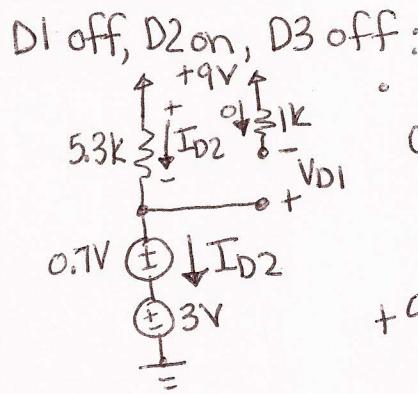
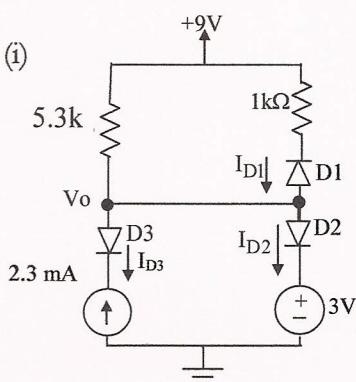


The material we have covered so far this semester is summarized (but NOT limited to) below:

1. Understand the difference between AC & DC signals.
2. Understand how to analyze circuit (with or without cap in it) to obtain transfer function.
3. Understand how to plot the Bode plots from an equation or circuit.
4. Amplifiers:
  - a) Understand how to apply Amplifier models (voltage, current, etc.) to multistage amplifiers
  - b) Analyze single input Amplifier (with model) for transfer function.
  - c) Analyze amplifier's gain in different configurations (inverting, noninverting, voltage follower)
  - d) Understand frequency response of amplifiers for single amplifiers
  - e) Compensation of real op-amp imperfections (Slew Rate, Clipping, Input bias currents, Voltage offset, frequency limitations, finite gain)
5. Diodes:
  - a) Analyze diode circuit using ideal model
  - b) Analyze diode circuit using constant voltage drop model
  - c) Analyze diode circuit with both DC and AC signals

1. (a) Find  $I_{D1}$ ,  $I_{D2}$ ,  $I_{D3}$ , and  $V_o$  using constant voltage drop method with  $n=1$ ,  $V_T=25mV$ , and  $V_{D0}=0.7V$ .

(b) Find  $V_{o\text{total}}$  if the 3V source has  $0.5\sin(\omega t)$  noise.



D3 has to be off because current is not allowed to flow in that direction through the diode.

$$+9 - I_{D2}(5.3k) - 3.7 = 0$$

$$I_{D2} = \frac{(9-3.7)}{5.3k} = \frac{5.3}{5.3k} = 1\text{mA} \checkmark > 0$$

$\therefore D2 \text{ on}$

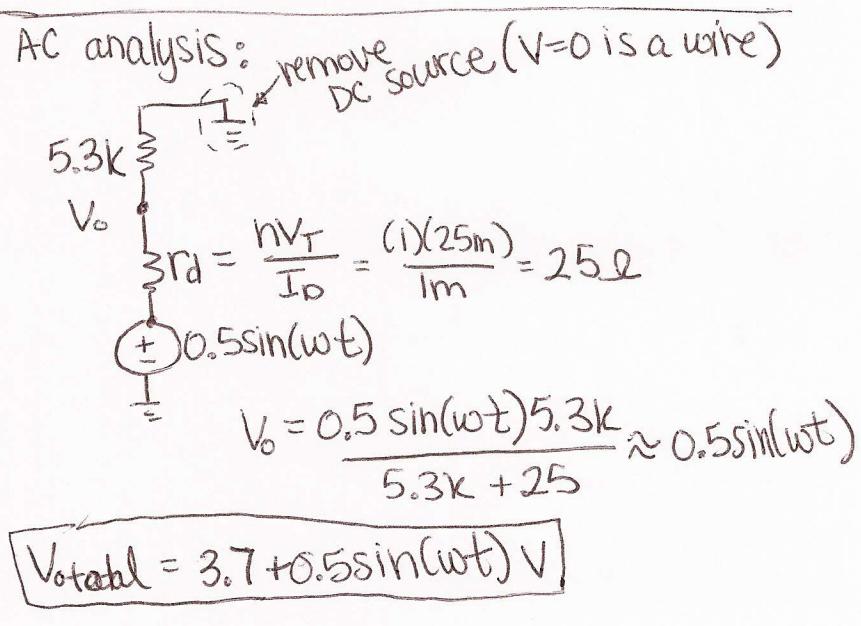
Check  $V_{D1}$ :

$$+3V + 0.7 - V_{D1} - 9 = 0$$

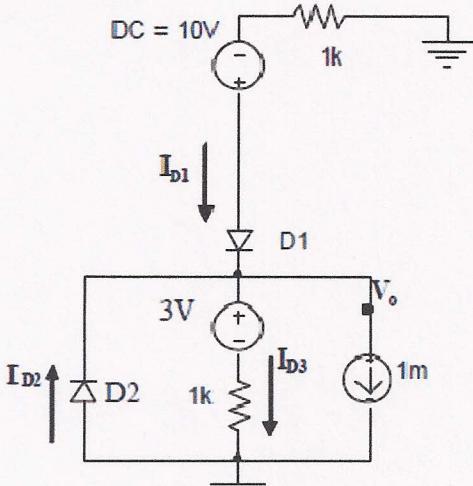
$$V_{D1} = -9 + 3 + 0.7 = -5.3 < 0.7$$

$\therefore D1 \text{ is off}$

$I_{D1} = 0$
$I_{D2} = 1\text{mA}$
$I_{D3} = 0$
$V_o = 3.7V$



(ii)



$$V_o = +3 + I_{D3}(1k)$$

V-loop:

$$-I_{D1}(1k) + 10 - 0.7 - 3 - I_{D3}(1k) = 0$$

 $\sum I:$ 

$$-I_{D1} + 1m + I_{D3} = 0$$

$$I_{D1} = (1m + I_{D3})$$

$$\therefore -(1m + I_{D3})(1k) + 6.3 - I_{D3}(1k) = 0$$

$$-1(-2k)I_{D3} + 6.3 = 0$$

$$I_{D3} = \frac{6.3}{2k} = 2.65m > 0$$

$$\therefore \underline{\text{D3 is on}} \Rightarrow V_o = 3 + 1k(2.65m)$$

$$V_o = 5.65V$$

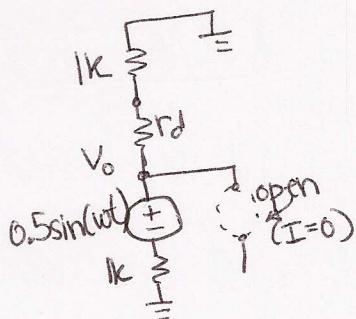
Check  $V_{D2}$ :

$$-V_{D2} - 3 - I_{D3}(1k) = 0$$

$$V_{D2} = -3 - 1k(2.65m) = -5.65 \angle 0.7$$

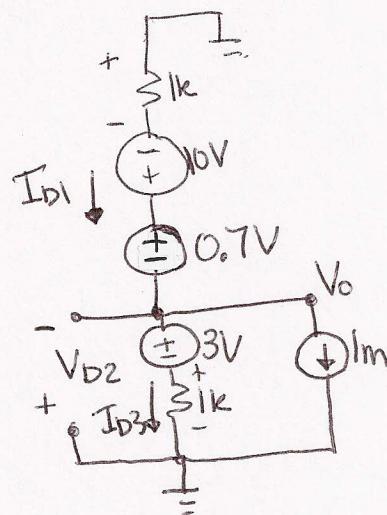
D2 is off

$$\text{AC analysis: } r_d = \frac{hV_T}{I_{D3}} = \frac{(1)(25m)}{2.65m} = 9.4\Omega$$



$$V_{o_{AC}} = \frac{0.5\sin(\omega t)(1k + r_d)}{(1k + r_d) + 1k} \approx 0.25\sin(\omega t)$$

$$V_{o_{total}} = 5.65 + 0.25\sin(\omega t)$$



OR use node-voltage

$$\frac{V_o - 3}{1k} + 1m + \frac{-(10 - 0.7) + V_o}{1k} = 0$$

$$V_o \left( \frac{1}{1k} + \frac{1}{1k} \right) = 3m - 1m + 9.3m$$

$$\therefore V_o = \frac{11.3m}{\frac{1}{1k} + \frac{1}{1k}} = 5.65V$$

$$I_{D3}(1k) + 3 - V_o = 0$$

$$I_{D3} = \frac{V_o - 3}{1k} = \frac{2.65}{1k} = 2.65m$$

$$I_{D1} = 1m + 2.65m = 3.65m$$

$$I_{D2} = 0$$

$$I_{D3} = 2.65m$$

$$V_o = 5.65V$$

2. Sketch Bode Plots for (a)  $H(s) = \frac{-1 \times 10^{14}(s^2)}{(s+10k)^2(s+100k)}$   $\Rightarrow$  Standard Form:  $\frac{-1 \times 10^{14}(s^2)}{(10k)^2\left(\frac{s}{10k}+1\right)^2(100k)\left(\frac{s}{100k}+1\right)}$

critical frequencies: origin( $\times 2$ ),  $10k(\times 2)$ ,  $100k$

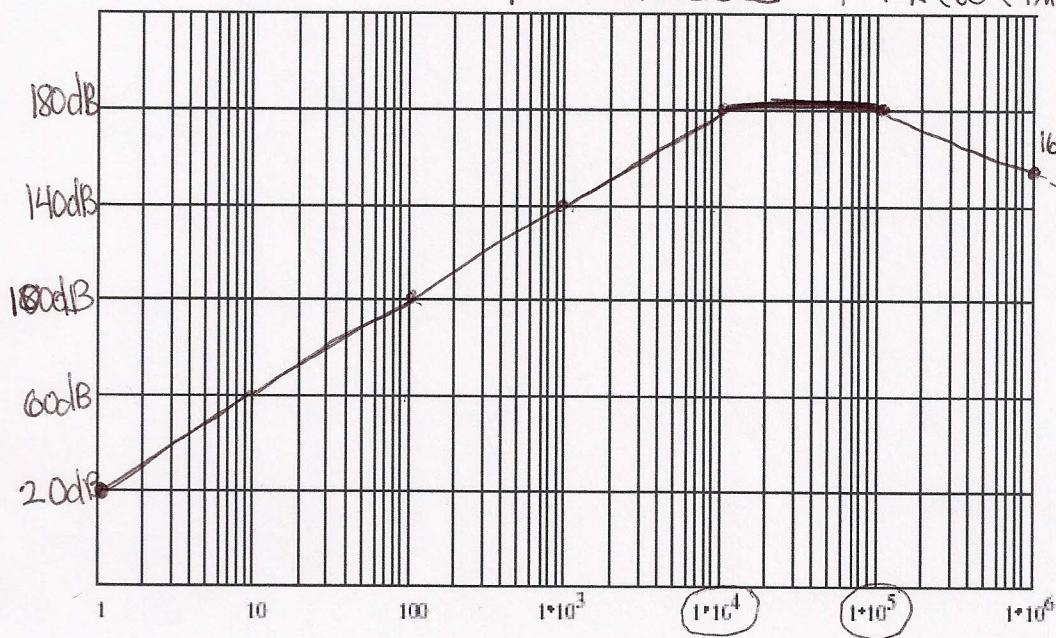
at  $\omega=1$ :  $\frac{-1 \times 10^{14}(1)}{(10k)^2(100k)\left(\sqrt{\frac{1}{10k}+1}\right)^2\left(\frac{1}{100k}+1\right)} = 10^{\sqrt{2}} \Rightarrow 20 \log(10) = \underline{20 \text{dB}}$

magnitude:

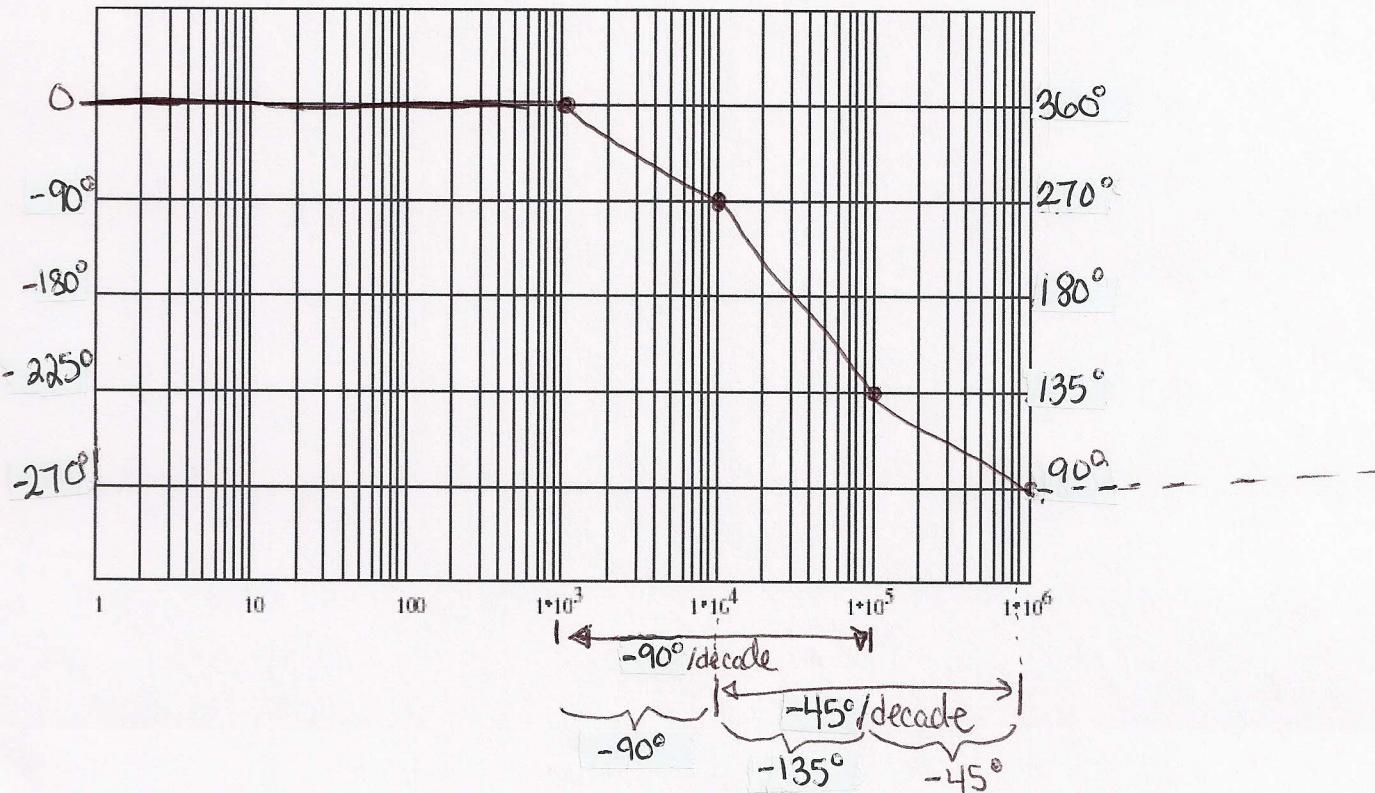
2 at origin	zeroes	+2 (+20dB/decade)
$10k(\times 2)$	pole	$2(-20 \text{dB/decade})$
$100k$	pole	-20dB/decade

phase:

(-) sign starts phase at $\pm 180^\circ$		
2 at origin	zeroes	$2(+90) = +180^\circ$
$10k < \omega < 100k$	pole	$< -45^\circ/\text{decade} \times 2$
$100k < \omega < 1 \text{ Meg}$	pole	-45°/decade

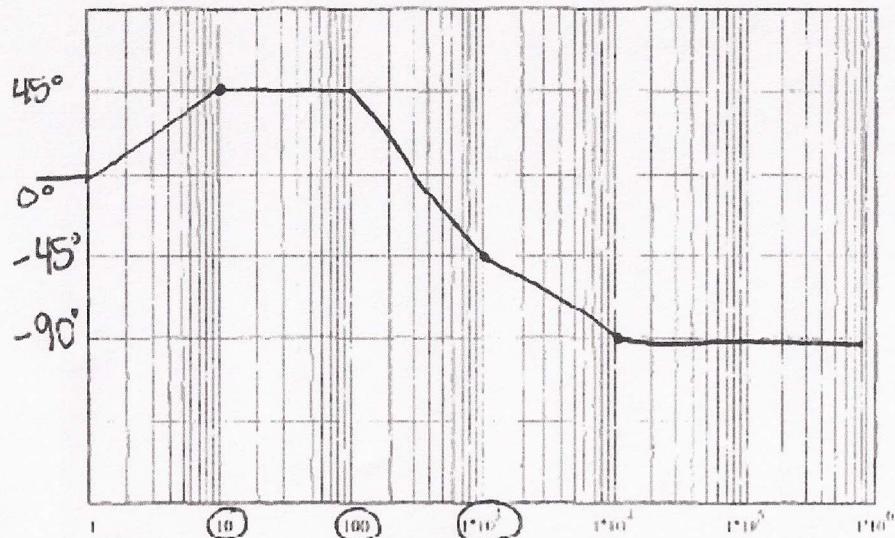
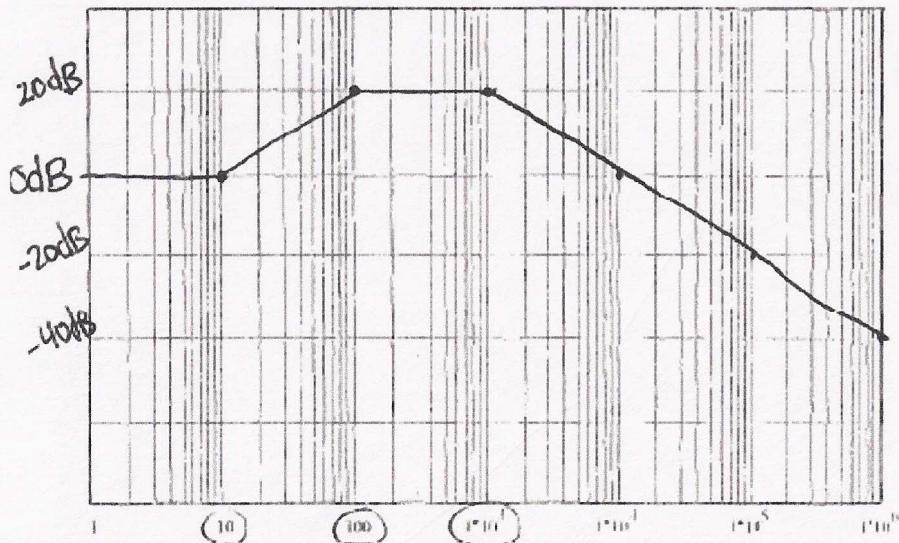


Bandwidth is from 10k to 100k



$$4. b \quad H(s) = \frac{10k(10)(\frac{s}{10} + 1)}{1k(\frac{s}{1k} + 1)(100)(\frac{s}{100} + 1)} = \frac{(\frac{s}{10} + 1)}{(\frac{s}{1k} + 1)(\frac{s}{100} + 1)}$$

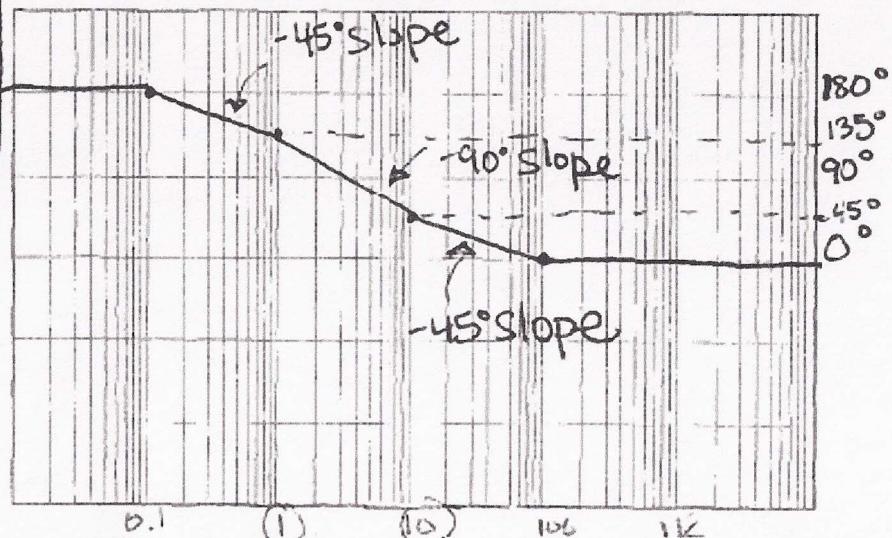
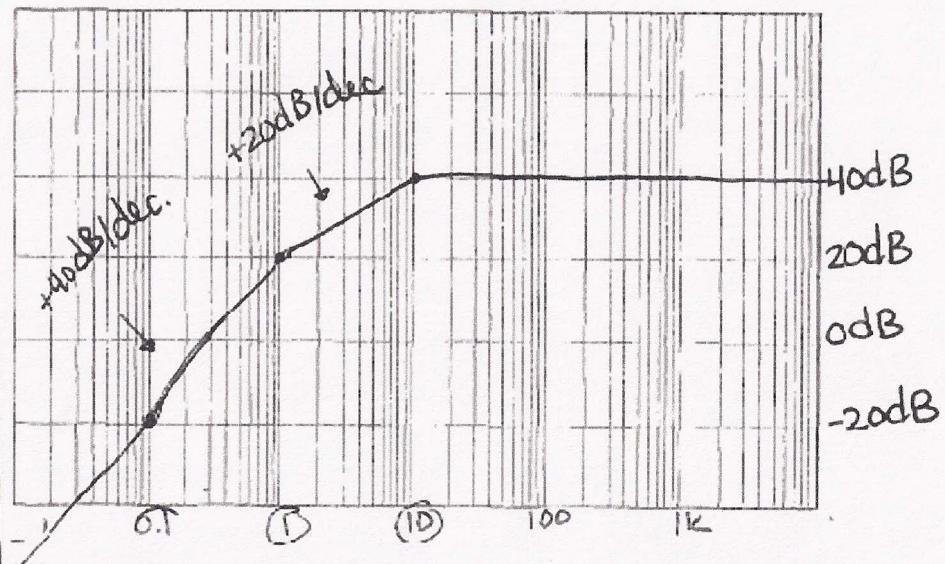
Break freq:  $\omega = 10, 100, 1k$



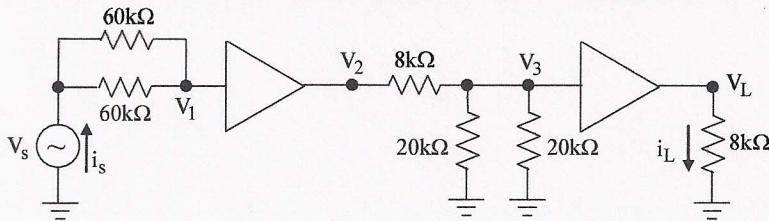
$$5. H(s) = \frac{100s^2}{(s+1)(10)(\frac{s}{10} + 1)} = \frac{10s^2}{(s+1)(\frac{s}{10} + 1)}$$

Break freq:  $\omega = 1, 10$

phase starts at  $180^\circ$   
slope starts at +40dB/dec.  
 $\Rightarrow \omega=1 \Rightarrow 20 \log(10) = 20 \text{dB}$



3. (a) Use voltage amplifier model to find voltage gain.  $A_{vo}=20$ ,  $R_i=10k\Omega$ ,  $R_o=2k\Omega$ .



whole number]

a)  $30k = (60k \parallel 60k)$

$V_s$   $\xrightarrow{i_s}$   $V_1 \xrightarrow{10k} 20V_1$   $V_2 \xrightarrow{2k} 8k$   $V_3 \xrightarrow{10k} 10k$   $V_L \xrightarrow{2k} 8k$

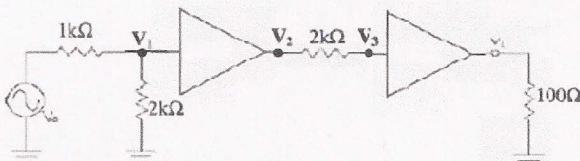
b)  $V_L = \frac{8k \cdot 20V_3}{10k} = \frac{4}{5} \cdot 20V_3$   $\frac{V_L}{V_s} = \left(\frac{4}{5}\right)(20)\left(\frac{1}{3}\right)(20)\left(\frac{1}{4}\right)$

$V_3 = \frac{5k \cdot 20V_1}{5k+10k} = \frac{1}{3} \cdot 20V_1$   $\frac{V_L}{V_s} = \frac{4}{3}(20) = \boxed{\frac{80V}{3}/V \approx 29dB}$

$V_1 = \frac{V_s \cdot 10k}{40k} = \frac{1}{4} \cdot V_s$

(b)

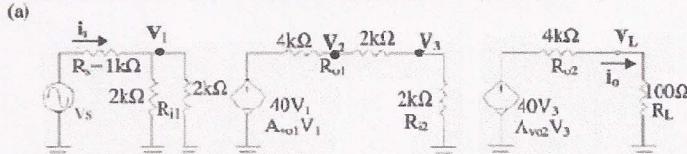
- Both amplifiers have the following characteristics:  
 $A_{vo}=40$        $R_i=2k\Omega$        $R_o=4k\Omega$       Clipping levels:  $L=\pm 12V$  (unloaded)



- (a) Redraw this 2 stage amplifier using the amplifier model. Make sure to label  $V_1$ ,  $V_2$ ,  $V_3$ , and  $V_L$  on the schematic.

(b) Find  $A_v = \frac{V_L}{V_s}$ . Express your answer as a ratio(V/V) and in dB. [Round the answer to a whole number]

- (c) For the input  $V_s = \sin(\omega t)$ . State the maximum output value at  $V_L$ .



(b) Find  $A_v = \frac{V_L}{V_s}$ . Express your answer as a ratio(V/V) and in dB. [Round to a whole number]

$$\frac{V_L}{V_s} = A_v = \left( \frac{R_{i1}}{R_s + R_{i1}} \right) \cdot A_{v1} \cdot \left( \frac{R_{o1}}{R_{o1} + R_{i2}} \right) \cdot A_{v2} \cdot \left( \frac{R_L}{R_{o2} + R_L} \right)$$

$$R_{i1} = \frac{1}{\frac{1}{2k} + \frac{1}{2k}} = 1k \quad R_{o1} = 4k + 2k - 6k$$

$$A_{v1} = \frac{1k \cdot 4k}{1k + 1k} \cdot \frac{2k}{2k + 2k} \cdot \frac{40}{40} = 4.878 \quad (\text{rounded } = 5V/V \text{ or } 20 \log(4) = 13.970)$$

(c) Amplitude  $\Rightarrow$  maximum output will be the peak value:  $\text{gain} = V_L/V_s = 5V/V \quad V_L = 5 \cdot V_{s\_peak} = \pm 5V$

Checking for clipping  $\Rightarrow \frac{V_2}{V_s} = \frac{40 \cdot (1k)}{2k} = \frac{4k}{8k} = 0.5V < \pm 2V$   
 $\rightarrow$  No clipping at 1st amp  
if  $V_s = 1V_{max}$ .

4 (a) You are given the following characteristics for a real amplifier:

Input offset voltage,  $V_{ios} = 4 \text{ mV}$

Input offset current,  $I_{ios} = 100 \text{nA}$

Input bias current,  $I_{ib} = 1 \mu\text{A}$

Input Resistance,  $R_i = 1 \text{ M}\Omega$

Output Resistance,  $R_o = 50 \Omega$

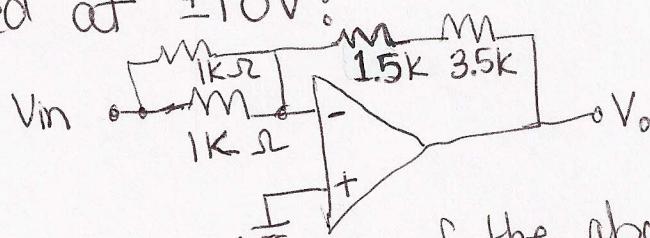
Open-loop gain,  $A_{vo} = 180 \text{ dB}$

Unity-gain bandwidth,  $f_T = 11 \text{ MHz}$

Output swing limits, within 2 Volts of supply

Settling Rate,  $SR = 2 \frac{\text{V}}{\mu\text{sec.}}$

Given the following circuit with the operational amplifier powered at  $\pm 10 \text{ V}$ :



i) Find the ideal gain of the above circuit:

$$\frac{V_o}{V_{in}} = \frac{-5k}{500} = -10 \frac{\text{V}}{\text{V}} \Rightarrow 20 \text{ dB}$$

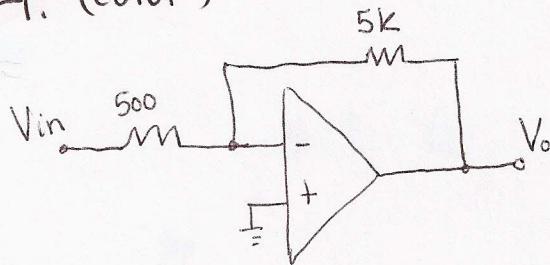
ii) For small input signals, what is the bandwidth of the circuit?

$$f_{3\text{dB}_{\text{new}}} = \frac{11 \text{ MHz}}{1 + (10)} = \underline{1 \text{ MHz}}$$

iii) If a circuit needs to operate up to 5 MHz, what is the maximum gain I can achieve using an inverting amplifier configuration?

$$f_{3\text{dB}_{\text{new}}} = 5 \text{ MHz} = \frac{11 \text{ MHz}}{(1 + \frac{R_2}{R_1})} \Rightarrow \left| \frac{R_2}{R_1} \right| = \frac{11}{5} - 1 = \underline{1.2 \frac{\text{V}}{\text{V}}}$$

4. (cont)



- iv) If the above circuit is operated to produce the maximum possible peak voltage,
- what is the bandwidth?
  - what is the maximum peak voltage value for the input?

$$a) f_{max} = \frac{SR}{V_p \cdot 2\pi} = \frac{2}{1 \times 10^6 \times (8)(2)\pi} = 39.8 \text{ KHz}$$

b) maximum peak voltage = +8V output max.

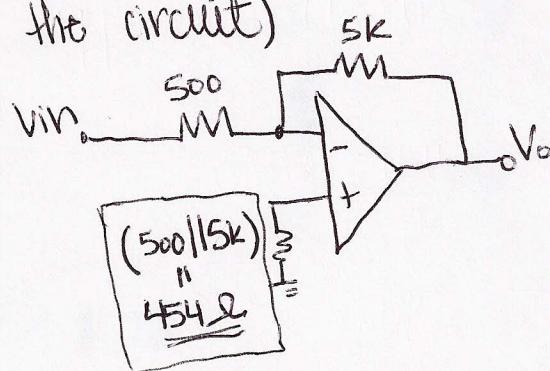
Using ideal gain =  $-10 \frac{V_o}{V_i}$

$$\left| \frac{V_o}{V_i} \right| = 10 \Rightarrow V_i = \frac{V_o}{10} = \frac{8}{10} = 0.8 \text{ V max}$$

- v) Find the effect of the input offset voltage when  $V_{in}=0V$ .

$$V_{out} = 4m(-10) = -40 \text{ mV}$$

- vi) How should the circuit above be modified to minimize the effect of the input bias current?  
(Draw the circuit)



4. (b) You are given the following characteristics for a real amplifier:

Input offset voltage,	$V_{ios}=3\text{mV}$
Input offset current,	$I_{ios}=100\text{nA}$
Input bias current,	$I_{ib}=1\mu\text{A}$
Input Resistance,	$R_i=1\text{M}\Omega$
Output Resistance,	$R_o=50\Omega$
Open-loop gain,	$A_{vo}=180\text{dB}$
Unity-gain bandwidth,	$f_T=15\text{MHz}$
Output swing limits,	within 2Volts of power supply
Slew Rate,	$\text{SR}=4 \frac{\text{V}}{\mu\text{sec}}$

If a circuit needs to operate at 5MHz using the above specifications for an operational amplifier and having a power supply of  $\pm 15\text{V}$ , what is the maximum gain that can be achieved using an:

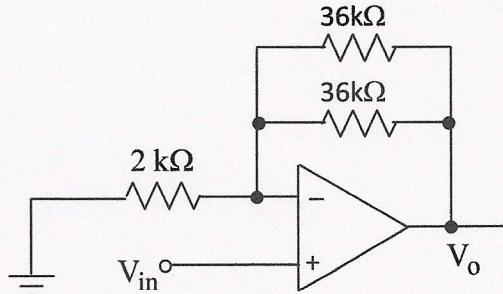
- i) Inverting amplifier

$$f_{3\text{dB}_{\text{new}}} = \frac{15 \times 10^6}{\left(1 + \frac{R_2}{R_1}\right)} = 5\text{MHz} \Rightarrow -\frac{R_2}{R_1} = -\left(\frac{15 \times 10^6}{5 \times 10^6} - 1\right) = -2 \frac{\text{V}}{\text{V}}$$

- ii) Non-inverting amplifier

$$f_{3\text{dB}_{\text{new}}} = \frac{15 \times 10^6}{\left(1 + \frac{R_2}{R_1}\right)} = 5\text{MHz} \Rightarrow \left(1 + \frac{R_2}{R_1}\right) = \frac{15 \times 10^6}{5 \times 10^6} = 3 \frac{\text{V}}{\text{V}}$$

Given the following circuit with the operational amplifier powered at  $\pm 15\text{V}$ .



- i) Find the ideal gain of the above circuit:

$$\frac{V_o}{V_{in}} = \left(1 + \frac{18k}{2k}\right) = 10 \frac{\text{V}}{\text{V}}$$

- ii) For small input signals, what is the bandwidth of the circuit

$$f_{3\text{dB}_{\text{new}}} = \frac{15 \times 10^6}{10} = 1.5\text{MHz}$$

- iii) What is the bandwidth when the circuit is operated to produce the maximum possible peak voltage value?

$$f = \frac{4}{10^{-6}(13 \cdot \pi)} = 98\text{kHz}$$

- iv) What is the maximum amplitude of the input ?

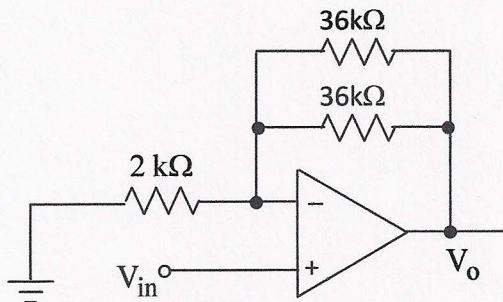
$$\frac{V_o}{V_{in}} = 10 \frac{\text{V}}{\text{V}} \Rightarrow V_{in} = \frac{V_o}{10} = \frac{13}{10} = 1.3\text{V}$$

- v) For  $V_{in}=0.001\sin(2\pi 90kt)$ , what is the ideal value for the peak to peak voltage value at the output?

$$V_{o_{pp}} = 2(0.001)(10) = 0.02\text{V}_{pp}$$

- vi) For  $V_{in}=0.001\sin(2\pi 90kt)$ , what is the peak to peak voltage value at the output considering the input offset voltage?

$$V_{o_{pp}} = 2(0.001)(10) + 3m(10) = .02 + .030 = .05\text{V}_{pp}$$



- vii) How should the circuit above be modified to minimize the effect of the input bias current? Draw the schematic of the modified circuit and state values of added component(s).

