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# Lecture15: CMOS amplifiers (2)

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# Turning point

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- We have only 12 lectures ahead!



← Tsukuba  
Marathon 2011  
turning point  
(from Wikimedia  
Commons )

# Things to be covered...

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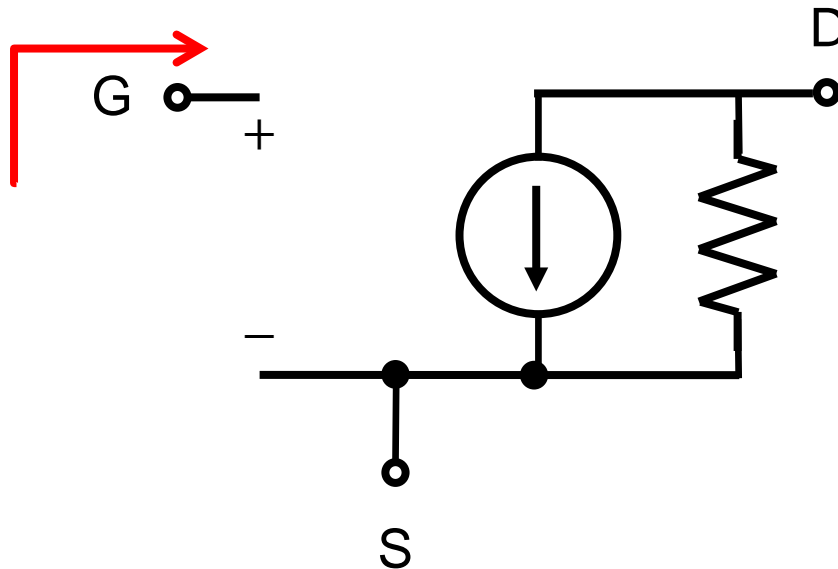
- We will cover:
  - Ch. 17
  - Ch. 4
  - Ch. 5
  - Ch. 8
  - Ch. 16 (It depends on the lecture progress.)

Mon	Tue	Wed	Thu	Fri	Sat	Sun
L15(4.27)		L16(4.29)				
L17(5.4)		L18(5.6)				
L19(5.11)		L20(5.13)				
L21(5.18)		L22(5.20)				
		L23(5.27)				
L24(6.1)		L25(6.3)				
L26(6.8)		Final				

# Impedances (1/3)

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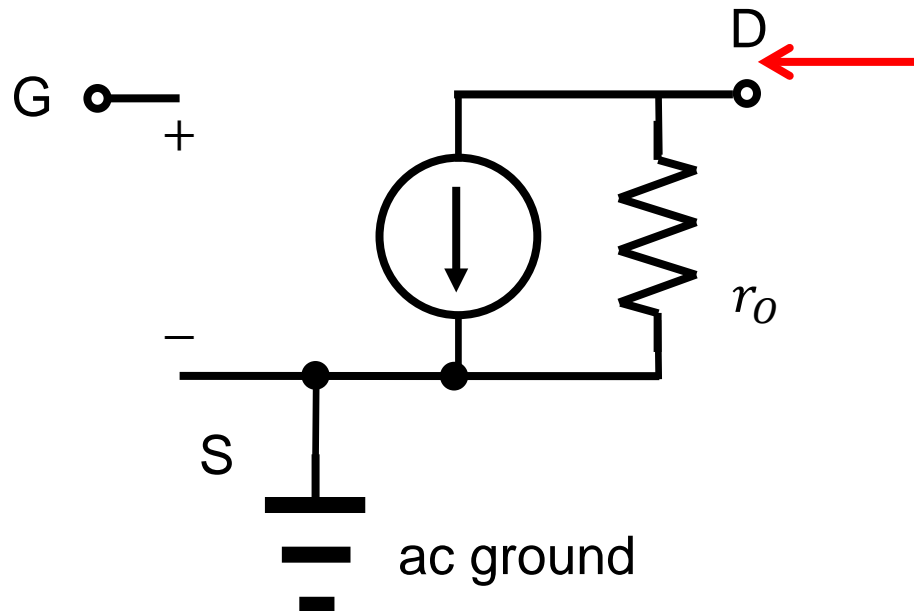
- A MOSFET with three terminals
  - Looking into the gate, we see the infinite impedance.
  - (Strictly valid at the low-frequency range)



# Impedances (2/3)

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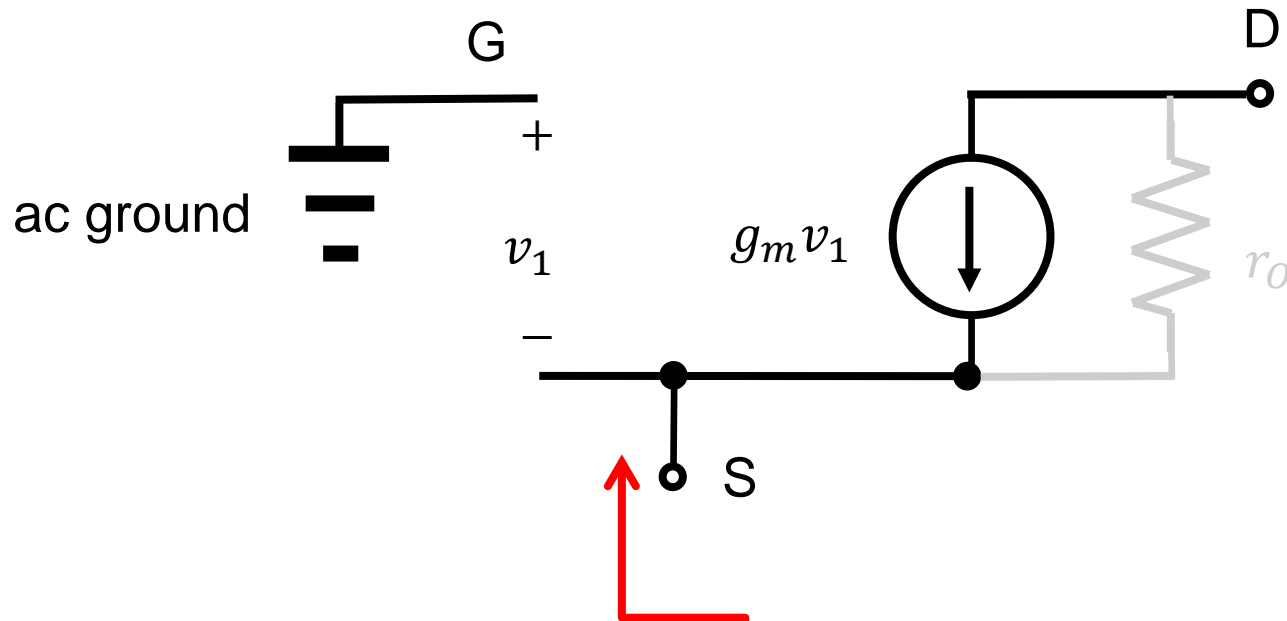
- A MOSFET with three terminals
  - Looking into the drain, we see  $r_o$  if the source is ac grounded.



# Impedances (3/3)

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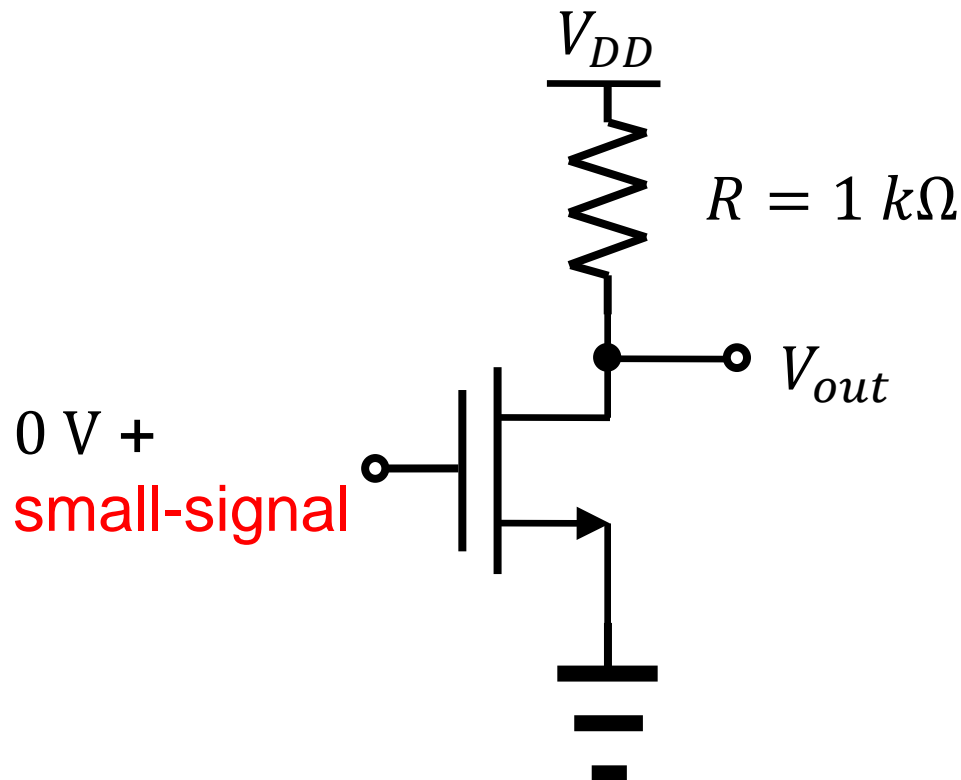
- A MOSFET with three terminals
  - Looking into the source, we see  $1/g_m$  if the gate is ac grounded and channel-length modulation is neglected.



# Transistor turned off

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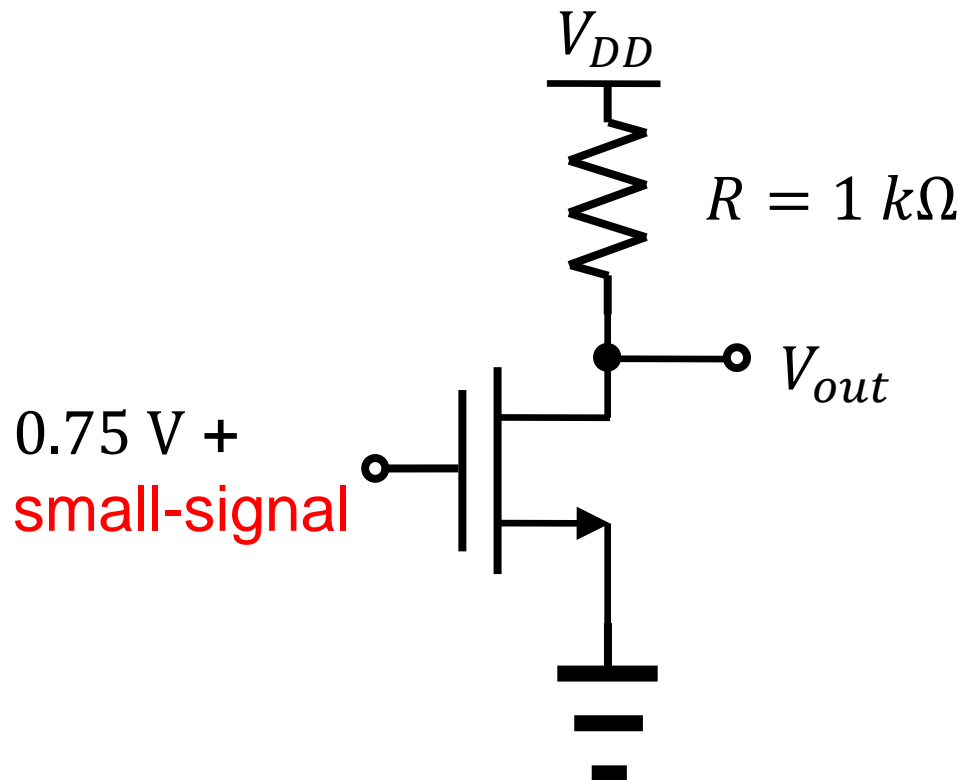
- The example 17.5 shows an amplifier circuit.
  - But, the transistor is not turned on.
  - The circuit generates no output signal.



# This is a solution.

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- The example 16.7 shows a revised circuit.
  - Then, how can we generate 0.75 V, for example?
  - Use of a separate battery can be a way.





# Simple biasing (1/2)

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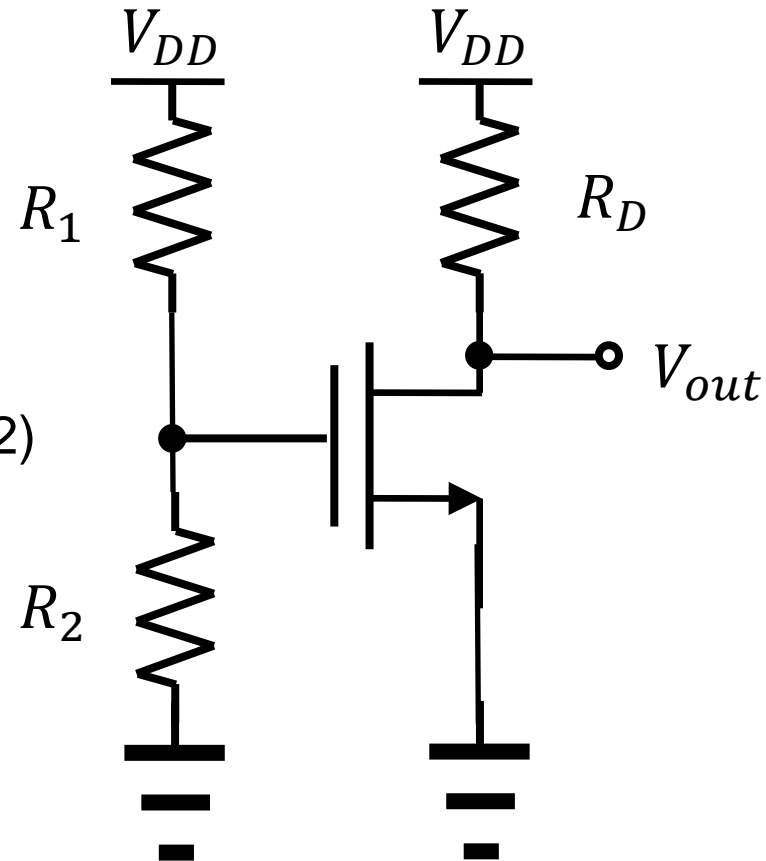
- A better way

- The gate bias voltage is

$$V_{GS} = \frac{R_2}{R_1 + R_2} V_{DD} \quad (17.10)$$

- The drain current is

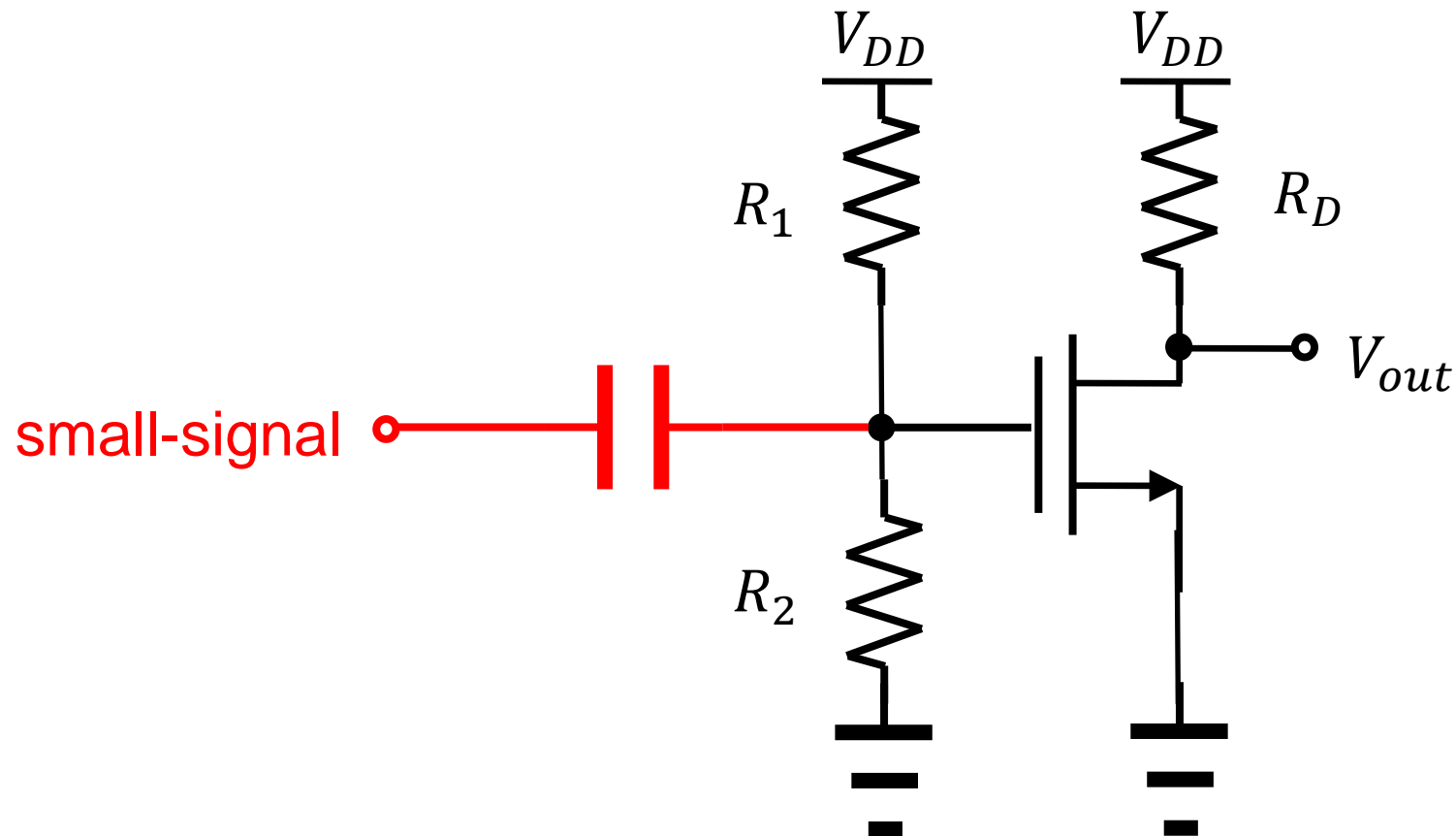
$$I_D = \frac{1}{2} \mu_n C_{ox} \frac{W}{L} \left( \frac{R_2}{R_1 + R_2} V_{DD} - V_{TH} \right)^2 \quad (17.12)$$



# Simple biasing (2/2)

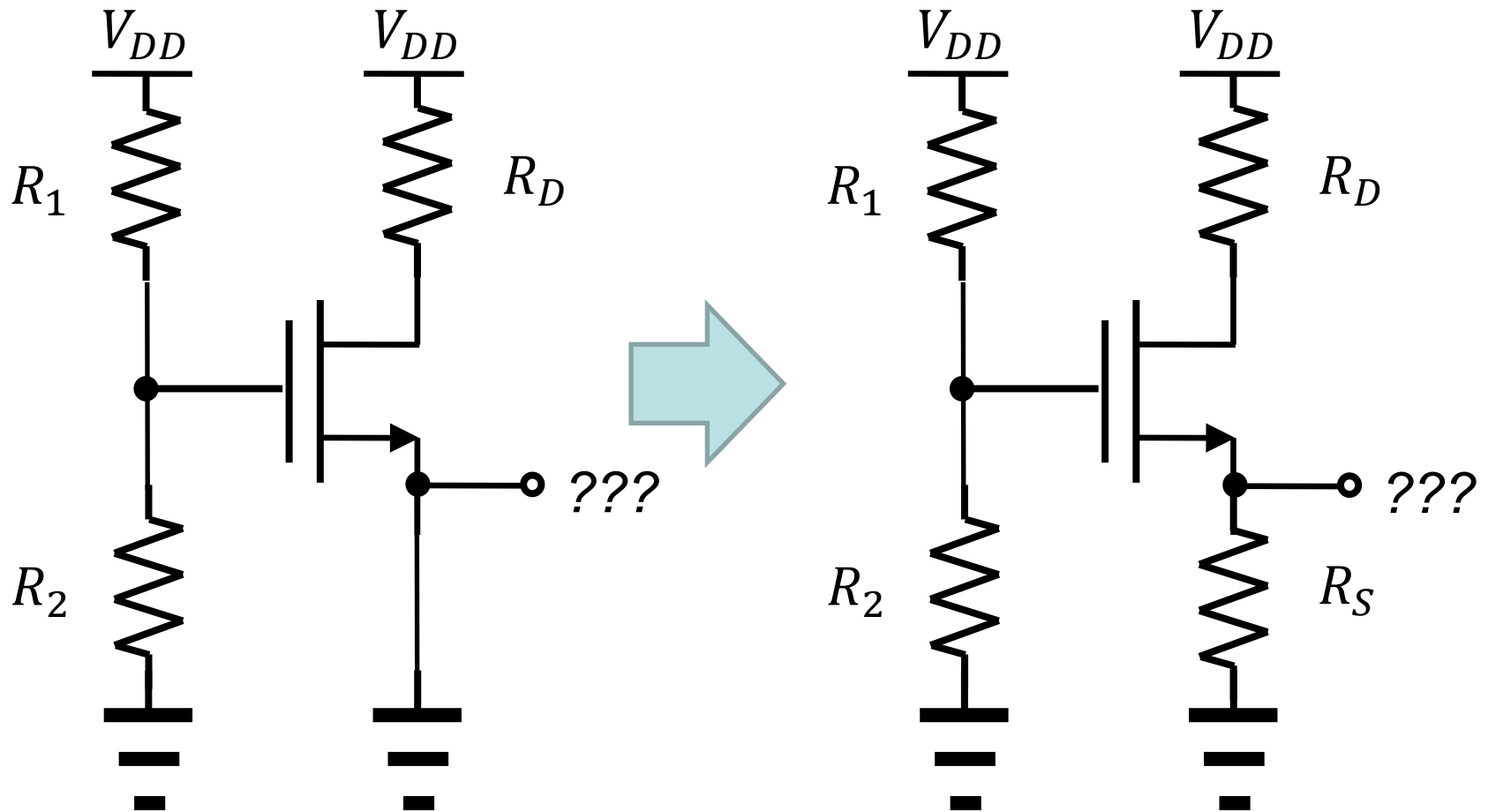
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- How to apply the small-signal input
  - Use a capacitor!



# Source degeneration (1/2)

- A resistor placed in series with the source terminal



# Source degeneration (2/2)

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- Now we have to find the source voltage.
  - (Saturation current of the MOSFET) = (Current flowing through  $R_S$ )
  - After a simple manipulation, we can find

$$V_S = V_G + V_1 - V_{TH} - \sqrt{V_1^2 + 2(V_G - V_{TH})V_1}$$

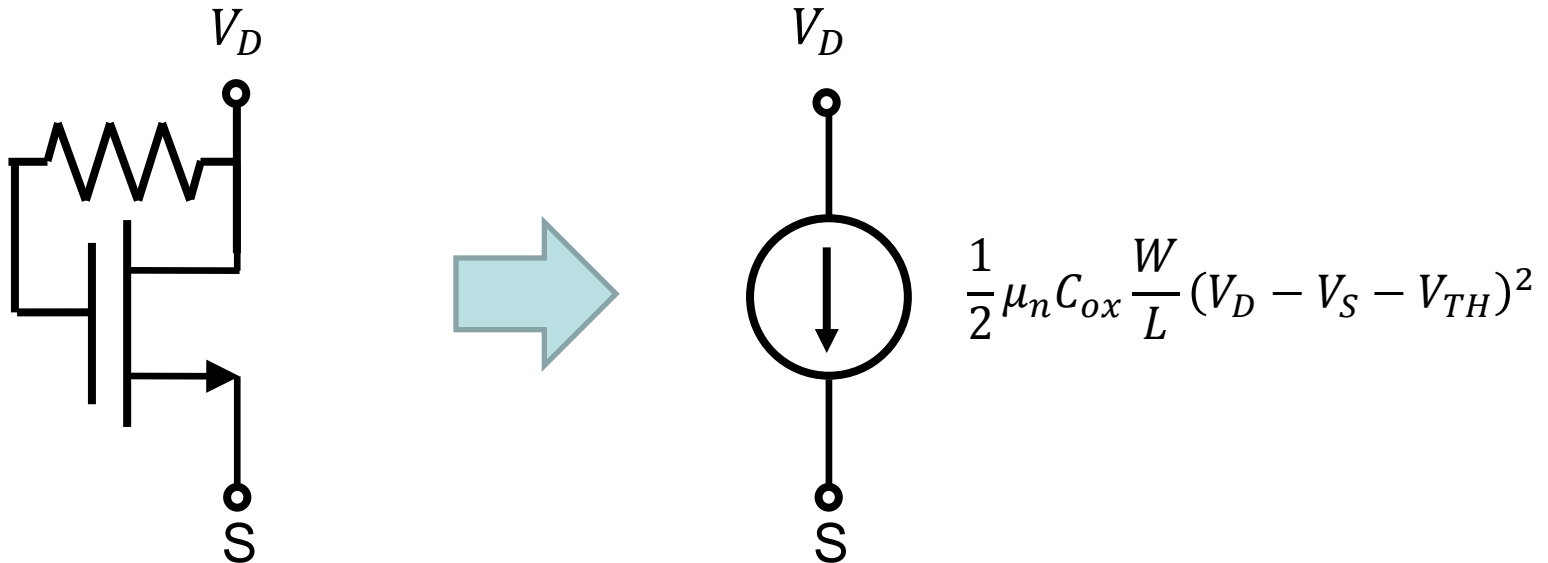
- Here,

$$V_1 = \frac{1}{\mu_n C_{ox} \frac{W}{L} R_S}$$

# Self-biasing

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- Already covered in Example 6.13.
  - Always in the saturation region.

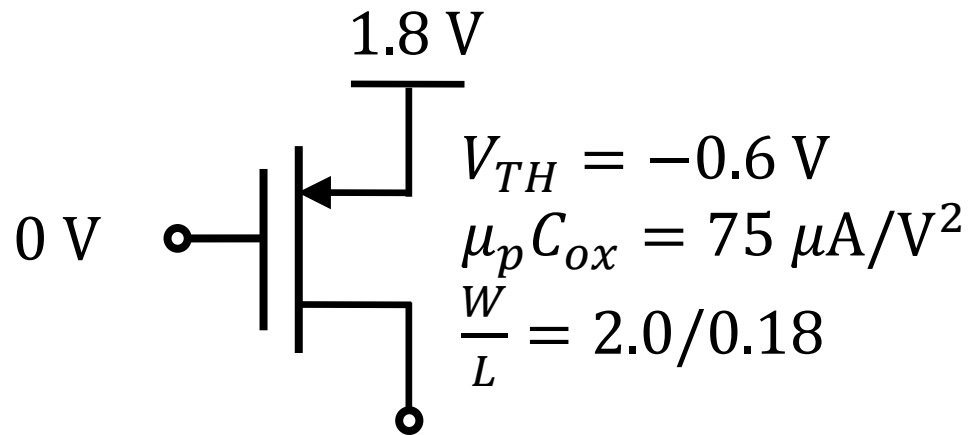


Gate and drain are tied.

# Biasing of PMOS devices

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- Let's recall the problem 18 of our mid-term exam.
  - The amount of “gate overdrive” is 1.2 V.
  - It is not 0.6 V.



# Read your textbook.

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- You can find today's lecture contents in Sec. 17. 2.
  - Up to p. 759
- We will cover Sec. 17. 3 and Sec. 17. 4.
  - Common source
  - Up to p. 766