer, the sensitivity decreases with an increasing data rate while maintaining roughly the same power consumption. Within the duration $T_{\rm on}$ it has to be possible to transmit at least a packet header with address information. Because of the different protocols this leads to a different quantity of necessary bits; 144 bits for IEEE802.15.4 and 80 bits for Bluetooth LE. Hence, the packet length and therefore the minimum ON time $T_{\rm on}$ is different for each protocol.

Using these characteristics from Table 2, in Figure 6 the mean power consumption of the commercially available transceivers are depicted together with wake-up receivers with a power consumption below $200 \,\mu\text{W}$. The delay in the case of WURs is equal to the time it takes them to send a wake-up message. For designs [16, 22, 28–30, 34, 35, 37, 50] that use spreading techniques with fixed spreading factors, the amount of bits to send is predefined. For other designs,

we assume that a wake-up message consists of 32 bits with the delay being directly proportional to their data rates.

With an increasing duty cycle period—while keeping the duration $T_{\rm on}$ constant—the mean power consumption for both commercially available radios decreases. Using a higher data rate decreases the sensitivity of the devices but does not change the power consumption. Hence, a shorter necessary ON time reduces the mean power consumption for a fixed duty-cycle

period. Until a certain limit—a period of about 10s for the Bluetooth LE radio and around 100s for the IEEE802.15.4 radio—the mean power consumption curves decrease linearly in the log-log scale. After the limit, the basic load of the devices becomes predominant, the mean power consumption values approach the device-specific horizontal asymptote.

A similar predominance effect can be observed when increasing the data rate. The necessary ON time will be reduced accordingly, but the switching time T_{switch} does not change. Hence, the switching time will become the predominant factor when a certain data rate, respectively necessary ON time is reached. Further increasing the data rate will not significantly decrease the mean power consumption.

Hence, duty cycling a radio decreases its mean power consumption but only until a data rate dependent threshold is reached. Then the mean power consumption approaches an horizontal asymptote.

As can be seen, all WUR concepts perform better than the commercially available transceivers regarding delay and power consumption. All of the concepts fulfill

Table 2. Characteristics for commercially available transceivers.								
	Data Rate	Sensitivity	P on	P off				
AT86RF212	20 kb/s	-110 dBm	30.36 mW	0.66 μW				
	40 kb/s	$-108\mathrm{dBm}$	30.36 mW	0.66 μW				
	250 kb/s	-101 dBm	30.36 mW	0.66 μW				
nRF51822	250 kb/s	$-96\mathrm{dBm}$	41.58 mW	1.98 μW				
	1 Mb/s	$-91\mathrm{dBm}$	42.90 mW	1.98 μW				
	2 Mb/s	- 85 dBm	44.22 mW	1.98 μW				

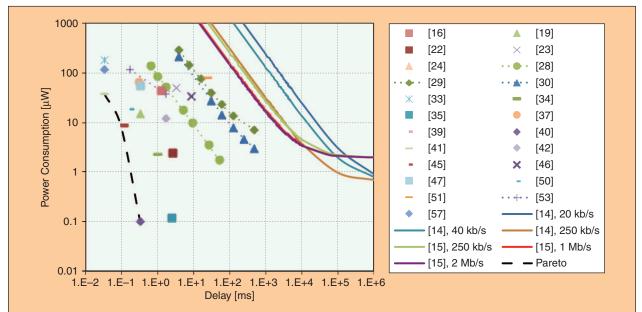


Figure 6. Wake-up delay of low-power receivers [16, 19, 22–24, 28–30, 33–35, 37, 39–42, 45–47, 50, 51, 53, 57] with a power consumption below $200\,\mu\text{W}$ together with duty-cycled transceivers Atmel AT86RF212 [59] and Nordic Semiconductor nRF51822 [60]; Pareto front included.

the delay requirements given in Table 1. Considering the sensitivity requirements as well, only a subset remains. For the aeronautical case only three concepts [23, 30, 51] fulfill the requirements, for the industrial only one [51]. A duty-cycled nRF51822 with a data rate of one or two Mb/s also meets the industrial requirements, but only with a sensitivity worse than the concept in [51].

For an application with a delay restriction of 1s, a duty-cycled transceiver is an option. The mean power consumption is in the order of $10\,\mu\text{W}$ to $100\,\mu\text{W}$ with a sensitivity of $-90\,\text{dBm}$ to $-110\,\text{dBm}$. This increase in sensitivity can improve the reliability of the communication system which outweighs the additional effort for the more complex communication protocol.

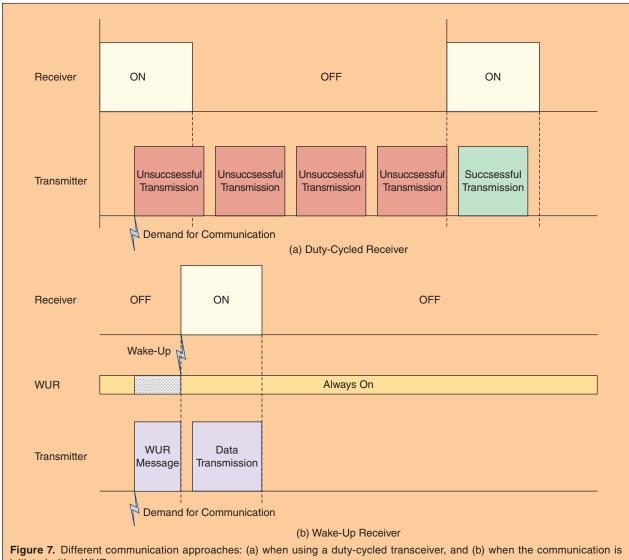
V. Using Wake-up Receivers

The most promising application for low-power receivers is to replace a "normal" receiver. If used for all communication purposes, the low-power design can yield the most energy savings. However, as outlined in Section III-B, low-power receivers do not meet all requirements for all applications. Another operational area is to use a low-power receiver as a wake-up receiver. The low-power receiver will not be used for communication itself but for initiating the communication via another receiver. Since the main transceiver does not have to be periodically duty-cycled, it can be kept turned off as long as no communication is required; it will be turned on only on demand.

This behavior can be seen in Figure 7. In the first case, the unsynchronized communication with a duty-cycled receiver is depicted in its simplest form. Once a communication demand arises, the transmitter wants to start communicating and repeatedly sends the data packet. This repetition must continue for the length of the duty cycle period to assure the meeting with the ON phase of the duty-cycled receiver. In the second case, instead of sending the data packet directly, first a wake-up message is transmitted. This message indicates the receiving node that a data packet will follow and therefore it will turn on its receiver. In this case, using a wake-up receiver will not only decrease the idle listening time but also shorten the mean communication delay.

This is beneficial in a typical aeronautic application: on the ground, the status of simple switches has to be evaluated. To accomplish this the monitored switch can be connected to a sensor board, which currently is using a duty-cycled transceiver. With this system, the status of one switch can be polled within one second while only consuming $<100\,\mu\text{W}$ on average. Using a wake-up receiver instead, the communication delay could be decreased by approximately one magnitude at the same power consumption. This increases the responsiveness of the system immensely; it is possible to inquire the status of several switches in the same time as only one with a duty-cycled transceiver.

Polling devices is not the best solution for a large number of devices. Gordon et. al [61] present a more



initiated with a WUR message.

sophisticated protocol where acknowledged wake-up messages are used to initiate, for example, a TDMA protocol. In doing so, the time to inquire the status of a large amount of sensors can be reduced roughly by half. A similar approach is pursued in [62], where many temperature values have to be monitored. To reduce the power consumption, sensor nodes measuring similar readings are clustered together. Only one of them is representing the cluster and is transmitting the actual reading. Wake-up receivers are used to establish the clusters in an energy efficient way and to inform about changes.

VI. Future Research and Conclusion

The development over the last 10 years showed a clear improvement of the performance of low-power receivers. The power consumption dropped from 1 mW to

almost only 100 nW, hence, a factor of 10000. Still there are no designs available which fulfill the combined requirements for energy consumption, sensitivity, and data rate for the aeronautical and for the WBAN applications. If the performance improvement continues to proceed, concepts for all application areas should be available soon.

Currently there are few applications which depend on the usage of a low-power receiver/wake-up receiver. Many applications could clearly benefit from using a wake-up receiver; either because of a reduced power consumption or because of a smaller communication delay. Some applications will only be enabled when using a low-power receiver. With the high potential for reducing the energy consumption, protocols that are dependent on wake-up receivers are expected to arise.

Depending on specific application requirements, it is already possible to replace a "normal" receiver with a low-power receiver. If the low-power receiver characteristics do not yet meet the requirements for the applications, they can be used as wake-up receivers, which, by keeping similar power consumption values, drastically reduces the communication delay in comparison with duty-cycled systems.

Acknowledgement

The authors would like to acknowledge the funding by the German Federal Ministry of Education and Research project "AETERNITAS".



Johannes Blanckenstein works on communication protocols in the general field of wireless sensor networks and especially on protocols and applications for low-power wake-up receivers. He writes his doctoral thesis at University Pader-

born in the computer networks group of Holger Karl. At the same time he works at Airbus Group Innovations in the field of WSN applications for aeronautics.



Jirka Klaue is a researcher at Airbus Group Innovations in the field of wireless communications. He received his diploma in computer science from the Technical University of Berlin in 1999. Afterwards he worked on wireless

video transmission at the Telecommunication Networks Group and DResearch. He is co-founder of a software company for web-based database applications. At Airbus he participated in research projects founded by the EU, BMBF, BMWi and Airbus. Currently he works on reliable and time-critical wireless sensor networks.



Holger Karl is a Full Professor of Computer Science at the University of Paderborn, Germany where he leads the Computer Networks group. Holger Karl received his MS and Ph.D degrees in 1996 and 1999, respectively. Later he

continued his research at the Technical University of Berlin before joining the University of Paderborn in 2004. He is coauthor of Protocols and Architectures for Wireless Sensor Networks (Wiley). His current research interests are architectural questions for future mobile communication systems and the future Internet concentrating on techniques like software-defined networking and network function virtualization, cross-layer optimization for mobile systems including the wireless/wire boundary, and wireless industrial automation.

References

- [1] P. Suriyachai, U. Roedig, and A. Scott, "A survey of MAC protocols for mission-critical applications in wireless sensor networks," *IEEE Commun. Surveys Tutorials*, vol. 14, no. 2, pp. 240–264, 2012.
- [2] J. Blanckenstein, J. Garcia-J imenez, and J. Klaue, "A scalable redundant TDMA protocol for high-density WSNs inside an aircraft," in *Real-World Wireless Sensor Networks* (Lecture Notes in Electrical Engineering, vol. 281), K. Langendoen, W. Hu, F. Ferrari, M. Zimmerling, and L. Mottola, Eds. New York: Springer, 2014, pp. 165–177.
- [3] A. Elefsiniotis, D. Samson, T. Becker, and U. Schmid, "Investigation of the performance of thermoelectric energy harvesters under real flight conditions," *J. Electron. Mater.*, vol. 42, no. 7, pp. 2301–2305, 2013.
- [4] M. Kluge, D. Samson, T. Becker, A. Gavrikov, B. Bennemann, and J. Nurnus, "Efficient power management for energy aware, self-sufficient wireless sensors in aeronautic applications," in *Proc. 10th Int. Workshop on Micro and Nanotechnology for Power Generation and Energy Conversion Applications (PowerMEMS)*, Leuven, Belgium, Dec. 2010, pp. 2–3.
- [5] N. Moraitis and P. Constantinou, "Radio channel measurements and characterization inside aircrafts for in-cabin wireless networks," in *Proc. IEEE 68th Vehicular Technology Conf.*, Calgary, Canada, Sept. 2008, pp. 1–5.
- [6] G. Hankins, L. Vahala, and J. H. Beggs, "Propagation prediction inside a B767 in the 2.4 GHz and 5 GHz radio bands," in *Proc. IEEE Antennas and Propagation Society Int. Symp.*, Washington, D.C., July 2005, vol. 1A, pp. 791–794.
- [7] G. Hankins, L. Vahala, and J. H. Beggs, "802.11ab propagation prediction inside a B777," in *Proc. 2005 IEEE/ACES Int. Conf. Wireless Communications and Applied Computational Electromagnetics*, Honolulu, HI, Apr. 2005, pp. 837–840. [8] R. D'Errico and L. Rudant, "UHF radio channel characterization for Wireless Sensor Networks within an aircraft," in *Proc. 5th European Conf. Antennas and Propagation (EUCAP)*, Rome, Italy, Apr. 2011, pp. 115–119.
- [9] B. Latré, B. Braem, I. Moerman, C. Blondia, and P. Demeester, "A survey on wireless body area networks," in *Wireless Networks*. New York: Springer, 2010, vol. 17, no. 1, pp. 1–18.
- [10] S. van Roy, F. Quitin, L. Liu, C. Oestges, F. Horlin, J.-M. Dricot, and P. De Doncker, "Dynamic channel modeling for multi-sensor body area networks," *IEEE Trans. Antennas Propag.*, vol. 61, no. 4, pp. 2200–2208, 2013.
- [11] T. Chrysikos, G. Georgopoulos, and S. Kotsopoulos, "Wireless channel characterization for a home indoor propagation topology at 2.4 GHz," in *Proc. Wireless Telecommunications Symp. (WTS)*, New York, Apr. 2011, pp. 1–10.
- [12] H. Tsuchiya, "Characterization of the radio propagation channel in residential environment for smart meter communications," in *Asia-Pacific Microwave Conf. Proc. (APMC)*, Melbourne, Australia, Dec. 2011, pp. 713–716. [13] E. Tanghe, W. Joseph, L. Verloock, L. Martens, H. Capoen, K. Herwegen, and W. Vantomme, "The industrial indoor channel: large-scale and tempo-
- and W. Vantomme, "The industrial indoor channel: large-scale and temporal fading at 900, 2400, and 5200 MHz," *IEEE Trans. Wireless Commun.*, vol. 7, no. 7, pp. 2740–2751, 2008.
- [14] J. L. Bohorquez, A. P. Chandrakasan, and J. L. Dawson, "A 350 μ W CMOS MSK transmitter and 400 μ W OOK super-regenerative receiver for medical implant communications," *IEEE J. Solid-State Circuits*, vol. 44, no. 4, pp. 1248–1259, 2009.
- [15] J.-Y. Chen, M. P. Flynn, and J. P. Hayes, "A fully integrated auto-calibrated super-regenerative receiver in 0.13- μ CMOS," *IEEE J. Solid-State Circuits*, vol. 42, no. 9, pp. 1976–1985, 2007.
- [16] J. Choi, K. Lee, S.-O. Yun, S.-G. Lee, and J. Ko, "An interference-aware 5.8 GHz wake-up radio for ETCS," in *Proc. IEEE Int. Solid-State Circuits Conf.*, San Francisco, CA, Dec. 2012, pp. 446–448.
- [17] D. C. Daly and A. P. Chandrakasan, "An energy-efficient OOK transceiver for wireless sensor networks," *IEEE J. Solid-State Circuits*, vol. 42, no. 5, pp. 1003–1011, 2007.
- [18] S. Drago, D. M. W. Leenaerts, F. Sebastiano, L. J. Breems, K. A. A. Makinwa, and B. Nauta, "4 2.4 GHz 830 pJ/bit duty-cycled wake-up receiver with -82 dBm sensitivity for crystal-less wireless sensor nodes," in *Proc. IEEE Int. Solid-State Circuits Conf. (ISSCC)*, San Francisco, CA, Feb. 2010, pp. 224–225
- [19] M. S. Durante and S. Mahlknecht, "An Ultra Low Power Wakeup Receiver for Wireless Sensor Nodes," in *Proc. 3rd Int. Conf. Sensor Technologies and Applications*, Athens, Greece, June 2009, pp. 167–170.
- [20] P. Favre, N. Joehl, A. Vouilloz, P. Deval, C. Dehollain, and M. J. Declercq, "A 2-V 600-µA 1-GHz BiCMOS super-regenerative receiver for ISM applications," *IEEE J. Solid-State Circuits*, vol. 33, no. 12, pp. 2186–2196, 1998.
- [21] S. Gambini, J. Crossley, E. Alon, and J. M. Rabaey, "A fully integrated, 300 pJ/bit, dual mode wireless transceiver for cm-range interconnects," in *Proc. Symp. VLSI Circuits*. Honolulu, HI, June 2010, pp. 31–32.

IEEE CIRCUITS AND SYSTEMS MAGAZINE THIRD QUARTER 2015

- [22] C. Hambeck, S. Mahlknecht, and T. Herndl, "A 2.4 µW wake-up receiver for wireless sensor nodes with -71 dBm sensitivity," in *Proc. IEEE Int. Symp. Circuits and Systems (ISCAS)*, Rio de Janeiro, Brazil, May 2011, pp. 534–537. [23] X. Huang, S. Rampu, X. Wang, G. Dolmans, and H. de Groot, "A 2.4 GHz/915 MHz 51 µW wake-up receiver with offset and noise suppression," in *Proc. IEEE Int. Solid-State Circuits Conf. (ISSCC)*, San Francisco, CA, Feb. 2010, pp. 222–223.
- [24] X. Huang, P. Harpe, G. Dolmans, and H. de Groot, "A 915 MHz ultra-low power wake-up receiver with scalable performance and power consumption," in *Proc. European Solid-State Circuits Conf. (ESSCIRC)*. Helsinki, Finland, Sept. 2011, pp. 543–546.
- [25] J. H. Jang, D. F. Berdy, J. Lee, D. Peroulis, and B. Jung, "A wireless sensor node for condition monitoring powered by a vibration energy harvester," in *Proc. IEEE Custom Integrated Circuits Conf. (CICC)*, San Jose, CA, Sept. 2011, pp. 1–4.
- [26] P. Kolinko and L. E. Larson, "Passive RF receiver design for wireless sensor networks," in *Proc. Int. Microwave Symp. (IEEE/MTT-S)*, Honolulu, HI, June 2007, pp. 567–570.
- [27] Y.-H. Liu, H.-H. Liu, and T.-H. Lin, "A super-regenerative ASK receiver with $\Delta\Sigma$ pulse-width digitizer and SAR-based fast frequency calibration for MICS applications," in *Proc. Symp. VLSI Circuits*, Kyoto, Japan, June 2009, pp. 38–39. [28] S. J. Marinkovic and E. M. Popovici, "Nano-power wireless wake-up receiver with serial peripheral interface," *IEEE J. Sel. Areas Commun.*, vol. 29, no. 8, pp. 1641–1647, 2011.
- [29] H. Milosiu, F. Oehler, and M. Eppel, "Sub-10 µA data reception with low latency using a 180-nm CMOS wake-up receiver at 868 MHz," in *Proc. Semiconductor Conf.*, Dresden, Germany, Sept. 2011, pp. 1–4.
- [30] H. Milosiu, F. Oehler, M. Eppel, D. Fruhsorger, S. Lensing, G. Popken, and T. Thones, "A 3-µW 868-MHz wake-up receiver with -83 dBm sensitivity and scalable data rate," in *Proc. European Solid-State Circuits Conf. (ES-SCIRC)*, Bucharest, Romania, Sept. 2013, pp. 387–390.
- [31] S. Moazzeni, G. E. R. Cowan, and M. Sawan, "A 28 $\,\mu$ W sub-sampling based wake-up receiver with -70 dBm sensitivity for 915 MHz ISM band applications," in *Proc. IEEE Int. Symp. Circuits and Systems*, Seoul, Korea, May 2012, pp. 2797–2800.
- [32] F. X. Moncunill-Geniz, P. Pala-Schonwalder, C. Dehollaini, N. Joehl, and M. Declercq, "An 11-Mb/s 2.1-mW synchronous superregenerative receiver at 2.4 GHz," *IEEE Trans. Microwave Theory Tech.*, vol. 55, no. 6, pp. 1355–1362, 2007.
- [33] P. M. Nadeau, A. Paidimarri, P. P. Mercier, and A. P. Chandrakasan, "Multichannel 180 pJ/b 2.4 GHz FBAR-based receiver," in *Proc. IEEE Radio Frequency Integrated Circuits Symp.*, Montreal, Canada, June 2012, pp. 381–384.
- [34] E. Nilsson and C. Svensson, "Ultra low power wake-up radio using envelope detector and transmission line voltage transformer," *IEEE J. Emerging Sel. Top. Circuits Syst.*, vol. 3, no. 1, pp. 5–12, 2013.
- [35] S. Oh, N. E. Roberts, and D. D. Wentzloff, "A 116 nW multi-band wake-up receiver with 31-bit correlator and interference rejection," in *Proc. IEEE Custom Integrated Circuits Conf.*, San Jose, CA, Sept. 2013, pp. 1–4.
- [36] B. Otis, Y. H. Chee, and J. M. Rabaey, "A 400 μW-RX, 1.6mW-TX superregenerative transceiver for wireless sensor networks," in *Proc. IEEE Int. Solid-State Circuits Conf. (ISSCC)*, San Francisco, CA, Feb. 2005, pp. 396–398. [37] N. M. Pletcher, S. Gambini, and J. M. Rabaey, "A 65 μW, 1.9 GHz RF to digital baseband wakeup receiver for wireless sensor nodes," in *Proc. IEEE Custom Integrated Circuits Conf.*, San Jose, CA, Sept. 2007, pp. 539–542.
- [38] N. M. Pletcher, S. Gambini, and J. M. Rabaey, "A 2 GHz 52 μW wake-up receiver with -72 dBm sensitivity using uncertain-if architecture," in *Proc. IEEE Int. Solid-State Circuits Conf.*, San Francisco, CA, Feb. 2008, pp. 524–633. [39] N. M. Pletcher, S. Gambini, and J. M. Rabaey, "A 52 μW wake-up receiver with -72 dBm sensitivity using an uncertain-IF architecture," *IEEE J. Solid-State Circuits*, vol. 44, no. 1, pp. 269–280, 2009.
- [40] N. E. Roberts and D. D. Wentzloff, "A 98 nW wake-up radio for wireless body area networks," in *Proc. IEEE Radio Frequency Integrated Circuits Symp.*, Montreal, Canada, June 2012, pp. 373–376.
- [41] A. Saito, K. Honda, Z. Yunfei, S. Iguchi, K. Watanabe, T. Sakurai, and M. Takamiya, "An all 0.5 V, 1 Mbps, 315 MHz OOK transceiver with 38-μW career-frequency-free intermittent sampling receiver and 52-μW class-F transmitter in 40-nm CMOS," in *Proc. Symp. VLSI Circuits (VLSIC)*, Honolulu, HI, June 2012, pp. 38–39.
- [42] T. Takiguchi, S. Saruwatari, T. Morito, S. Ishida, M. Minami, and H. Morikawa, "A novel wireless wake-up mechanism for energy-efficient ubiquitous networks," in *Proc. IEEE Int. Conf. Communications Workshops*, Dresden, Germany, June 2009, pp. 1–5.

- [43] M. Vidojkovic, X. Huang, P. Harpe, S. Rampu, C. Zhou, L. Huang, J. van de Molengraft, K. Imamura, B. Büsze, F. Bouwens, M. Konijnenburg, J. Santana, A. Breeschoten, J. Huisken, K. Philips, G. Dolmans, and H. de Groot, "A 2.4 GHz ULP OOK single-chip transceiver for healthcare applications," *IEEE Trans. Biomed. Circuits Syst.*, vol. 5, no. 6, pp. 523–34, 2011
- [44] A. Vouilloz, M. Declercq, and C. Dehollaini, "A low-power CMOS super-regenerative receiver at 1 GHz," *IEEE J. Solid-State Circuits*, vol. 36, no. 3, pp. 440–451, 2001.
- [45] T. Wada, M. Ikebe, and E. Sano, "60-GHz, 9- μ W wake-up receiver for short-range wireless communications," in *Proc. European Solid-State Circuits Conf. (ESSCIRC)*, Bucharest, Romania, Sept. 2013, pp. 383–386.
- [46] D.-Y. Yoon, C.-J. Jeong, J. Cartwright, H.-Y. Kang, S.-K. Han, N.-S. Kim, D.-S. Ha, and S.-G. Lee, "A new approach to low-power and low-latency wake-up receiver system for wireless sensor nodes," *IEEE J. Solid-State Circuits*, vol. 47, no. 10, pp. 2405–2419, 2012.
- [47] X. Yu, J.-s. Lee, C. Shu, and S.-g. Lee, "A 53 µW super-regenerative receiver for 2.4 GHz wake-up application," in *Proc. Asia Pacific Microwave Conf.*, Macau, Dec. 2008, pp. 1–4.
- [48] J. Ayers, K. Mayaram, and T. S. Fiez, "A 0.4 nJ/b 900 MHz CMOS BFSK super-regenerative receiver," in *Proc. IEEE Custom Integrated Circuits Conf.*, San Jose, CA, Sept. 2008, pp. 591–594.
- [49] B. W. Cook, A. Berny, A. Molnar, S. Lanzisera, and K. S. J. Pister, "Low-power 2.4-GHz transceiver with passive RX front-end and 400-mV supply," *IEEE J. Solid-State Circuits*, vol. 41, no. 12, pp. 2757–2766, 2006.
- [50] P. Le-Huy and S. Roy, "Low-power 2.4 GHz wake-up radio for wireless sensor networks," in *Proc. IEEE Int. Conf. Wireless and Mobile Computing, Networking and Communications*, Avignon, France, Oct. 2008, pp. 13–18.
- [51] E. Lee, P. Hess, J. Gord, H. Stover, and P. Nercessian, "A 400 MHz RF transceiver for implantable biomedical micro-stimulators," in *Proc. IEEE Custom Integrated Circuits Conf.*, San Jose, CA, Sept. 2007, pp. 173–176.
- [52] M. Lont, D. Milosevic, A. H. M. van Roermund, and G. Dolmans, "Ultralow power FSK receiver for body area networks with automatic frequency control," in *European Solid-State Circuits Conf. (ESSCIRC)*, Bordeaux, France, Sept. 2012, pp. 430–433.
- [53] J. Pandey, J. Shi, and B. Otis, "A 120 $\,\mu$ W MICS/ISM-band FSK receiver with a 44 $\,\mu$ W low-power mode based on injection-locking and 9x frequency multiplication," in *Proc. IEEE Int. Solid-State Circuits Conf.*, San Francisco, CA, Feb. 2011, pp. 460–462.
- [54] G. Papotto, F. Carrara, A. Finocchiaro, and G. Palmisano, "A 90 nm CMOS 5 Mb/s crystal-less RF transceiver for RF-powered WSN nodes," in *Proc. IEEE Int. Solid-State Circuits Conf.*, San Francisco, CA, Feb. 2012, pp. 452–454.
- [55] R. van Langevelde, M. van Elzakker, D. van Goor, H. Termeer, J. Moss, and A. J. Davie, "An ultra-low-power 868/915 MHz RF transceiver for wireless sensor network applications," in *Proc. IEEE Radio Frequency Integrated Circuits Symp.*, Boston, MA, June 2009, pp. 113–116.
- [56] A. C. W. Wong, G. Kathiresan, C. K. T. Chan, O. Eljamaly, and A. J. Burdett, "A 1 V wireless transceiver for an ultra low power SoC for biotelemetry applications," in *Proc. European Solid-State Circuits Conf. (ESSCIRC)*, Munich, Germany, Sept. 2007, pp. 127–130.
- [57] H. Yan, J. G. Macias-Montero, A. Akhnoukh, L. C. N. de Vreede, J. R. Long, J. J. Pekarik, and J. N. Burghartz, "A 120 μ W fully-integrated BPSK receiver in 90 nm CMOS," in *Proc. IEEE Radio Frequency Integrated Circuits Symp.*, Anaheim, CA, May 2010, pp. 277–280.
- [58] R.-F. Ye, T.-S. Horng, and J.-M. Wu, "Highly sensitive and low power injection-locked FSK receiver for short-range wireless applications," in *Proc. IEEE Radio Frequency Integrated Circuits Symp.*, Montreal, Canada, June 2012, pp. 377–380.
- [59] Atmel Corporation. AT86RF212 IEEE802.15.4 transceiver data sheet. [Online]. Available: http://www.atmel.com/images/doc8168.pdf
- [60] Nordic Semiconductor. (2013). nRF51822 Bluetooth LE transceiver data sheet. [Online]. Available: http://www.nordicsemi.com/eng/Products/Bluetooth-R-low-energy/nRF51822
- [61] D. Gordon, M. Berning, R. E. Masri, J. Blanckenstein, J. Klaue, and M. Beigl, "WoR-MAC: Combining wake-on-radio with quality-of-service for intelligent environments," in *Proc. 2012 9th Int. Conf. Networked Sensing Systems (INSS)*, Antwerp, June 2012, pp. 1–8.
- [62] J. Blanckenstein, J. Klaue, and H. Karl, "Energy efficient clustering using a wake-up receiver," in *Proc. 18th European Wireless Conf. (EW 2012)*, Poznan, Apr. 2012, pp. 1–8.