

Datasheet

MM32SPIN120B

32-bit Micro controller based on ARM Cortex M0

Ver: 1.02_q

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1

Introduction

Introduction

1.1 Description

MM32SPIN120Bqa(named as "the device" throughout this document) is ARM® Cortex™-M0 32-bit RISC core based micro controller family. The device has high speed embedded memory and the CPU, memory and AHB bus subsystem speed can attain up to 72MHz. The device also has integrated with extensive range of enhanced I/Os, two APB buses peripherals, 1 12-bit ADC, 1 Comparator, 1 general purpose 16-bit timer, 1 general purpose 32-bit timer, 3 Basic timers, 1 Advanced 16-bit timer, and standard communication interfaces device: 1 I2C, 2 SPIs, and 1 UART.

This product is equipped with two N-channel half-bridge gate drivers with bootstrap diodes. It is available in QFN32 package and is suitable for a variety of applications:

- Single-phase permanent magnet brushless motor
- Fast wireless charging

1.2 Product Features

- Core and system
 - ARM® Cortex™-M0 CPU
 - Maximum operating frequency is up to 72MHz
 - Hardware divider(32bit)
 - Single cycle 32-bit hardware multiplier
- Two sets of N-type half-bridge gate drivers (GATE-DRIVER)
 - Operating voltage range: 4.5V to 15V
 - Two-phase gate driver
 - Drive capacity: 5000PF load, rise time and fall time is 75nS
 - UVLO protection
 - Upper and lower bridge short circuit automatic protection function
- Memories
 - 32K Bytes of Flash memory
 - 4K Bytes of SRAM
 - Boot loader support Chip Flash
- Clock, reset and power management
 - 2.0V to 5.5V application supply
 - Power-on/Power-down reset (POR/PDR), Programmable voltage detector (PVD)

- External 2 to 24MHz high speed crystal oscillator
 - Embedded factory-tuned 48/72MHz high speed oscillator
 - Embedded 40KHz low speed oscillator
- Low-power
 - Sleep, Stop and Standby modes
- 1 12-bit ADC, 1 μ S A/D converters (up to 8 channels)
 - Conversion range: 0 to V_{DDA}
 - Support sampling time and resolution configuration
 - On-chip temperature sensor
 - On-chip voltage sensor
- 1 Comparator
- 5 DMA controller
 - Supported peripherals: Timer、UART、I2C、SPI and ADC
- Up to 16 fast I/Os:
 - All mappable on 16 external interrupt vectors
 - Almost all can work on 5V
- Debug mode
 - Serial wire debug (SWD)
- Up to 9 timers
 - 1 16-bit 4-channel advanced-control timer for 4 channels PWM output, with deadtime generation and emergency stop
 - 1 16-bit timer and 1 32-bit timer, with up to 4 IC/OC, usable for IR control decoding
 - 2 16-bit timer, with 1 IC/OC, 1 OCN, deadtime generation and emergency stop and modulator gate for IR control
 - 1 16-bit timer, with 1 IC/OC
 - 2 watchdog timers (independent and window type)
 - SysTick timer: 24-bit downcounter
- Up to 5 Communication interfaces
 - 1 UART
 - 1 I2C
 - 2 SPIs
- 96-bit unique ID (UID)
- Packages QFN32

For more information about the complete product, refer to Section 2.2 of the data sheet. The relevant information about the CortexTM-M0, please refer to CortexTM-M0 technical reference manual.

2

Specification

Specification

2.1 Device contrast

Table 1. MM32SPIN120B device features and peripheral counts

Peripheral		MM32SPIN120B
Flash memory -K Bytes		32
SRAM -K Bytes		4
Timers	General purpose (16 bit)	1
	General purpose (32 bit)	1
	basic	3
	Advanced control	1
Common interfaces	UART	1
	I2C	1
	SPI	2
GPIOs		16
12-bit ADC (number of channels)		1 8 channels
Comparators		1
Max CPU frequency		72 MHz
Operating voltage	VCC	3.0V ~ 15V
	VCDDA	2.0V ~ 5.5V
Packages		QFN32

2.2 Summary

2.2.1 ARM® Cortex™-M0 and SRAM

The ARM® Cortex™-M0 is a generation of ARM processors for embedded systems. It has been developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while delivering outstanding computational performance and an advanced system response to interrupts.

The ARM® Cortex™-M0 processors feature exceptional code-efficiency, delivering the high performance expected from an ARM core, with memory sizes usually associated with 8- and 16-bit devices.

The devices have embedded ARM core and are compatible with all ARM tools and software.

2.2.2 Memory

32K Bytes of embedded Flash memory.

2.2.3 SRAM

4K Bytes of embedded SRAM.

2.2.4 Clocks and startup

Several prescalers allow the application to configure the frequency of the AHB and the APB domains. The maximum frequency of the AHB and the APB domains is 72MHz. Refer to figure 3 for the clock drive block diagram.

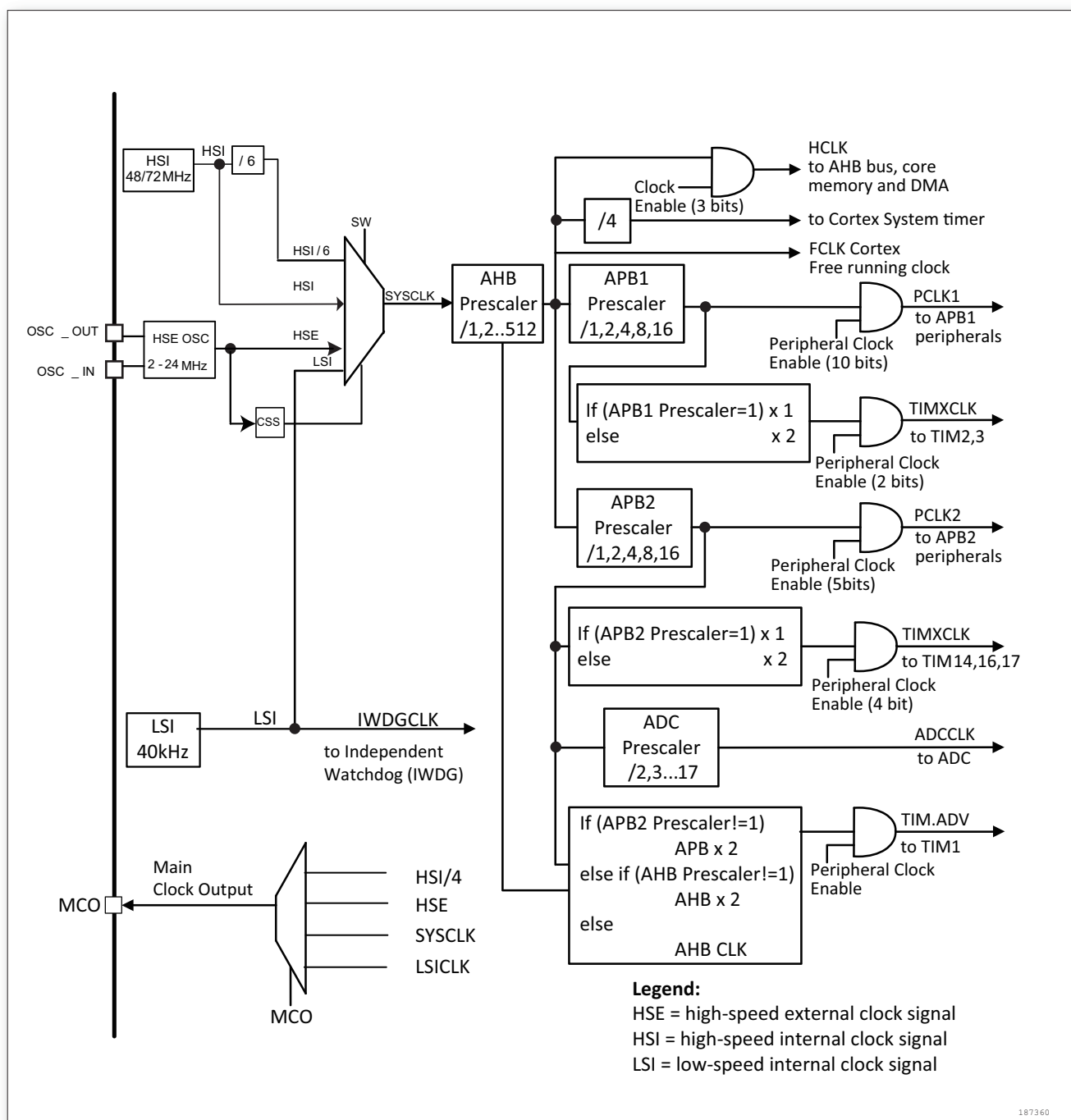


Figure 1. Clock tree

2.2.5 Nested vectored interrupt controller (NVIC)

The device embeds a nested vectored interrupt controller and is able to handle up to 68 maskable interrupt channels (not including the 16 interrupt lines of Cortex™-M0) with 16 priority levels.

- Closely coupled NVIC gives low latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Closely coupled NVIC core interface

- Allows early processing of interrupts
- Processing of late arriving higher priority interrupts
- Support for tail-chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

This hardware block provides flexible interrupt management features with minimal interrupt latency.

2.2.6 Extended interrupt/event controller (EXTI)

The extended interrupt/event controller consists of many edge detector lines are used to generate interrupt/event requests for waking up the system. Each line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the internal APB2 clock period. All GPIOs can be connected to the 16 external interrupt lines.

2.2.7 Boot modes

At startup, the boot pin and boot selector option bit are used to select one of the three boot options:

- Boot from User Flash memory
- Boot from System Memory
- Boot from embedded SRAM

2.2.8 Power supply schemes

- $V_{DD} = 2.0V \sim 5.5V$: external power supply for I/Os and the internal regulator. Provided externally through V_{DD} pins.
- $V_{SSA}, V_{DDA} = 2.0V \sim 5.5V$: 为复位模块 和 振荡器提供供电. V_{DDA} and V_{SSA} must be connected to V_{DD} and V_{SS} .
- Gate-Drive = $4.5V \sim 15V$.

2.2.9 Power supply supervisors

The device has integrated power-on reset (POR) and power-down reset (PDR) circuits. They are always active, and ensure proper operation above a threshold of 1.8V. The device remains in reset mode when the monitored supply voltage is below a specified threshold $V_{POR/PDR}$, without the need for an external reset circuit.

The device features an embedded programmable voltage detector (PVD) that monitors the V_{DD}/V_{DDA} power supply and compares it to the V_{PVD} threshold. An interrupt can be generated when VDD drops below the V_{PVD} threshold and/or when VDD is higher than the V_{PVD} threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

2.2.10 Voltage regulator

The voltage regulator converts the external voltage to the internal digital logic and it is always enabled after reset.

2.2.11 Low-power modes

The device support three low-power modes to achieve the best compromise between low power consumption, short startup time and available wakeup sources.

Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

Stop mode

Stop mode achieves very low power consumption while retaining the content of SRAM and registers. the HSI and the HSE crystal oscillators are disabled. The voltage regulator can also be put either in normal or in low power mode.

Standby mode

Standby mode achieves the lowest power consumption of the system. This mode turns off the voltage regulator in CPU deep sleep mode. The entire 1.5V power supply area is powered down. HSI and HSE oscillators are also powered down. SRAM and register contents are missing.

2.2.12 Direct memory access controller (DMA)

The 5-channel general-purpose DMAs manage memory-to-memory, peripheral-to-memory and memory-to-peripheral transfers. The DMA supports circular buffer management, removing the need for user code intervention when the controller reaches the end of the buffer.

Each channel is connected to dedicated hardware DMA requests, with support for software trigger on each channel. Configuration is made by software and transfer sizes between source and destination are independent.

DMA can be used with the main peripherals: UART、I2C、SPI、ADC general-purpose and advanced-control timers TIMx.

2.2.13 Timers and watchdogs

Medium capacity device include 1 advanced control、2 general-purpose timers、3 base-timer、2 watchdog timers and 1 SysTick timer.

The following table compares the features of the different timers:

Table 2. Timer feature comparison

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/-compare channels	Complementary outputs
Advanced control	TIM1	16-bit	Up, down, up/down	integer from 1 to 65536	Yes	4	Yes
General purpose	TIM2	32-bit	Up, down, up/down	integer from 1 to $2^{32} - 1$	Yes	4	No
	TIM3	16-bit	Up, down, up/down	integer from 1 to 65536	Yes	4	No
basic	TIM14	16-bit	Up	integer from 1 to 65536	Yes	1	No
	TIM16 / TIM17	16-bit	Up	integer from 1 to 65536	Yes	1	Yes

Advanced-control timer (TIM1)

The advanced-control timer can be seen as a three-phase PWM multiplexed on six channels. It has complementary PWM outputs with programmable inserted dead times. It can also be seen as a complete general-purpose timer. The four independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge or center-aligned modes)
- One-pulse mode output

If configured as a standard 16-bit timer, it has the same features as the TIMx timer. If configured as the 16-bit PWM generator, it has full modulation capability (0 ~ 100%).

In debug mode, the counter can be frozen and the PWM output is disabled to cut off the switches controlled by these outputs.

Many features are shared with those of the standard timers which have the same architecture. The advanced control timer can therefore work together with the other timers via the Timer Link feature for synchronization or event chaining.

General-purpose timers (TIMx)

There are 2 synchronizable general-purpose timers (TIM2、TIM3).

General-purpose timers 32-bit

TIM2

The timer is based on a 32-bit auto-reload up/downcounter and a 16-bit prescaler. The feature is 4 independent channels each for input capture/output compare, PWM or one-pulse mode output.

General-purpose timers 16-bit

TIM3

TIM3

The timer is based on a 16-bit auto-reload up/downcounter and a 16-bit prescaler. The feature is 4 independent channels each for input capture/output compare, PWM or one-pulse mode output.

The timer can work together or with the TIM1 advanced-control timer via the Timer Link feature for synchronization or event chaining. Their counter can be frozen in debug mode. Any of the general-purpose timers can be used to generate PWM outputs. They all have independent DMA request generation.

These timers are capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 3 hall-effect sensors.

Basic timer

TIM14

This timer is based on a 16-bit auto-reload upcounter and a 16-bit prescaler. TIM14 features one single channel for input capture/output compare, PWM or one-pulse mode output. Their counter can be frozen in debug mode.

TIM16/TIM17

Every timer is based on a 16-bit auto-reload upcounter and a 16-bit prescaler. They each have a single channel for input capture/output compare, PWM or one-pulse mode output. TIM16 and TIM17 have a complementary output with dead-time generation and independent DMA request generation. Their counters can be frozen in debug mode.

Independent watchdog (IWDG)

The independent watchdog is based on an 8-bit prescaler and 12-bit downcounter with user-defined refresh window. It is clocked from an independent 40 KHz internal oscillator and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free running timer for application timeout management. It is hardware or software configurable through the option bytes. The counter can be frozen in debug mode.

System window watchdog (WWDG)

The system window watchdog is based on a 7-bit downcounter that can be set as free running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the APB clock (PCLK). It has an early warning interrupt capability and the counter can be frozen in debug mode.

SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard down counter. It features:

- A 24-bit down counter
- Autoreload capability
- Maskable system interrupt generation when the counter reaches 0
- Programmable clock source

2.2.14 Universal asynchronous receiver/transmitter (UART)

UART provides hardware management of the CTS, RTS.

Support LIN master-slave function.

All UART interface can be served by the DMA controller.

2.2.15 I2C interface

The I2C interface can operate in multimaster or slave modes. It can support Standard mode, and Fast Mode.

It supports 7-bit and 10-bit addressing modes, multiple 7-bit slave addresses (two addresses, one with configurable mask).

2.2.16 Serial peripheral interface (SPI)

The SPI interface, in slave or master mode, can be configured to 1 ~ 32 bits per frame.

All SPI interface can be served by the DMA controller.

2.2.17 General-purpose inputs/outputs (GPIO)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. The I/O configuration can be locked if needed following a specific sequence in order to avoid spurious writing to the I/Os registers.

2.2.18 Analog-to-digital converter (ADC)

The one 12-bit analog-to-digital converters is embedded into microcontrollers and the ADC shares up to 10 external channels, performing conversions in single-shot or scan modes. In scan mode, automatic conversion is performed on a selected group of analog inputs. The ADC can be served by the DMA controller.

The analog watchdog function allows very precise monitoring of all the way, multiple or all selected channels, and an interruption occurs when the monitored signal exceeds the preset threshold. The events generated by the general-purpose timers (TIMx) and the advanced-control timer (TIM1) can be internally connected to the ADC start trigger to allow the application to synchronize A/D conversion and timers.

2.2.19 Hardware Division

The hardware division unit consists of four 32-bit data registers, which are dividend, divisor, quotient and remainder, and can be done with signed or unsigned 32-bit division. The hardware division control register USIGN can choose whether to have signed division or unsigned division.

Each time the divisor register is written, the division operation is automatically triggered. After the operation is completed, the result is written to the quotient and remainder registers. If the reader register, remainder register, or status register is read before the end, the read operation is suspended until the end of the operation.

If the divisor is zero, an overflow interrupt flag will be generated.

2.2.20 Temperature sensor

The temperature sensor has to generate a voltage that varies linearly with temperature. The temperature sensor is internally connected to the input channel which is used to convert the sensor output voltage into a digital value.

2.2.21 Serial single line SWD debug port (SW-DP)

Built-in ARM two-wire serial debug port (SW-DP).

An ARM SW-DP interface is provided to allow a serial wire debugging tool to be connected to the MCU.

2.2.22 Comparator (COMP)

The devices embed 1 general purpose comparators. that can be used either as standalone devices (all terminal are available on I/Os) or combined with the timers. The comparators can be used for a variety of functions including:

- Wake-up from low-power mode triggered by an analog signal,
- Analog signal conditioning,

- Cycle-by-cycle current control loop when combined with the PWM output from a timer.
- Rail-to-rail comparators
- Each comparator has positive and configurable negative inputs used for flexible voltage
- Selection:
 - Reusable I/O pins
 - Internal comparison voltage CRV selects the voltage divider value of AVDD or internal reference voltage
- Programmable hysteresis
- Programmable speed/consumption
- The outputs can be redirected to an I/O or to timer inputs for triggering:
 - Capture events
 - OCref_clr events (for cycle-by-cycle current control)
 - Break events for fast PWM shutdowns

2.2.23 GateDriver

Two GateDrivers are integrated into the chip to support external power N-type MOSFET drivers.

The GateDriver also supports voltage UVLO protection.

The N-type half-bridge gate driver (GateDriver) supports the following features:

- Operating voltage range: 3V ~ 15V
- Two-phase gate driver
- Drive capacity: 5000pF load, rise time and fall time is 75n seconds
- UVLO protection
- Upper and lower bridge short circuit automatic protection function

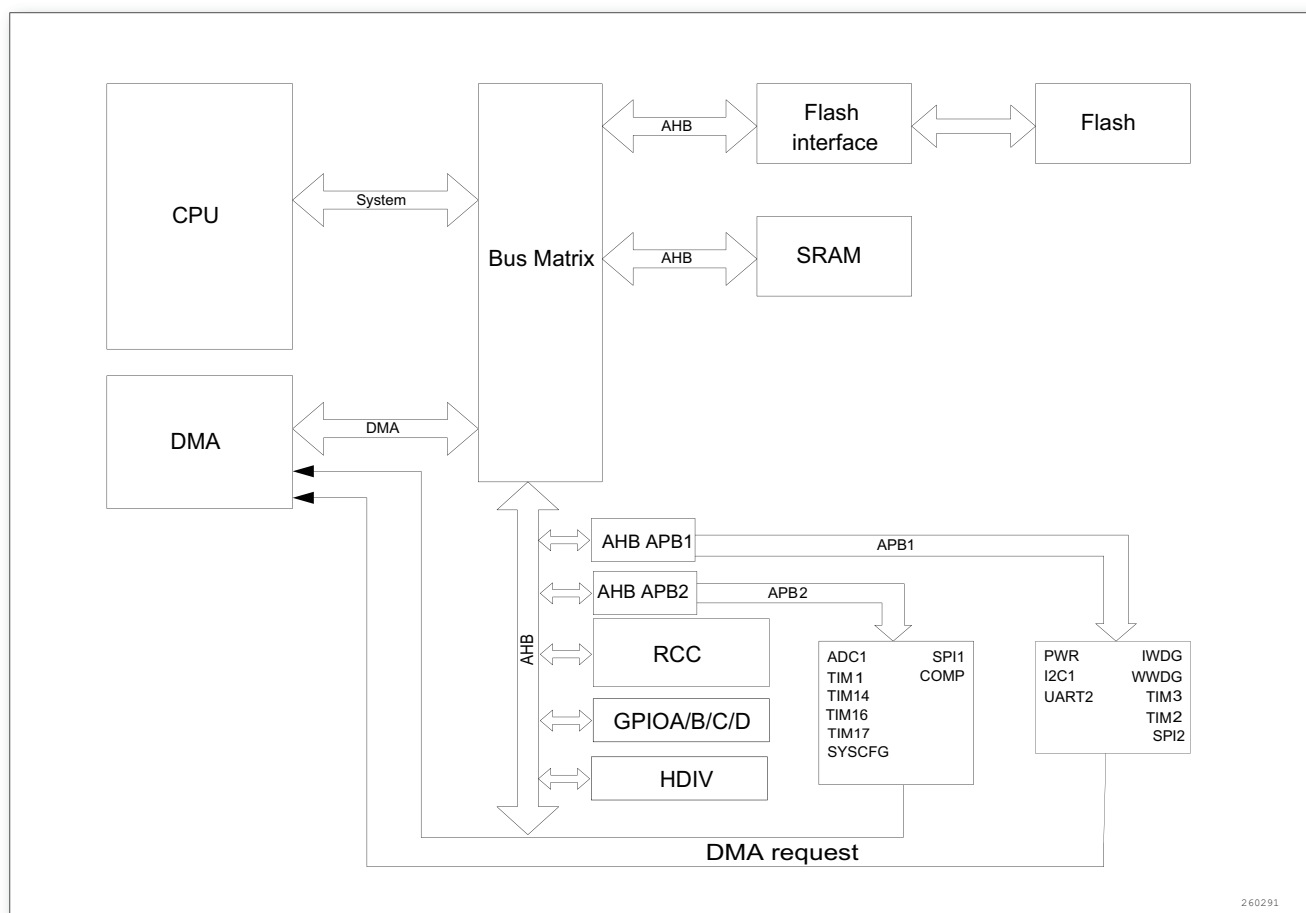


Figure 2. Block diagram

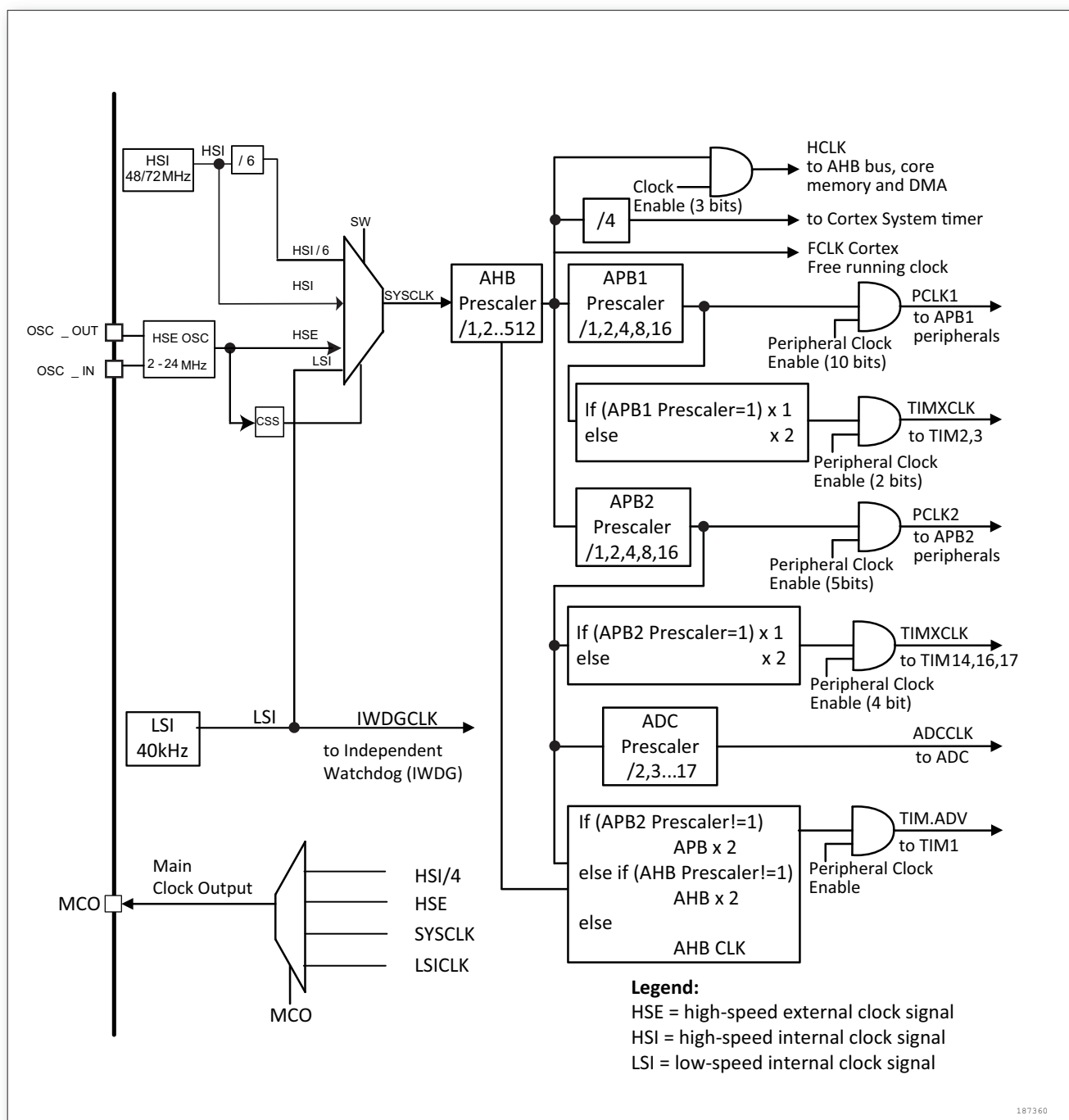


Figure 3. Clock tree

3

Pin definition

Pin definition

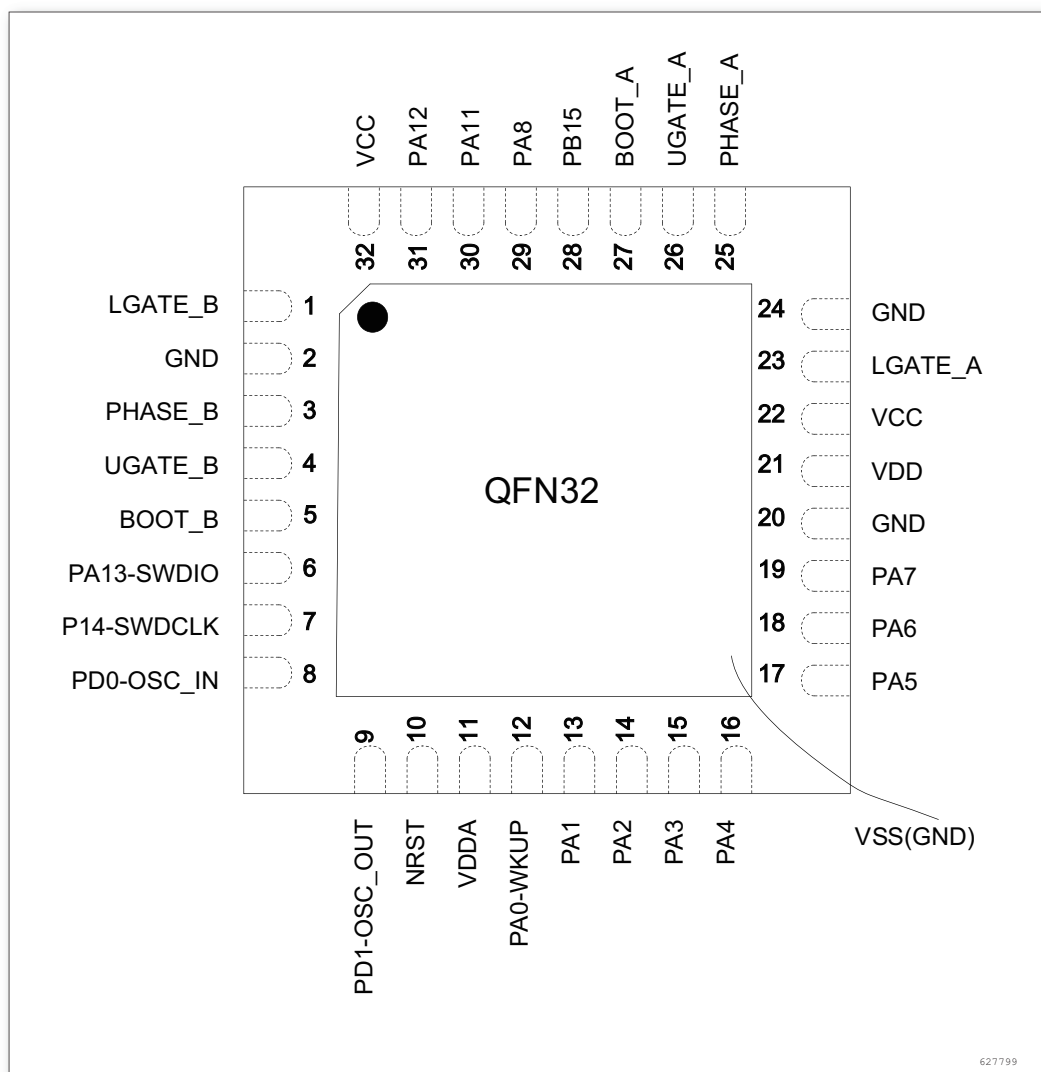


Figure 4. QFN32 packet pinout

3.1 Pin definitions and package file

Table 3. Pin definitions

Pin number	Pin name	Type ⁽¹⁾	I/O structure ⁽²⁾	Main function	Alternate functions	Additional functions
QFN32						
1	LGATE_B	-	-	LGATE_B	-	-
2	GND	S	-	GND	-	-
3	PHASE_B	-	-	PHASE_B	-	-

Pin number QFN32	Pin name	Type ⁽¹⁾	I/O structure ⁽²⁾	Main function	Alternate functions	Additional functions
4	UGATE_B	-	-	UGATE_B	-	-
5	BOOT_B	-	-	BOOT_B	-	-
6	PA13	I/O	FT	PA13	SWDIO/ SPI2_MISO/ MCO/ TIM1_CH2/ TIM1_BKIN	-
7	PA14	I/O	FT	PA14	SWDCLK/ UART2_TX/ SPI1_NSS	-
8	PD0- OSC_IN	I/O	FT	OSC_IN	I2C1_SDA	-
9	PD1- OSC_OUT	I/O	FT	OSC_OUT	I2C1_SCL	-
10	NRST	I	-	NRST	-	-
0	VSSA	S	-	VSSA	-	-
11	VDDA	S	-	VDDA	-	-
12	PA0- WKUP	I/O	TC	PA0	UART2_CTS/ TIM2_CH1_ETR/ SPI2_NSS/ TIM2_CH3/ COMP1_OUT	ADC1_VIN[0]
13	PA1	I/O	TC	PA1	UART2_RTS/ TIM2_CH2	ADC1_VIN[1]/ COMP1_INP[0]
14	PA2	I/O	TC	PA2	UART2_TX/ TIM2_CH3/ SPI2_NSS	ADC1_VIN[2]/ COMP1_INP[1]
15	PA3	I/O	TC	PA3	UART2_RX/ TIM2_CH4	ADC1_VIN[3]/ COMP1_INP[2]
16	PA4	I/O	TC	PA4	SPI1_NSS/ TIM1_BKIN/ TIM14_CH1/ I2C1_SDA	ADC1_VIN[4]/ COMP1_INP[3]

Pin number QFN32	Pin name	Type ⁽¹⁾	I/O structure ⁽²⁾	Main function	Alternate functions	Additional functions
17	PA5	I/O	TC	PA5	SPI1_SCK/ TIM2_CH1_ETR/ TIM1_ETR/ I2C1_SCL/ TIM1_CH3N	ADC1_VIN[5]/ COMP1_INM[0]
18	PA6	I/O	TC	PA6	SPI1_MISO/ TIM3_CH1/ TIM1_BKIN/ UART2_RX/ TIM1_ETR/ TIM16_CH1/ TIM1_CH3/ COMP1_OUT	ADC1_VIN[6]/ COMP1_INM[1]
19	PA7	I/O	TC	PA7	SPI1_MOSI/ TIM3_CH2/ TIM1_CH1N/ TIM14_CH1/ TIM17_CH1/ TIM1_CH2N/ TIM1_CH3N	ADC1_VIN[7]/ COMP1_INM[2]
20	GND	S	-	GND	-	-
21	VDD	S	-	VDD	-	-
22	VCC	S	-	VCC	-	-
23	LGATE_A	-	-	LGATE_A	-	-
24	GND	S	-	GND	-	-
25	PHASE_A		-	PHASE_A	-	-
26	UGATE_A		-	UGATE_A	-	-
27	BOOT_A		-	BOOT_A	-	-
28	PB15	I/O	FT	PB15	SPI2_MOSI/ SPI2_NSS/ TIM1_CH3N/ SPI2_MISO/ SPI2_SCK/ TIM1_CH2N/ TIM1_CH2	-

Pin number QFN32	Pin name	Type ⁽¹⁾	I/O structure ⁽²⁾	Main function	Alternate functions	Additional functions
29	PA8	I/O	FT	PA8	MCO/ TIM1_CH1/ TIM1_CH2/ TIM1_CH3	-
30	PA11	I/O	FT	PA11	TIM1_CH4/ SPI2_MOSI/ I2C1_SCL/ COMP1_OUT	-
31	PA12	I/O	FT	PA12	TIM1_ETR/ SPI2_MISO/ I2C1_SDA/ TIM1_CH2	-
32	VCC	S	-	VCC	-	-
0	VSS(GND)	S	-	VSS(GND)	-	-

A

1. I = input, O = output, S = power supply, HiZ = high resistance.

2. FT: 5V tolerant, Input signal should be between VDD and 5V.

TC: Standard I/O, Input signal does not exceed VDD.

Table 4. Alternate functions for port A

Pin Name	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PA0	-	UART2_CTS	TIM2_CH1 _ETR	SPI2_NSS	TIM2_CH3	-	-	COMP1_OUT
PA1	-	UART2_RTS	TIM2_CH2	-	-	-	-	-
PA2	-	UART2_TX	TIM2_CH3	SPI2_NSS	-	-	-	-
PA3	-	UART2_RX	TIM2_CH4	-	-	-	-	-
PA4	SPI1_NSS	-	-	TIM1_BKIN	TIM14_CH1	I2C1_SDA	-	-
PA5	SPI1_SCK	-	TIM2_CH1 _ETR	TIM1_ETR	-	I2C1_SCL	TIM1_CH3N	-
PA6	SPI1_MISO	TIM3_CH1	TIM1_BKIN	UART2_RX	TIM1_ETR	TIM16_CH1	TIM1_CH3	COMP1_OUT
PA7	SPI1_MOSI	TIM3_CH2	TIM1_CH1N	-	TIM14_CH1	TIM17_CH1	TIM1_CH2N	TIM1_CH3N
PA8	MCO	-	TIM1_CH1	-	-	-	TIM1_CH2	TIM1_CH3
PA11	-	-	TIM1_CH4	-	SPI2_MOSI	I2C1_SCL	-	COMP1_OUT
PA12	-	-	TIM1_ETR	-	SPI2_MISO	I2C1_SDA	-	TIM1_CH2
PA13	SWDIO	-	-	-	SPI2_MISO	MCO	TIM1_CH2	TIM1_BKIN
PA14	SWDCLK	UART2_TX	-	SPI1_NSS	-	-	-	-

Table 5. Alternate functions for port B

Pin Name	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PB15	SPI2_MOSI	SPI2_NSS	TIM1_CH3N	SPI2_MISO	SPI2_SCK	-	TIM1_CH2N	TIM1_CH2

Table 6. Alternate functions for port D

Pin Name	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PD0	-	I2C1_SDA	-	-	-	-	-	-
PD1	-	I2C1_SCL	-	-	-	-	-	-

3.2 Expansion pin description of sealed chip

Table 7. GateDriver Pin description

Pin number	Pin name	Pin function
1	LGATE_B	Gate driver lower arm output B. Connect this pin to the gate of the low-side MOSFET. This pin is monitored by an injection protection circuit to determine when the MOSFET is turned off.
2	GND	The ground wire of the IC.
3	PHASE_B	The power switching circuit outputs phase node B. Connect this pin to the source above the MOSFET and the drain of the MOSFET lower arm. This pin is used as the return path for the UGATE driver. This pin is also monitored by the injection protection circuit to determine when the upper MOSFET is turned off.
4	UGATE_B	The bridge arm output B on the gate driver. Connect this pin to the gate of the upper MOSFET. This pin is monitored by an injection protection circuit to determine when the upper MOSFET is turned off.
5	BOOT_B	Gate driver bootstrap power output B. Used to boost the voltage of the upper arm driver. The boot capacitor CBOOT is connected between the BOOT pin and the PHASE pin for extraction from the bootstrap circuit. The bootstrap capacitor supplies charge to the upper MOSFET. The circuit design tries to ensure that the CBOOT is placed near the IC.
20	GND	The ground wire of the IC.
21	VDD	MCU power supply inside the IC. This pin provides the bias voltage for the IC. This pin can be connected to a 3.3 5V voltage source and bypassed with an R/C filter.
22	VCC	The gate driver power supply inside the IC. This pin provides the bias voltage for the IC. This pin can be connected to a 5 13V voltage source and bypassed with an R/C filter.

Pin number	Pin name	Pin function
23	LGATE_A	Gate driver lower arm output A. Connect this pin to the gate of the low-side MOSFET. This pin is monitored by an injection protection circuit to determine when the MOSFET is turned off.
24	GND	The ground wire of the IC.
25	PHASE_A	The power switching circuit outputs phase node A. Connect this pin to the source above the MOSFET and the drain below the MOSFET. This pin is used as the return path for the UGATE driver. This pin is also monitored by a pass-through protection circuit to determine when the upper MOSFET is turned off.
26	UGATE_A	The upper arm of the gate driver outputs A. Connect this pin to the gate of the upper MOSFET. This pin is monitored by an injection protection circuit to determine when the upper MOSFET is turned off.
27	BOOT_A	Gate driver bootstrap power output A. Used to boost the voltage of the upper arm driver. The boot capacitor CBOOT is connected between the BOOT pin and the PHASE pin for extraction from the bootstrap circuit. The bootstrap capacitor supplies charge to the upper MOSFET. The circuit design tries to ensure that the CBOOT is placed near the IC.
32	VCC	The gate driver power supply inside the IC. This pin provides the bias voltage for the IC. This pin is connected to a 3 15V voltage source and bypassed with an R/C filter.

3.3 Functional block diagram and application reference circuit

3.3.1 Functional block diagram

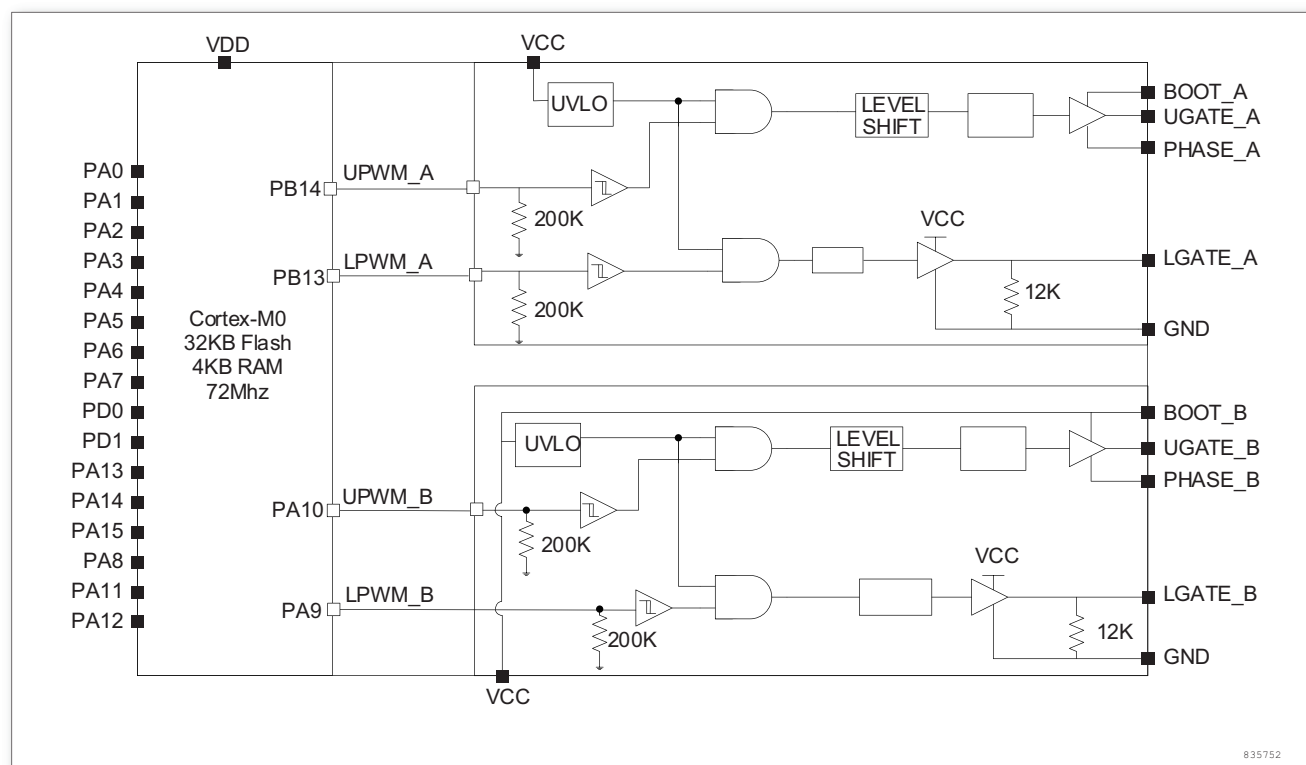


Figure 5. Functional block diagram

3.3.2 Application reference circuit

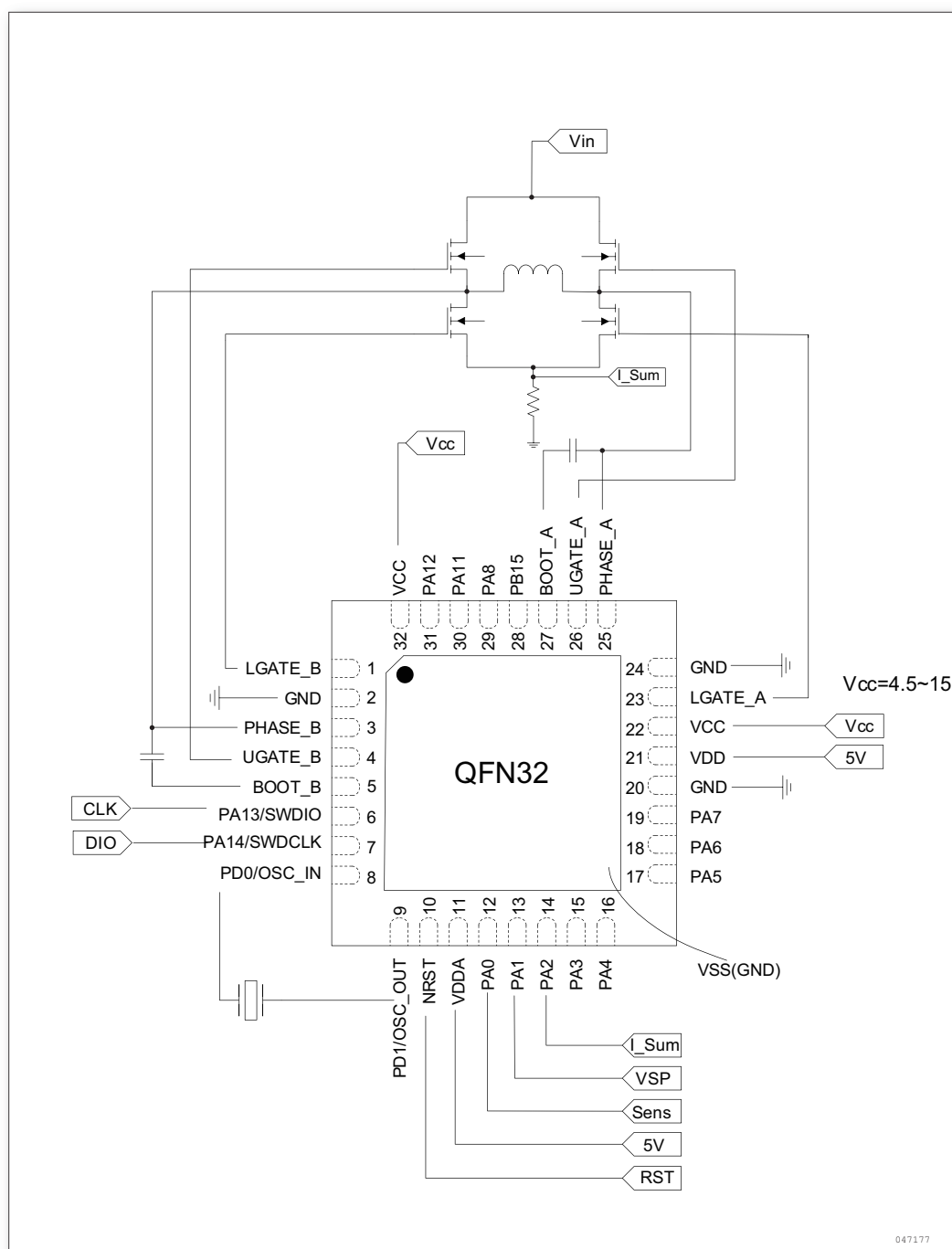


Figure 6. Application reference circuit

4

Memory mapping

Memory mapping

Table 8. Memory mapping

Bus	Boundaryaddress	Size	Peripheral	Notes
Flash	0x0000 0000 - 0x0001 FFFF	128 KB	Main flash memory, system memory, or SRAM, depends on the configuration of BOOT	
	0x0002 0000 - 0x07FF FFFF	~ 128 MB	Reserved	
	0x0800 0000 - 0x0800 7FFF	32 KB	Main Flash memory	
	0x0800 8000 - 0x1FFD FFFF	~ 256 MB	Reserved	
	0x1FFE 0000 - 0x1FFE 01FF	0.5 KB	Reserved	
	0x1FFE 0200 - 0x1FFE 0FFF	3 KB	Reserved	
	0x1FFE 1000 - 0x1FFE 1BFF	3 KB	Reserved	
	0x1FFE 1C00 - 0x1FFF F3FF	~ 256 MB	Reserved	
	0x1FFF F400 - 0x1FFF F7FF	1 KB	System memory	
	0x1FFF F800 - 0x1FFF F80F	16 B	Option bytes	
	0x1FFF F810 - 0x1FFF FFFF	~ 2 KB	Reserved	
SRAM	0x2000 0000 - 0x2000 0FFF	4 KB	SRAM	
	0x2000 1000 - 0x2FFF FFFF	~ 512 MB	Reserved	
APB1	0x4000 0000 - 0x4000 03FF	1 KB	TIM2	
	0x4000 0400 - 0x4000 07FF	1 KB	TIM3	
	0x4000 0800 - 0x4000 27FF	8 KB	Reserved	
	0x4000 2800 - 0x4000 2BFF	1 KB	Reserved	
	0x4000 2C00 - 0x4000 2FFF	1 KB	WWDG	
	0x4000 3000 - 0x4000 33FF	1 KB	IWDG	
	0x4000 3400 - 0x4000 37FF	1 KB	Reserved	
	0x4000 3800 - 0x4000 3BFF	1 KB	SPI2	
	0x4000 4000 - 0x4000 43FF	1 KB	Reserved	
	0x4000 4400 - 0x4000 47FF	1 KB	UART2	
	0x4000 4800 - 0x4000 4BFF	3 KB	Reserved	
	0x4000 5400 - 0x4000 57FF	1 KB	I2C1	
	0x4000 5800 - 0x4000 5BFF	1 KB	Reserved	
	0x4000 5C00 - 0x4000 5FFF	1 KB	Reserved	
	0x4000 6000 - 0x4000 63FF	1 KB	Reserved	
	0x4000 6400 - 0x4000 67FF	1 KB	Reserved	
	0x4000 6800 - 0x4000 6BFF	1 KB	Reserved	

Bus	Boundaryaddress	Size	Peripheral	Notes
APB1	0x4000 6C00 - 0x4000 6FFF	1 KB	Reserved	
	0x4000 7000 - 0x4000 73FF	1 KB	PWR	
	0x4000 7400 - 0x4000 FFFF	35 KB	Reserved	
APB2	0x4001 0000 - 0x4001 03FF	1 KB	SYSCFG	
	0x4001 0400 - 0x4001 07FF	1 KB	EXTI	
	0x4001 0800 - 0x4001 23FF	7 KB	Reserved	
	0x4001 2400 - 0x4001 27FF	1 KB	ADC1	
	0x4001 2800 - 0x4001 2BFF	1 KB	Reserved	
	0x4001 2C00 - 0x4001 2FFF	1 KB	TIM1	
	0x4001 3000 - 0x4001 33FF	1 KB	SPI1	
	0x4001 3400 - 0x4001 37FF	1 KB	DBGMCU	
	0x4001 3800 - 0x4001 3BFF	1 KB	UART1	
	0x4001 3C00 - 0x4001 3FFF	1 KB	COMP	
	0x4001 4000 - 0x4001 43FF	1 KB	TIM14	
	0x4001 4400 - 0x4001 47FF	1 KB	TIM16	
	0x4001 4800 - 0x4001 4BFF	1 KB	TIM17	
	0x4001 4C00 - 0x4001 7FFF	13 KB	Reserved	
AHB	0x4002 0000 - 0x4002 03FF	1 KB	DMA	
	0x4002 0400 - 0x4002 0FFF	3 KB	Reserved	
	0x4002 1000 - 0x4002 13FF	1 KB	RCC	
	0x4002 1400 - 0x4002 1FFF	3 KB	Reserved	
	0x4002 2000 - 0x4002 23FF	1 KB	Flash interface	
	0x4002 2400 - 0x4002 5FFF	15 KB	Reserved	
	0x4002 6000 - 0x4002 63FF	1 KB	Reserved	
	0x4002 6400 - 0x4002 FFFF	39 KB	Reserved	
	0x4003 0000 - 0x4003 03FF	1 KB	HDIV	
	0x4003 0400 - 0x47FF FFFF	~ 128 MB	Reserved	
	0x4800 0000 - 0x4800 03FF	1 KB	GPIOA	
	0x4800 0400 - 0x4800 07FF	1 KB	GPIOB	
	0x4800 0800 - 0x4800 0BFF	1 KB	GPIOC	
	0x4800 0C00 - 0x4800 0FFF	1 KB	GPIOD	
	0x4800 1000 - 0x5FFF FFFF	~ 384 MB	Reserved	

5

Electrical characteristics

Electrical characteristics

5.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V_{SS} .

5.1.1 Minimum and maximum values

Unless otherwise specified, the minimum and maximum values are guaranteed with an ambient temperature at $T_A = 25^\circ\text{C}$, $V_{DD} = 3.3\text{V}$.

5.1.2 Typical values

Unless otherwise specified, typical data are based on $T_A = 25^\circ\text{C}$ and $V_{DD} = 3.3\text{V}$. They are given only as design guidelines and are not tested.

5.1.3 Typical curves

Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

5.1.4 Loading capacitor

The load conditions used for pin parameter measurement are shown in the figure below.

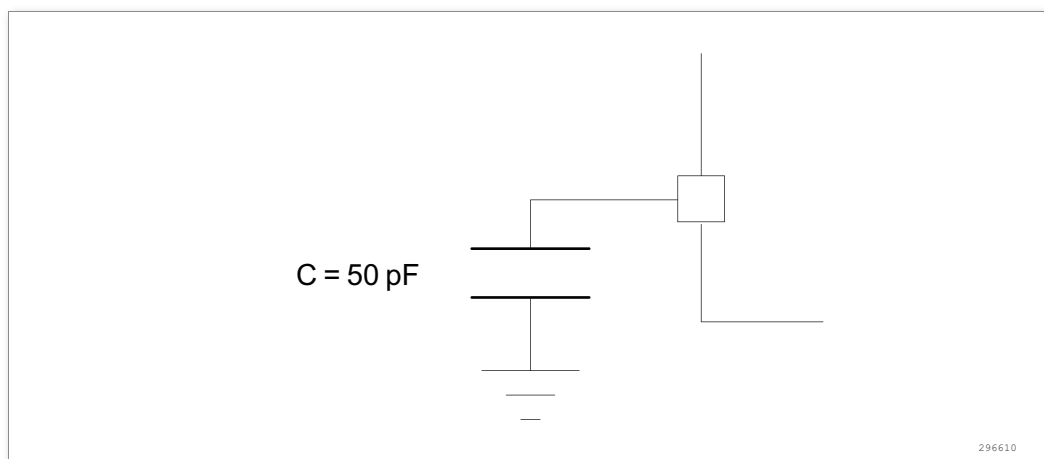


Figure 7. Pin loading conditions

5.1.5 Pin input voltage

The input voltage measurement on a pin of the device is shown in the figure below.

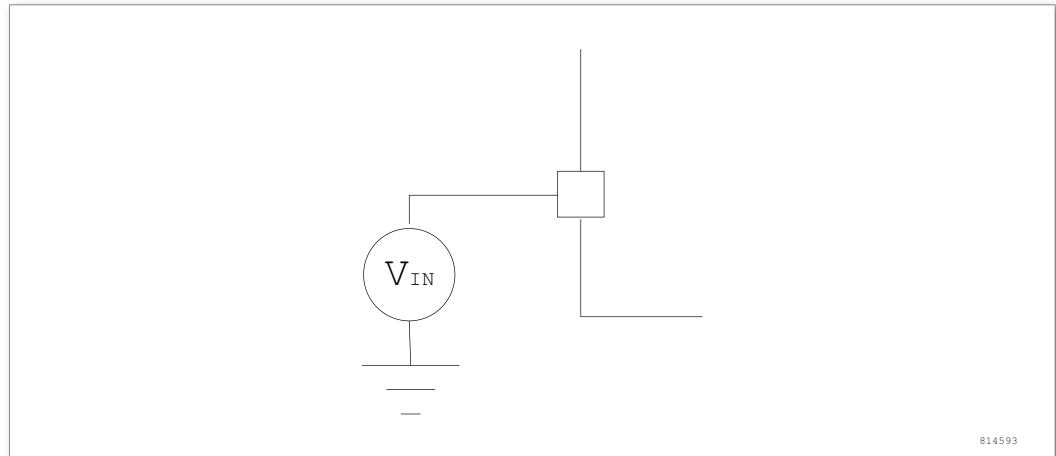


Figure 8. Pin input voltage

5.1.6 Power supply scheme

The power supply design scheme is shown in the figure below.

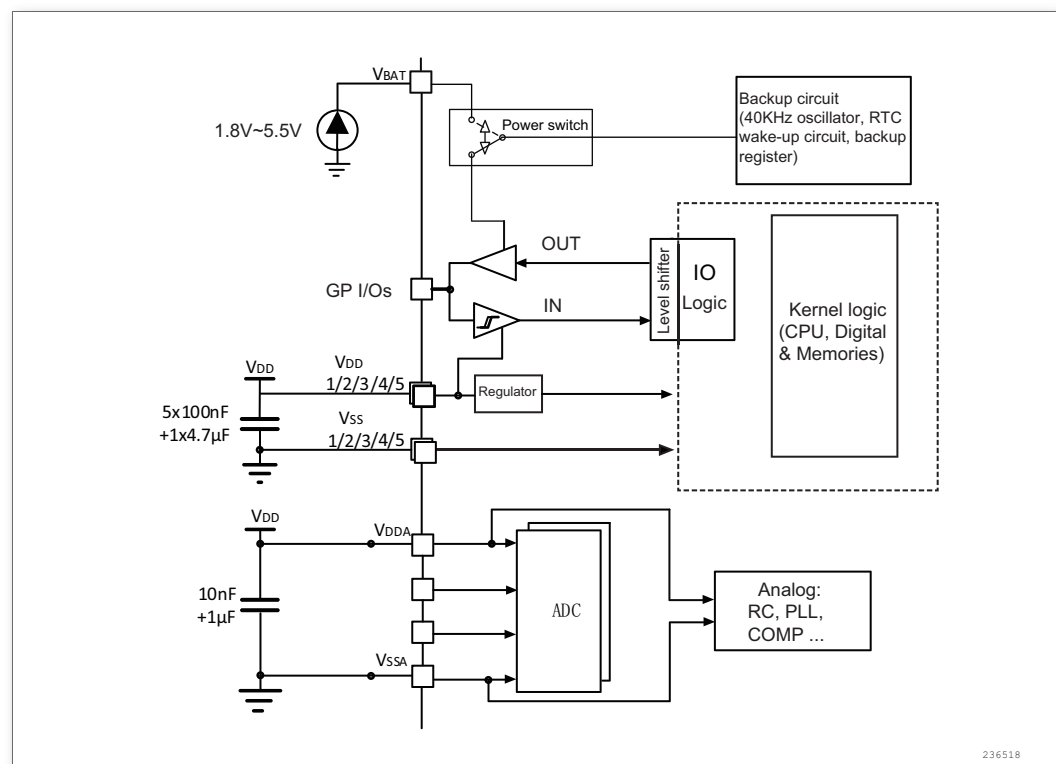


Figure 9. Power supply scheme

5.1.7 Current consumption measurement

The measurement of the current consumption on the pin is shown in the figure below.

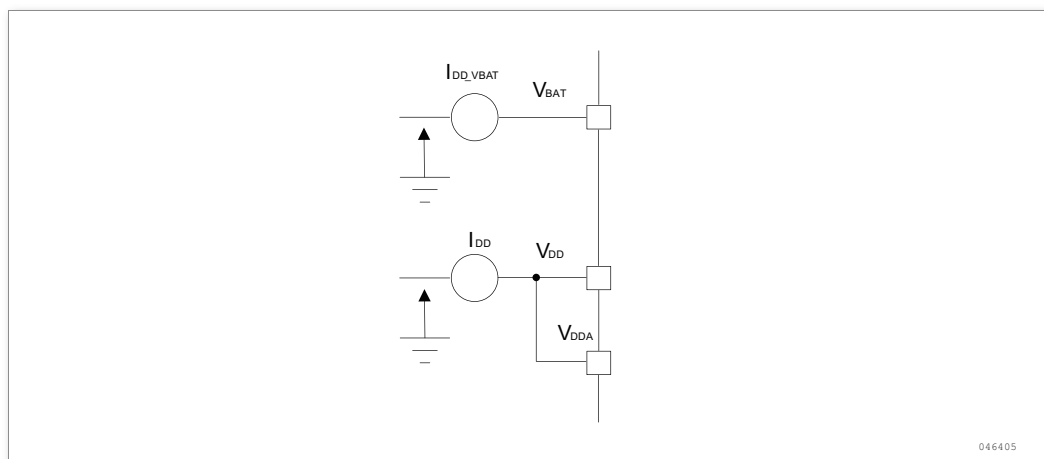


Figure 10. Current consumption measurement scheme

5.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in Tables (Table 9, Table 10, Table 11) may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Table 9. Voltage characteristics

Symbol	Definition	Min	Max	Unit
$V_{DD} - V_{SS}$	External main supply voltage (including V_{DDA} and V_{SSA}) ⁽¹⁾	- 0.3	5.5	V
V_{IN}	Input voltage on FT and FTf pins ⁽²⁾	$V_{SS} - 0.3$	5.5	
	Input voltage on other pins ⁽²⁾	$V_{SS} - 0.3$	5.5	
$ \Delta V_{DDx} $	Variations between different V_{DD} power pins		50	mV
$ V_{SSx} - V_{SS} $	Variations between all the different ground pins		50	

1. All main power (V_{DD} , V_{DDA}) and ground (V_{SS} , V_{SSA}) pins must always be connected to the external power supply, in the permitted range.
2. V_{IN} maximum must always be respected. Refer to Table below for maximum allowed injected current values.

Table 10. Current characteristics

Symbol	Definition	Max	Unit
I_{VDD}	Total current into sum of all V_{DD}/V_{DDA} power lines (source) ⁽¹⁾	150	mA
I_{VSS}	Total current out of sum of all V_{SS} ground lines (sink) ⁽¹⁾	150	mA

Symbol	Definition	Max	Unit
I_{IO}	Output current sunk by any I/O and control pin	20	mA
I_{IO}	Output current source by any I/O and control pin	-18	mA
$I_{INJ(PIN)}^{(2)(3)}$	Injected current on NRST pins	± 5	mA
$I_{INJ(PIN)}^{(2)(3)}$	Injected current on OSC_IN pin of HSE and OSC_IN pin of LSE	± 5	mA
$I_{INJ(PIN)}^{(2)(3)}$	Injected current on other pins ⁽⁴⁾	± 5	mA
$\Sigma I_{INJ(PIN)}^{(2)}$	Total injected current(sum of all I/O and control pins) ⁽⁴⁾	± 25	mA

1. All main power(V_{DD} , V_{DDA}) and ground(V_{SS} , V_{SSA}) pins must always be connected to the external power supply, in the permitted range.
2. $I_{INJ(PIN)}$ cannot exceed its limit, that is, to ensure that the V_{IN} does not exceed its maximum value. If V_{IN} does not guarantee that its maximum value is not exceeded, ensure that $I_{INJ(PIN)}$ does not exceed its maximum value under external restrictions. When $V_{IN} > V_{DD}$, there is a forward injection current; when $V_{IN} < V_{SS}$, there is a reverse injection current.
3. Negative injection disturbs the analog performance of the device.
4. When several inputs are submitted to a current injection, the maximum $I_{INJ(PIN)}$ is the absolute sum of the positive and negative injected currents (instantaneous values).

Table 11. Thermal characteristics

Symbol	Definition	Max	Unit
T_{STG}	Storage temperature range	- 45 ~ + 150	°C
T_J	Maximum junction temperature	125	°C

5.3 Operating conditions

5.3.1 General operating conditions

Table 12. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit
f_{HCLK}	Internal AHB clock frequency		0	96	MHz
f_{PCLK1}	Internal APB1 clock frequency		0	f_{HCLK}	
f_{PCLK2}	Internal APB2 clock frequency		0	f_{HCLK}	
$V_{DD}^{(1)}$	Standard operating voltage		2.0	5.5	V

Symbol	Parameter	Conditions	Min	Max	Unit
V_{DDA}	Analog operating voltage	Must be the same voltage as V_{DD} ⁽¹⁾	2.5	5.5	V
V_{BAT}	Backup part of working voltage		1.8	5.5	V
P_D	Power dissipation temperature: $T_A = 85^\circ\text{C}$ ⁽²⁾	LQFP64		203	mW
		LQFP48			
		LQFP32/QFN32			
T_A	Ambient temperature: $T_A = 85^\circ\text{C}$	Maximum power dissipation	-40	85	$^\circ\text{C}$
		Low power dissipation ⁽³⁾	-40	105	
	Ambient temperature: $T_A = 105^\circ\text{C}$	Maximum power dissipation	-40	105	$^\circ\text{C}$
		Low power dissipation ⁽³⁾	-40	125	

1. It is recommended to use the same power supply for V_{DD} and V_{DDA} , the maximum permissible difference between V_{DD} and V_{DDA} is 300mV during power up and normal operation.
2. If T_A is low, higher P_D values are allowed as long as T_J does not exceed T_{Jmax} (See subsec 5.1).
3. In low power dissipation state, T_A can be extended to this range as long as T_J does not exceed T_{Jmax} (See subsec 5.1).

5.3.2 Operating conditions at power-up/power-down

The parameters given in the table below are based on tests under normal operating conditions.

Table 13. Operating conditions at power-up/power-down

Symbol	Parameter	Conditions	Min	Max	Unit
t_{VDD}	V_{VDD} rise time rate	$T_A = 27^\circ\text{C}$	100	∞	$\mu\text{S/V}$
	V_{VDD} fall time rate		100	∞	

5.3.3 Embedded reset and power control block characteristics

The parameters given in the table below are based on the ambient temperature and the V_{DD} supply voltage listed in Table 12.

Table 14. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
		PLS[3: 0]=0000 (Rising edge)	1.813	1.819	1.831	V
		PLS[3: 0]=0000 (Falling edge)		1.705		V
		PLS[3: 0]=0001 (Rising edge)	2.112	2.116	2.124	V
		PLS[3: 0]=0001 (Falling edge)		2.0		V

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
		PLS[3: 0]=0010 (Rising edge)	2.411	2.414	2.421	V
		PLS[3: 0]=0010 (Falling edge)		2.297		V
		PLS[3: 0]=0011 (Rising edge)	2.711	2.714	2.719	V
		PLS[3: 0]=0011 (Falling edge)		2.597		V
		PLS[3: 0]=0100 (Rising edge)	3.011	3.013	3.018	V
		PLS[3: 0]=0100 (Falling edge)		2.895		V
		PLS[3: 0]=0101 (Rising edge)	3.311	3.313	3.317	V
		PLS[3: 0]=0101 (Falling edge)		3.194		V
		PLS[3: 0]=0110 (Rising edge)	3.611	3.613	3.616	V
		PLS[3: 0]=0110 (Falling edge)		3.494		V
		PLS[3: 0]=0111 (Rising edge)	3.91	3.913	3.916	V
		PLS[3: 0]=0111 (Falling edge)		3.793		V
		PLS[3: 0]=1000 (Rising edge)	4.21	4.212	4.215	V
		PLS[3: 0]=1000 (Falling edge)		4.092		V
		PLS[3: 0]=1001 (Rising edge)	4.51	4.512	4.515	V
		PLS[3: 0]=1001 (Falling edge)		4.391		V
		PLS[3: 0]=1010 (Rising edge)	4.809	4.811	4.813	V
		PLS[3: 0]=1010 (Falling edge)		4.69		V
$V_{PVDhyst}^{(2)}$	PVD hysteresis			100		mV
$V_{POR/PDR}$	Power on/down reset threshold	Falling edge	1.63 ⁽¹⁾	1.66	1.68	V
		Rising edge		1.75		V
$V_{PDRhys}^{(2)}$	PDR hysteresis			90.9		mV
$T_{RSTTEMPO}^{(2)}$	Reset duration			20		ms

1. The product behavior is guaranteed by design down to the minimum value $V_{POR/PDR}$.
2. Guaranteed by design, not tested in production.

Note: The reset duration is measured from power-on (POR reset) to the time when the user application code reads the first instruction.

5.3.4 Embedded internal reference voltage

The parameters given in the table below are based on the ambient temperature and the V_{DD} supply voltage listed in Table 12.

Table 15. Embedded internal reference voltage⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{REFINT}	Internal reference voltage	$-40^{\circ}\text{C} < T_A < +105^{\circ}\text{C}$		1.2		V
		$-40^{\circ}\text{C} < T_A < +85^{\circ}\text{C}$		1.2		V

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$T_{S_vrefint}^{(1)}$	ADC sampling time when reading the internal reference voltage		40			μS

1. Shortest sampling time can be determined in the application by multiple iterations.

5.3.5 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

All Run-mode current consumption measurements given in this section are performed with a reduced code.

Maximum current consumption

The MCU is placed under the following conditions:

- All I/O pins are in analog input mode, and are connected to a static level — V_{DD} or V_{SS} (no load)
- All peripherals are disabled except when explicitly mentioned
- The Flash memory access time is adjusted to the f_{HCLK} (0 ~ 24 MHz is 0 waiting period , 24 ~ 48 MHz is 1 waiting period, 48 ~ 72 MHz is 2 waiting period, 72 ~ 96 MHz is 3 waiting period).
- The instruction prefetching function is on. When the peripherals are enabled:
 $f_{PCLK1} = f_{HCLK}$.

Note: The instruction prefetching function must be set before setting the clock and bus divider.

The parameters given in the table (Table 16、Table 17、Table 18) are based on the ambient temperature and the V_{DD} supply voltage listed in Table 12 .

Table 16. Typical current consumption in Run mode, code executing from Flash memory

Symbol	Parameter	Conditions	f_{HCLK}	Typ ⁽¹⁾		Unit
				All peripherals enabled ⁽²⁾	All peripherals disabled	
I_{DD}	Supply current in run mode	Internal clock ⁽²⁾	96MHz	26.23	15.2	mA
			72MHz	20.52	12.19	
			48MHz	14.71	9.13	
			36MHz	11.76	7.58	
			24MHz	8.84	6.03	
I_{DD}	Supply current in run mode	External clock ⁽²⁾	8MHz	4.1	3.14	

1. The typical value is tested at $T_A = 25^\circ\text{C}$. Guaranteed by design, not tested in production.
2. External clock is 8MHz, when $f_{\text{HCLK}} > 8\text{MHz}$ enable PLL.

Table 17. Typical current consumption in Sleep mode, code executing from Flash memory or RAM

Symbol	Parameter	Conditions	f_{HCLK}	Typ ⁽¹⁾		Unit
				All peripherals enabled ⁽²⁾	All peripherals disabled	
I_{DD}	Supply current in Sleep mode	External clock ⁽²⁾	96MHz	22.41	10.92	mA
			72MHz	17.57	8.96	
			48MHz	12.68	6.96	
			36MHz	10.29	5.95	
			24MHz	7.79	4.9	
			8MHz	3.46	2.8	

1. The typical value is tested at $T_A = 25^\circ\text{C}$. From a comprehensive evaluation, it is tested in terms of V_{DDmax} and f_{HCLKmax} enable peripherals in production.
2. External clock is 8MHz, when $f_{\text{HCLK}} > 8\text{MHz}$ enable PLL.

Table 18. Maximum current consumption in Stop and Standby mode, code executing from Flash memory

Symbol	Parameter	Conditions	Max	Unit
			$T_A = 25^\circ\text{C}$	
I_{DD}	Supply current in Stop mode	Enter the stop mode after reset, $V_{\text{DD}} = 3.3\text{V}$	402	μA
	Supply current in Standby mode	Enter the standby mode after reset, $V_{\text{DD}} = 3.3\text{V}$	0.4	
$I_{\text{DD_VBAT}}$	Supply current in the backup area	Low speed oscillator and RTC are on, $V_{\text{DD}}/V_{\text{BAT}} = 3.3\text{V}$	0.2	μA

1. I/O status is analog input.

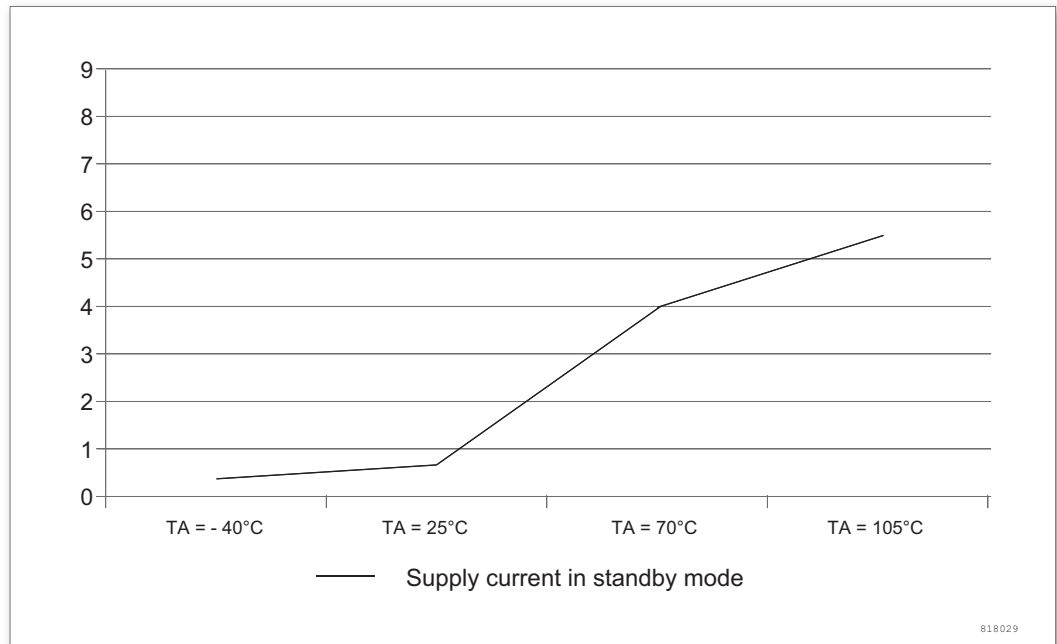


Figure 11. Typical current consumption in standby mode vs. temperature at $V_{DD} = 3.3V$

Typical current consumption

The MCU is placed under the following conditions:

- All I/O pins are in analog input configuration, and are connected to a static level — V_{DD} or V_{SS} (no load).
- All the peripherals are closed, unless otherwise specified.
- The Flash memory access time is adjusted to the f_{HCLK} (0 ~ 24 MHz is 0 waiting period, 24 ~ 48 MHz is 1 waiting period, 48 ~ 72 MHz is 2 waiting period, 72 ~ 96 MHz is 3 waiting period).
- The ambient temperature and V_{DD} supply voltage conditions are summarized in Table 12.
- The instruction prefetching function is on. When the peripherals are enabled:
 $f_{PCLK1} = f_{HCLK}$.

Note: The instruction prefetch function must be set before the clock is set and the bus is divided.

On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in Table 19. The MCU is placed under the following conditions:

- All I/O pins are in analog input mode, and are connected to a static level — V_{DD} or V_{SS} (no load).
- All peripherals are disabled except when explicitly mentioned.
- The given value is calculated by measuring the current consumption.
 - with all peripherals clocked OFF
 - with only one peripheral clocked on

- Ambient operating temperature and supply voltage conditions V_{DD} summarized in Table 12.

Table 19. On-chip peripheral current consumption⁽¹⁾

Peripheral		Typical consumption at 25 °C	Unit	Peripheral		Typical consumption at 25 °C	Unit
APB1	TIM2	0.098	mA	APB2	GPIOA	0.045	mA
	TIM3	0.062			GPIOB	0.046	
APB1	TIM4	0.055	mA	APB2	GPIOC	0.052	mA
	SPI2	0.133			GPIOD	0.046	
	UART2	0.077			ADC1	0.051	
	UART3	0.078			ADC2	0.052	
	I2C1	0.132			TIM1	0.121	
	I2C2	0.134			SPI1	0.122	
	USB	0.058			UART1	0.078	
	CAN	0.033					

1. $f_{HCLK} = 96\text{MHz}$, $f_{APB1} = f_{HCLK}/2$, $f_{APB2} = f_{HCLK}$, the prescale coefficient for each device is the default value.

5.3.6 External clock source characteristics

High-speed external user clock generated from an external source

The characteristic parameters given in the following table are measured using a high-speed external clock source, ambient temperature and power supply voltage meet the conditions of General operating conditions.

Table 20. High-speed external user clock characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{HSE_ext}	User external clock source frequency ⁽¹⁾		2	8	24	MHz
V_{HSEH}	OSC_IN input pin high level voltage		$0.7V_{DD}$		V_{DD}	V
V_{HSEL}	OSC_IN input pin low level voltage		V_{SS}		$0.3V_{DD}$	
$t_{w(HSE)}$	OSC_IN high or low time ⁽¹⁾		16			nS
$t_{r(HSE)}$ $t_{f(HSE)}$	OSC_IN rise or fall time ⁽¹⁾				20	
$C_{in(HSE)}$	OSC_IN input capacitance ⁽¹⁾			5		pF
$DuCy_{(HSE)}$	Duty cycle		45		55	%
I_L	OSC_IN input leakage current	$V_{SS} \leq V_{IN} \leq V_{DD}$			± 1	uA

1. Guaranteed by design, not tested in production.

Low-speed external user clock characteristics

The characteristic parameters given in the following table are measured using a low-speed external clock source, ambient temperature and power supply voltage meet the conditions of General operating conditions.

Table 21. Low-speed external user clock characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{LSE_ext}	User external clock source frequency ⁽¹⁾		16	32.768	200	KHz
V_{LSEH}	OSC_IN input pin high level voltage				1.2	V
V_{LSEL}	OSC_IN input pin low level voltage		0.25			V
$t_{w(LSE)}$	OSC_IN high or low time ⁽¹⁾			15259		nS
$t_{r(LSE)}$	OSC_IN rise time ⁽¹⁾			30		nS
$t_{f(LSE)}$	OSC_IN fall time ⁽¹⁾			30		nS
$C_{in(LSE)}$	OSC_IN input capacitance ⁽¹⁾			5		pF
$DuCy_{(LSE)}$	Duty cycle			50		%
I_L	OSC_IN input leakage current	$V_{SS} \leq V_{IN} \leq V_{DD}$		0.03		uA

1. Guaranteed by design, not tested in production.

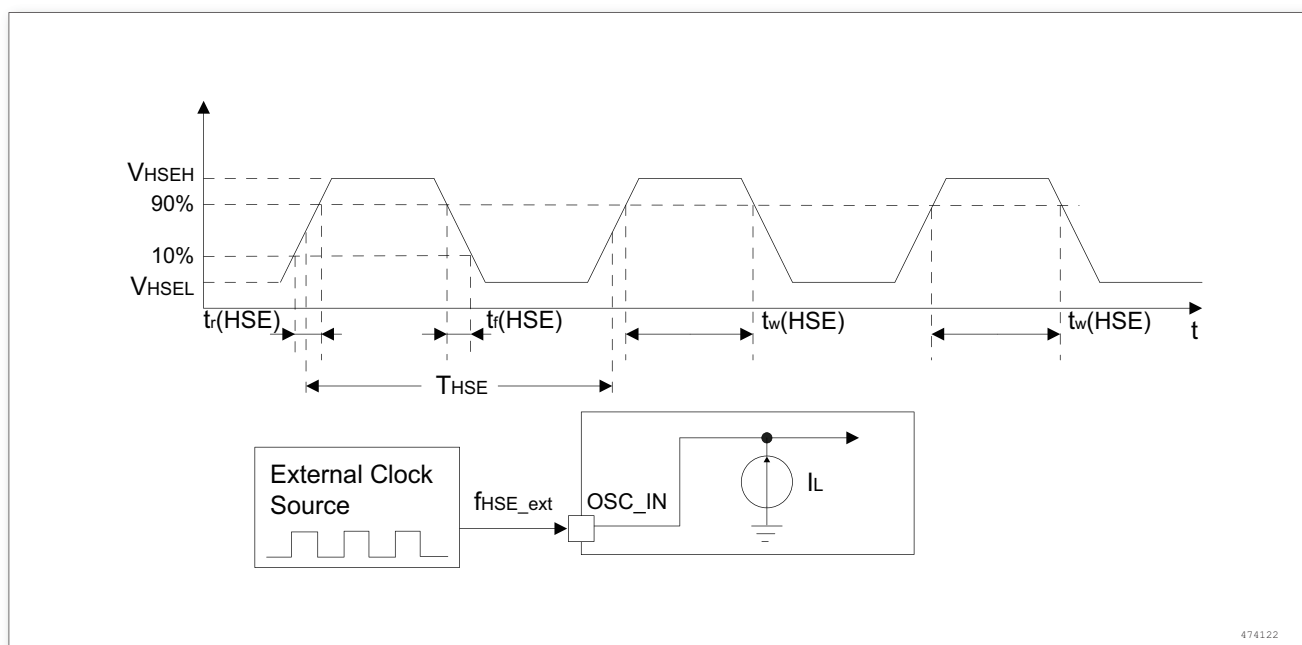


Figure 12. High-speed external clock source AC timing diagram

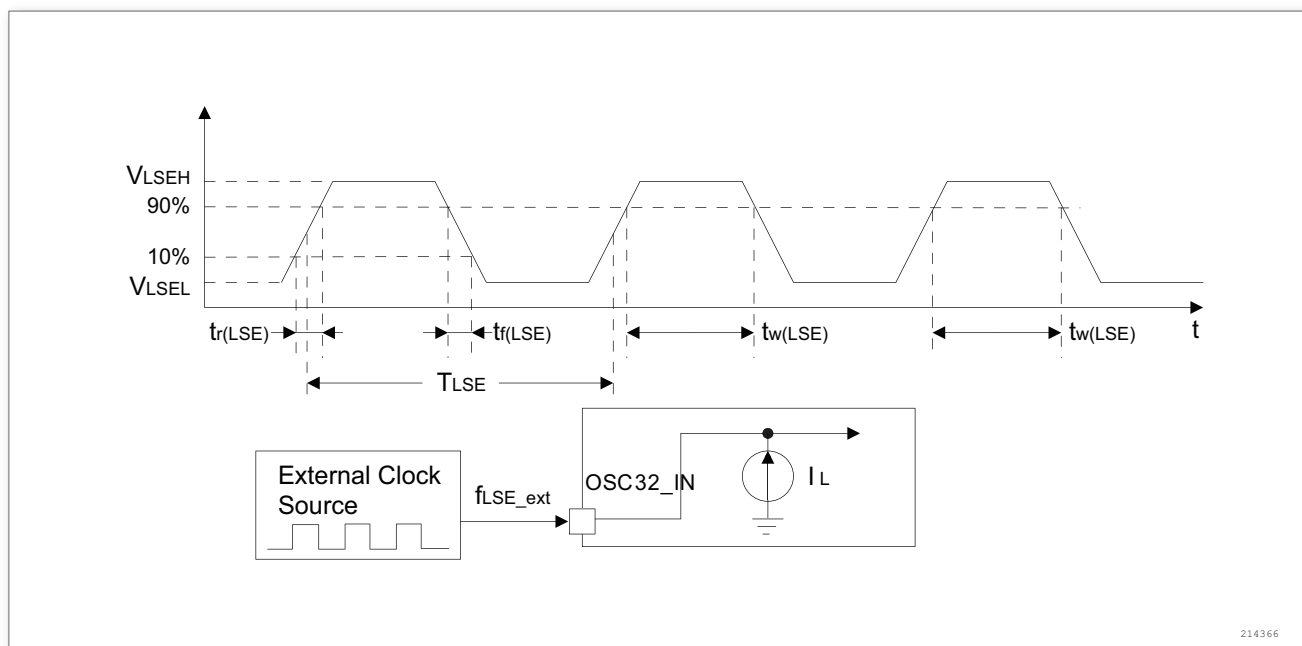


Figure 13. Low-speed external clock source AC timing diagram

High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with an 2 to 24 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in the table below. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy...).

Table 22. HSE oscillator characteristics⁽¹⁾⁽²⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{OSC_IN}	Oscillator frequency		2	8	24	MHz
R_F	Feedback resistor	$R_S = 30\Omega$		1000		k Ω
C_{L1} $C_{L2}^{(3)}$	The proposed load capacitance corresponds to the crystal serial impedance (R_S) ⁽⁴⁾	$V_{DD} = 3.3V$ $V_{IN} = V_{SS}$ 30pF load		30		pF
I_2	HSE current consumption	Startup			1	mA
g_m	Oscillator transconductance	V_{DD} is stabilized	25			mA/V
$t_{SU(HSE)}^{(5)}$	Startup time	$R_S = 30\Omega$		2		mS

1. Resonator characteristics given by the crystal/ceramic resonator manufacturer characteristics Parameter.

2. Guaranteed by design, not tested in production.
3. For C_{L1} and C_{L2} , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 25 pF range (Typ.) , designed for high-frequency applications, and selected to match the requirements of the crystal or resonator. C_{L1} and C_{L2} are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of C_{L1} and C_{L2} . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing C_{L1} and C_{L2} .
4. The relatively low value of the RF resistance can be used to avoid problems arising from the use of wet conditions to provide protection, this environment resulting in leakage and bias conditions have changed. However, if the MCU is applied in bad wet conditions, the design needs to take this parameter into account.
5. $t_{SU(HSE)}$ is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.

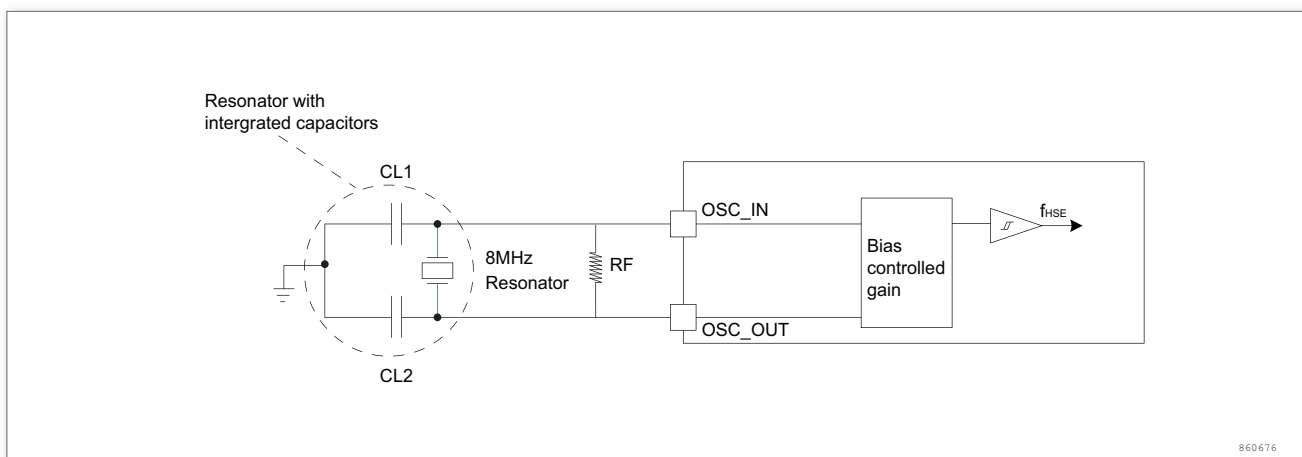


Figure 14. Typical application with an 8 MHz crystal

Low-speed external clock generated from a crystal/ceramic resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in the table below. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy). Note: For C_{L1} and C_{L2} , it is recommended to use a high quality ceramic capacitor between 5pF and 15pF and select the crystal or resonator that meets the requirements. Usually C_{L1} and C_{L2} have the same parameters. Crystal manufacturers typically give the parameters of the load capacitance in a serial combination of C_{L1} and C_{L2} . The load capacitance C_L is calculated by: $C_L = C_{L1} \times C_{L2} / (C_{L1} + C_{L2}) + C_{stray}$, where C_{stray} is the capacitance of the pin and the capacitance associated with the PCB or PCB. Its typical value is between 2pF and 7pF. WARNING: To avoid exceeding

the maximum value of C_{L1} and C_{L2} (15pF), it is highly recommended to use a resonator with a load capacitance of $C_L \leq 7\text{pF}$. A resonator with a load capacitance of 12.5pF cannot be used. For example, if a resonator with a load capacitance of $C_L = 6\text{pF}$ is selected and $C_{\text{stray}} = 2\text{pF}$, then $C_{L1} = C_{L2} = 8\text{pF}$.

Table 23. LSI oscillator characteristics ($f_{\text{LSE}}=32.768\text{KHz}$) ⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
R_F	Feedback resistor			25		$\text{M}\Omega$
C_{L1} C_{L2} ⁽²⁾	The proposed load capacitance corresponds to the crystal serial impedance (R_S) ⁽³⁾	$R_S = 30\Omega$			4	pF
I_2	LSE current consumption	$V_{DD} = 3.3\text{V}$ $V_{IN} = V_{SS}$		0.08		μA
g_m	Oscillator transconductance			0.5		$\mu\text{A/V}$
$t_{\text{SU(HSE)}}^{(4)}$	Startup time	V_{DD} is stabilized		1	4	S

1. Guaranteed by design, not tested in production.
2. See the note and warning paragraphs above this table.
3. Selecting a high quality oscillator with a small R_S value (such as MSIV-TIN 32.768KHz) optimizes current consumption. Please consult the crystal manufacturer for details.
4. $t_{\text{SU(HSE)}}$ is the start-up time, which is measured from the time the software enables HSE until a stable 8MHz oscillation is obtained. This value is measured on a standard crystal resonator, which may vary greatly depending on the crystal manufacturer.

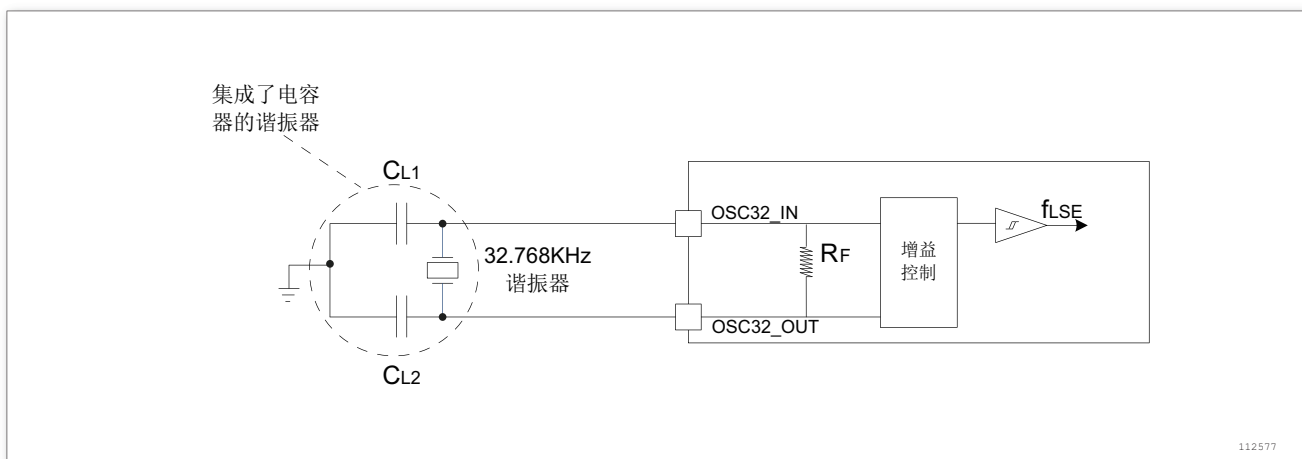


Figure 15. Typical application with a 32.768 kHz crystal

5.3.7 Internal clock source characteristics

The characteristic parameters given in the table below are measured using ambient temperature and supply voltage in accordance with general operating conditions.

High-speed internal (HSI) oscillator

Table 24. HSI oscillator characteristics⁽¹⁾⁽²⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{HSI}	Frequency		40	48	64	MHz
ACC_{HSI}	Accuracy of the HSI oscillator	$T_A = -40^{\circ}\text{C} \sim 105^{\circ}\text{C}$	-5		5	%
ACC_{HSI}	Accuracy of the HSI oscillator	$T_A = -10^{\circ}\text{C} \sim 85^{\circ}\text{C}$	-3		3	%
ACC_{HSI}	Accuracy of the HSI oscillator	$T_A = 0^{\circ}\text{C} \sim 70^{\circ}\text{C}$	-2		2	%
ACC_{HSI}	Accuracy of the HSI oscillator	$T_A = 25$	-1		1	%
$t_{\text{SU(HSI)}}$	HSI oscillator startup time				2	μS
$I_{\text{DD(HSI)}}$	HSI oscillator power consumption			81	200	μA

1. $V_{\text{DD}} = 3.3\text{V}$, $T_A = -40^{\circ}\text{C} \sim 105^{\circ}\text{C}$, unless otherwise specified.
2. Guaranteed by design, not tested in production.

Low-speed internal (LSI) oscillator

Table 25. LSI oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{\text{LSI}}^{(2)}$	Frequency		31	40	75	KHz
$t_{\text{SU(LSI)}}^{(2)}$	LSI oscillator startup time				1	μS
$I_{\text{DD(LSI)}}^{(3)}$	LSI oscillator power consumption			1.1	1.7	μA

1. $V_{\text{DD}} = 3.3\text{V}$, $T_A = -40^{\circ}\text{C} \sim 105^{\circ}\text{C}$, Unless otherwise stated
2. Comprehensive assessment, not tested in production.
3. Guaranteed by design, not tested in production.

Wake-up times from low power mode

The wake-up times listed in the table below are measured during the wake-up phase of the internal clock HSI. The clock source used when waking up depends on the current operating mode:

- Stop or Standby mode: The clock source is the oscillator
- Sleep mode: The clock source is the clock used when entering sleep mode

All times are measured using ambient temperature and supply voltage in accordance with common operating conditions.

Table 26. Low-power mode wakeup timings

Symbol	Parameter	Conditions	Max	Unit
$t_{WUSLEEP}^{(1)}$	Wakeup from Sleep mode	HSI clock wakeup	4.2	μS
$t_{WUSTOP}^{(1)}$	Wakeup from Stop mode (The regulator is in run mode)	HSI clock wakeup = $2\mu\text{S}$	6.3	
$t_{WUSTDBY}^{(1)}$	Wakeup from Standby mode	HSI clock wakeup = $2\mu\text{S}$ The regulator wakes up from the off mode = $38\mu\text{S}$	47	mS

1. The wake-up time is measured from the start of the wake-up event to the user program to read the first instruction.

5.3.8 PLL characteristics

The parameters listed in the table below are measured using ambient temperature and supply voltage in accordance with common operating conditions.

Table 27. PLL characteristics⁽¹⁾

Symbol	Parameter	Min	Typ	Max	Unit
$f_{\text{PLL_IN}}$	PLL input clock ⁽²⁾	2		24	MHz
	PLL input clock duty cycle	40		60	%
$f_{\text{PLL_OUT}}$	PLL multiplier output clock	40		200	MHz
t_{LOCK}	PLL lock time			100	μS

1. Guaranteed by design, not tested in production.
 2. Take care to use the appropriate multiplier factors to obtain PLL input clock values compatible with the range defined by $f_{\text{PLL_OUT}}$.

5.3.9 Memory characteristics

Flash memory

The characteristics are given at $T_A = -40^\circ\text{C} \sim 105^\circ\text{C}$ unless otherwise specified.

Table 28. Flash memory characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
t_{prog}	8-bit programming time	$T_A = -40^\circ\text{C} \sim 125^\circ\text{C}$	4			μS

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
t_{ERASE}	Page (512K bytes) erase time	$T_A = -40^{\circ}\text{C} \sim 125^{\circ}\text{C}$		4	5	mS
t_{ME}	Mass erase time	$T_A = -40^{\circ}\text{C} \sim 125^{\circ}\text{C}$	20		40	mS
I_{DD}	Supply current	Read mode, $f_{HCLK} = 48\text{MHz}$		5	6	mA
		Write mode, $f_{HCLK} = 48\text{MHz}$			7	mA
		Erase mode, $f_{HCLK} = 48\text{MHz}$			2	mA
I_{SB}	Standby current			1@25°C	50@125°C	μA
I_{DEP}	Deep Standby current			0.5	15@125°C	μA

Table 29. Flash memory endurance and data retention⁽¹⁾⁽²⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
NEND	Endurance (Annotation: Erase number of times)	$T_A = -40^{\circ}\text{C} \sim 85^{\circ}\text{C}$ $T_A = -40^{\circ}\text{C} \sim 105^{\circ}\text{C}$	10			K cycle
t_{RET}	Data retention	1 K cycle ⁽²⁾ at $T_A = 85^{\circ}\text{C}$	30			Year
		1 K cycle ⁽¹⁾⁽²⁾ at $T_A = 105^{\circ}\text{C}$	10			
		10 K cycle ⁽¹⁾⁽²⁾ at $T_A = 55^{\circ}\text{C}$	20			

1. Guaranteed by design, not tested in production.
2. Cycle tests are carried out in the whole temperature range.

5.3.10 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports), the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- Electrostatic discharge (ESD) (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to VDD and VSS through a 100 pF capacitor, until a functional disturbance occurs. This test is

compliant with the IEC 1000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in the following table. They are based on the EMS levels and classes defined in application note.

Table 30. EMS characteristics

Symbol	Parameter	Conditions	Level/Class
V_{EFT}	Fast transient voltage burst limits to be applied through 100 pF on V_{DD} and V_{SS} pins to induce a functional disturbance	$V_{DD} = 3.3V, T_A = +25^{\circ}C$, $f_{HCLK} = 48MHz$. Conforming to IEC 1000-4-4	TBD

Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and pre-qualification tests in relation with the EMC level requested for his application.

Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical Data corruption (for example control registers)

Prequalification trials

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors.

5.3.11 Absolute Maximum (Electrical Sensitivity)

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts \times (n+1) supply pins). This test conforms to the JESD22-A114/C101 standard.

Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin
- A current injection is applied to each input, output and configurable I/O pin

These tests are compliant with EIA/JESD78A IC latch-up standard.

Table 31. ESD characteristics

Symbol	Parameter	Conditions	Type	Max	Unit
$V_{ESD(HBM)}$	Electrostatic discharge voltage (Human body model)	$T_A = +25^\circ\text{C}$, Conforming to JESD22-A114		2000	V
$V_{ESD(CDM)}$	Electrostatic discharge voltage (Charging device model)	$T_A = +25^\circ\text{C}$, Conforming to JESD22-C101		500	
I_{LU}	Latch-up current	$T_A =$ $+25^\circ\text{C}$, Conforming to JESD78A		200	mA

5.3.12 I/O port characteristics

General input/output characteristics

Unless otherwise specified, the parameters given in Table 9 are derived from tests.

Table 32. I/O static characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{IL}	Low level input voltage	CMOS Port	-0.5		1.1	V
V_{IH}	High level input voltage	CMOS Port	2.08			V
V_{hys}	Schmitt trigger hysteresis ⁽¹⁾		500	700	800	mV
I_{lkg}	Input leakage current ⁽²⁾				1	μA
R_{PU}	Weak pull-up equivalent resistor ⁽³⁾	$V_{IN} = V_{SS}$	30	50	100	k Ω
R_{PD}	Weak pull-down equivalent resistor ⁽³⁾	$V_{IN} = V_{DD}$	30	50	100	

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
C_{IO}	I/O pin capacitance				5	pF

1. Schmitt Trigger switching hysteresis voltage level. Data based on design simulation only. Not tested in production.
2. The leakage could be higher than the maximum value, if negative current is injected on adjacent pins.
3. Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This PMOS/NMOS contribution to the series resistance is minimal (10% order).

All I/Os are CMOS (no software configuration required). Their characteristics cover more than the strict CMOS-technology.

- For V_{IH} :
 - If V_{DD} is between [2.50V~ 3.08V]; use CMOS features.
 - If V_{DD} is between [3.08V~ 3.60V]; include CMOS.
- For V_{IL} :
 - Use CMOS features.

Output driving current

The GPIOs (general purpose input/outputs) can sink or source up to $\pm 20\text{mA}$.

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in 5.2:

- The sum of the currents obtained from V_{DD} for all I/O ports, plus the maximum operating current that the MCU obtains on V_{DD} , cannot exceed the absolute maximum rating I_{VDD} .
- The sum of the currents drawn by all I/O ports and flowing out of V_{SS} , plus the maximum operating current of the MCU flowing out on V_{SS} , cannot exceed the absolute maximum rating I_{VSS} .

Output voltage levels

Unless otherwise stated, the parameters listed in the table below are measured using the ambient temperature and V_{DD} supply voltage in accordance with the condition Table 9. All I/O ports are CMOS compatible.

Table 33. Output voltage characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
$V_{OL}^{(1)}$	Output low level voltage for an I/O pin, when 8 pins absorb current	CMOS Port, $I_{IO} = +8\text{mA}$ $2.7\text{V} < V_{DD} < 3.6\text{V}$		0.4	

Symbol	Parameter	Conditions	Min	Max	Unit
$V_{OH}^{(2)}$	Output high level voltage for an I/O pin, when 8 pins output current	CMOS Port, $I_{IO} = +8mA$ $2.7V < V_{DD} < 3.6V$	$0.8V_{DD}$		V
$V_{OL}^{(1)(3)}$	Output low level voltage for an I/O pin, when 8 pins absorb current	$I_{IO} = +20mA$ $2.7V < V_{DD} < 3.6V$		0.4	
$V_{OH}^{(2)(3)}$	Output high level voltage for an I/O pin, when 8 pins output current		$0.8V_{DD}$		
$V_{OL}^{(2)(3)}$	Output low level voltage for an I/O pin, when 8 pins absorb current	$I_{IO} = +6mA$ $2V < V_{DD} < 2.7V$		TBD	
$V_{OH}^{(2)(3)}$	Output high level voltage for an I/O pin, when 8 pins output current	$I_{IO} = +6mA$ $2V < V_{DD} < 2.7V$	TBD		

1. The current absorbed by the chip I_{IO} must always follow the absolute maximum ratings given in the table, and the sum of I_{IO} (all I/O feet and control pins) must not exceed I_{VSS} .
2. The current output I_{IO} of the chip must always follow the absolute maximum rating given in the table, and the sum of I_{IO} (all I/O pins and control pins) must not exceed I_{VDD} .
3. Data based on characterization results. Not tested in production.

Input/output AC characteristics

The definitions and values of the input and output AC characteristics are given in figure 16 and Table 34, respectively.

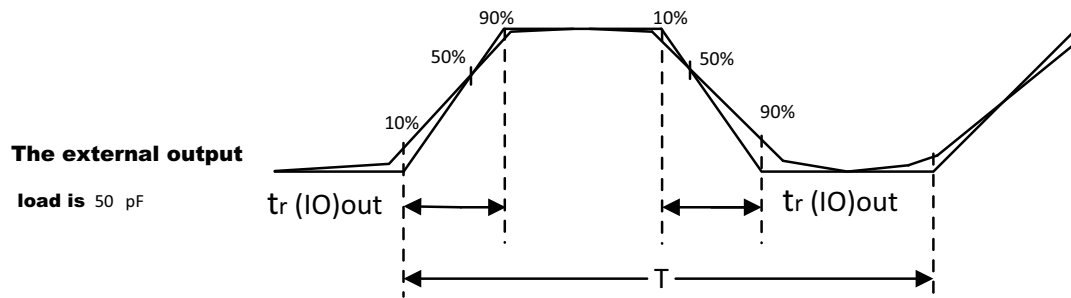
Unless otherwise stated, the parameters listed in Table 34 are measured using the ambient temperature and supply voltage in accordance with the condition Table 9.

Table 34. I/O AC characteristics⁽¹⁾

OSPEEDRy [1:0] value (1)	Symbol	Parameter	Conditions	Min	Max	Unit
01 (10MHz)	$f_{max(IO)out}$	Maximum frequency ⁽²⁾	$C_L = 50pF$, $V_{DD} = 2V \sim 3.6V$		10	MHz
	$t_{f(IO)out}$	Output fall time	$C_L = 50pF$, $V_{DD} = 2V \sim 3.6V$		25 ⁽³⁾	nS
	$t_{r(IO)out}$	Output rise time			25 ⁽³⁾	
10 (20MHz)	$f_{max(IO)out}$	Maximum frequency ⁽²⁾	$C_L = 50pF$, $V_{DD} = 2V \sim 3.6V$		20	MHz
	$t_{f(IO)out}$	Output fall time	$C_L = 50pF$, $V_{DD} = 2V \sim 3.6V$		125 ⁽³⁾	nS

OSPEEDRy [1:0] value (1)	Symbol	Parameter	Conditions	Min	Max	Unit
	$t_{r(IO)out}$	Output rise time			125 ⁽³⁾	
11 (50MHz)	$f_{max(IO)out}$	Maximum frequency ⁽²⁾	$C_L = 30pF$, $V_{DD} = 2.7V \sim 3.6V$		50	MHz
			$C_L = 50pF$, $V_{DD} = 2.7V \sim 3.6V$		30	
			$C_L = 50pF$, $V_{DD} = 2V \sim 2.7V$		20	
	$t_{f(IO)out}$	Output fall time	$C_L = 30pF$, $V_{DD} = 2.7V \sim 3.6V$		5	nS
			$C_L = 50pF$, $V_{DD} = 2.7V \sim 3.6V$		8	
			$C_L = 50pF$, $V_{DD} = 2V \sim 2.7V$		12	
	$t_{r(IO)out}$	Output rise time	$C_L = 30pF$, $V_{DD} = 2.7V \sim 3.6V$		5	
			$C_L = 50pF$, $V_{DD} = 2.7V \sim 3.6V$		8	
			$C_L = 50pF$, $V_{DD} = 2V \sim 2.7V$		12	
	t_{EXTIpw}	Pulse width of external signals detected by the EXTI controller		10		nS

1. The speed of the I/O port can be configured via MODEx[1:0]. See the description of the GPIO Port Configuration Register in this chip reference manual.
2. The maximum frequency is defined in figure 16.



Maximum frequency is achieved if $((t_r + t_f) \leq 2/3)T$, and if the duty cycle is (45 ~ 55%) when loaded by C_L (see the i/O AC characteristics definition)

868304

Figure 16. I/O AC characteristics

5.3.13 NRST pin characteristics

The NRST pin input driver uses the CMOS technology. It is connected to a permanent pullup resistor, R_{PU} .

Unless otherwise stated, the parameters listed in the table below are measured using the ambient temperature and V_{DD} supply voltage in accordance with the condition Table 9.

Table 35. NRST pin characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IL(NRST)}^{(1)}$	NRST input low level voltage		-0.5		0.8	V
$V_{IH(NRST)}^{(1)}$	NRST input high level voltage		2		V_{DD}	
$V_{hys(NRST)}$	NRST Schmitt trigger voltage hysteresis			$0.2V_{DD}$		V
R_{PU}	Weak pull-up equivalent resistor ⁽²⁾	$V_{IN} = V_{SS}$		15		k Ω
$V_{F(NRST)}^{(1)}$	NRST input filtered pulse				100	ns
$V_{NF(NRST)}^{(1)}$	NRST input not filtered pulse		300			

1. Data based on design simulation only. Not tested in production.
2. The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is minimal (10% order).

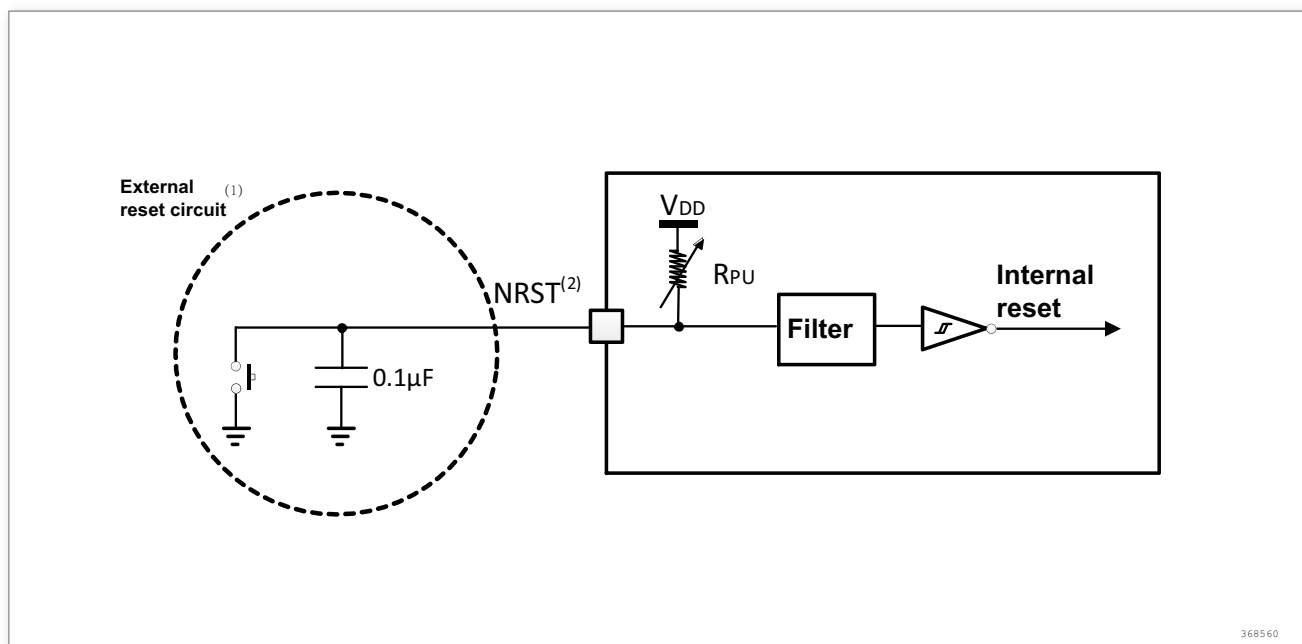


Figure 17. Recommended NRST pin protection

1. The reset network is to prevent parasitic reset
2. The user must ensure that the potential of the NRST pin is below the maximum $V_{IL(NRST)}$ listed in Table 35, otherwise the MCU cannot be reset.

5.3.14 Timer characteristics

The parameters given in the following tables are guaranteed by design.

For details on the characteristics of the I/O multiplexing function pins (output compare, input capture, external clock, PWM output) , see subsubsec 5.3.12.

Table 36. TIMx⁽¹⁾ characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
$t_{res(TIM)}$	Timer resolution time		1		$t_{TIMxCLK}$
		$f_{TIMxCLK} = 96MHz$	10.4		nS
f_{EXT}	Timer external clock frequency on CH1 to CH4		0	$f_{TIMxCLK}/2$	MHz
		$f_{TIMxCLK} = 96MHz$	0	48	
Res_{TIM}	Timer resolution			16	Bit
$t_{COUNTER}$	16-bit timer maximum period		1	65536	$t_{TIMxCLK}$
		$f_{TIMxCLK} = 96MHz$	0.0104	682	µS
t_{MAX_COUNT}	The maximum possible count			65536×65536	$t_{TIMxCLK}$
		$f_{TIMxCLK} = 96MHz$		44.7	S

1. TIMx is a generic name, representing TIM4.

5.3.15 Communication interfaces

I2C interface characteristics

Unless otherwise specified, the parameters given in Table 37 are derived from tests performed under the ambient temperature, f_{PCLK1} frequency and supply voltage conditions summarized in Table 12: General operating conditions.

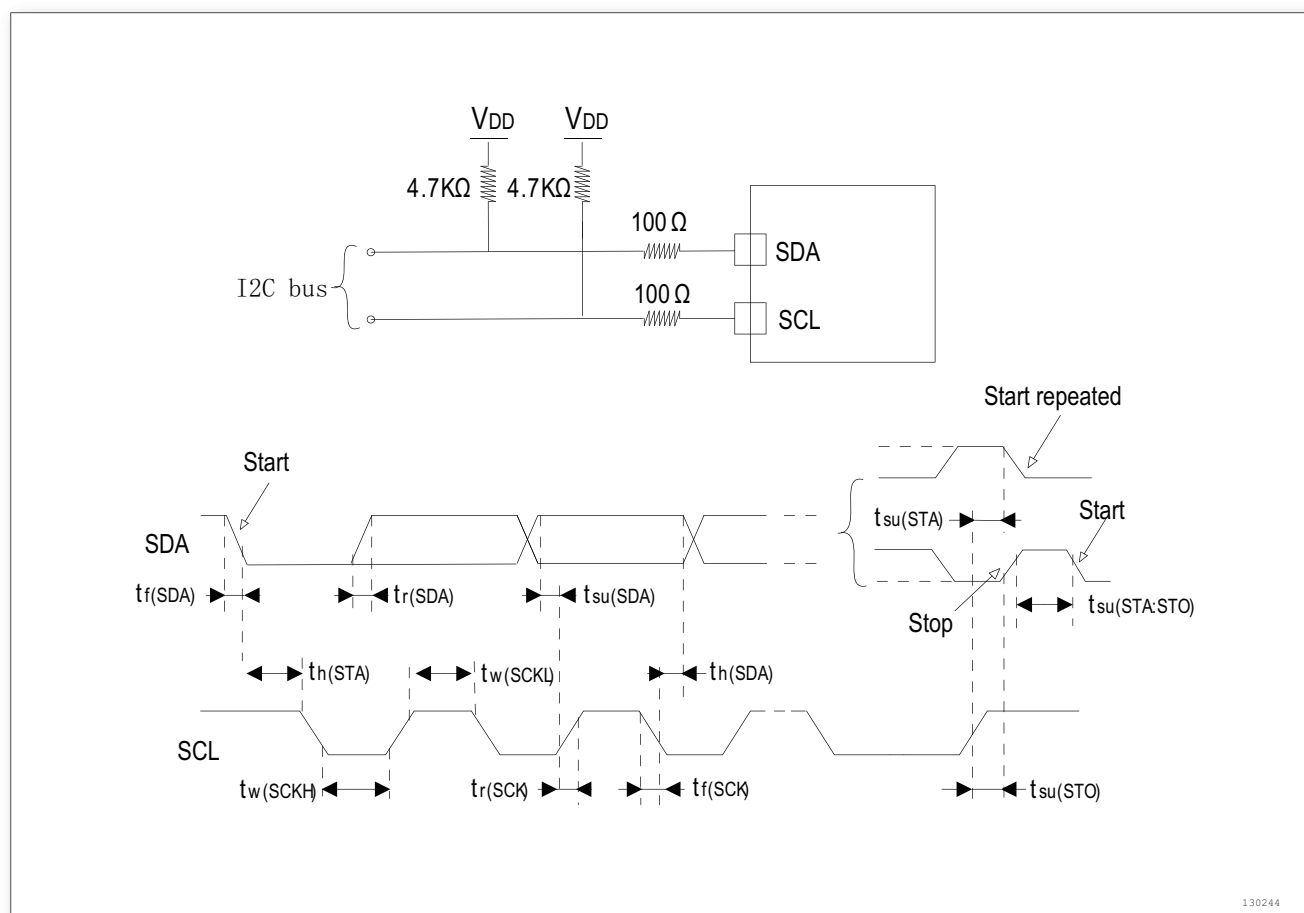
The I2C interface conforms to the standard I2C communication protocol, but has the following limitations: SDA and SCL are not true pins. When configured as open-drain output, the PMOS transistor between the pin and V_{DD} Was closed but still exists.

The I2C I/Os characteristics are listed in Table 37, the alternate function characteristics of I/Os (SDA and SCL) refer to subsubsec 5.3.12.

Table 37. I2C characteristics

Symbol	Parameter	Standard I2C ⁽¹⁾		Fast I2C ⁽¹⁾⁽²⁾		Unit
		Min	Max	Min	Max	
$t_{w(SCL)}$	SCL clock fall time	4.7		1.3		μs
$t_{w(SCLH)}$	SCL clock rise time	4.0		0.6		μs
$t_{su(SDA)}$	SDA setup time	250		100		ns
$t_{h(SDA)}$	SDA data hold time	0 ⁽³⁾		0 ⁽⁴⁾	900 ⁽³⁾	
$t_{r(SDA)} \ t_{r(SDL)}$	SDA and SCL rise time		1000	$2.0+0.1C_b$	300	
$t_{f(SDA)} \ t_{f(SDL)}$	SDA and SCL fall time		300		300	
$t_{h(STA)}$	Start condition hold time	4.0		0.6		μs
$t_{su(STA)}$	Start condition setup time	4.7		0.6		
$t_{su(STO)}$	Stop condition setup time	4.0		0.6		
$t_{w(STO:STA)}$	Time from Stop condition to Start condition	4.7		1.3		
C_b	Capacitive load of each bus		400		400	pF

1. Guaranteed by design, not tested in production.
2. f_{PCLK1} must be at least 3MHz to achieve standard mode I2C frequencies. It must be at least 12MHz to achieve fast mode I2C frequencies.
3. The maximum Data hold time has only to be met if the interface does not stretch the low period of SCL signal.
4. In order to span the undefined area of the falling edge of SCL, it must ensure that the SDA signal has a hold time of at least 300ns.

Figure 18. I2C bus AC waveform and measurement circuit⁽¹⁾

1. Measurement point is set to the CMOS level: $0.3V_{DD}$ and $0.7V_{DD}$.

SPI characteristics

Unless otherwise specified, the parameters given in Table 38 are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in Table 12.

Refer to subsubsec 5.3.12 for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO).

Table 38. SPI characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Max	Unit
$f_{SCK} 1/t_{c(SCK)}$	SPI clock frequency	Master mode	0	36	MHz
		Slave mode	0	18	
$t_{r(SCK)}$ $t_{f(SCK)}$	SPI clock rise and fall time	Load capacitance: $C = 30pF$		8	ns
$t_{su(NSS)}^{(2)}$	NSS setup time	Slave mode	$4t_{PCLK}$		
$t_{h(NSS)}^{(2)}$	NSS hold time	Slave mode	73		
$t_{w(SCKH)}^{(2)}$ $t_{w(SCKL)}^{(2)}$	SCK high and low time	Master mode, $f_{PCLK} = 36MHz$, prescale coefficient = 4	50	60	

Symbol	Parameter	Conditions	Min	Max	Unit
$t_{su(MI)}^{(2)}$	Data input setup time, Master mode	SPI1	1		
$t_{su(SI)}^{(2)}$	Data input setup time, Slave mode		1		

Symbol	Parameter	Conditions	Min	Max	Unit
$t_{h(MI)}^{(2)}$	Data input hold time, Master mode	SPI1	1		ns
$t_{h(SI)}^{(2)}$	Data input hold time, Slave mode		3		
$t_{a(SO)}^{(2)(3)}$	Data output access time	Slave mode, $f_{PCLK} = 36MHz$, prescale coefficient = 4	0	55	
		Slave mode, $f_{PCLK} = 24MHz$		$4t_{PCLK}$	
$t_{dis(SO)}^{(2)(4)}$	Data output disable time	Slave mode	10		
$t_{v(SO)}^{(2)(1)}$	Data output valid time	Slave mode (after enable edge)		25	
$t_{v(MO)}^{(2)(1)}$	Data output valid time	Master mode (after enable edge)		3	
$t_{h(SO)}^{(2)}$	Data output hold time	Slave mode (after enable edge)	25		
$t_{h(MO)}^{(2)}$		Master mode (after enable edge)	4		

1. Remapping SPI1 characteristics needs to be further determined.
2. Data based on characterization results. Not tested in production.
3. Min time is for the minimum time to drive the output and the max time is for the maximum time to validate the data.
4. Min time is for the minimum time to invalidate the output and the max time is for the maximum time to put the data in Hi-Z.

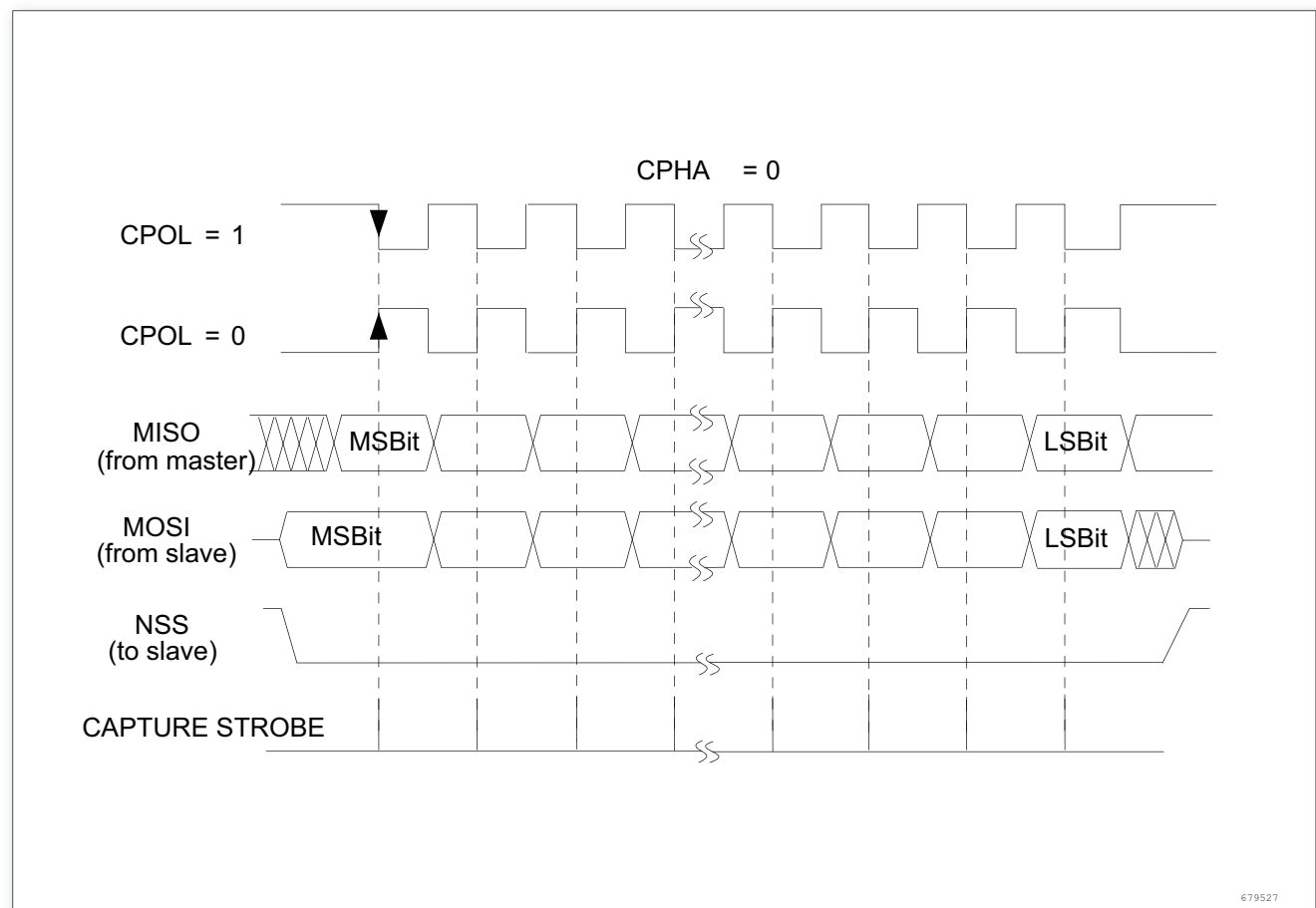
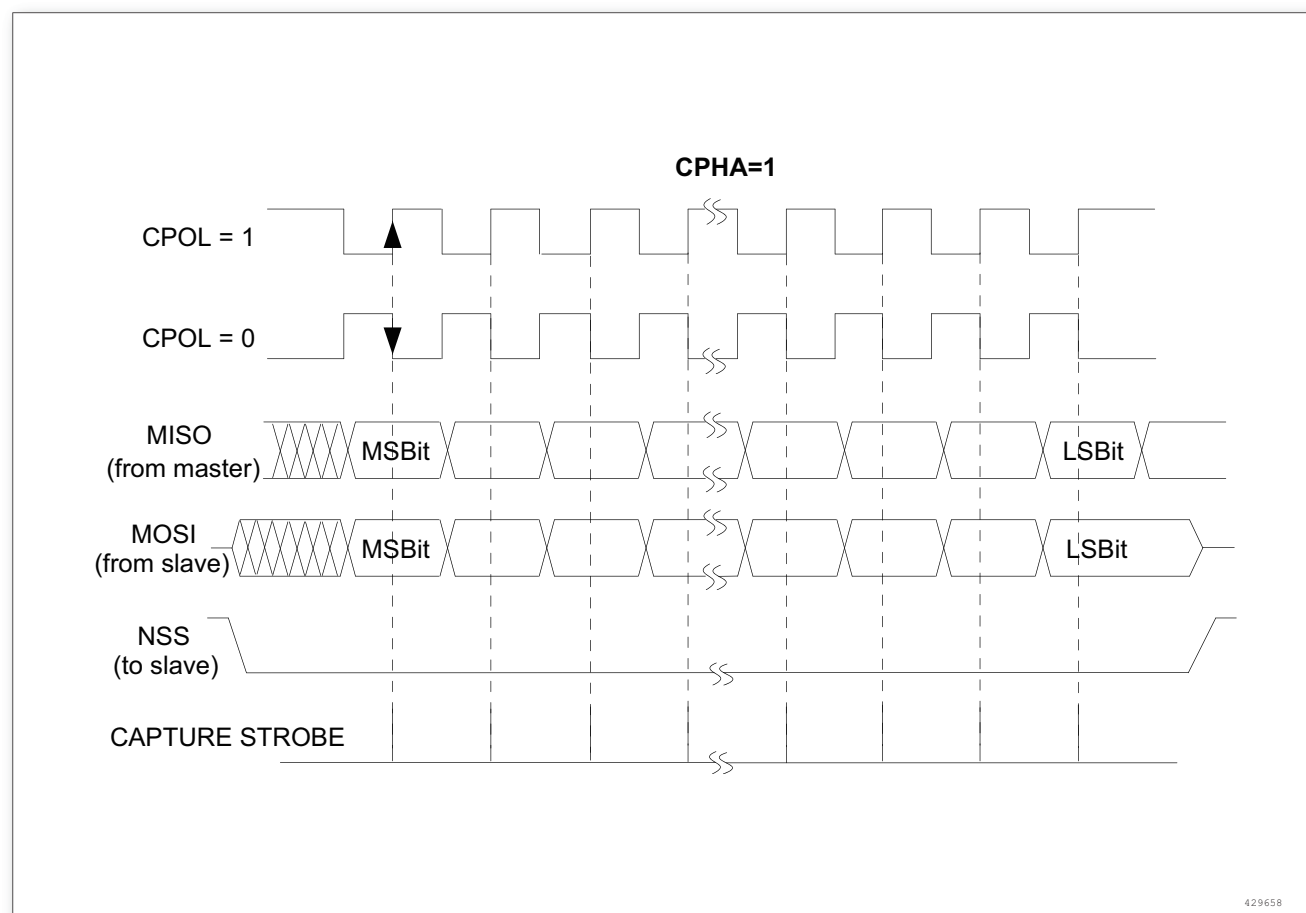
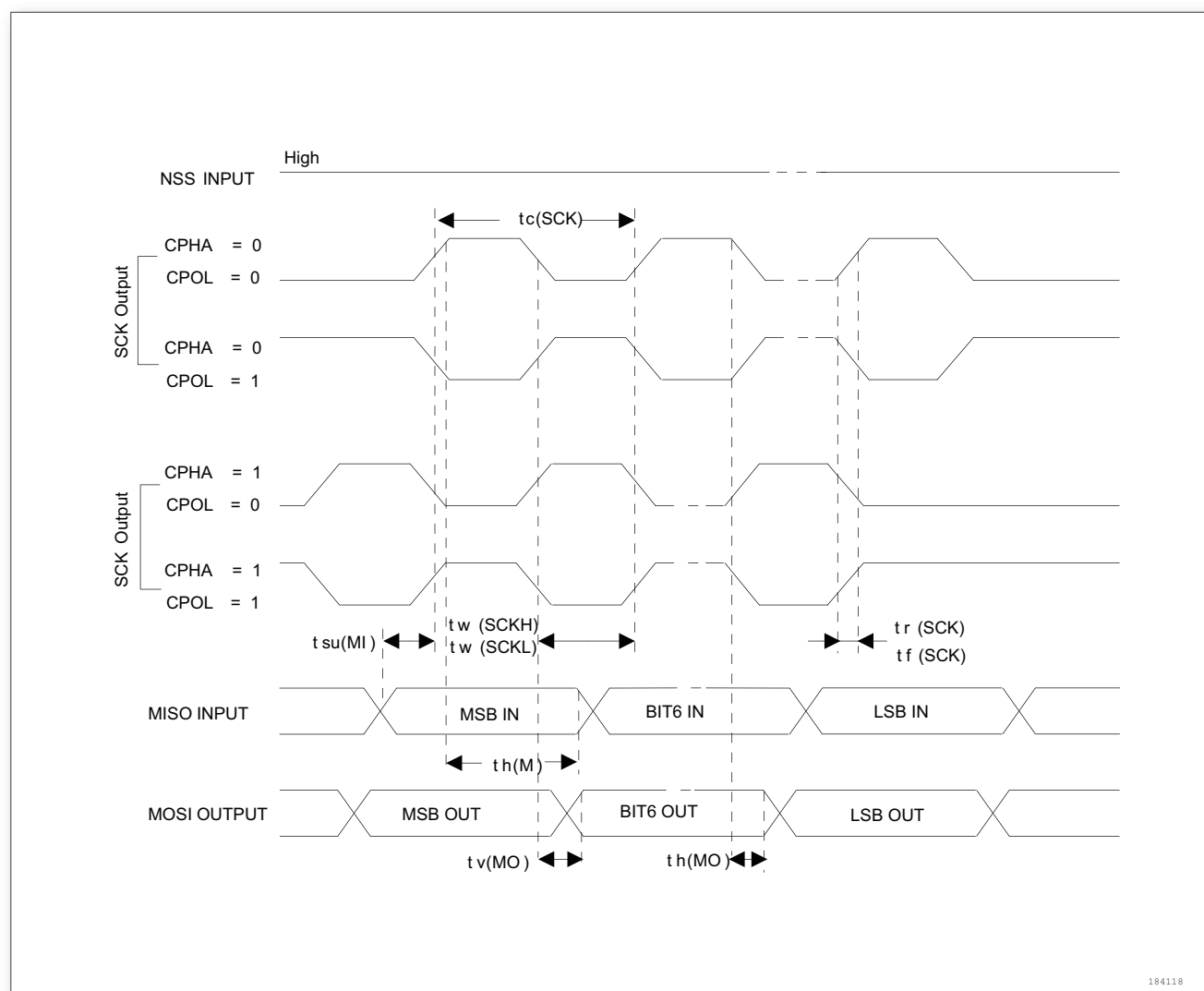


Figure 19. SPI timing diagram-slave mode and CPHA = 0

Figure 20. SPI timing diagram-slave mode and CPHA = 1⁽¹⁾

1. Measurement points are done at CMOS levels: 0.3V_{DD} and 0.7V_{DD}.

Figure 21. SPI timing diagram-master mode⁽¹⁾

1. Measurement points are done at CMOS levels: $0.3V_{DD}$ and $0.7V_{DD}$.

5.3.16 12-bit ADC characteristics

Unless otherwise specified, The parameters in the table below are measured using the ambient temperature, f_{PCLK2} frequency and V_{DDA} supply voltage in accordance with the conditions of Table 12.

Note: It is recommended to perform a calibration after each power-up

Table 39. ADC characteristics

Symbol	Parameter	Conditions	Min	Type	Max	Unit
V_{DDA}	Supply voltage		2.5	5	5.5	V
V_{REF+}	Positive reference voltage			V_{DDA}		V

Symbol	Parameter	Conditions	Min	Type	Max	Unit
$f_{ADC}^{(1)(3)}$	ADC clock frequency				15	MHz
$f_S^{(1)(3)}$	Sampling rate				1	MHz
$f_{TRIG}^{(1)}$	External trigger frequency	$f_{ADC} = 15\text{MHz}$			833	KHz
					18	$1/f_{ADC}$
$V_{AIN}^{(2)}$	Conversion voltage range		0 (V_{SSA} or V_{REF-} connected to ground)		V_{REF+}	V
$R_{AIN}^{(1)}$	External sample and hold capacitor		See Formulas 1 and Table 40			$k\Omega$
$R_{ADC}^{(1)}$	Sampling switch resistance				0.75	$k\Omega$
$C_{ADC}^{(1)}$	Internal sample and hold capacitor			10		pF
$t_S^{(1)}$	Sampling time	$f_{ADC} = 15\text{MHz}$	0.1		16	μs
			1.5		239.5	$1/f_{ADC}$
$t_{STAB}^{(1)}$	Stabilization time			1		μs
$t_{conv}^{(1)}$	Total conversion time (including Sampling time)	$f_{ADC} = 15\text{MHz}$	1		17.44	μs
			15 ~ 253 (sampling t_{S+}) stepwise approximation 13.5			$1/f_{ADC}$

1. Guaranteed by design. Not tested in production.
2. In this series of products, V_{REF+} is internally connected to DDA , V_{REF-} is internally connected to SSA .
3. f_{ADC} Maximum support 15MHz, f_S Maximum support 1MHz ($f_{PCLK2} = 60\text{MHz}$, ADC Prescaler = 4, $f_{ADC} = 15\text{MHz}$, $TS = 1.5$)

Formula 1: Maximum R_{AIN} Formula

$$R_{AIN} < \frac{T_S}{f_{ADC} \times C_{ADC} \times (N + 3) \times \ln(2)} - R_{ADC}$$

The above formula (Equation 1) is used to determine the maximum external impedance so that the error can be less than 1/4 LSB. Where $N = 12$ (representing 12-bit resolution) . $\ln(2) = 0.69314718$.

Table 40. Maximum R_{AIN} at $f_{ADC} = 15\text{MHz}^{(1)}$

T_S (cycles)	t_S (μs)	R_{AIN} max ($k\Omega$)
1.5	0.1	0.2
7.5	0.5	4.1
13.5	0.9	7.9
28.5	1.9	17.5
41.5	2.8	25.9

T_S (cycles)	t_S (μs)	R_{AIN} max (kΩ)
55.5	3.7	34.8
71.5	4.8	NA
239.5	16.0	NA

1. Guaranteed by design. Not tested in production.

Table 41. ADC Accuracy - Limit Test Conditions⁽¹⁾⁽²⁾

Symbol	Parameter	Test Conditions	Type	Max	Unit
ET	Comprehensive error	f _{PCLK2} = 60MHz, f _{ADC} = 15MHz, R _{AIN} < 10KΩ, V _{DDA} = 5V, T _A = 25°C	8	10	LSB
EO	Offset error		3	3	
EG	Gain error		1	1	
ED	Differential linearity error		6.5	7	
EL	Integral linearity error		8	8	

1. ADC Accuracy vs. Negative Injection Current: Injecting negative current on any of the standard (non-robust) analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to standard analog pins which may potentially inject negative current.

Any positive injection current within the limits specified for I_{INJ(PIN)} and ΣI_{INJ(PIN)} in subsubsec 5.3.13 does not affect the ADC accuracy.

2. Guaranteed based on test during characterization. Not tested in production.

ET = Total unadjusted error: The maximum deviation between the actual and ideal transmission curves.

EO = Offset error: The deviation between the first actual conversion and the first ideal conversion.

EG = Gain error: The deviation between the last ideal transition and the last actual transition.

ED = Differential linearity error: The maximum deviation between the actual step and the ideal value.

EL = Integral linearity error: The maximum deviation between any actual conversion and the associated line of the endpoint.

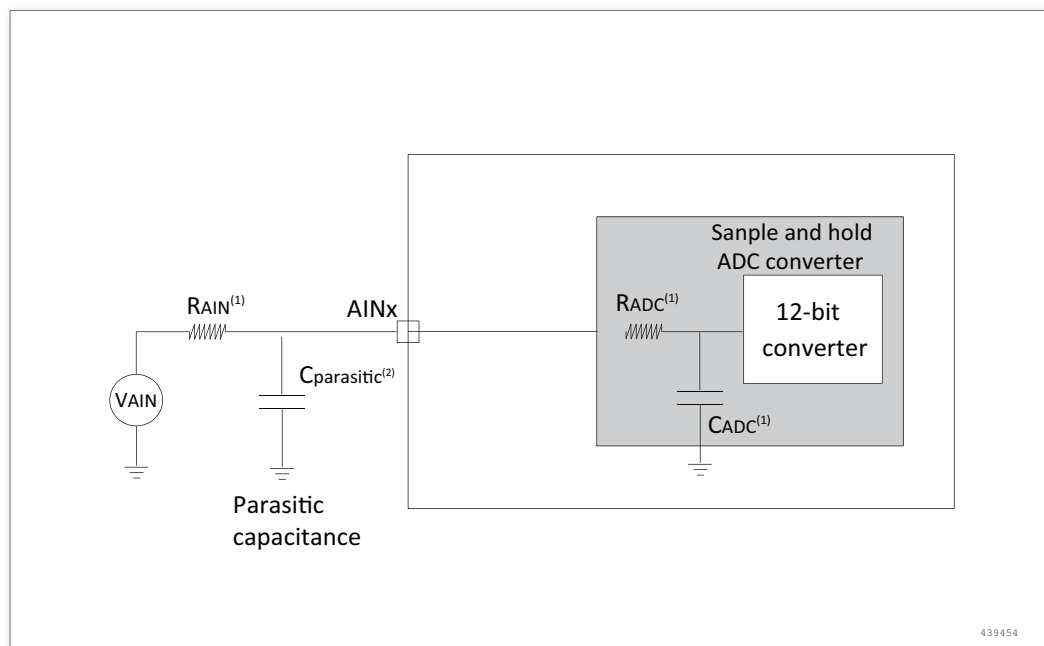


Figure 22. Typical connection diagram using the ADC

1. See Table 41 for the values of R_{AIN} , R_{ADC} and C_{ADC} .
2. $C_{parasitic}$ represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 7pF) . A high $C_{parasitic}$ value will downgrade conversion accuracy. To remedy this, f_{ADC} should be reduced.

PCB design recommendations

The power supply must be connected as shown below. The 10nF capacitor in the figure must be a ceramic capacitor (good quality) , and they should be as close as possible to the MCU chip.

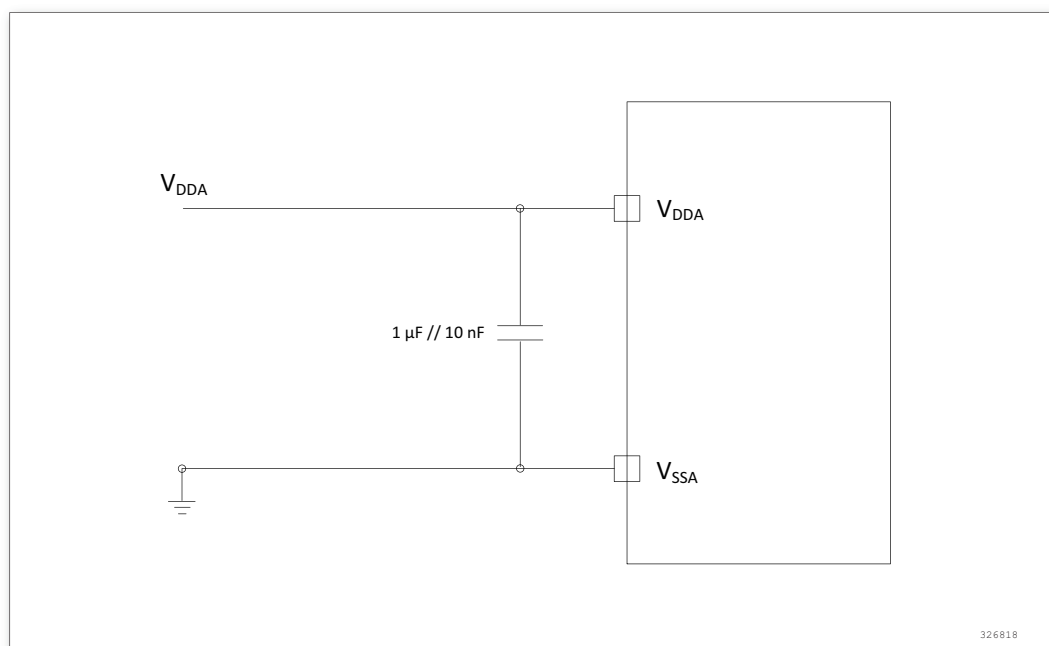


Figure 23. Power supply and reference power supply decoupling circuit

5.3.17 Temperature sensor characteristics

Table 42. Temperature sensor characteristics⁽³⁾ ⁽⁴⁾

Symbol	Parameter	Min	Type	Max	Unit
$T_L^{(1)}$	V_{SENSE} linearity with respect to temperature		± 5		$^{\circ}\text{C}$
Avg_Slope ⁽¹⁾	Average slope	4.571	4.801	5.984	mV/ $^{\circ}\text{C}$
$V_{25}^{(1)}$	Voltage at 25 $^{\circ}\text{C}$	1.433	1.451	1.467	V
$t_{start}^{(2)}$	Setup time			10	μs
$T_{S_temp}^{(2)}$	ADC sampling time when reading temperature	10			μs

1. Guaranteed based on test during characterization. Not tested in production.
2. Guaranteed by design. Not tested in production.
3. The shortest Sampling time can be determined by the application through multiple iterations.
4. $V_{DD} = 3.3\text{V}$.

5.3.18 Comparator characteristics

Table 43. Comparator characteristics

Symbol	Parameter	Register configuration	Min	Type	Max	Unit
HYST	Hysteresis	00		0		mV
HYST	Hysteresis	01		15		mV
HYST	Hysteresis	10		30		mV

Symbol	Parameter	Register configuration	Min	Type	Max	Unit
HYST	Hysteresis	11		90		mV
OFFSET	Offset voltage	00	0.091	0.213	0.358	mV
OFFSET	Offset voltage	01	3.23	7.51	12.08	mV
OFFSET	Offset voltage	10	9.79	15	20.8	mV
OFFSET	Offset voltage	11	34.25	47.4	62.22	mV
DELAY ⁽¹⁾	Propagation delay	00		80		nS
DELAY ⁽¹⁾	Propagation delay	01		51		nS
DELAY ⁽¹⁾	Propagation delay	10		26		nS
DELAY ⁽¹⁾	Propagation delay	11		9		nS
I _q ⁽²⁾	Operating current mean	00		4.5		uA
I _q ⁽²⁾	Operating current mean	01		4.4		uA
I _q ⁽²⁾	Operating current mean	10		4.4		uA
I _q ⁽²⁾	Operating current mean	11		4.4		uA

1. The output flips 50% of the time and the time difference between the input and the flip.
2. Total current consumption, operating current.

5.3.19 Gate driver characteristics

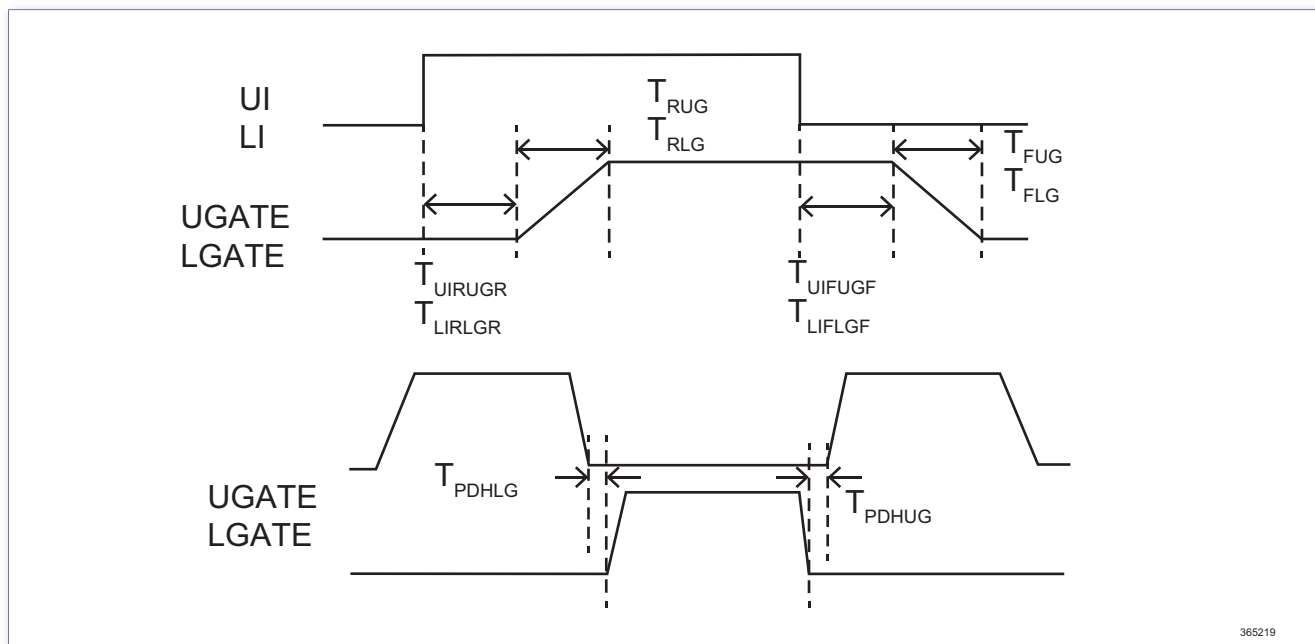


Figure 24. Power-on timing state diagram

Shoot-Through Protection

The GateDriver is equipped with a Shoot-Through protection circuit. The figure below shows that when HI and LI are turned on simultaneously, the output protection will cut off both the high-side and low-side switches. This is to prevent high high-side and low

low-side outputs from turning on at the same time.

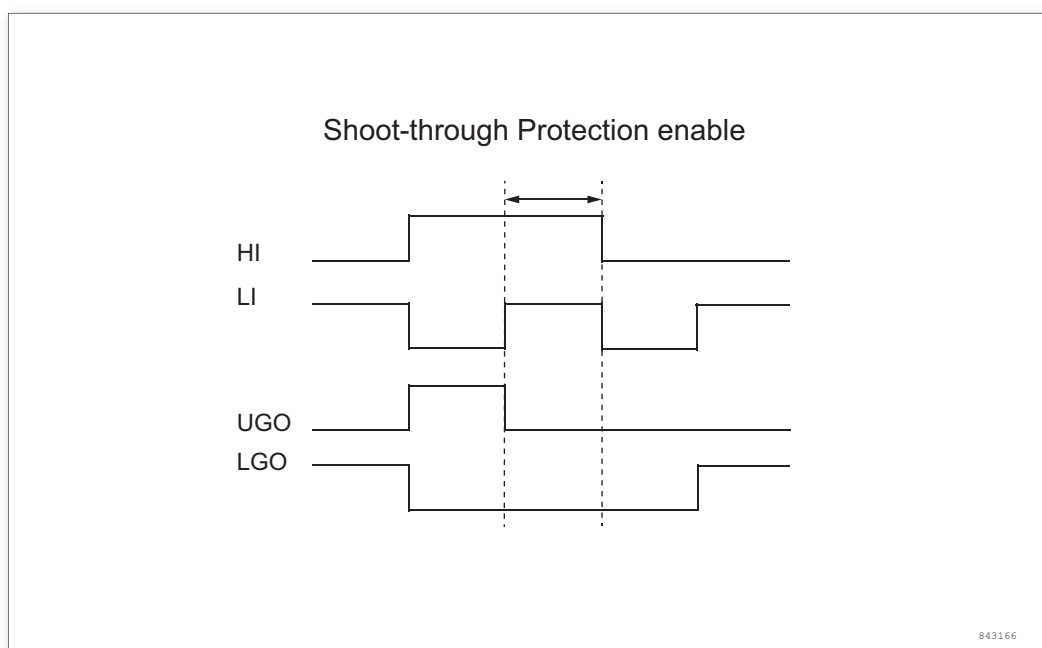


Figure 25. Shoot-Through Protection timing diagram

The table below indicates the pwm input/output status.

Table 44. PWM input and output status table

PWM State			
HI	LI	UGATE	LGATE
ON	OFF	ON	OFF
OFF	ON	OFF	ON
ON	ON	OFF	OFF
OFF	OFF	OFF	OFF

Undervoltage protection

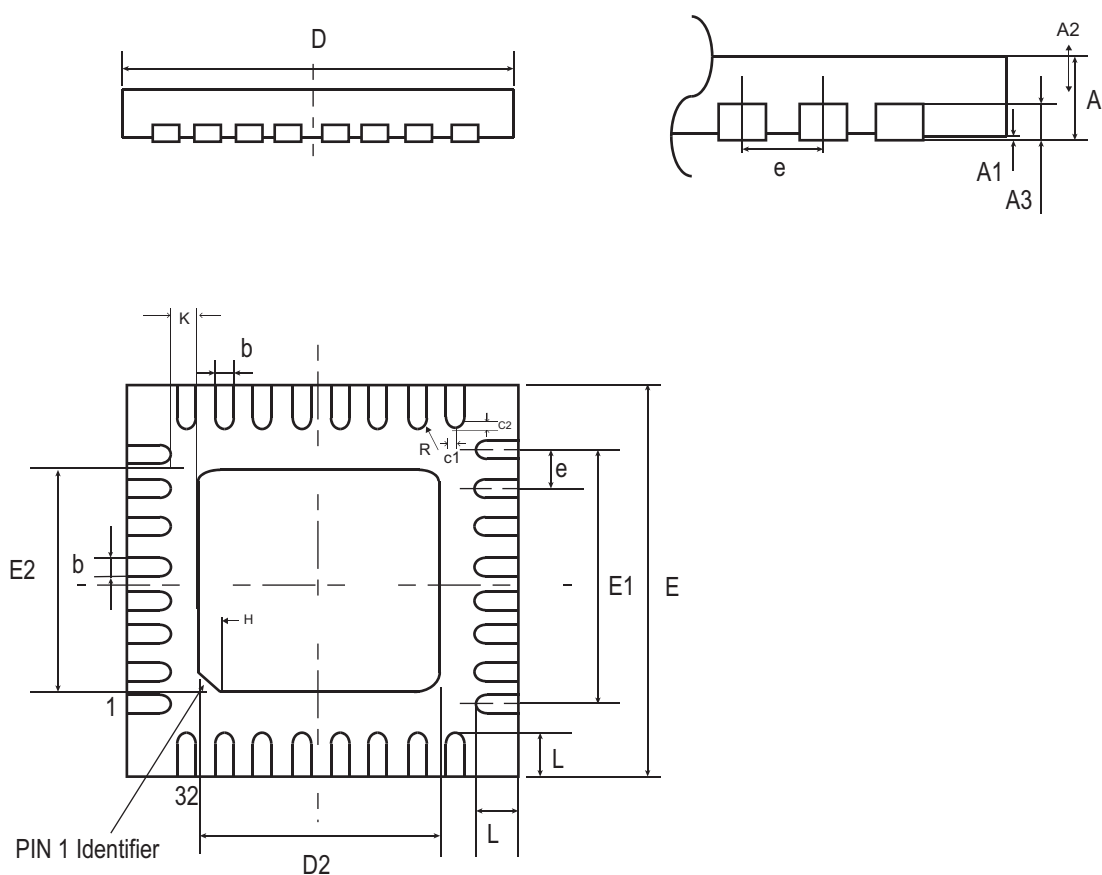
Undervoltage protection is triggered if the voltage $V_{CC} < 2.5V$. When the undervoltage protection function is activated, GateDriver turns off the LDO5V and the drive output.

6

Package information

Package information

6.1 QFN32 Package information



978941

Figure 26. QFN32 - 32-pin quad flat no-leads package outline

1. Drawing is not to scale.
2. Dimensions are expressed in millimeters.

Table 45. QFN32 mechanical data

Symbol	Millimeters		
	Min	Typ	Max
A	0.7	0.75	0.80
A1	0.00	0.02	0.05
A2	0.50	0.55	0.60
A3	0.20REF		
b	0.20	0.25	0.30
D	4.90	5.00	5.10
E	4.90	5.00	5.10
D2	3.40	3.50	3.60
E2	3.40	3.50	3.60
e		0.5	
H	0.30REF		
K	0.35REF		
L	0.35	0.40	0.45
R	0.09		
c1		0.08	
c2		0.08	
N	Number of pins = 32		

7

Ordering information

Ordering information

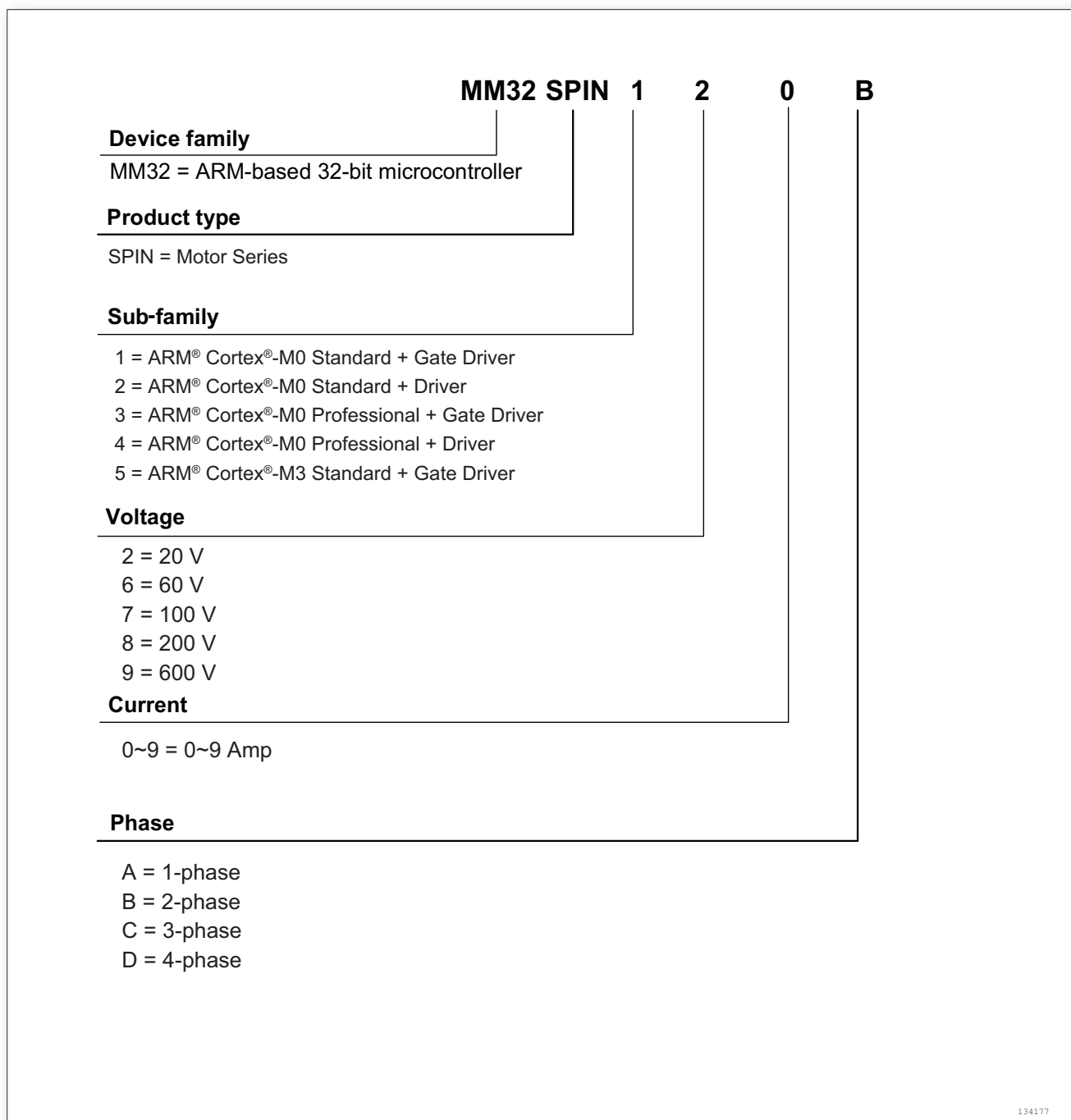


Figure 27. Ordering information scheme

8

Revision history

Revision history

Table 46. Document revision history

Revision	Changes	Date
Rev1.02	Modify the package parameters.	2019/3/11
Rev1.01	Modify ADC electrical parameters.	2019/1/7
Rev1.00	Initial release.	2018/12/6