



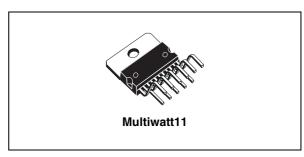
10 + 10 W stereo amplifier for car radio

Features

- Low distortion
- Low noise
- Protection against:
 - Output AC short circuit to ground
 - Overrating chip temperature
 - Load dump voltage surge
 - Fortuitous open ground
 - Very inductive loads

Description

The TDA2004R is a class B dual audio power amplifier in Multiwatt11 package specifically designed for car radio applications.



Power booster amplifiers can be easily designed using this device that provides a high current capability (up to 3.5 A) and can drive very low impedance loads (down to 1.6 Ω).

The TDA2004R allows very compact applications because few external components are required and it doesn't need electrical insulation between the package and the heatsink.

Table 1. Device summary

| Order code | Package | Packing |
|------------|-------------|---------|
| TDA2004R | Multiwatt11 | Tube |

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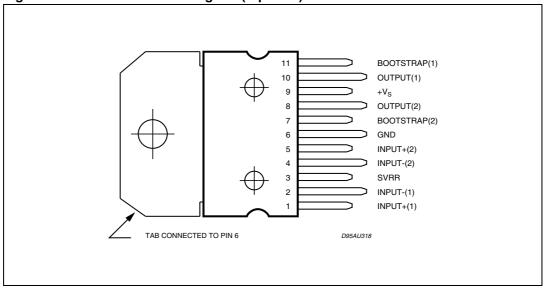
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TDA2004R Pins description

1 Pins description

Figure 1. Pins connection diagram (top view)



2 Electrical specifications

2.1 Absolute maximum ratings

Table 2. Absolute maximum ratings

| Symbol | Parameter | Value | Unit |
|-----------------------------------|---|------------|------|
| | Operating supply voltage | 18 | |
| V _S | DC supply voltage | 28 | V |
| | Peak supply voltage (50 ms) | 40 | |
| lo ⁽¹⁾ | Output peak current (non repetitive t = 0.1 ms) | 4.5 | ۸ |
| 10(1) | Output peak current (repetitive f ≥10 Hz) | 3.5 | А |
| P _{tot} | Power dissipation at T _{case} = 60 °C | 30 | W |
| T _{stg} , T _j | Storage and junction temperature | -40 to 150 | °C |

^{1.} The max. output current is internally limited.

2.2 Thermal data

Table 3. Thermal data

| Symbol | Parameter | | Value | Unit |
|------------------------|-------------------------------------|-----|-------|------|
| R _{th-j-case} | Thermal resistance junction-to-case | max | 3 | °C/W |

2.3 Electrical characteristics

Refer to the stereo application circuit T_{amb} = 25 °C; G_v = 50 dB; $R_{th(heatsink)}$ = 4 °C/W unless otherwise specified

Table 4. Electrical characteristics

| Symbol | Parameter | Test condition | Min. | Тур. | Max. | Unit |
|----------------|-------------------------------|--|-------------------|----------------------|------------|----------|
| V _S | Supply voltage | | 8 | | 18 | V |
| V _o | Quiescent offset voltage | V _S = 14.4 V V _S = 13.2 V | 6.6 6 | 7.2 6.6 | 7.8 7.2 | V V |
| I _d | Total quiescent drain current | V _S = 14.4 V V _S = 13.2 V | - | 65 62 | 120 120 | mA mA |
| P _o | Output power (each channel) | $f = 1 \text{ kHz}; \text{ THD} = 10 \%$ $V_S = 14.4 \text{ V}; \text{ R}_L = 4 \Omega$ $V_S = 14.4 \text{ V}; \text{ R}_L = 3.2 \Omega$ $V_S = 14.4 \text{ V}; \text{ R}_L = 2 \Omega$ $V_S = 14.4 \text{ V}; \text{ R}_L = 1.6 \Omega$ | 6 7 9 10 | 6.5 8 10 11 | - | W |

Table 4. Electrical characteristics (continued)

| Symbol | Parameter Test condition | | Min. | Тур. | Max. | Unit |
|-----------------|---------------------------------------|---|----------|--------------------------------|----------------------|----------|
| P _o | Output power (each channel) | $f = 1 \text{ kHz; THD} = 10 \%$ $V_{S} = 13.2 \text{ V; R}_{L} = 3.2 \Omega$ $V_{S} = 13.2 \text{ V; R}_{L} = 1.6 \Omega$ $V_{S} = 16 \text{ V; R}_{L} = 2 \Omega$ | 6 9 | 6.5 10 ⁽¹⁾ 12 | - | W |
| | | $f = 1 \text{ kHz}; V_S = 14.4 \text{ V};$ $R_L = 4 \Omega; P_O = 50 \text{ mW to } 4 \text{ W};$ | - | 0.2 | 1 | % |
| TUD | Total house onic alieta stice. | $f = 1 \text{ kHz}; V_S = 14.4 \text{ V};$ $R_L = 2 \Omega; P_0 = 50 \text{ mW to } 6 \text{ W};$ | - | 0.3 | 1 | % |
| THD | Total harmonic distortion | $f = 1 \text{ kHz}; V_S = 13.2 \text{ V}; \\ R_L = 3.2 \Omega; P_o = 50 \text{ mW to 3W};$ | - | 0.2 | 1 | % |
| | | $f = 1KHz; V_S = 13.2V;$ $R_L = 1.6Ω; P_o = 40mW to 6W;$ | - | 0.3 | 1 | % |
| СТ | Cross talk | $V_S = 14.4 \text{ V; } V_o = 4 \text{ V}_{RMS};$ $R_g = 5 \text{ k}\Omega; \text{ R}_L = 4 \Omega;$ $f = 1 \text{ kHz}$ $f = 10 \text{ kHz}$ | 50 40 | 60 45 | - | mW mW |
| V _i | Input saturation voltage | - | 300 | | - | mW |
| R _i | Input resistance | f = 1 kHz | 70 | 200 | - | kΩ |
| f _L | Low frequency roll off (-3 dB) | $R_L = 4 \Omega$ $R_L = 2 \Omega$ $R_L = 3.2 \Omega$ $R_L = 1.6 \Omega$ | - | - | 35 50 40 55 | Hz |
| f _H | High frequency roll off (-3 dB) | $R_L = 1.6 \Omega \text{ to } 4 \Omega$ | 15 | _ | - | kHz |
| ••• | Open loop voltage gain | f = 1 kHz | - | 90 | _ | |
| G_{v} | Closed loop voltage gain | f = 1 kHz | 48 | 50 | 51 | dB |
| ΔG _v | Closed loop gain matching | - | - | 0.5 | - | dB |
| e _N | Total input noise voltage | $R_{q} = 10 \text{ k}\Omega^{(2)}$ | - | 1.5 | 5 | μV |
| SVR | Supply voltage rejection | V_{ripple} = 0.5 Vrms; f_{ripple} =100 Hz; R _g = 10 kΩ; C_3 = 10 μF | 35 | 45 | - | dB |
| | | $f = 1 \text{ kHz}; V_S = 14.4 \text{ V};$ $R_L = 4 \Omega; P_o = 6.5 \text{ W};$ $R_L = 2\Omega; P_o = 10 \text{ W};$ | - | 70 60 | - | % |
| η | Efficiency | $f = 1 \text{ kHz}; V_S = 13.2 \text{ V};$ $R_L = 3.2 \Omega; P_0 = 6.5 \text{ W};$ $R_L = 1.6 \Omega; P_0 = 10 \text{ W};$ | - | 70 60 | - | % |
| T_J | Thermal shutdown junction temperature | - | - | 145 | - | °C |

^{1. 9.3} W without bootstrap.

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^{2.} Bandwidth filter: 22 Hz to 22 kHz.

2.4 Test and application circuit

Figure 2. Test and application circuit

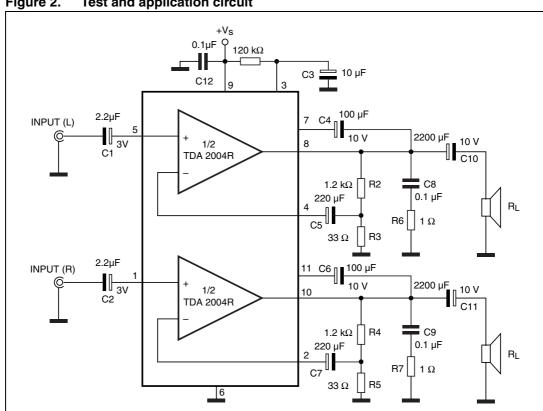
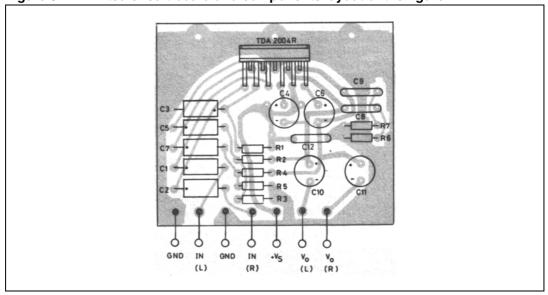


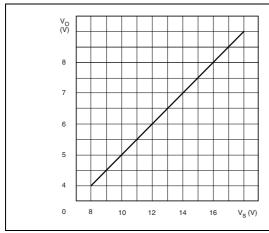
Figure 3. Printed circuit board and components layout of the figure 2



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2.5 Electrical characteristics curves

Figure 4. Quiescent output voltage vs. sup- Figure 5. Quiescent drain current vs. supply ply voltage voltage



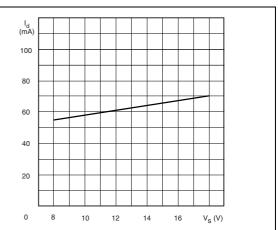
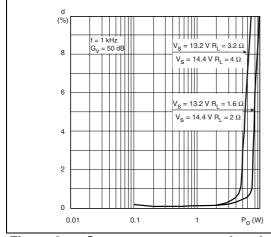


Figure 6. Distortion vs. output power

Figure 7. Output power vs. supply voltage, R_L = 2 and 4 Ω



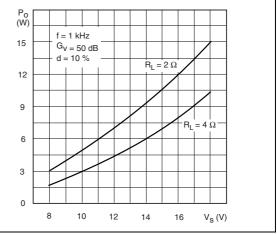
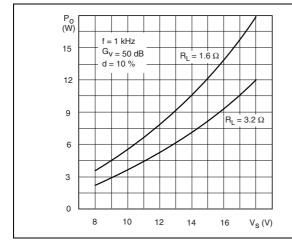
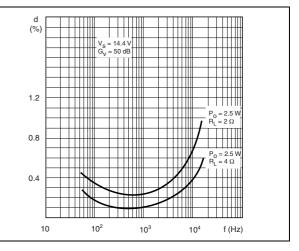


Figure 8. Output power vs. supply voltage, $R_L = 1.6$ and 3.2Ω

Figure 9. Distortion vs. frequency, R_L = 2 and 4 Ω





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Figure 10. Distortion vs. frequency, R_L = 1.6 Figure 11. Supply voltage rejection vs. C3 and 3.2 Ω

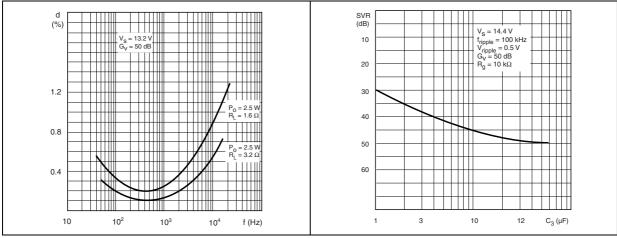


Figure 12. Supply voltage rejection vs. frequency

Figure 13. Supply voltage rejection vs. C2 and C3, $G_V = 390/1\Omega$

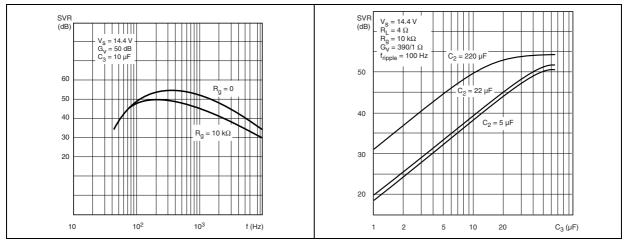
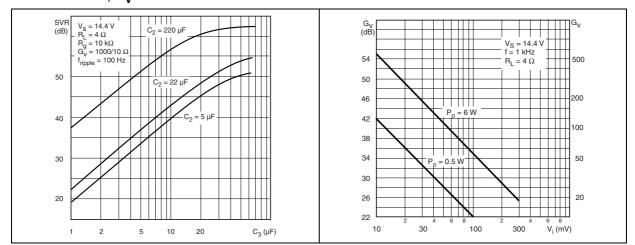


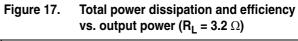
Figure 14. Supply voltage rejection vs. C2 and C3, $G_V = 1000/10\Omega$

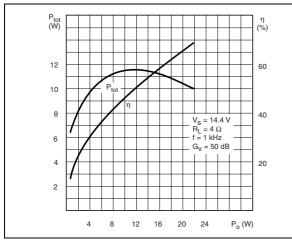
Figure 15. Gain vs. input sensitivity



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Figure 16. Total power dissipation and efficiency vs. output power ($R_L = 2 \Omega$)





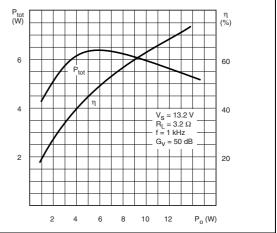
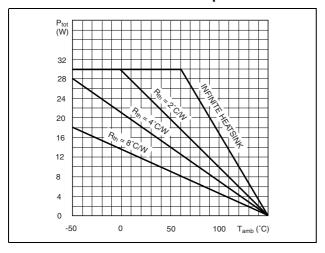


Figure 18. Maximum allowable power dissipation vs. ambient temperature



3 Application suggestion

The recommended values of the components are those shown on application circuit of *Figure 2*. Different values can be used; the following table can help the designer.

Table 5. Recommended values of the component of the application circuit

| Component | Recommended value | Purpose | Larger than | Smaller than r |
|-----------|-------------------|--|---|--|
| R1 | 120 kΩ | Optimization of the output signal symmetry | Smaller P _{omax} | Smaller P _{omax} |
| R2, R4 | 1 kΩ | Closed loop gain setting | Increase of gain | Decrease of gain |
| R3, R5 | 3.3 Ω | (1) | Decrease of gain | Increase of gain |
| R6, R7 | 1 Ω | Frequency stability | Danger of oscillation at high frequency with inductive load | |
| C1, C2 | 2.2 μF | Input DC decoupling | High turn-on delay | High turn-on pop, higher low frequency cutoff. Increase of noise |
| C3 | 10 μF | Ripple rejection | Increase of SVR, Increase of the switch-on time | Degradation of SVR |
| C4, C6 | 100 μF | Bootstrapping | - | Increase of distortion at low frequency |
| C5, C7 | 100 μF | Feedback input DC decoupling | - | - |
| C8, C9 | 0.1 μF | Frequency stability | - | Danger of oscillation |
| C10, C11 | 1000 to 2200 μF | Output DC decoupling | - | Higher low-frequency cut-off |

^{1.} The closed loop gain must be higher than 26 dB.

3.1 Built-in protection systems

3.1.1 Load dump voltage surge

The TDA2004R has a circuit which enables it to withstand voltage pulse train, on Pin 9, of the type shown in *Figure 20*. If the supply voltage peaks to more than 40 V, then an LC filter must be inserted between the supply and pin 9, in order to assure that the pulses at pin 9 will be held within the limits shown.

A suggested LC network is shown in *Figure 19*. With this network, a train of pulses with amplitude up to 120 V and width of 2 ms can be applied at point A. This type of protection is ON when the supply voltage (pulse or DC) exceeds 18 V. For this reason the maximum operating supply voltage is 18 V.

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Figure 19. Suggested LC network circuit

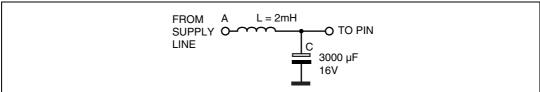
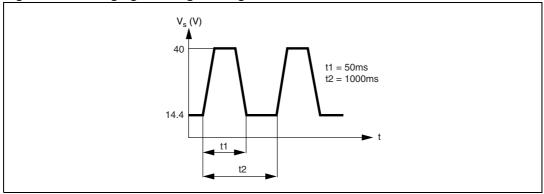


Figure 20. Voltage gain bridge configuration



3.1.2 Short circuit (AC condition)

The TDA2004R can withstand a permanent short-circuit from the output to ground caused by a wrong connection during normal working.

3.1.3 Polarity inversion

High current (up to 10 A) can be handled by the device with no damage for a longer period than the blow-out time of a quick 2 A fuse (normally connected in series with the supply). This feature is added to avoid destruction, if during fitting to the car, a mistake on the connection of the supply is made.

3.1.4 Open ground

When the ratio is in the ON condition and the ground is accidentally opened, a standard audio amplifier will be damaged. On the TDA2004R protection diodes are included to avoid any damage.

3.1.5 Inductive load

A protection diode is provided to allow use of the TDA2004R with inductive loads.

3.1.6 DC voltage

The maximum operating DC voltage for the TDA2004R is 18 V. However the device can withstand a DC voltage up to 28 V with no damage. This could occur during winter if two batteries are series connected to crank the engine.

3.1.7 Thermal shut-down

The presence of a thermal limiting circuit offers the following advantages:

- 1. an overload on the output (even if it is permanent), or an excessive ambient temperature can be easily withstood.
- 2. the heatsink can have a smaller factor of safety compared with that of a conventional circuit. There is no device damage in the case of excessive junction temperature: all that happens is that P_o (and therefore P_{tot}) and I_d are reduced.

The maximum allowable power dissipation depends upon the size of the external heatsink (i.e. its thermal resistance); *Figure 18* shows the power dissipation as a function of ambient temperature for different thermal resistance.

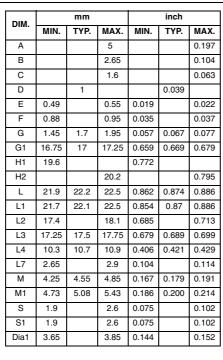
TDA2004R Package information

4 Package information

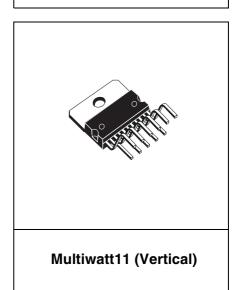
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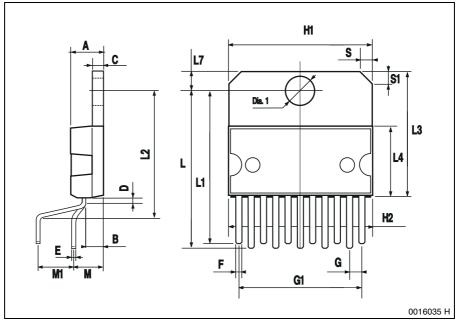
 $\mathsf{ECOPACK}^{\mathbb{R}}$ is an ST trademark.

Figure 21. Multiwatt11 mechanical data and package dimensions



OUTLINE AND MECHANICAL DATA





Revision history TDA2004R

5 Revision history

Table 6. Document revision history

| Date | Revision | Changes |
|-------------|----------|---------------------|
| 18-Jun-2010 | 1 | Initial release. |
| 18-Sep-2013 | 2 | Updated Disclaimer. |

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