

Introduction to Massive MIMO

Communication system concepts

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Outline

- Introduction
- How to Achieve Higher Spectral Efficiency
- Basic Properties of Massive MIMO
- Massive MIMO Transmission Protocol
- Summary

INTRODUCTION

Incredible Success of Wireless Communications

Martin Cooper's law

The number of voice/data connections has doubled every 2.5 years (+32% per year) since the beginning of wireless

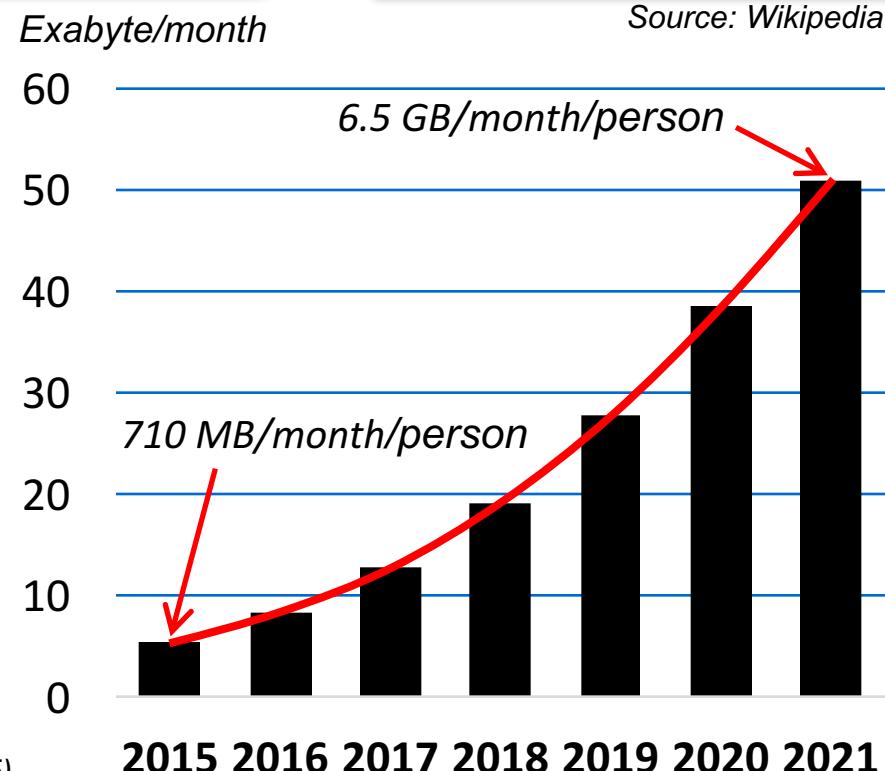
Last 45 years: 1 Million increase in wireless traffic

Two-way radio, FM radio, satellites, cellular, WiFi, etc.



Source: Wikipedia

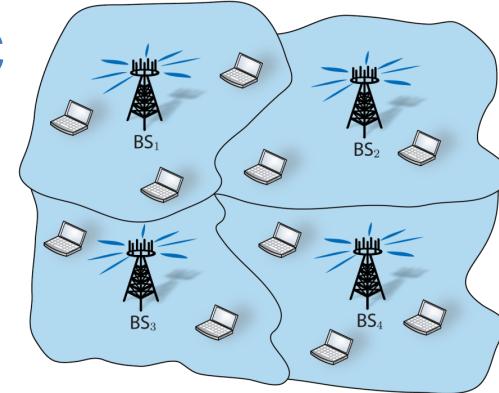
- Future Network Traffic Growth
 - 45% annual data traffic growth
 - Slightly faster than in the past!
 - Exponential increase
 - Extrapolation: 6x in 5 years
40x in 10 years
260x in 15 years



Evolving Networks for Higher Traffic

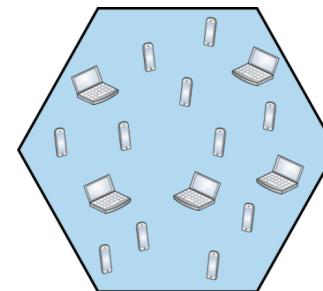
Cellular networks

Coverage area divided into cells
Users served by a base station



- Increase Network Throughput [bit/s/km²]

- Consider a given area →



- Simple Formula for Network Throughput:

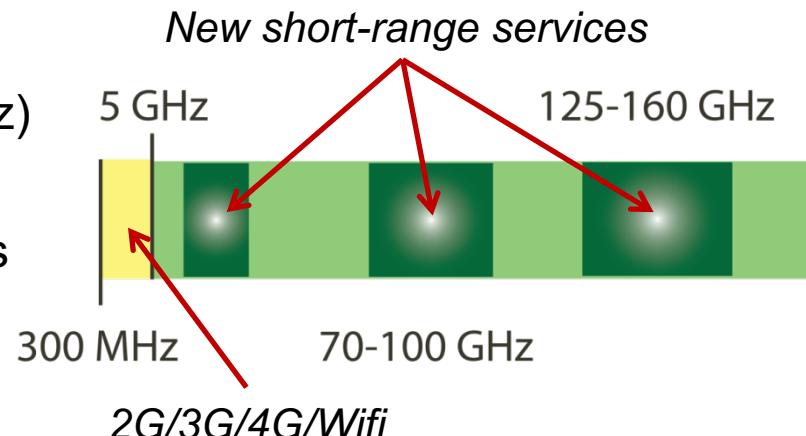
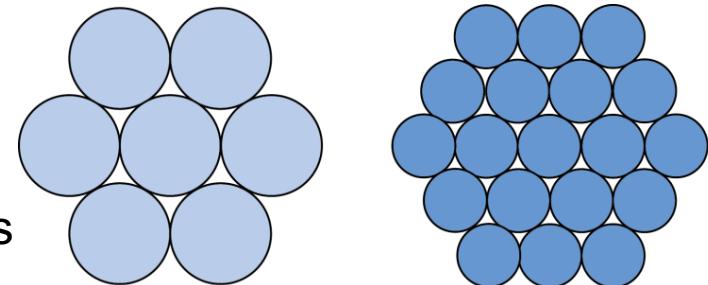
$$\text{Throughput} = \underbrace{\text{Cell density}}_{\text{bit/s/km}^2} \cdot \underbrace{\text{Available spectrum}}_{\text{Cell/km}^2} \cdot \underbrace{\text{Spectral efficiency}}_{\text{Hz}} \cdot \underbrace{\text{Spectral efficiency}}_{\text{bit/s/Hz/Cell}}$$

- Ways to achieve 1000x improvement:

	Higher cell density	More spectrum	Higher spectral efficiency
Nokia (2011)	10x	10x	10x
SK Telecom (2012)	56x	3x	6x

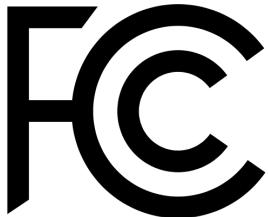
Conventional Solutions

- Higher Cell Density
 - Traditional way to improve throughput
 - Cut cell radius by $z \rightarrow z^2$ times more cells
 - Issues: High rent and deployment costs
Interference is getting worse
WiFi + Cellular is already dense: *Coverage is the issue!*
- More Spectrum
 - Range suitable for coverage: < 5 GHz
 - Already allocated for services!
(Cellular: ~550 MHz, WiFi: ~550 MHz)
 - Far above 5 GHz: High propagation losses \rightarrow Mainly short-range hotspots



Higher Spectral Efficiency

“Imagine that we decided to reward the first person who finds a way to make spectrum use below 5 GHz 50 or 100 times more efficient over the next decade. The reward could be something simple—say 10 megahertz of spectrum suitable for mobile broadband.”



FCC Commissioner Jessica Rosenworcel
Marconi Society Anniversary Symposium, Oct. 2, 2014.

Price of sub-5 GHz Spectrum

January 2015: FCC sold 65 MHz at 1.7-2.1 GHz for \$45 billion

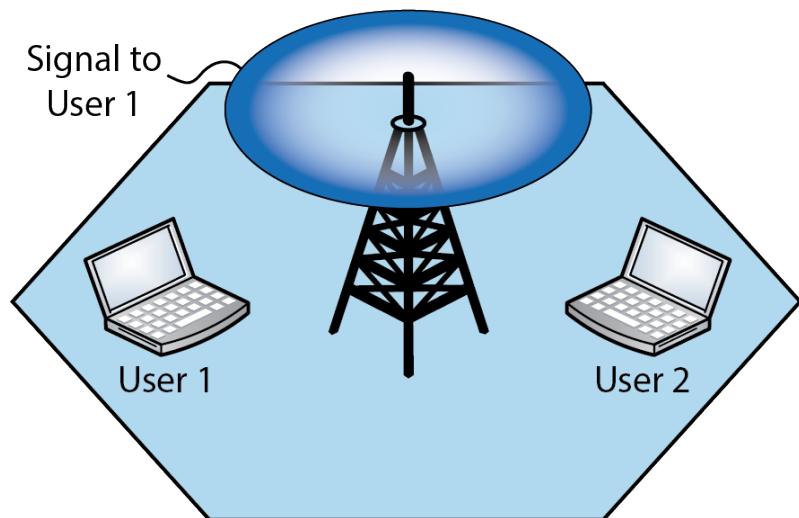
Can FCC's 50× goal be reached?

How to achieve

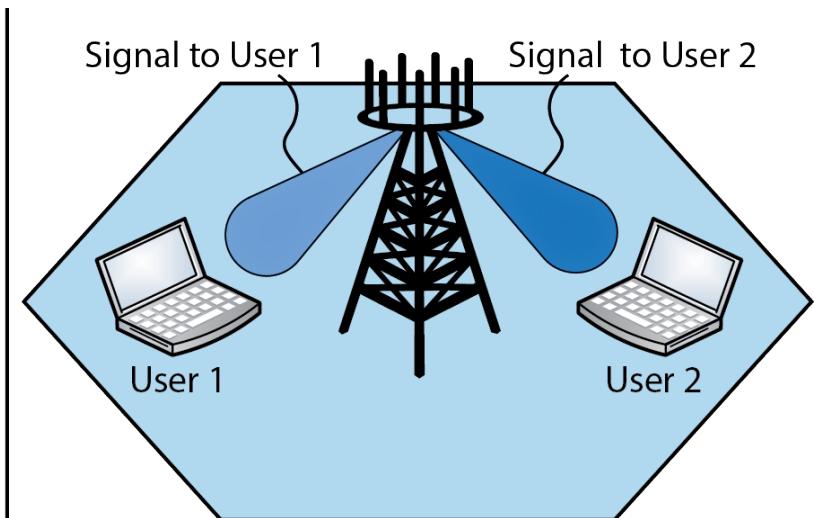
HIGHER SPECTRAL EFFICIENCY

Higher Spectral Efficiency

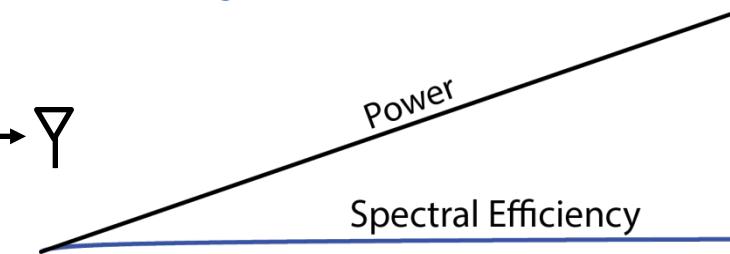
- Point-to-Point Spectral Efficiency: $\Uparrow \rightarrow \Uparrow$
 - Governed by Shannon's capacity limit:
$$\log_2 \left(1 + \frac{\text{Received Signal Power}}{\text{Interference Power} + \text{Noise Power}} \right) \text{ [bit/s/Hz/user]}$$
 - Issue: 4 bit/s/Hz \rightarrow 8 bit/s/Hz requires 17 \times more power!
- Many Parallel Transmissions: *Spatially focused to each desired user*



Single-Antenna Transmission

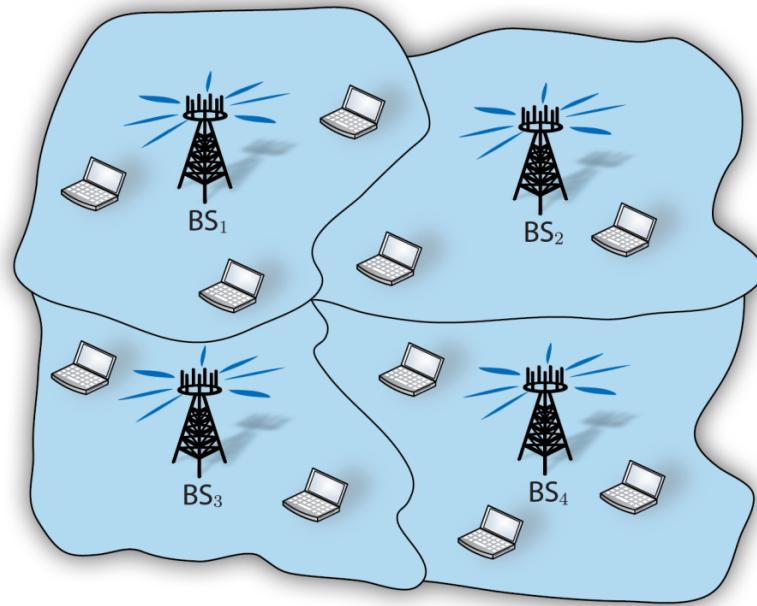


Multi-Antenna Transmission



Multi-User MIMO (Multiple-input Multiple-output)

- Cellular Multi-User MIMO
 - Base stations (BSs) with M antennas
 - Parallel uplink/downlink for K users
 - Channel coherence block: τ_c symbols
- Theory: Hardware is Limiting
 - Spectral efficiency roughly prop. to
$$\min\left(M, K, \frac{\tau_c}{2}\right)$$
 - $2\times$ improvement = $2\times$ antennas and users $(\tau_c \in [100, 10000])$
- Practice: Co-User Interference is Limiting
 - Multi-user MIMO in LTE-A: Up to 8 antennas
 - Small gains: Hard to learn users' channels
Hard to coordinate BSs



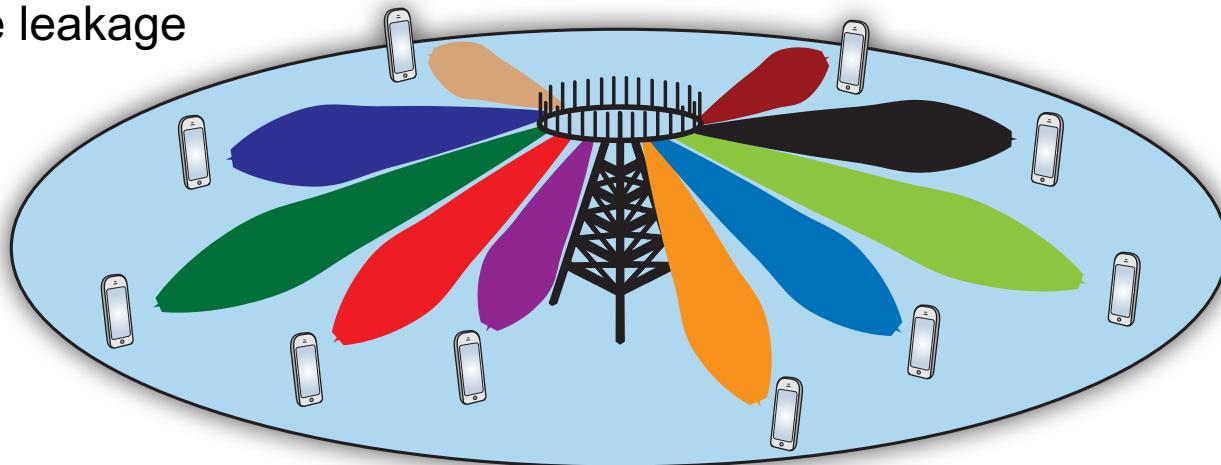
} **End of the MIMO road?**
No reason to add more antennas/users?

Taking Multi-User MIMO to the Next Level

- Network Architecture: Massive MIMO
 - Many BS antennas; e.g., $M \approx 200$ antennas, $K \approx 40$ single-antenna users
 - Key: Many more antennas than users: $M \gg K$
 - Very directive signals
 - Little interference leakage

*Spectral efficiency prop.
to number of users!*

$$\min\left(M, K, \frac{\tau_c}{2}\right) \approx K$$



- Seminal work:
 - T. Marzetta, “Noncooperative Cellular Wireless with Unlimited Numbers of Base Station Antennas,” IEEE Trans. Wireless Communications, 2010.
 - 2013 IEEE Guglielmo Marconi Prize Paper Award
 - 2015 IEEE W. R. G. Baker Award

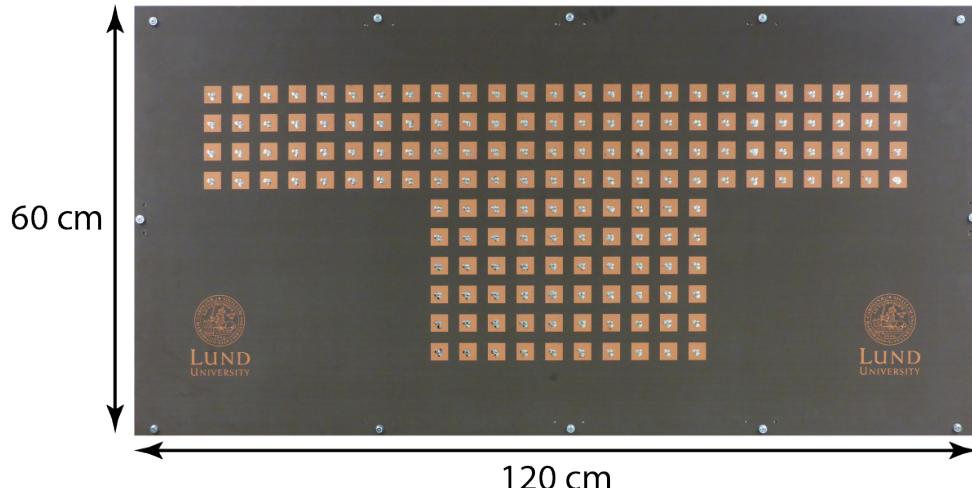
What is the Key Difference from Today?

- Number of Antennas? **No, we already have many antennas!**
 - 3G/UMTS: 3 sectors x 20 element-arrays = 60 antennas
 - 4G/LTE-A: 8-MIMO x 30 = 240 antennas

Massive MIMO Characteristics

Many fully digital steerable antennas

Massive in numbers – not massive in size



160 antenna elements, LuMaMi testbed, Lund University

Typical vertical array:

10 antennas x 2 polarizations

Only 2 antenna ports



3 sectors, 4 vertical arrays per sector

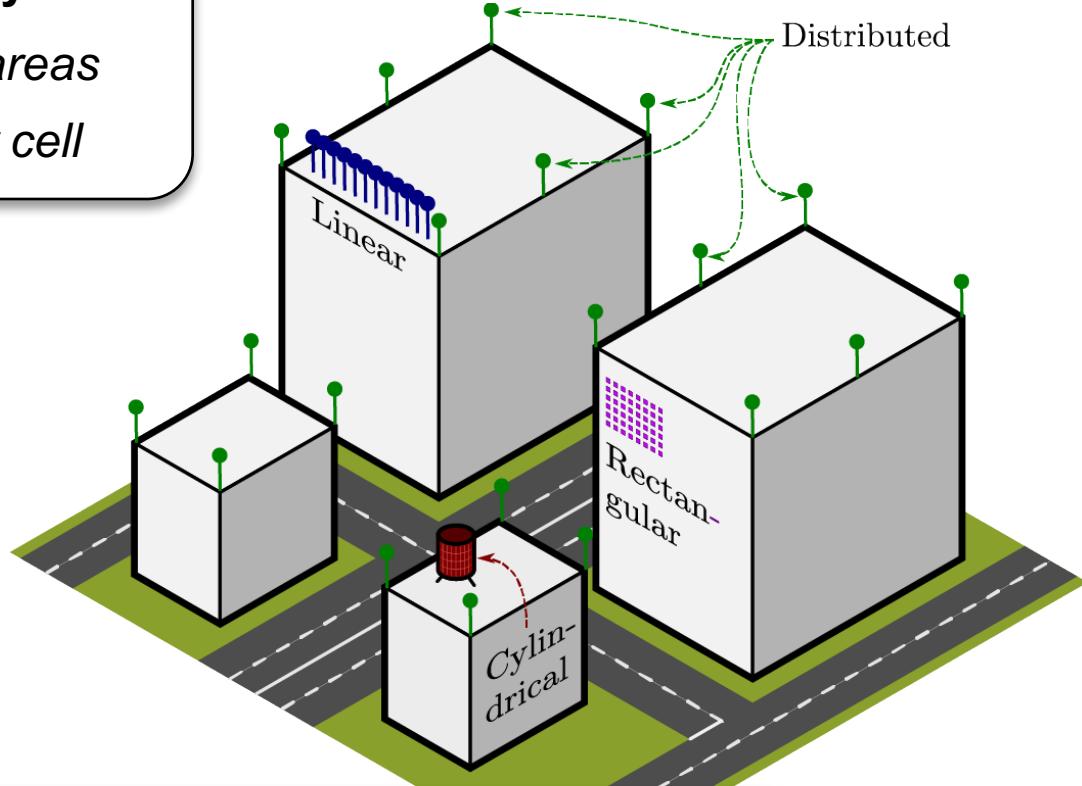
How to Deploy Massive MIMO?

Conventional Cellular Deployment

Non-overlapping coverage areas

One or multiple sectors per cell

- Co-located Deployment
 - 1D, 2D, or 3D arrays
 - No need for sectors
- Distributed Deployment
 - Remote radio heads

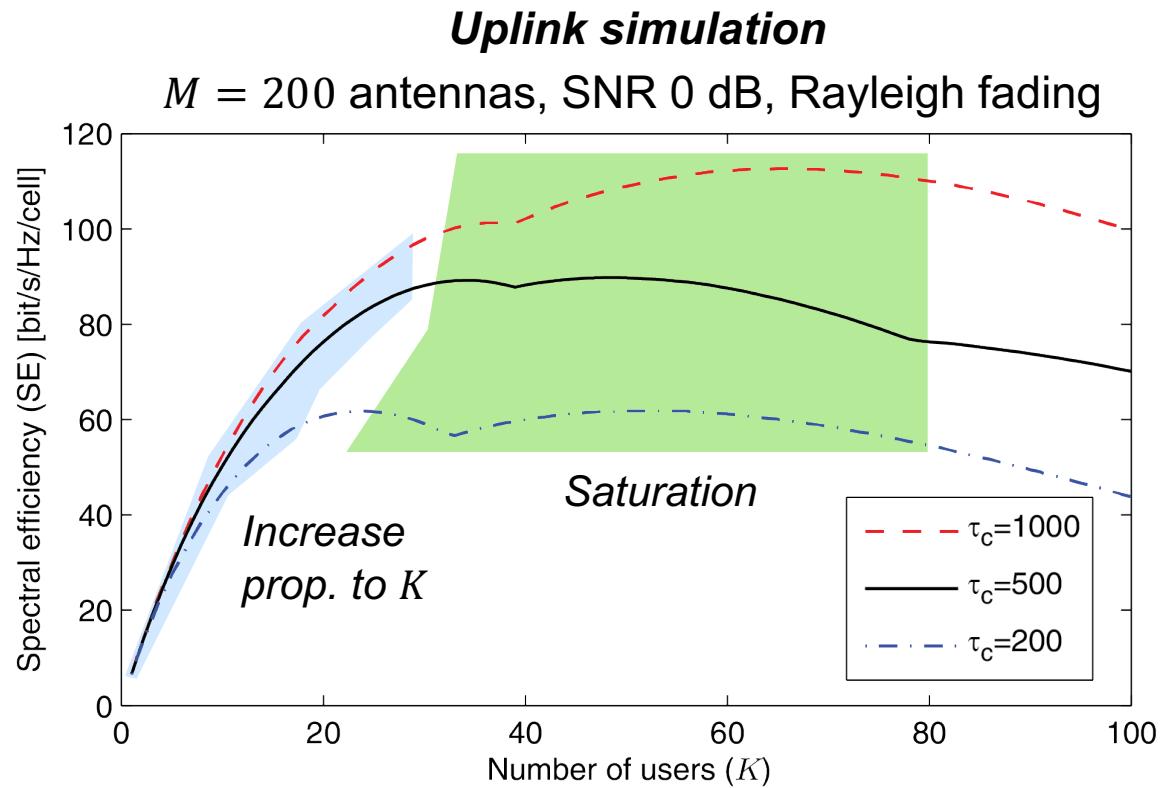
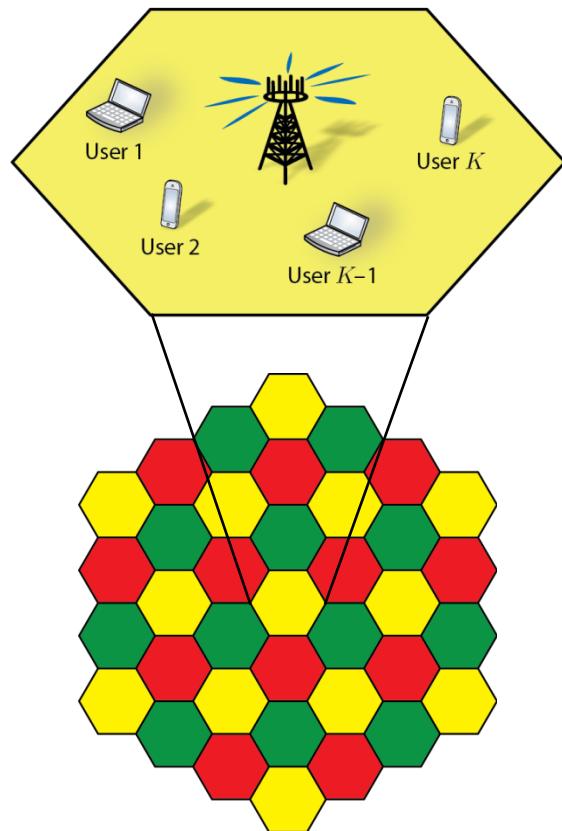


Key Benefits of Massive MIMO

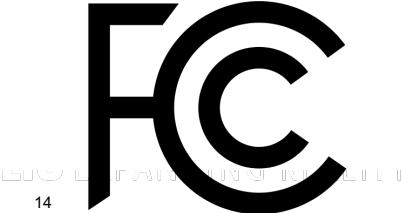
Outdoor users: Handle mobility and guarantee coverage

Indoor users: No need to put BSs inside buildings

How Much can Spectral Efficiency be Improved?



- *Baseline:* 2.25 bit/s/Hz/cell (IMT-Advanced)
- *Massive MIMO:* 25×–50× improvement
- Large coherence τ_c is key → Use lower frequencies



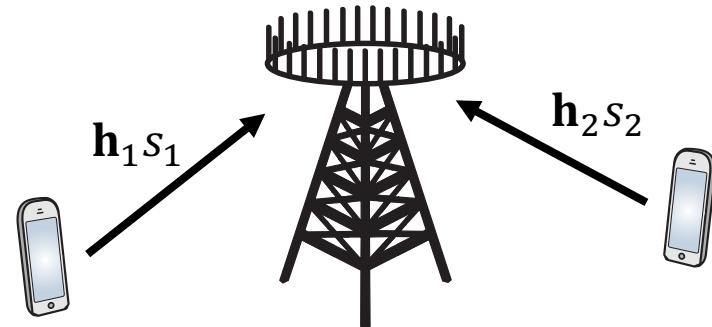
Yes, FCC's 50× goal is within reach!

Basic Properties of
MASSIVE MIMO

Asymptotic Channel Orthogonality

- Example: Uplink with i.i.d. Rayleigh Fading

- Two users, send signals s_k for $k = 1, 2$
- Channels: $\mathbf{h}_k = [h_{k1} \dots h_{kM}]^T \sim \mathcal{CN}(\mathbf{0}, \mathbf{I}_M)$
- Noise: $\mathbf{n} \sim \mathcal{CN}(\mathbf{0}, \mathbf{I}_M)$
- Received: $\mathbf{y} = \mathbf{h}_1 s_1 + \mathbf{h}_2 s_2 + \mathbf{n}$



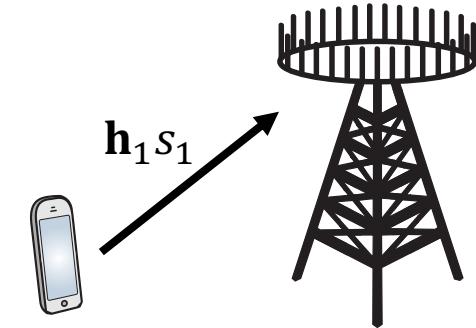
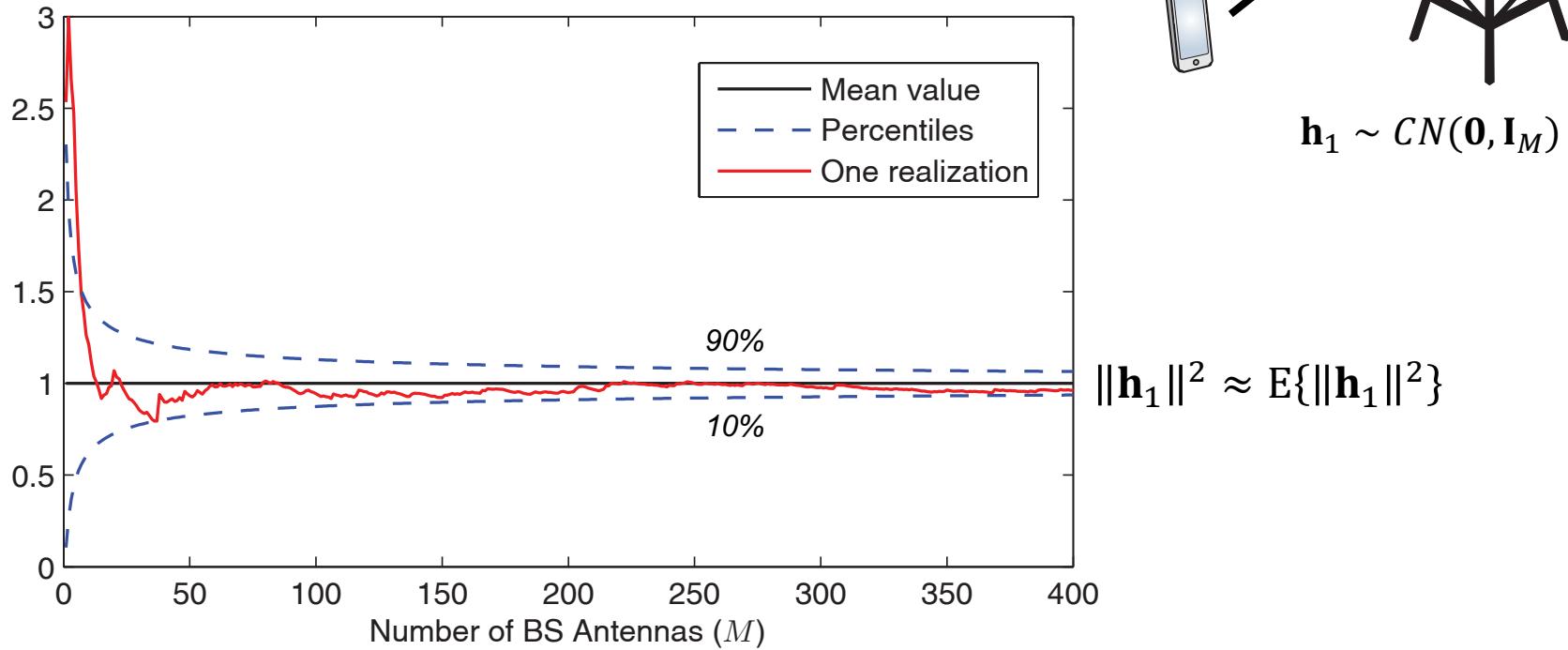
- Linear Detector \mathbf{w}_1 for User 1: $\tilde{y}_1 = \mathbf{w}_1^H \mathbf{y} = \boxed{\mathbf{w}_1^H \mathbf{h}_1} s_1 + \boxed{\mathbf{w}_1^H \mathbf{h}_2} s_2 + \boxed{\mathbf{w}_1^H \mathbf{n}}$
 - Maximum ratio filter: $\mathbf{w}_1 = \frac{1}{M} \mathbf{h}_1$
 - Signal remains: $\mathbf{w}_1^H \mathbf{h}_1 = \frac{1}{M} \|\mathbf{h}_1\|^2 \xrightarrow{M \rightarrow \infty} \text{E}[|h_{11}|^2] = 1$
 - Interference vanishes: $\mathbf{w}_1^H \mathbf{h}_2 = \frac{1}{M} \mathbf{h}_1^H \mathbf{h}_2 \xrightarrow{M \rightarrow \infty} \text{E}[h_{11}^H h_{21}] = 0$
 - Noise vanishes: $\mathbf{w}_1^H \mathbf{n} = \frac{1}{M} \mathbf{h}_1^H \mathbf{n} \xrightarrow{M \rightarrow \infty} \text{E}[h_{11}^H n_1] = 0$

Asymptotically noise/interference-free communication: $\tilde{y}_1 \xrightarrow{M \rightarrow \infty} s_1$

Asymptotic Channel Hardening

Variations of effective channel reduce with M :

$$\frac{1}{M} \|\mathbf{h}_1\|^2 \text{ has } \begin{cases} \text{Mean: 1} \\ \text{Variance: } 1/M \end{cases}$$



$$\mathbf{h}_1 \sim \mathcal{CN}(\mathbf{0}, \mathbf{I}_M)$$

Double benefits: $\|\mathbf{h}_1\|^2$ scales with M , variations reduces

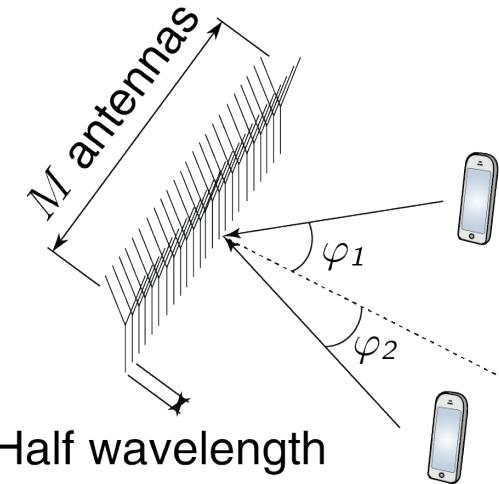
Orthogonality Only in Isotropic Fading?

- Assumptions in i.i.d. Rayleigh Fading
 - No dominant directivity
 - Very many scattering objectives

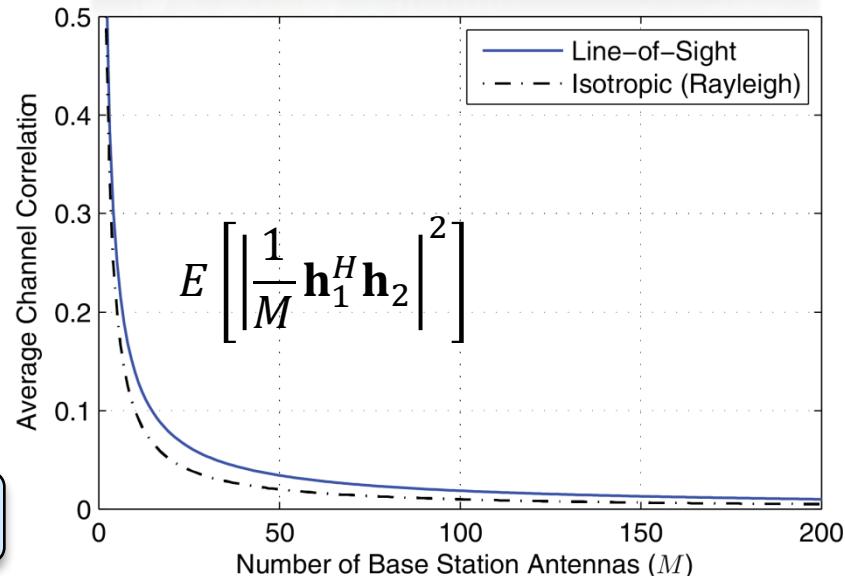
Less true as $M \rightarrow \infty$



Example: Line-of-Sight Channels



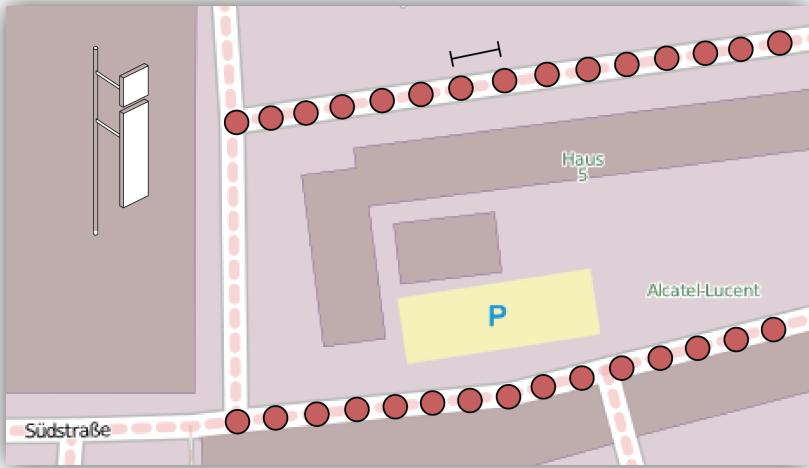
- Random user angles
- M observations:
 - Stronger signal
 - Suppressed noise
- What is $\mathbf{h}_1^H \mathbf{h}_2$?



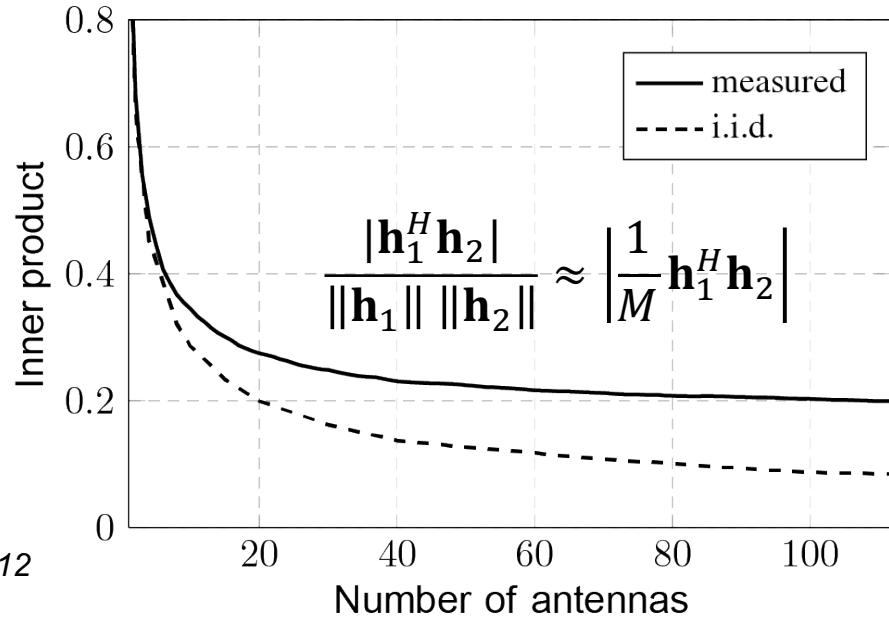
Difference: How quickly orthogonality appears

How Do Practical Channels Behave?

- Measurements show similar results



Source: J. Hoydis, C. Hoek, T. Wild, and S. ten Brink,
"Channel Measurements for Large Antenna Arrays," ISWCS 2012

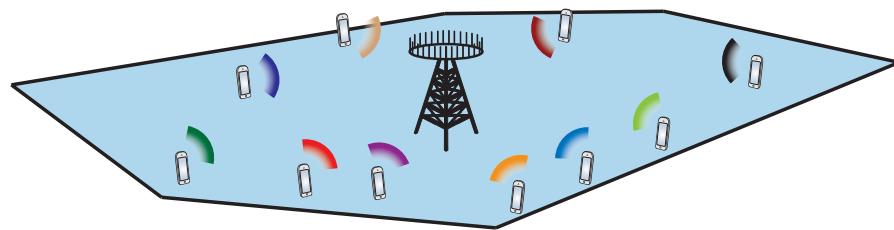


- Asymptotic Favorable Propagation: $\frac{1}{M} \mathbf{h}_1^H \mathbf{h}_2 \rightarrow 0$ as $M \rightarrow \infty$
 - Achieved in Rayleigh fading and line-of-sight – two extremes!
 - Same behavior expected and observed in practice

Massive MIMO

TRANSMISSION PROTOCOL

Classical Multi-User MIMO vs. Massive MIMO

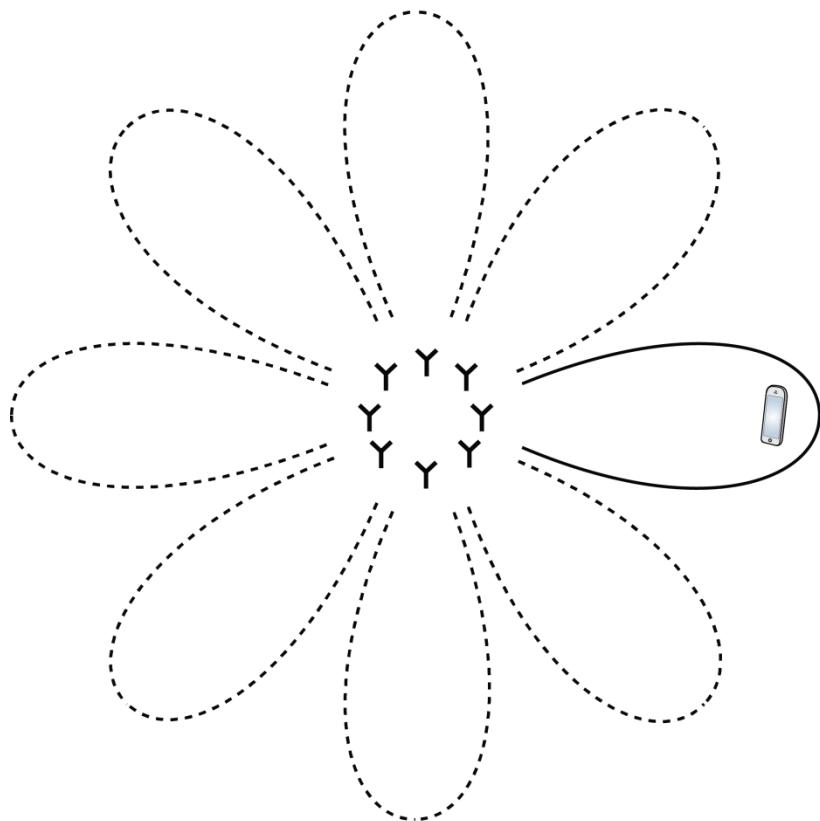


	Classic multi-user MIMO	Massive MIMO (Canonical)
Antennas M, users K	$M \approx K$	$M \gg K$
Signal processing	Non-linear is preferred	Linear is near optimal
Duplexing mode	Designed for TDD and FDD	Designed for TDD w. reciprocity
Instantaneous channel	Known at BS and user	Only needed at BS (hardening)
Channel quality	Affected by frequency-selective and fast fading	Almost no channel quality variations (hardening)
Variations in user load	Scheduling needed if $K > M$	Scheduling seldom needed
Resource allocation	Rapid due to fading	Only on a slow time scale
Cell-edge performance	Only good if BSs cooperate	Improved by array gain of M
BS cooperation	Highly beneficial if rapid	Only long-term coordination

FDD = Frequency-division duplex,

TDD = Time-division duplex

Downlink MIMO Precoding



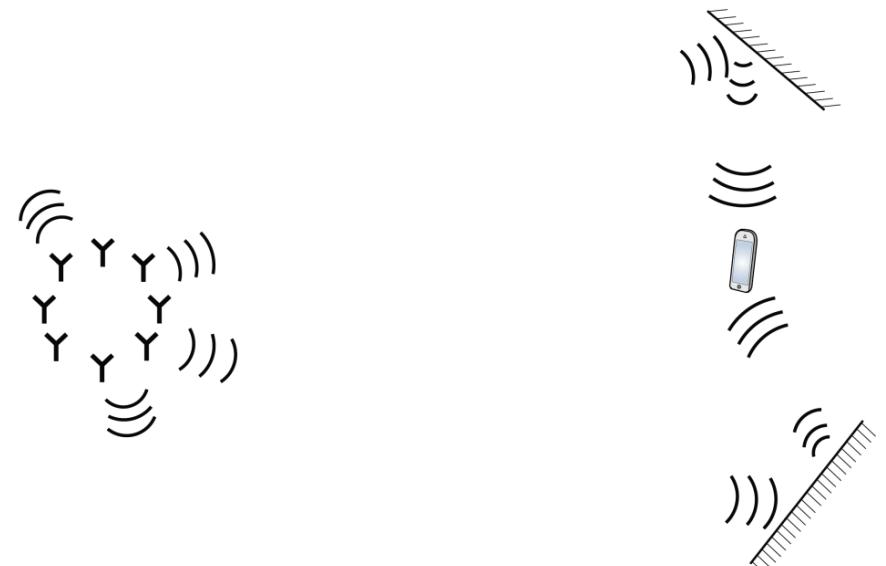
Line-of-Sight

Channels characterized by angles

1-2 parameters to estimate per user

Precoding = Angular beamforming

Same principle for MIMO detection



Non-Line-of-Sight

Rich multipath propagation

M parameters to estimate per user

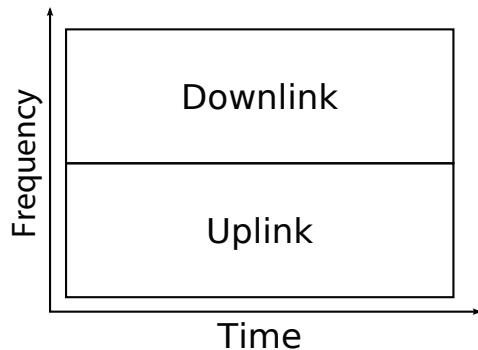
Precoding \neq Angular beamforming

How to Limit the Pilot Overhead?

Frequency-division duplex (FDD)

Downlink: M pilots + K feedback

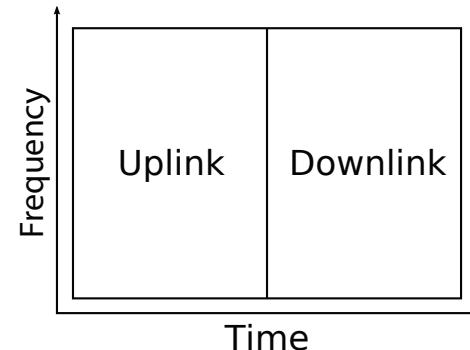
Uplink: K pilots + M feedback



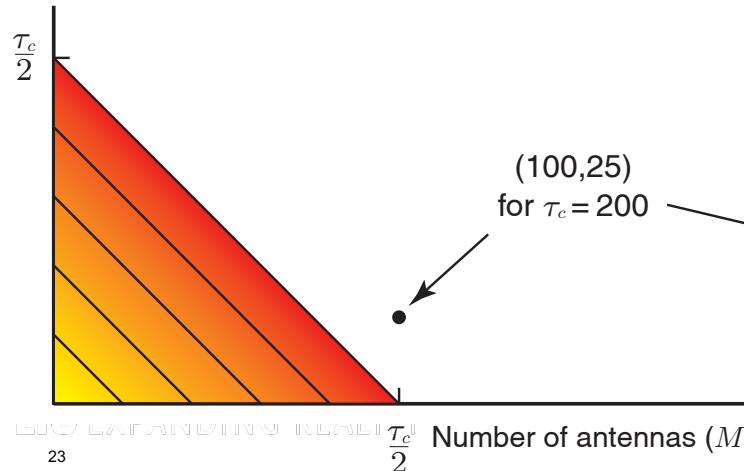
Time-division duplex (TDD)

Uplink: K pilots, exploit channel reciprocity

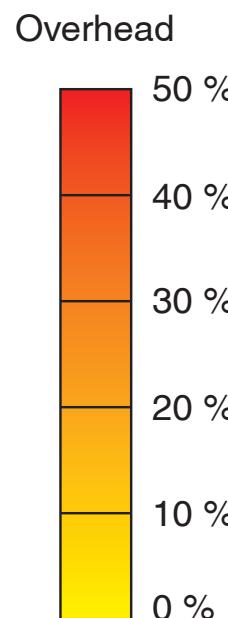
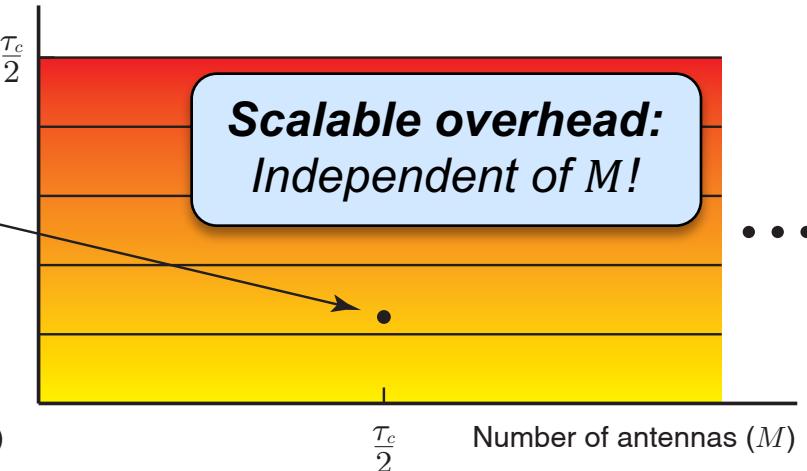
Downlink: K precoded pilots (optional)



Number of users (K)



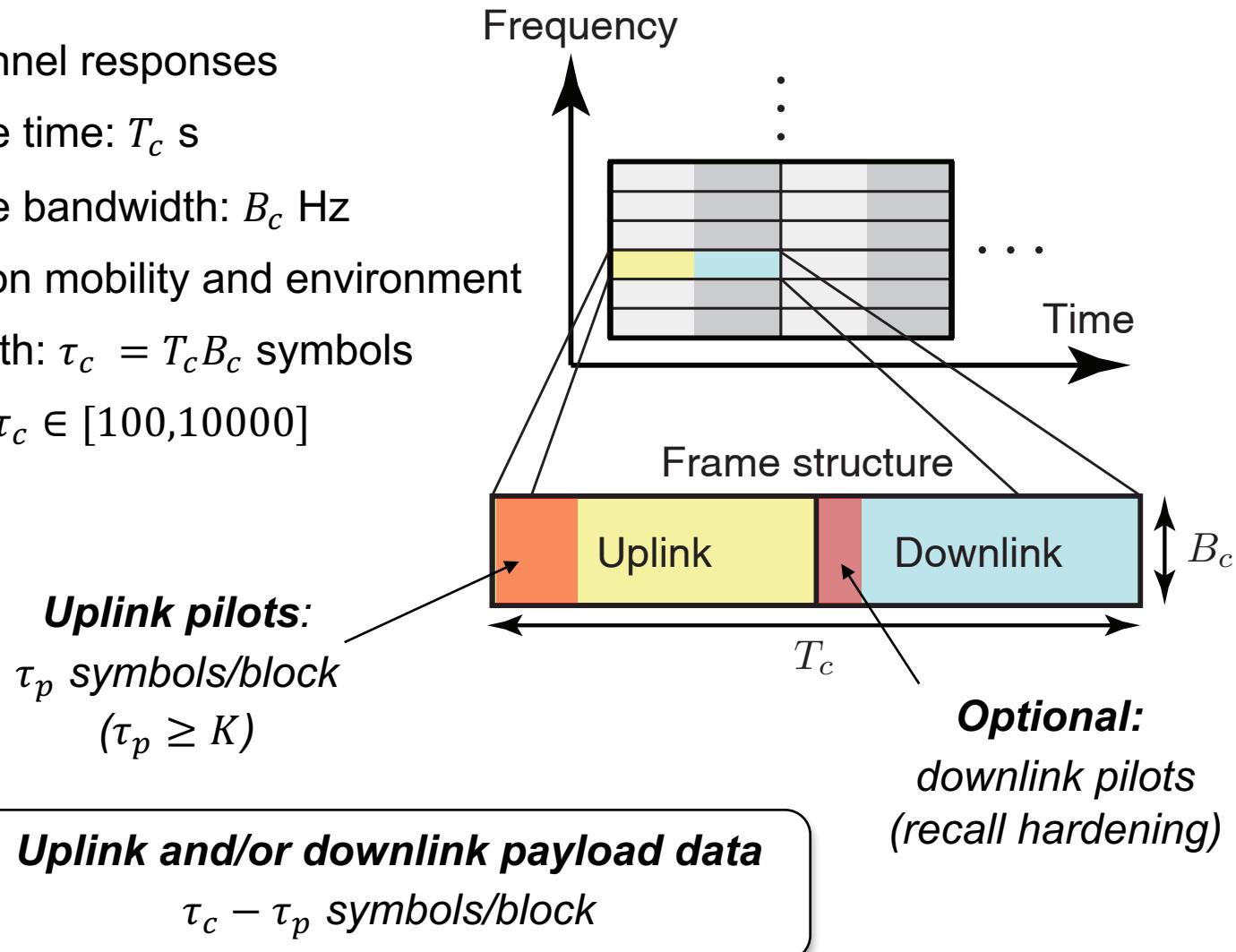
Number of users (K)



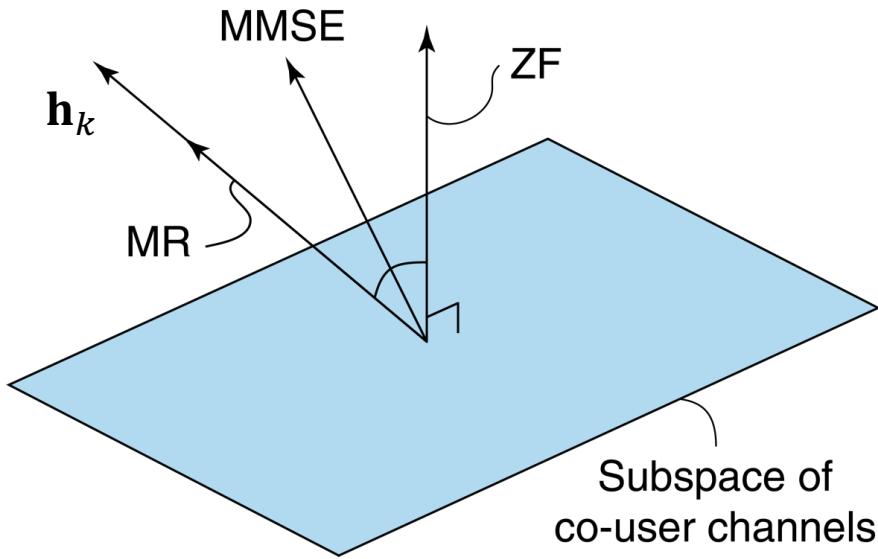
Massive MIMO TDD Protocol

- Coherence Blocks

- Fixed channel responses
- Coherence time: T_c s
- Coherence bandwidth: B_c Hz
- Depends on mobility and environment
- Block length: $\tau_c = T_c B_c$ symbols
- Typically: $\tau_c \in [100, 10000]$



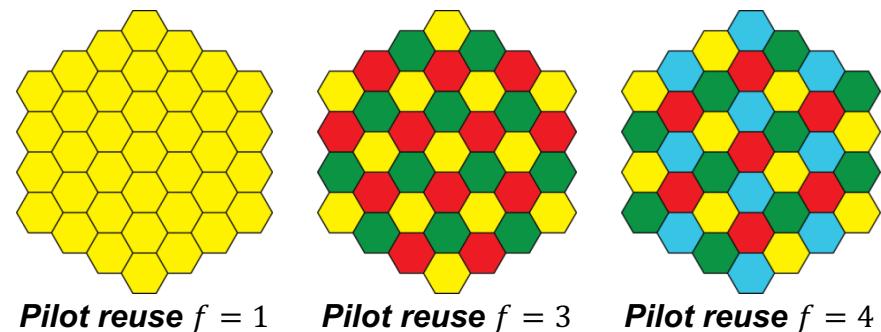
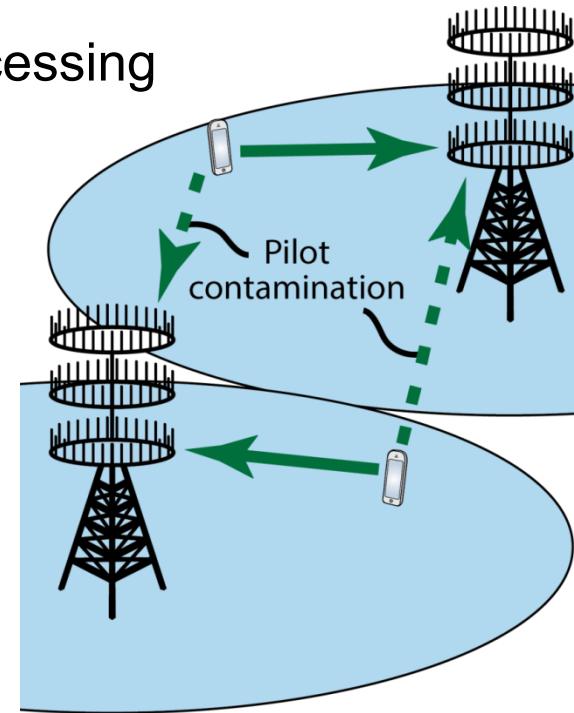
Linear Processing Schemes



- Three Options for Downlink Precoding and Uplink Detection
 - Maximum ratio (MR): Maximize received signal power
 - Zero-forcing (ZF): Minimize interference
 - Maximum mean-squared error (MMSE): Balance to signal/interference to minimize uplink MSE

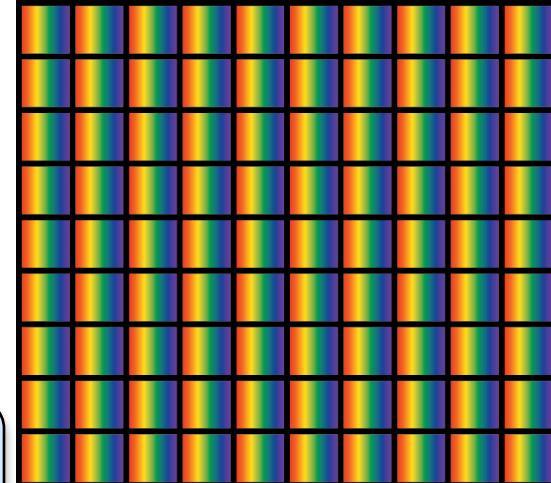
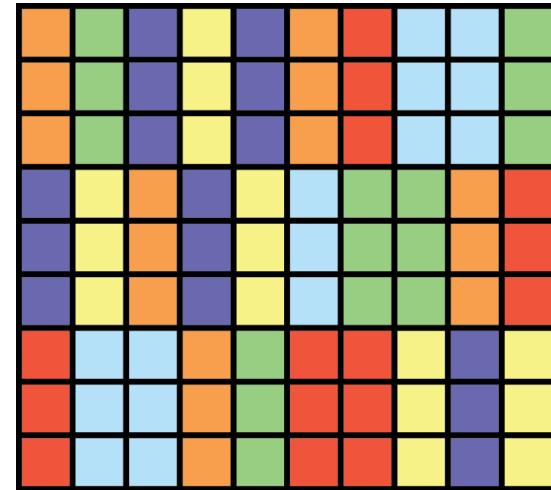
Channel Acquisition in Massive MIMO

- BS Needs Channel Responses for Linear Processing
 - Estimate using $\tau_p \leq \tau_c$ pilot symbols
 - Must reuse pilot sequences in different cells
- Called: Pilot Contamination
 - BSs cannot tell some users apart
 - Recall: Noise and interference vanish as $M \rightarrow \infty$
 - Not interference between users with same pilot!
- Scalable Solution: Select how often pilots are reused
 - Pilot reuse factor $f \geq 1$
 - Users per cell: $K \leq \tau_p/f$
 - Higher $f \rightarrow$ Fewer users per cell,
but interferers further away



Simple Resource Allocation

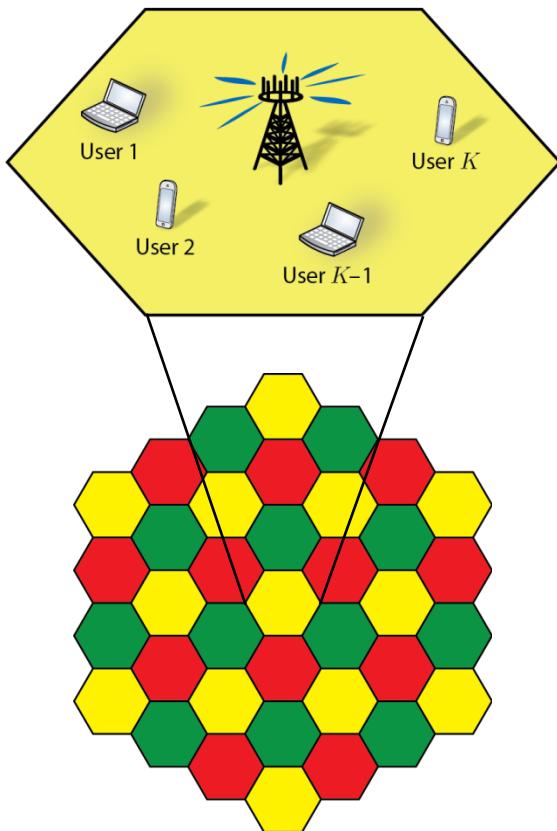
- Resource Allocation in 4G
 - Give each time/frequency block to one user
 - Utilize current fading realization
 - *Not needed in Massive MIMO*
- Spatial Resource Allocation
 - Each user get the whole bandwidth, whenever needed!
 - Separate users spatially
 - Same channel quality in all blocks



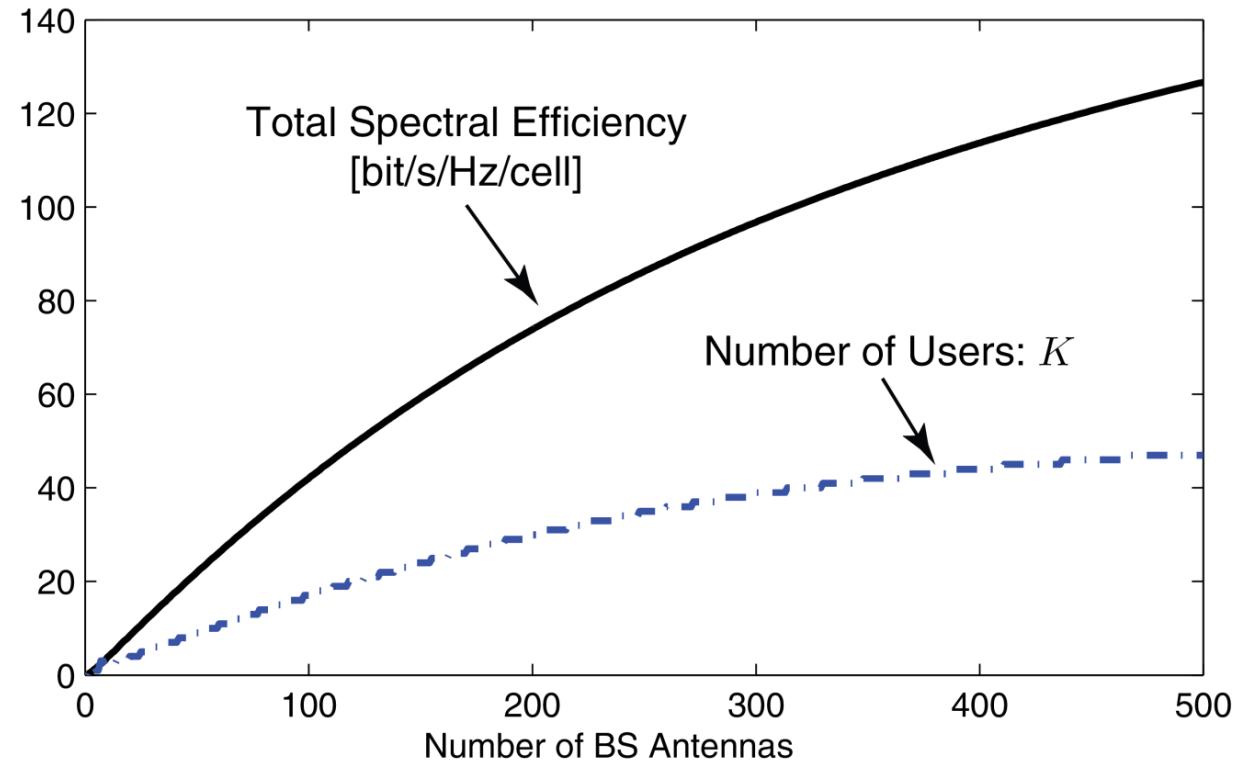
Scalable resource allocation:

Only power control based on long-term channel quality

How Many Antennas Are Needed?



Uplink simulation
SNR 5 dB, Rayleigh fading,
ZF detection, $\tau_c = 500$, pilot reuse $f = 3$

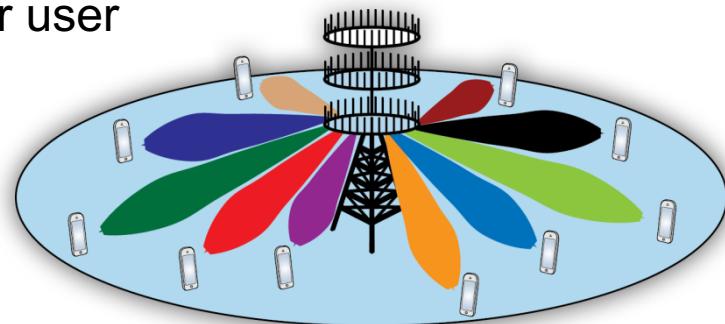


Massive MIMO is an incredibly scalable technology!

SUMMARY

Summary

- Massive MIMO: The way to increase spectral efficiency in 5G networks
 - >20x gain over IMT-Advanced are foreseen
 - BSs with many small antennas and transceiver chains
 - Higher spectral efficiency per cell, not per user
 - Many potential deployment strategies
- Facts to Remember
 - Massive MIMO \neq Massive size: TV sized panels at cellular frequencies
 - Favorable propagation in most propagation environments
 - Resource allocation and processing are simplified, not complicated
- Further Reading
 - Emil Björnson, Erik G. Larsson, Thomas L. Marzetta, “Massive MIMO: 10 Myths and One Critical Question,” IEEE Commun. Magazine, Feb. 2016.





Thank you!

Emil Björnson

Slides and papers available online:

<http://www.commsys.isy.liu.se/en/staff/emibj29>