

# Analog Transmission of Vocoder Features

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Given a vector of vocoder features  $\mathbf{f}$ , use an autoencoder  $E$  to map them to a dimension  $d$  latent vector  $\mathbf{z}$  where  $d$  is even. The magnitude of each element  $z_i$  of  $\mathbf{z}$  is constrained to a maximum of 1, but unlike digital modulation is continuously valued and not constrained to a discrete set of points. For bandwidth efficient transmission over the channel the elements of  $\mathbf{z}$  are mapped to  $d/2$  complex symbols  $\mathbf{q}$ . Compared to classical digital modulation, the elements of  $\mathbf{z}$  can be considered BPSK symbols (continuously valued, analog bits), and the elements of  $\mathbf{q}$  analog QPSK symbols.

Our goal is to determine if reasonable speech quality can be obtained over a channel of bandwidth  $B < 3000$  Hz and SNR between 0 and 6dB, roughly the lower limit of Single Side Band (SSB) - a common power and bandwidth efficient form of analog radio communication.

## 1 Noise Simulation

The autoencoder output  $\mathbf{z}$  is updated every  $T_z = 1/R_z$  seconds, giving a BPSK symbol rate of:

$$R_b = d/T_z \quad (1)$$

For example with  $T_z = 0.04$ ,  $d = 80$ ,  $R_b = 2000$  symbols/s. The QPSK symbol rate is given by:

$$R_q = \frac{d}{2T_z} \quad (2)$$

For example with  $T_z = 0.04$ ,  $d = 80$ ,  $R_q = 1000$  symbols/s.

We wish to simulate a channel with a user-defined  $E_b/N_0$ , where  $E_b$  is the energy of each BPSK symbol, and  $N_0$  is the noise power per unit bandwidth. The energy of each BPSK symbol  $E_b$  is the signal power  $S$  divided by the symbol rate  $R_b = 1/T_b$ . The noise per unit bandwidth is the total noise power  $N$  divided by the bandwidth  $B$  of the system. If we are simulating at one sample

per symbol,  $B = R_b$ :

$$\begin{aligned}\frac{E_b}{N_0} &= \frac{S/R_b}{N/R_b} \\ &= \frac{S}{N} \\ &= \frac{A^2}{\sigma^2}\end{aligned}\tag{3}$$

where  $A$  is the amplitude of each BPSK symbol and  $\sigma^2 = N$  is the noise variance (mean noise energy per sample). If the noise is zero mean, we can estimate  $\sigma^2$  over  $K$  noise samples  $r_i$  as:

$$\sigma^2 = \frac{1}{K} \sum_{i=0}^{K-1} r_i^2\tag{4}$$

We generate noise samples  $r_i$  by sampling a unit variance, zero mean Gaussian noise source  $\mathcal{N}_i(0, 1)$  that is scaled by  $\sigma$ :

$$\begin{aligned}r_i &= \sigma \mathcal{N}_i(0, 1) \\ \sigma &= \frac{A}{\sqrt{Eb/N_0}}\end{aligned}\tag{5}$$

In practice we simulate the channel by adding noise to the QPSK symbols. Due to orthogonality, this can be simulated as two independent channels that have the same noise power:

$$r_i = \sigma \mathcal{N}_{2i}(0, 1) + j\sigma \mathcal{N}_{2i+1}(0, 1)\tag{6}$$

## 2 SNR Measurement

In order to compare with other methods of speech communication, it is useful to formulate expressions for estimating SNR from the BPSK and QPSK symbols. The Signal to Noise ratio (SNR) is given by:

$$\begin{aligned}\frac{S}{N} &= \frac{E_b R_b}{N_0 B} \\ &= \frac{E_q R_q}{N_0 B}\end{aligned}\tag{7}$$

A noise bandwidth  $B$  needs to be selected; common choices for HF radio are  $B = 3000$  Hz to compare with existing analog and digital voice modes, and  $B = 1$  to obtain a normalised  $C/N_0$  carrier power to noise density ratio.

At one sample per symbol, the power, the mean energy of each QPSK symbol over a window of  $K$  samples is given by:

$$E_q = Var(q_i) = \frac{1}{K} \sum_{i=0}^{K-1} |q_i|^2\tag{8}$$

As each QPSK symbol contains 2 BPSK symbols, the energy is split evenly:

$$E_b = E_q/2 \quad (9)$$

For example if the symbol amplitude is  $A = 1$ ,  $E_b = A^2 = 1$ , then  $E_q = 1+1 = 2$ .

For transmission over multipath channels we arrange the QPSK symbols as  $N_c$  parallel carriers, each running at a symbol rate of  $R_s = R_q/N_c$  symbols/s, where  $R_s$  is chosen based on delay spread considerations. Typical values for HF modems are  $N_c = 20$  and  $R_s = 50$  Hz. However the OFDM carriers are arranged such that the total symbol rate over the channel remains constant. So for a given signal power  $E_q$  and  $E_b$  remain constant (Table 1).

Waveform	$N_c$	$R_s$	$R_q$	$R_b$	$E_q$	$E_b$
Single Carrier BPSK	1	-	-	2000	-	$S/2000$
Single Carrier QPSK	1	-	1000	2000	$S/1000$	$S/2000$
OFDM QPSK	20	50	1000	2000	$S/1000$	$S/2000$

Table 1:  $E_b$  and  $E_q$  examples for single and multi-carrier OFDM waveforms for constant carrier power  $S$

In order to evaluate the ML system early in the development process it is important to ensure the noise is correctly calibrated. The expressions above can be used to check the noise injection process:

1. Set a target  $E_b/N_0$  for the simulation run, and calculate  $\sigma$  using (5).
2. Establish the equivalent target SNR from (7) evaluated using the target  $E_b/N_0$ .
3. After the simulation run measure  $E_q = \text{Var}(q_i)$  over a sample of transmitted symbols.
4. Calculate measured SNR using (7) and compare.

### 3 Glossary

Symbol	Explanation	Units
$B$	noise or signal bandwidth	Hz
$d$	dimension of latent vector $\mathbf{z}$	
$E_b/N_0$	energy per BPSK symbol on spectral noise density	
$E_q/N_0$	energy per QPSK symbol on spectral noise density	
$N$	total noise power	Watts
$N_c$	Number of carriers	
$\mathbf{q}$	vector of QPSK symbols	
$q_i$	single QPSK symbol, element of $\mathbf{q}$	
$R_b$	BPSK symbol rate	symbols/second
$R_q$	QPSK symbol rate	symbols/second
$R_s$	OFDM per carrier QPSK symbol rate	symbols/second
$R_z$	latent vector update rate	Hz
$SNR$	signal to noise Ratio	
$S$	total signal (carrier) power	Watts
$T_b$	BPSK symbol period	seconds
$T_q$	QPSK symbol period	seconds
$T_s$	OFDM per carrier QPSK symbol period	seconds
$T_z$	time between latent vector updates	seconds
$r_i$	noise sample	
$\mathbf{z}$	Autoencoder output latent vector	
$z_i$	single latent vector element of $\mathbf{z}$ , a BPSK symbol	

Table 2: Glossary of Symbols