

V. Homomorphic Signal Processing

◎ 5-A Homomorphism

Homomorphism is a way of “carrying over” operations from one algebra system into another.

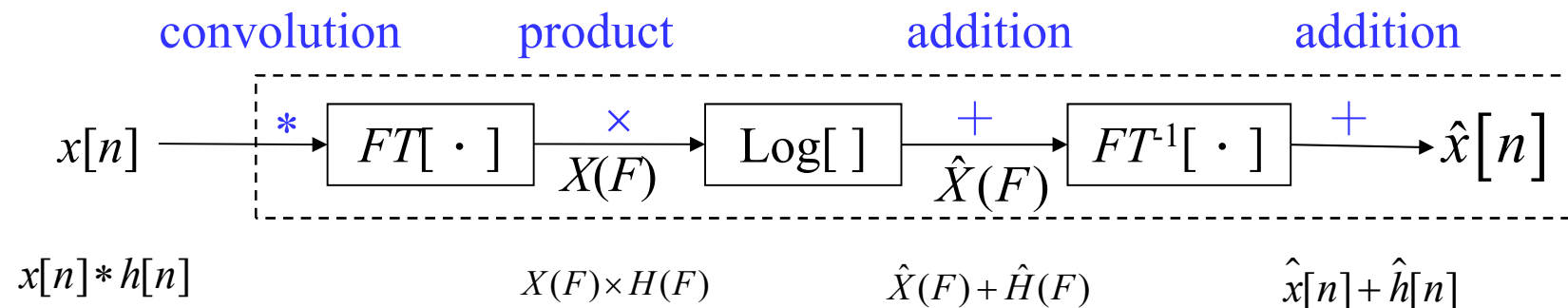
Ex. convolution $\xrightarrow{\text{Fourier}}$ multiplication $\xrightarrow{\log}$ addition

把複雜的運算，變成效能相同但較簡單的運算

◎ 5-B Cepstrum

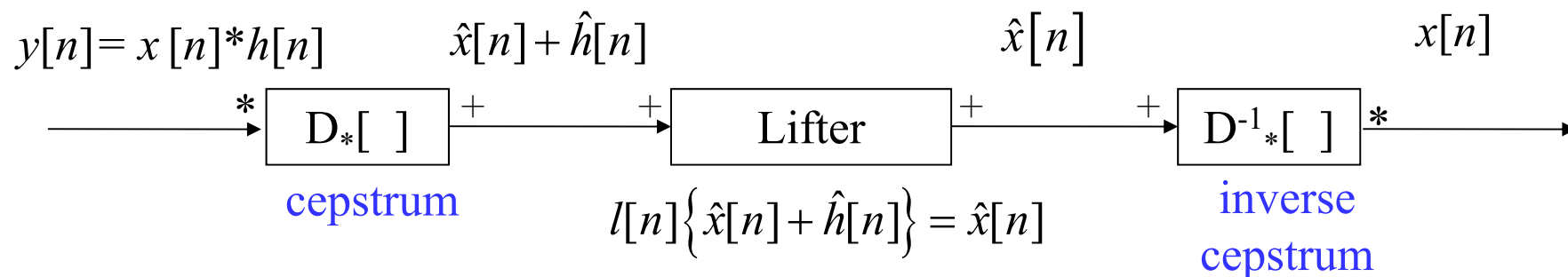
$$\hat{X}(Z)\Big|_{z=e^{j2\pi F}} = \log X(Z)\Big|_{z=e^{j2\pi F}} = \log|X(Z)\Big|_{z=e^{j2\pi F}} + j \arg[X(e^{j2\pi F})]$$

For the process of cepstrum (denoted by $D_*[\cdot]$)

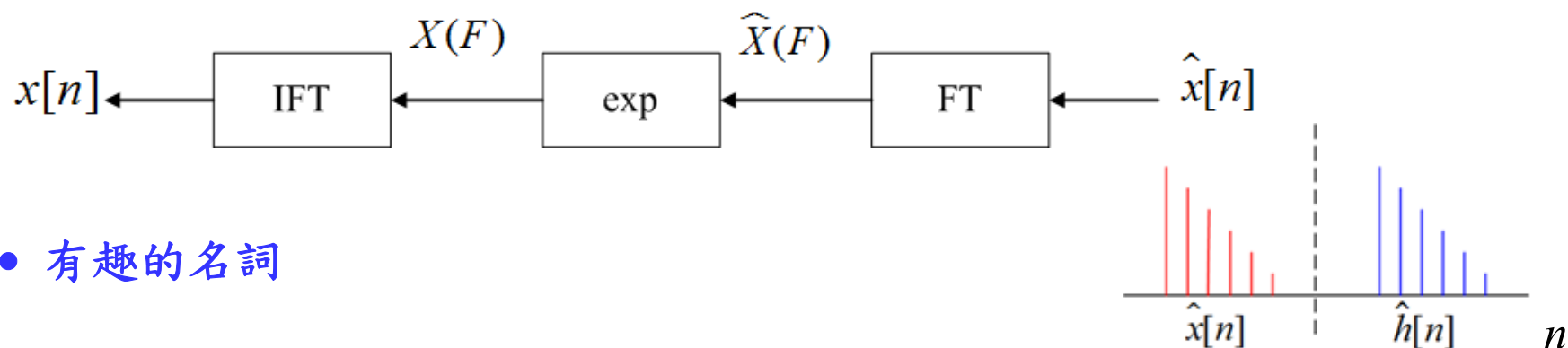


FT : discrete-time Fourier transform

- 由 $y[n] = x[n] * h[n]$ 重建 $x[n]$



For the process of the inverse cepstrum $D^{-1}_*[\]$



- 有趣的名詞

$\hat{x}[n]$ cepstrum

n quefrency

$l[n]$ lifter

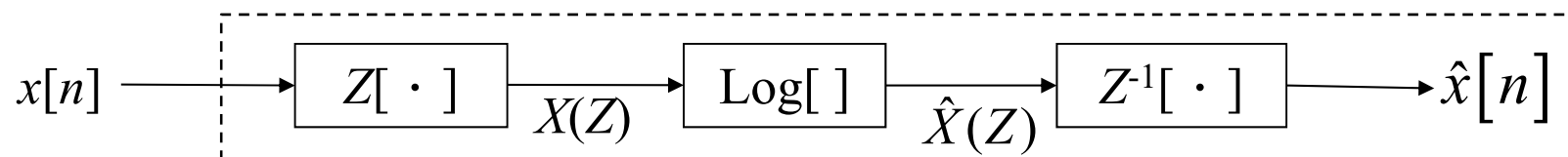
lifter



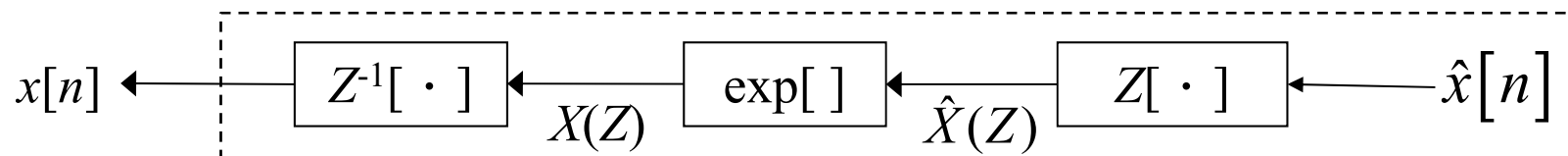
$$l[n] \{ \hat{x}[n] + \hat{h}[n] \} = \hat{x}[n]$$

Using the Z transforms instead of the Fourier transforms:

For the process of [cepstrum](#)



For the process of the [inverse cepstrum](#)




© 5-C Methods for Computing the Cepstrum

- **Method 1:** Compute the inverse discrete time Fourier transform:

$$\hat{x}[n] = \int_{-1/2}^{1/2} \hat{X}(F) e^{i2\pi nF} dF \quad : \text{inverse F.T}$$

$$\text{where } \hat{X}(F) = \log|X(F)| + j \arg[X(F)]$$

ambiguity for phase



Problems: (1)

(2)

Actually, the COMPLEX Cepstrum is REAL for real input

• Method 2 (From Poles and Zeros of the Z Transform)

實際上計算
cepstrum的方法

186

$$X(Z) = \frac{A \cancel{Z^r} \prod_{k=1}^{m_i} (1 - a_k Z^{-1}) \prod_{k=1}^{m_0} (1 - b_k Z)}{\prod_{k=1}^{P_i} (1 - c_k Z^{-1}) \prod_{k=1}^{P_0} (1 - d_k Z)}$$

time delay

where

$$|a_k|, |b_k|, |c_k|, |d_k| \leq 1$$

a_k : zeros inside unit circle

c_k : poles inside unit circle

b_k^{-1} : zeros outside unit circle

d_k^{-1} : poles outside unit circle

$$\begin{aligned} \therefore \hat{X}(Z) = \log X(Z) &= \log A + \cancel{r \cdot \log Z} + \sum_{k=1}^{m_i} \log (1 - a_k Z^{-1}) + \sum_{k=1}^{m_0} \log (1 - b_k Z) \\ &\quad - \sum_{k=1}^{P_i} \log (1 - c_k Z^{-1}) - \sum_{k=1}^{P_0} \log (1 - d_k Z) \end{aligned}$$

$$\therefore \hat{X}(Z) = \log X(Z) = \log A + \cancel{r \cdot \log Z} + \sum_{k=1}^{m_i} \log(1 - a_k Z^{-1}) + \sum_{k=1}^{m_0} \log(1 - b_k Z)$$

$$- \sum_{k=1}^{P_i} \log(1 - c_k Z^{-1}) - \sum_{k=1}^{P_0} \log(1 - d_k Z)$$

Z^{-1}
 \downarrow
 (inverse Z transform)
 \downarrow
 ?

Taylor series

$$f(t) = f(t_0) + \sum_{n=1}^{\infty} \frac{f^{(n)}(t_0)}{n!} (t - t_0)^n$$

Taylor series expansion

Z^{-1}

(Suppose that $r = 0$)

$$\hat{x}[n] = \begin{cases} \log(A) & , n = 0 \\ -\sum_{k=1}^{m_i} \frac{a_k^n}{n} + \sum_{k=1}^{P_i} \frac{c_k^n}{n} & , n > 0 \\ \sum_{k=1}^{m_0} \frac{b_k^{-n}}{n} - \sum_{k=1}^{P_0} \frac{d_k^{-n}}{n} & , n < 0 \end{cases}$$

Poles & zeros inside unit circle, right-sided sequence

Poles & zeros outside unit circle, left-sided sequence

Note:

- (1) $\hat{x}[n]$ always decays with $|n|$.
- (2) 在 complex cepstrum domain
Minimum phase 及 maximum phase 之貢獻以 $n = 0$ 為分界切開
- (3) For FIR case, there is no c_k and d_k
- (4) The complex cepstrum is unique and of infinite duration for both positive & negative n , even though $x[n]$ is causal & of finite durations

$\hat{x}[n]$ is always IIR

• **Method 3**

$$Z \cdot \hat{X}'(Z) = Z \cdot \frac{X'(Z)}{X(Z)}$$

$$\therefore ZX'(Z) = Z\hat{X}'(Z) \cdot X(Z)$$

$$\downarrow Z^{-1}$$

$$n x[n] = \sum_{k=-\infty}^{\infty} k \hat{x}[k] x[n-k]$$

$$\therefore x[n] = \sum_{k=-\infty}^{\infty} \frac{k}{n} \hat{x}[k] x[n-k] \quad \text{for } n \neq 0$$

Suppose that $x[n]$ is causal and has minimum phase, i.e. $x[n] = \hat{x}[n] = 0, n < 0$

$$x[n] = \sum_{k=-\infty}^{\infty} \frac{k}{n} \hat{x}[k] x[n-k] \quad \text{for } n \neq 0$$

$$\Rightarrow x[n] = \sum_{k=0}^n \frac{k}{n} \hat{x}[k] x[n-k] \quad \text{for } n > 0 \quad (\text{causal sequence})$$

$$x[n] = \hat{x}[n] x[0] + \sum_{k=0}^{n-1} \frac{k}{n} \hat{x}[k] x[n-k]$$

For a minimum phase sequence $x[n]$

$$\hat{x}[n] = \begin{cases} 0 & , n < 0 \\ \frac{x[n]}{x[0]} - \sum_{k=0}^{n-1} \left(\frac{k}{n}\right) \hat{x}[k] \frac{x[n-k]}{x[0]}, & n > 0 \\ \log A & , n = 0 \end{cases} \quad \text{recursive method}$$

Determining $\hat{x}[n]$ from $\hat{x}[0], \hat{x}[1], \dots, \hat{x}[n-1]$

For anti-causal and maximum phase sequence, $x[n] = \hat{x}[n] = 0, n > 0$

$$\begin{aligned} x[n] &= \sum_{k=n}^0 \frac{k}{n} \hat{x}[k] x[n-k] \quad , n < 0 \\ &= \hat{x}[n] x[0] + \sum_{k=n+1}^0 \frac{k}{n} \hat{x}[k] x[n-k] \end{aligned}$$

For maximum phase sequence,

$$\hat{x}[n] = \begin{cases} 0 & , n > 0 \\ \log A & , n = 0 \\ \frac{x[n]}{x[0]} - \sum_{k=n+1}^0 \left(\frac{k}{n}\right) \hat{x}[k] \frac{x[n-k]}{x[0]} & , n < 0 \end{cases}$$

◎ 5-D Properties

P.1) The complex cepstrum decays at least as fast as $\frac{1}{n}$

$$|\hat{x}[n]| < c \left| \frac{\alpha^n}{n} \right| \quad -\infty < n < \infty$$

$$\alpha = \max(|a_k|, |b_k|, |c_k|, |d_k|)$$

P.2) If $X(Z)$ has no poles and zeros outside the unit circle, i.e. $x[n]$ is minimum phase, then

$$\hat{x}[n] = 0 \quad \text{for all } n < 0$$

because of no b_k, d_k

P.3) If $X(Z)$ has no poles and zeros inside the unit circle, i.e. $x[n]$ is maximum phase, then

$$\hat{x}[n] = 0 \quad \text{for all } n > 0$$

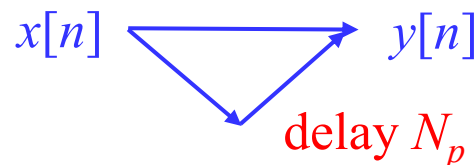
because of no a_k, c_k

P.4) If $x[n]$ is of finite duration, then
 $\hat{x}[n]$ has infinite duration

◎ 5-E Application of Homomorphic Deconvolution

(1) Equalization for Echo

$$y[n] = x[n] + \alpha x[n - N_p]$$



Let $p[n]$ be $p[n] = \delta[n] + \alpha \delta[n - N_p]$

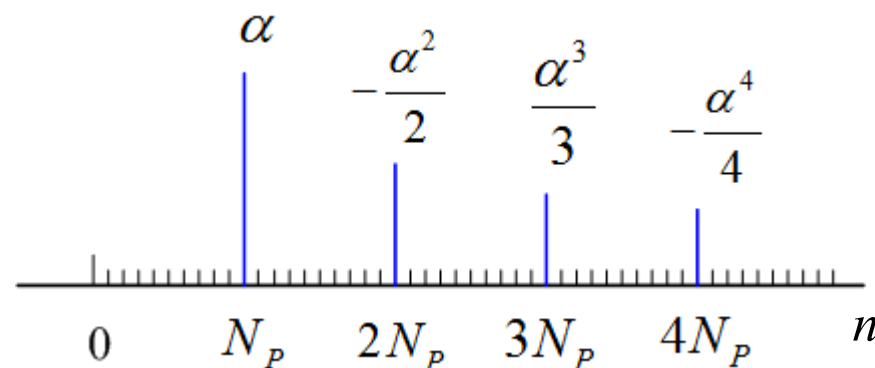
$$y[n] = x[n] + \alpha x[n - N_p] = x[n] * p[n]$$

$$P(Z) = 1 + \alpha Z^{-N_p}$$

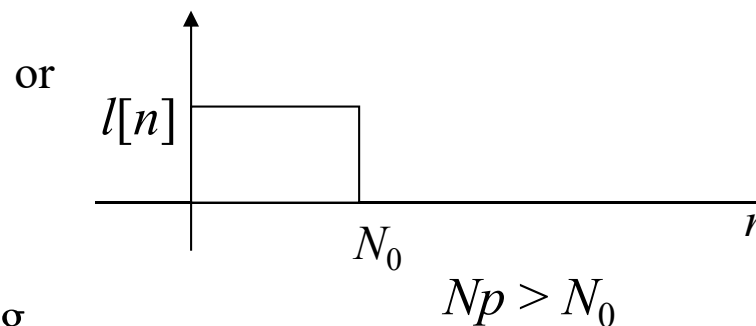
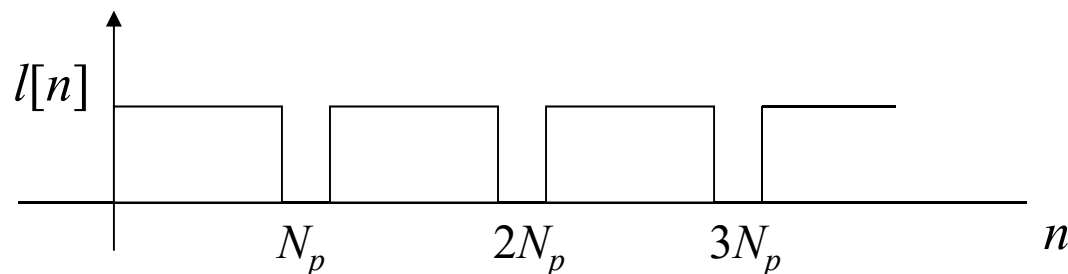
$$\hat{P}(Z) = \log(1 + \alpha Z^{-N_p}) = \sum_{k=1}^{\infty} (-1)^{k+1} \frac{\alpha^k}{k} Z^{-kN_p}$$

$\downarrow Z^{-1}$

$$\hat{p}[n] = \sum_{k=1}^{\infty} (-1)^{k+1} \frac{\alpha^k}{k} \delta(n - k \cdot N_p)$$



Filtering out the echo by the following “lifter”:



Q: For the case where N_p is unknown

(2) Representation of acoustic engineering

$$y[n] = x[n] * h[n]$$

Synthesized music

building effect : e.g. 羅馬大教堂的 impulse response

(3) Speech analysis

$$s[n] = g[n] * v[n] * p[n]$$

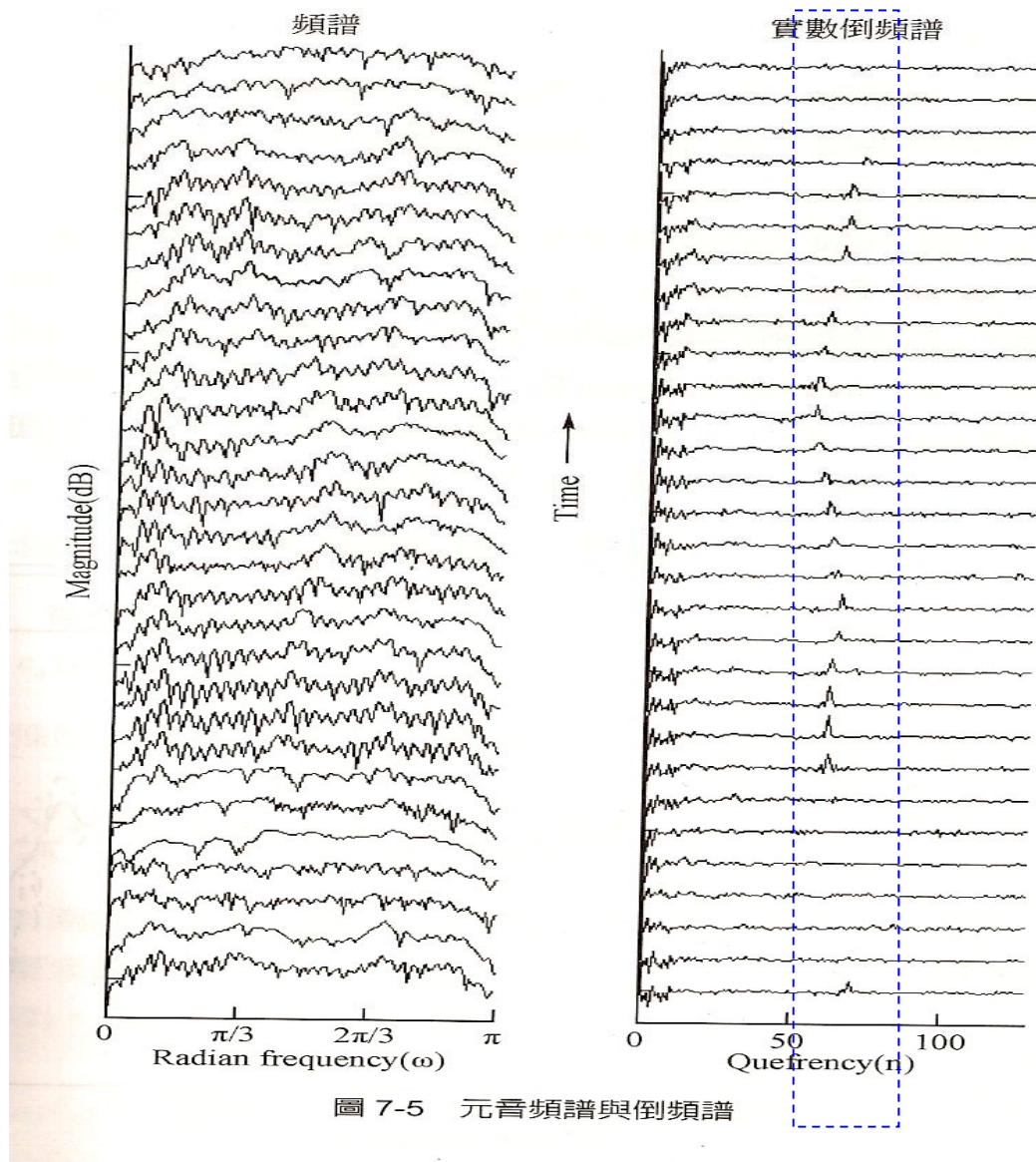
$$\begin{array}{cccc} \text{Speech} & \text{Global} & \text{Vocal tract} & \text{Pitch} \\ \text{wave} & \text{wave} & \text{impulse} & \\ \text{shape} & & & \end{array}$$

They can be separated by filtering in the complex cepstrum domain

(4) Seismic Signals

(5) Multiple-path analysis for any wave-propagation problem

用 cepstrum 將 multipath 的影響去除



From 王小川，「語音訊號處理」，全華出版，台北，民國94年。

From 王小川，「語音訊號處理」，全華出版，台北，民國94年。

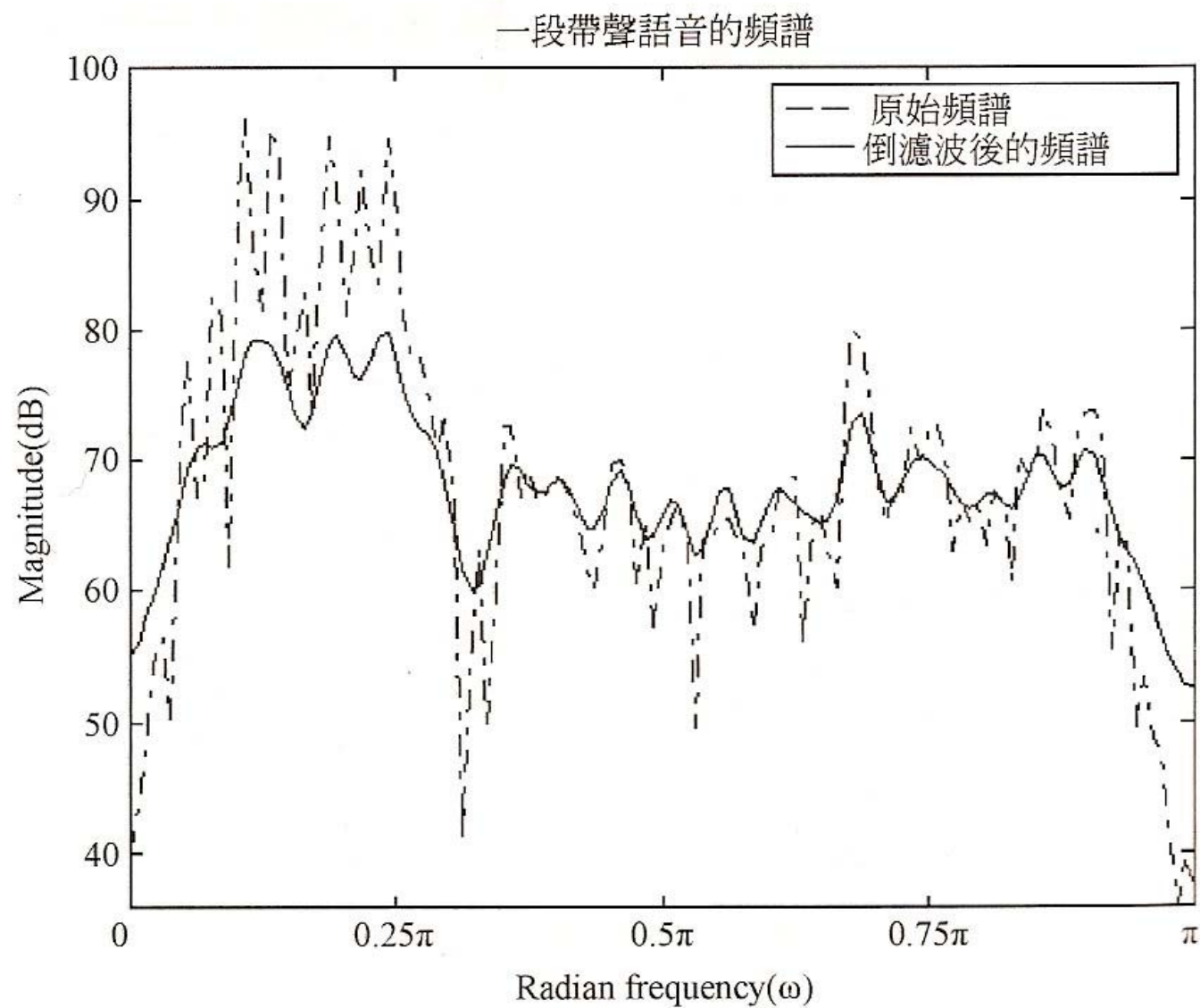


圖 7-6 經過倒濾波器作平滑處理的頻譜

◎ 5-F Problems of Cepstrum

(1) $|\log(X(Z))|$

(2) Phase

(3) Delay Z^{-k}

(4) Only suitable for the multiple-path-like problem

© 5-G Differential Cepstrum

$$\hat{x}_d(n) = Z^{-1} \left[\frac{X'(Z)}{X(Z)} \right] \quad \text{或} \quad \hat{x}_d[n] = \int_{-1/2}^{1/2} \frac{X'(F)}{X(F)} e^{i2\pi F} dF$$

↑
inverse Z transform

Note: $\frac{d}{dZ} \hat{X}(Z) = \frac{d}{dZ} \log(X(Z)) = \frac{X'(Z)}{X(Z)}$

If $x(n) = x_1(n) * x_2(n)$

$$X(Z) = X_1(Z) \cdot X_2(Z)$$

$$X'(Z) = X_1'(Z) \cdot X_2(Z) + X_1(Z) \cdot X_2'(Z)$$

$$\frac{X'(Z)}{X(Z)} = \frac{X_1'(Z)}{X_1(Z)} + \frac{X_2'(Z)}{X_2(Z)}$$

$$\therefore \hat{x}_d(n) = \hat{x}_{1d}(n) + \hat{x}_{2d}(n)$$

Advantages: no phase ambiguity

able to deal with the delay problem

• Properties of Differential Cepstrum

(1) The differential Cepstrum is shift & scaling invariant

不只適用於 multi-path-like problem

也適用於 pattern recognition

If $y[n] = A X[n - r]$

$$\Rightarrow \hat{y}_d(n) = \begin{array}{ll} \hat{x}_d(n) & , \quad n \neq 1 \\ -r + \hat{x}_d(1) & , \quad n = 1 \end{array}$$

(Proof): $Y(z) = Az^{-r} X(z)$

$$Y'(z) = Az^{-r} X'(z) - rAz^{-r-1} X(z)$$

$$\frac{Y'(z)}{Y(z)} = \frac{X'(z)}{X(z)} - rz^{-1}$$

$$\hat{y}_d(n) = \hat{x}_d(n) - r\delta(n-1)$$

(2) The complex cepstrum $\hat{C}[n]$ is closely related to its differential cepstrum $\hat{x}_d[n]$ and the signal original sequence $x[n]$

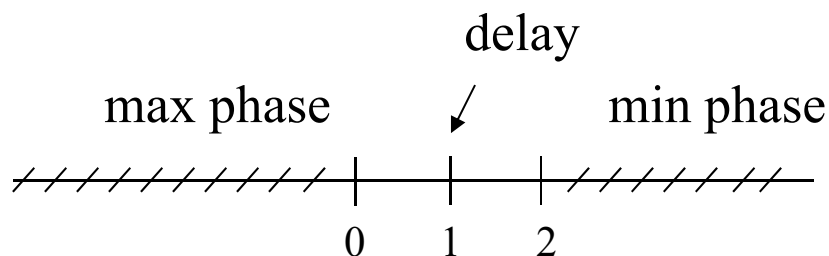
$$\hat{C}(n) = \frac{-\hat{x}_d(n+1)}{n} \quad n \neq 0 \quad \text{diff cepstrum}$$

$$\text{and} \quad -(n-1)x(n-1) = \sum_{k=-\infty}^{\infty} \hat{x}_d(n)x(n-k) \quad \text{recursive formula}$$

Complex cepstrum 做得到的事情, differential cepstrum 也做得到 !

(3) If $x[n]$ is minimum phase (no poles & zeros outside the unit circle), then $\hat{x}_d[n] = 0$ for $n \leq 0$

(4) If $x[n]$ is maximum phase (no poles & zeros inside the unit circle), then $\hat{x}_d[n] = 0$ for $n \geq 2$



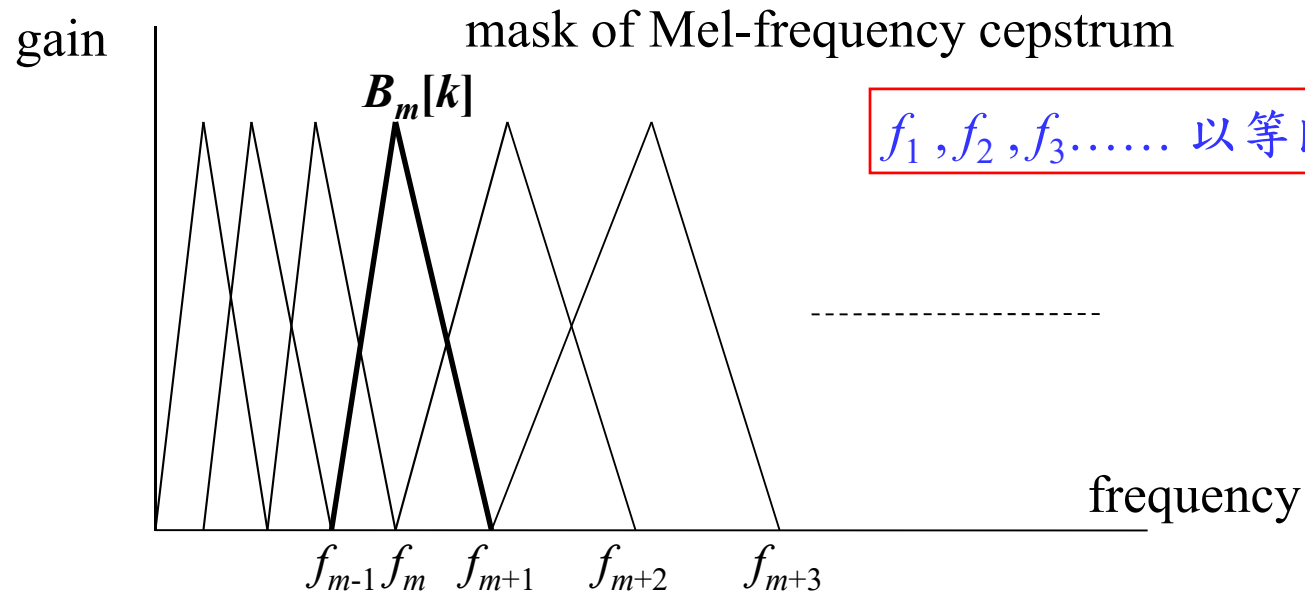
(5) If $x(n)$ is of finite duration, $\hat{x}_d[n]$ has infinite duration

Complex cepstrum decay rate $\propto \frac{1}{n}$

Differential Cepstrum decay rate 變慢了, $\therefore \hat{x}_d(n+1) = n \cdot \hat{c}(n) \propto n \cdot \frac{1}{n} = 1$

◎ 5-H Mel-Frequency Cepstrum (梅爾頻率倒頻譜)

Take log in the frequency mask



f_1, f_2, f_3, \dots 以等比級數增加

$$f_m = \alpha^m f_0$$

$$B_m[k] = 0 \quad \text{for } f < f_{m-1} \text{ and } f > f_{m+1}$$

$$B_m[k] = (k - f_{m-1}) / (f_m - f_{m-1}) \quad \text{for } f_{m-1} \leq f \leq f_m$$

$$B_m[k] = (f_{m+1} - k) / (f_{m+1} - f_m) \quad \text{for } f_m \leq f \leq f_{m+1}$$

$$f = k f_s / N$$

Process of the Mel-Frequency Cepstrum

$$(1) \quad x[n] \xrightarrow{FT} X[k]$$

$$(2) \quad Y[m] = \log \left\{ \sum_{k=f_{m-1}}^{f_{m+1}} |X[k]|^2 B_m[k] \right\}$$

$$(3) \quad c_x[n] = \frac{1}{M} \sum_{m=1}^M Y[m] \cos \left(\frac{\pi n(m-1/2)}{M} \right)$$

summation of the effect
inside the m^{th} mask

Q: What are the difference between the Mel-frequency cepstrum and the original cepstrum?

Advantages :

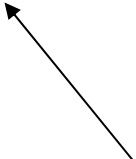
Mel-frequency cepstrum 更接近人耳對語音的區別性
用 $c_x[1], c_x[2], c_x[3], \dots, c_x[13]$ 即足以描述語音特徵

◎ 5-I References

- R. B. Randall and J. Hee, “Cepstrum analysis,” *Wireless World*, vol. 88, pp. 77-80. Feb. 1982
- 王小川， “語音訊號處理”，全華出版，台北，民國94年。
- A. V. Oppenheim and R. W. Schaffer, *Discrete-Time Signal Processing*, London: Prentice-Hall, 3rd ed., 2010.
- S. C. Pei and S. T. Lu, “Design of minimum phase and FIR digital filters by differential cepstrum,” *IEEE Trans. Circuits Syst. I*, vol. 33, no. 5, pp. 570-576, May 1986.
- S. Imai, “Cepstrum analysis synthesis on the Mel-frequency scale,” *ICASSP*, vol. 8, pp. 93-96, Apr. 1983.

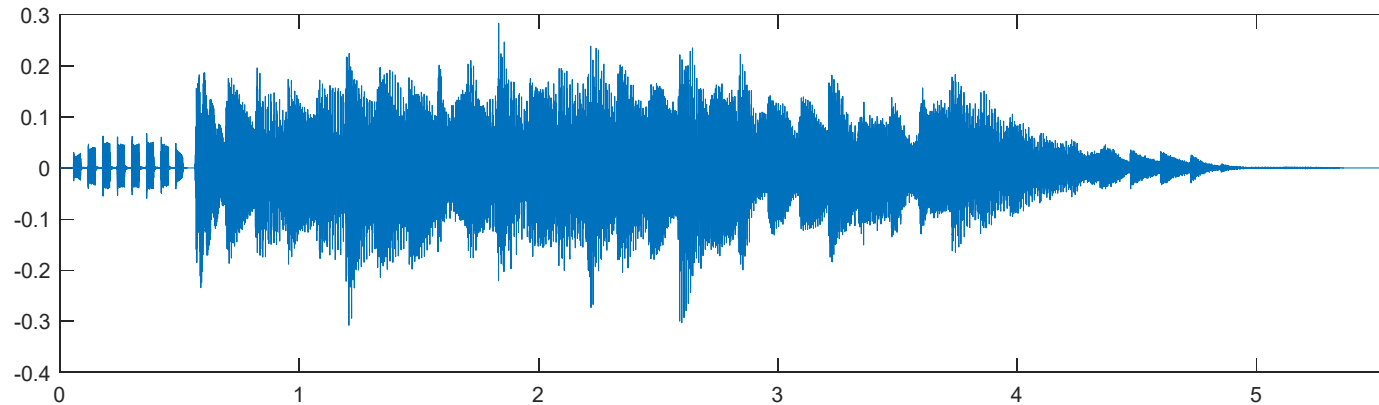
附錄六：聲音檔和影像檔的處理 (by Matlab)

A. 讀取聲音檔

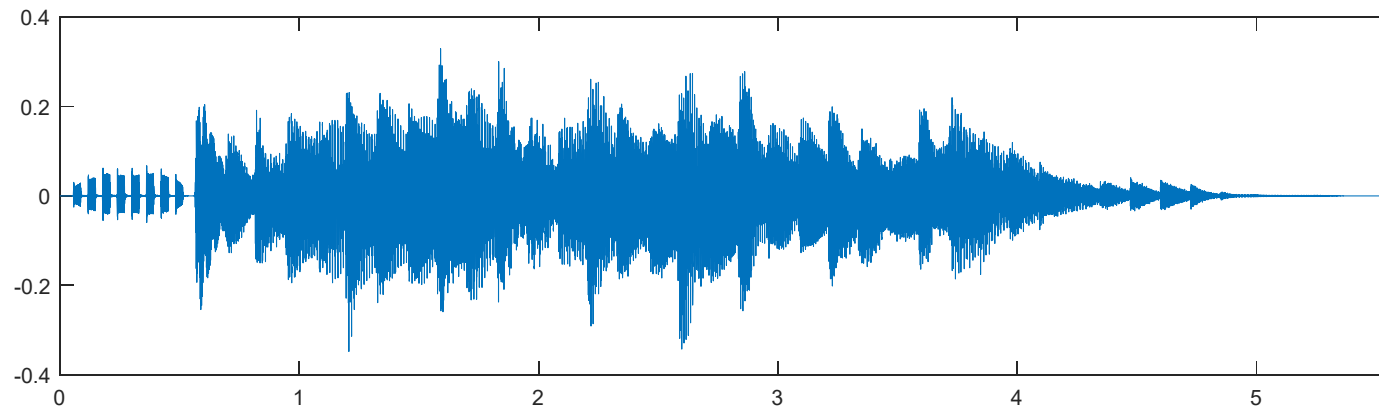
- 電腦中，沒有經過壓縮的聲音檔都是 *.wav 的型態
有經過壓縮的聲音檔是 *.mp3 的型態
 - 讀取：audioread
註：2015版本以後的 Matlab，wavread 將改為 **audioread**
 - 例：[**x**, **fs**] = audioread(C:\WINDOWS\Media\Alarm01.wav');
可以將 Alarm01.wav 以數字向量 **x** 來呈現。 **fs**: sampling frequency
這個例子當中 size(**x**) = 122868 2 fs = 22050

 - 思考：所以，取樣間隔多大？
 - 這個聲音檔有多少秒？
- 雙聲道 (Stereo，俗稱立體聲)

畫出聲音的波型

```
time = [0:size(x,1)-1]/fs; % x 是前頁用 audioread 所讀出的向量  
subplot(2,1,1); plot(time, x(:,1)); xlim([time(1),time(end)])
```



```
subplot(2,1,2); plot(time, x(:,2)); xlim([time(1),time(end)])
```



注意：*.wav 檔中所讀取的資料，值都在 -1 和 +1 之間

B. 繪出頻譜 (詳細方法請參考附錄二)

`X = fft(x(:,1));` % 只做這一步無法得出正確的頻譜

`X=X.';`

`N=length(X); N1=round(N/2);`

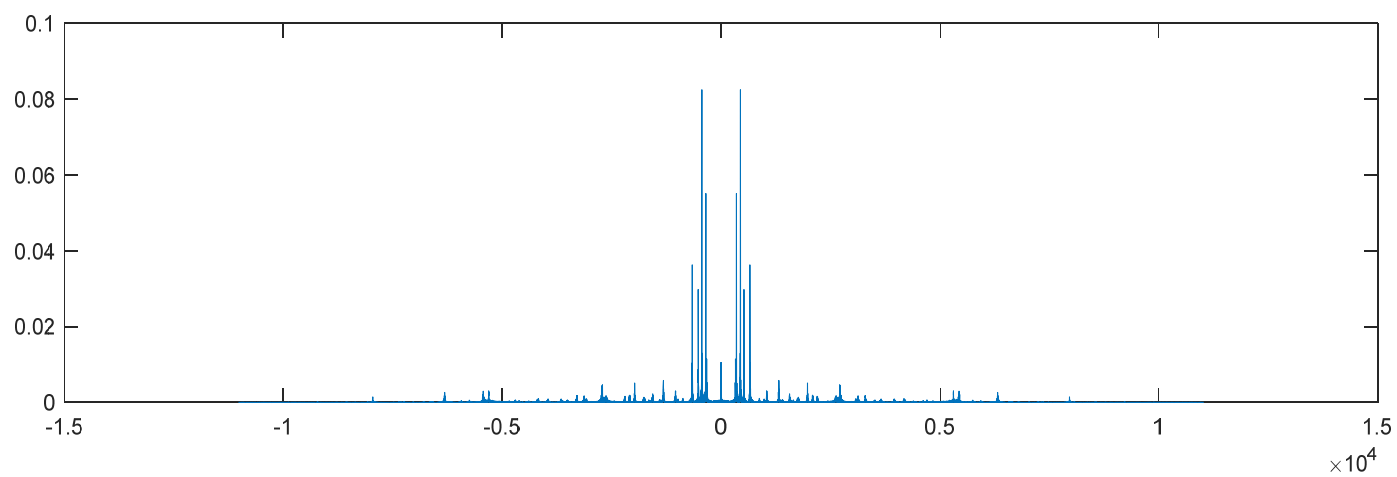
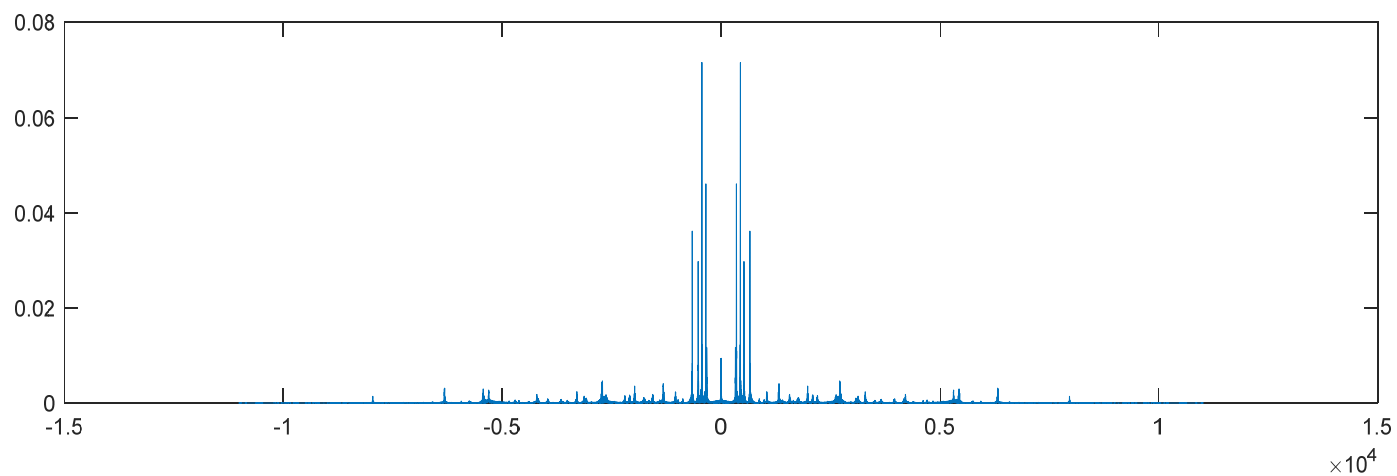
`dt=1/fs;`

`X1=[X(N1+1:N),X(1:N1)]*dt;` % shifting for spectrum

`f=[(N1:N-1)-N,0:N1-1]/N*fs;` % valid f

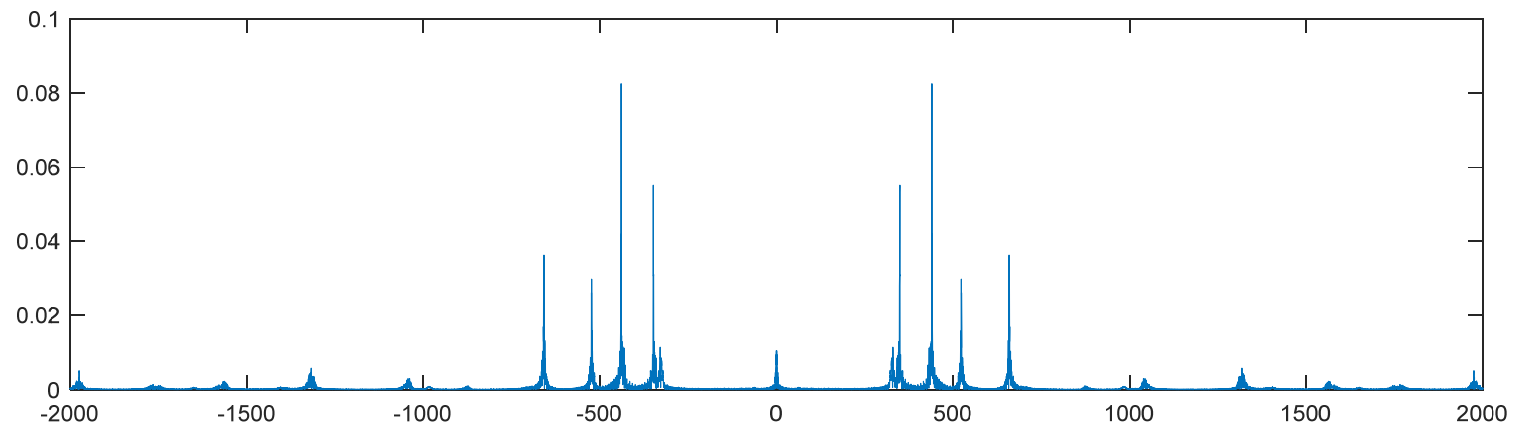
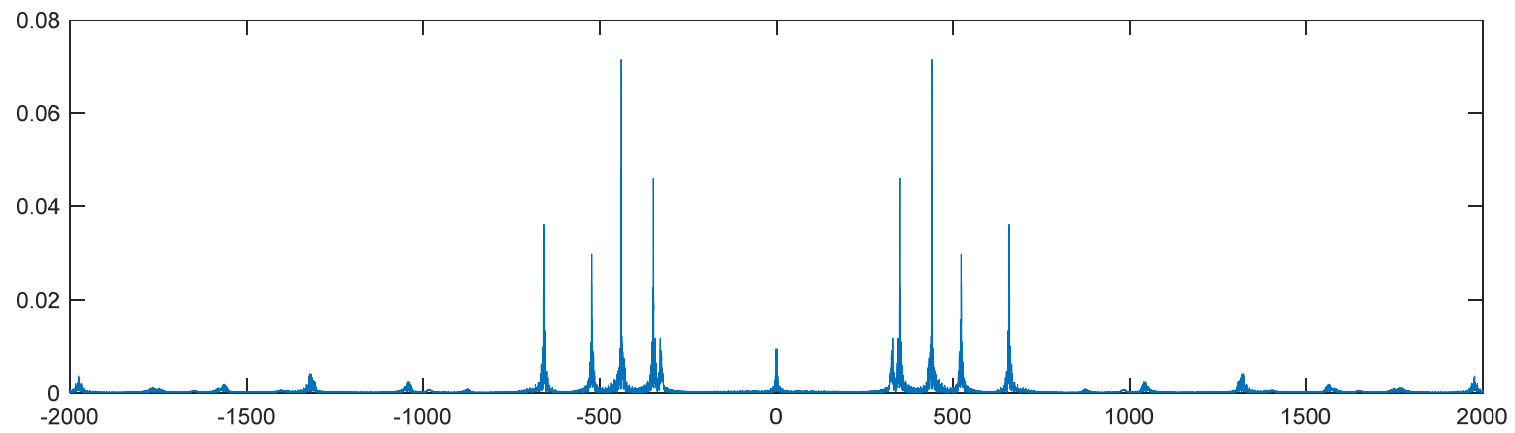
`plot(f, abs(X1));`

Alarm01.wav 的頻譜



Alarm01.wav 的頻譜

xlim([-2000,2000]) % 只看其中 -2000Hz ~ 2000Hz 的部分



C. 聲音的播放

(1) `sound(x)`: 將 **x** 以 8192Hz 的頻率播放

(2) `sound(x, fs)`: 將 **x** 以 **fs** Hz 的頻率播放

Note: (1)~(3) 中 **x** 必需是1 個column (或2個 columns)，且 **x** 的值應該 介於 -1 和 +1 之間

(3) `soundsc(x, fs)`: 自動把 **x** 的值調到 -1 和 +1 之間 再播放

D. 用 Matlab 製作 *.wav 檔：audiowrite

`audiowrite(filename, x, fs)`

將數據 **x** 變成一個 *.wav 檔，取樣速率為 **fs** Hz

① **x** 必需是1 個column (或2個 columns) ② **x** 值應該 介於 -1 和 +1

E. 用 Matlab 錄音的方法

錄音之前，要先將電腦接上麥克風，且確定電腦有音效卡
(部分的 notebooks 不需裝麥克風即可錄音)

範例程式：

```
Sec = 3;  
Fs = 8000;  
recorder = audiorecorder(Fs, 16, 1);  
recordblocking(recorder, Sec);  
audioarray = getaudiodata(recorder);
```

執行以上的程式，即可錄音。

錄音的時間為三秒，sampling frequency 為 8000 Hz

錄音結果為 audioarray，是一個 column vector (如果是雙聲道，則是兩個 column vectors)

範例程式 (續)：

```
sound(audioarray, Fs);           % 播放錄音的結果  
t = [0:length(audioarray)-1]./Fs;  
plot(t, audioarray);            % 將錄音的結果用圖畫出來  
xlabel('sec','FontSize',16);  
audiowrite('test.wav', audioarray, Fs) % 將錄音的結果存成 *.wav 檔
```

指令說明：

`recorder = audiorecorder(Fs, nb, nch);` (提供錄音相關的參數)

Fs: sampling frequency,

nb: using nb bits to record each data

nch: number of channels (1 or 2)

`recordblocking(recorder, Sec);` (錄音的指令)

recorder: the parameters obtained by the command “audiorecorder”

Sec: the time length for recording

`audioarray = getaudiodata(recorder);`

(將錄音的結果，變成 audioarray 這個 column vector，如果是雙聲道，則 audioarray 是兩個 column vectors)

以上這三個指令，要並用，才可以錄音

F：影像檔的處理

Image 檔讀取: `imread`

Image 檔顯示: `imshow`, `image`, `imagesc`

Image 檔製作: `imwrite`

基本概念：灰階影像在 Matlab 當中是一個矩陣

彩色影像在 Matlab 當中是三個矩陣，分別代表 Red, Green, Blue

*.bmp: 沒有經過任何壓縮處理的圖檔

*.jpg: 有經過 JPEG 壓縮的圖檔

Video 檔讀取: `aviread`

範例一：(黑白影像)

```
im=double(imread('C:\Program Files\MATLAB\pic\Pepper.bmp'));
```

(注意，如果 Pepper.bmp 是個灰階圖，im 將是一個矩陣)

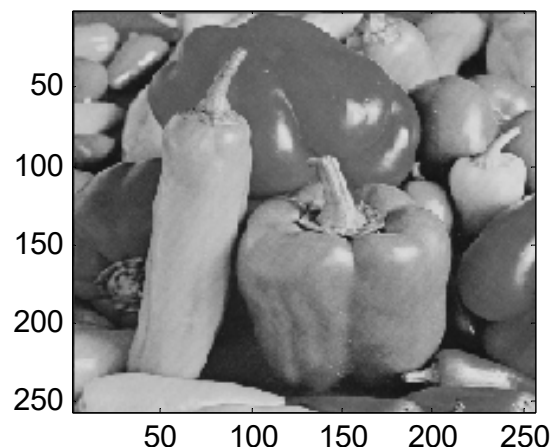
`size(im)` (用 size 這個指令來看 im 這個矩陣的大小)

```
ans =
```

```
256 256
```

```
image(im);
```

```
colormap(gray(256))
```



範例二：(彩色影像)

```
im2=double(imread('C:\Program Files\MATLAB\pic\Pepper512c.bmp'));
```

```
size(im2)
```

(注意，由於這個圖檔是個彩色的，所以 im2 將由三個矩陣複合而成)

```
ans =
```

```
512 512 3
```

```
imshow(im); or image(im/255);
```

注意：要對影像做運算時，要先變成 double 的格式

否則電腦會預設影像為 integer 的格式，在做浮點運算時會產生誤差

例如，若要對影像做 2D Discrete Fourier transform

```
im=imread('C:\Program Files\MATLAB\pic\Pepper.bmp');
```

```
im=double(im);
```

```
Imf=fft2(im);
```

附錄七 聲音檔和影像檔的處理 (by Python)

可以先安裝幾個模組

```
pip install numpy  
pip install scipy  
pip install matplotlib # plot  
pip install pipwin  
pipwin install simpleaudio # vocal files  
pipwin install pyaudio
```

PS: 謝謝2021年擔任助教的蔡昌廷同學協助製作

A. 讀音訊檔

要先import 相關模組：`import scipy.io.wavfile as wavfile`

讀取音檔：

```
fs, wave_data = wavfile.read('C:/WINDOWS/Media/Alarm01.wav')
```

```
# fs: sampling frequency
```

```
# If the audio file has one channel, then wave_data is a column vector
```

```
# If the audio file has two channels, then wave_data has two column  
vectors
```

```
num_frame = len(wave_data) # 音訊長度:
```

```
n_channel = int(wave_data.size/ num_frame) # channels 數量
```

```
>>> fs
```

```
22050
```

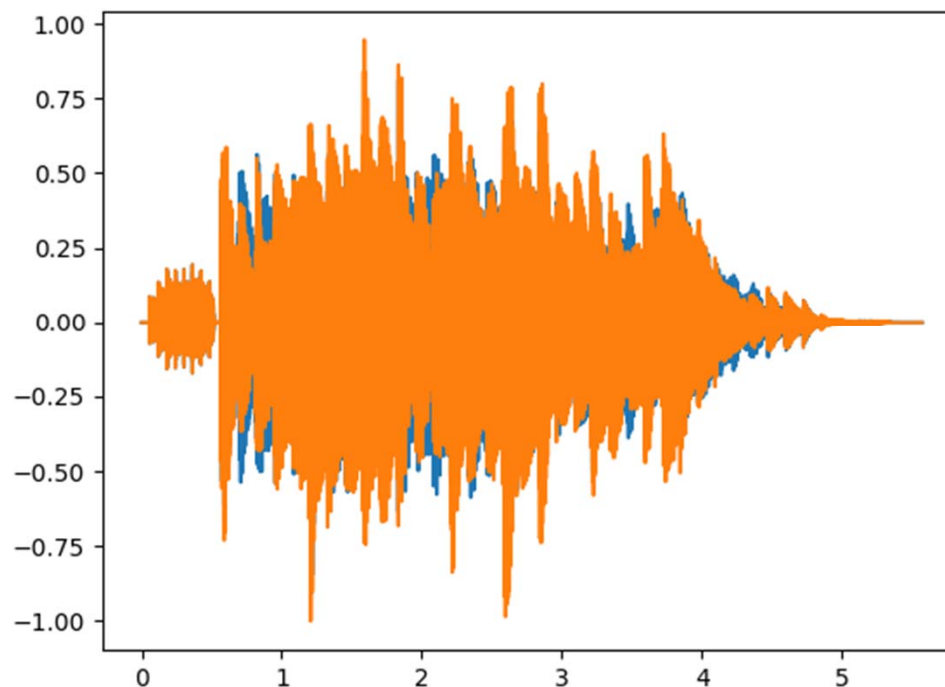
```
>>> num_frame, n_channel
```

```
(1150416, 2)
```


畫出音訊波形圖

要先import 相關模組：`import matplotlib.pyplot as plt`

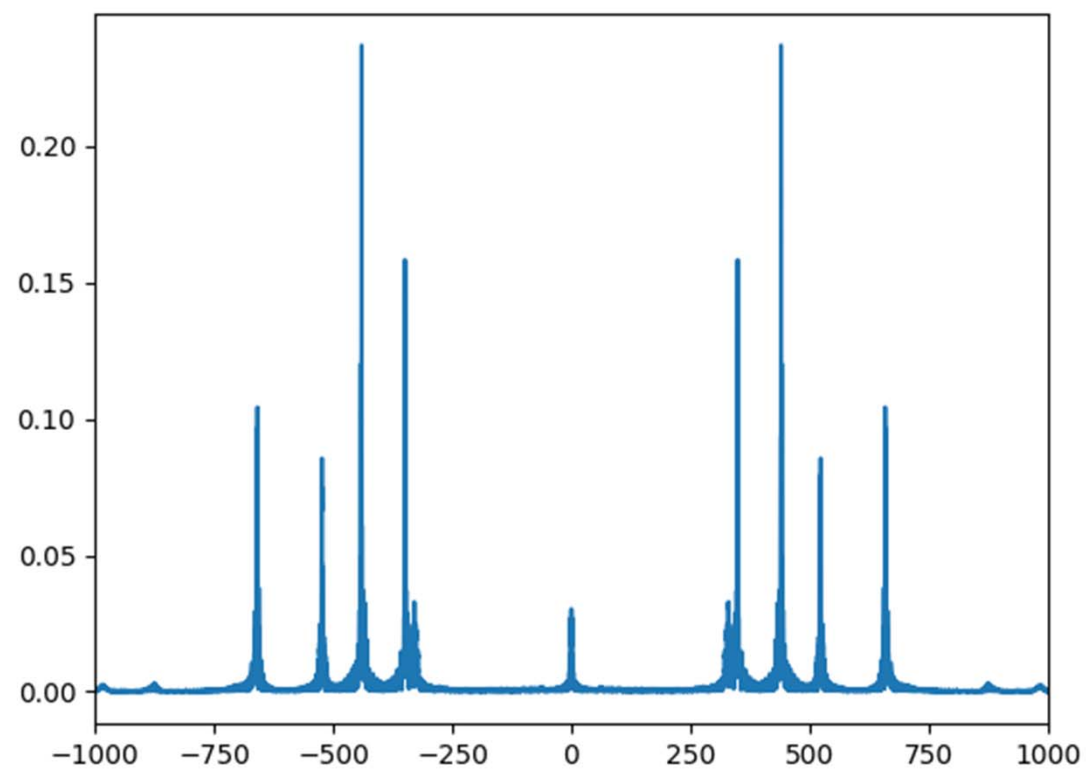
- `time = np.arange(0, num_frame)*1/fs`
- `plt.plot(time, wave_data)`
- `plt.show()`



B. 畫出頻譜

要先import 相關模組：`from scipy.fftpack import fft`

- `fft_data = abs(fft(wave_data[:,1]))/fs` # only choose the 1st channel
注意要乘上 $1/fs$
- `n0=int(np.ceil(num_frame/2))`
- `fft_data1=np.concatenate([fft_data[n0:num_frame],fft_data[0:n0]])`
將頻譜後面一半移到前面
- `freq=np.concatenate([range(n0-num_frame,0),range(0,n0)])*fs/num_frame`
頻率軸跟著調整
- `plt.plot(freq,fft_data1)`
- `plt.xlim(-1000,1000)` # 限制頻率的顯示範圍
- `plt.show()` # 如後圖



C. 播放聲音

要先import 相關模組： `import simpleaudio as sa`

- `n_bytes = 2` # using two bytes to record a data
- `wave_data = (2**15-1)* wave_data`
change the range to $-2^{15} \sim 2^{15}$
- `wave_data = wave_data.astype(np.int16)`
- `play_obj = sa.play_buffer(wave_data, n_channel, n_bytes, fs)`
- `play_obj.wait_done()`

D. 製作音檔

- `wavfile.write(file name, fs, data)`
fs means the sampling frequency
data should be a one-column or two column array

Example:

- `wavfile.write('Alarm01_test.wav', 22050, wave_data)`

E. 錄音

要先import 相關模組： `import pyaudio`

範例程式

```
import pyaudio
pa=pyaudio.PyAudio()
fs = 44100
chunk = 1024
stream = pa.open(format=pyaudio.paInt16, channels=1,
rate=fs, input=True, frames_per_buffer=chunk)

vocal=[]
count=0
```

```
while count<200: #控制錄音時間
    audio = stream.read(chunk) #一次性錄音取樣位元組大小
    vocal.append(audio)
    count +=1

save_wave_file('testrecord.wav',vocal)
stream.close()
```

參考

<https://codertw.com/%E7%A8%8B%E5%BC%8F%E8%AA%9E%E8%A8%80/491427/>

F. 影像檔的處理

可以先安裝幾個模組

```
pip install numpy
```

```
pip install matplotlib
```

1. 讀取影像檔

```
import cv2
```

```
image = cv2.imread('D:/Pic/peppers.bmp')
```

或

```
import matplotlib.pyplot as plt
```

```
image = plt.imread('D:/Pic/peppers.bmp')
```


注意

(1) 寫入圖片若為彩色圖片，需要注意 `cv2.imread` 預設 channels 順序為 BGR,

`image[:, :, 0] => B, image[:, :, 1] => G, image[:, :, 2] => R`

(2) 若使用 `plt.imread`, 則 3 個 channels 的順序仍為 RGB

`image[:, :, 0] => R, image[:, :, 1] => G, image[:, :, 2] => B`

(3) 若讀檔讀不出來，有時要將路徑的 `\` 改為 `/`

(4) `image.shape` 可以看出影像之大小

```
>>> image.shape
```

```
(512, 512, 3)
```

(5) 可讀 jpg, bmp, png 檔，但不能讀 gif 檔

2. 顯示影像

Case 1: 圖的值格式為 int

以下的指令要配合使用 (彩色，灰階皆可)

```
cv2.imshow('test', image) # 以 test 為顯示的圖的名稱
```

```
cv2.waitKey(0)
```

```
cv2.destroyAllWindows()
```

亦可用以下之指令

```
import matplotlib.pyplot as plt
```

```
plt.imshow(image)
```

```
plt.show()
```

若一開始用 cv.imread 讀圖但要用 plt.imshow 來顯示彩色圖，要先將 BGR 的順序轉回 RGB，將第二行改為

```
plt.imshow(image[:, :, [2, 1, 0]])
```

Case 2: 圖的值格式為 double (非整數)

此時，無論用 `cv2.imshow` 或是 `plt.imshow`，皆要先除以255，再將圖顯示出來

Example 1:

```
image = cv2.imread('D:/Pic/peppers.bmp')
```

```
Image1 = image*0.5 + 127.5 # lighten the image
```

```
cv2.imshow('test', image) # int 不用除255
```

```
cv2.waitKey(0)
```

```
cv2.destroyAllWindows()
```

```
cv2.imshow('test', image/255) # 非整數要除255
```

```
cv2.waitKey(0)
```

```
cv2.destroyAllWindows()
```



Example 2:

```
import matplotlib.pyplot as plt  
image = plt.imread('D:/Pic/peppers.bmp')  
image1 = image*0.5 + 127.5 # lighten the image  
plt.imshow(image) # int 不用除255  
plt.show()  
plt.imshow(image1/255) # 非整數要除255  
plt.show()
```



3. 寫入圖片檔

```
cv2.imwrite('D:/Pic/jpg', image)
```

或

```
plt.imsave('D:/Pic/jpg', image)
```

注意

(1) 寫入圖片若為彩色圖片，在使用 cv2.imwrite 時需要注意

$\text{image[:, :, 0]} \Rightarrow \text{B}$, $\text{image[:, :, 1]} \Rightarrow \text{G}$, $\text{image[:, :, 2]} \Rightarrow \text{R}$

若用 plt.imsave 則

$\text{image[:, :, 0]} \Rightarrow \text{R}$, $\text{image[:, :, 1]} \Rightarrow \text{G}$, $\text{image[:, :, 2]} \Rightarrow \text{B}$

(2) 若用 cv2.imwrite('D:\Pic\jpg', image) 可能無法存檔，要將 \ 改為 /

(3) 若是使用 plt.imshow 和 plot.show() 來顯示，可以用右下角的“save the figure”來存檔