## **Datasheet**

## MM32W0xxBnc

## 32-bit Micro controller based on ARM Cortex M0

Ver: 1.82\_n

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## Introduction

Introduction

#### 1.1 Description

This product is an ultra-low-power single-mode Bluetooth chip with a frequency of 2.4 GHz ISM band and a 2 MHz channel spacing, which complies with the Bluetooth specification. MM32W0xxBnc(named as "the device" throughout this document) is ARM® Cortex<sup>TM</sup>-M0 32-bit RISC core based micro controller family. The device has high speed embedded memory and the CPU, memory and AHB bus subsystem speed can attain up to 48MHz. The device also has integrated with extensive range of enhanced I/Os, two APB buses peripherals, 1 12-bit ADC, 2 Comparators, 1 general purpose 16-bit timer, 1 general purpose 32-bit timer, 1 PWM timer, 3 Basic timers, and standard communication interfaces device: 1 I2C, 1 SPI, 1 USB, 1 CAN, and 2 UARTs.

The device operates from a 2.3V to 3.6V power supply. They are available in both the -40°C to +85°C temperature range. A comprehensive set of power-saving mode allows the design of low-power applications.

The devices are available in 2 different packages: LQFP48 and QFN32. Depending on the device chosen, different sets of peripherals are included.

The abundant peripheral configurations enable the device to fit wide range of applications in difference industries, Few examples are as follows:

- Beacon
- · Wireless keyboard, mouse
- · Industrial applications: industrial remote control, telemetry
- Alarm system, access control system, data acquisition and transmission system

#### 1.2 Product Features

- Core and system
  - ARM® Cortex<sup>TM</sup>-M0 CPU
  - Maximum operating frequency is up to 48MHz
  - Single instruction cycle 32-bit hardware multiplier
- Memories
  - 128K Bytes of Flash memory
  - 8K Bytes of SRAM
  - Boot loader support Chip Flash and ISP (In-System Programming)

- · Single mode BLE RF transceiver
  - Packet processing engine
  - GFSK coding method
  - Internal voltage regulator guarantees PSRR
  - Programmable transmit power range: -28dBm to +4dBm
  - 1Mbps air data transmission
  - Excellent RF link budget: up to -80dBm
- · Clock, reset and power management
  - 2.3V to 3.6V application supply
  - Power-on/Power-down reset (POR/PDR), Programmable voltage detector (PVD)
  - External 16MHz high speed crystal oscillator
  - Embedded factory-tuned 48MHz high speed oscillator
  - Embedded 40KHz low speed oscillator
  - PLL supports CPU running at 48MHz
- · Low-power
  - Sleep, Stop and Standby modes
- 1 12-bit ADC, 1µS A/D converters (up to 10 channels)
  - Conversion range: 0 to V<sub>DDA</sub>
- · 2 Comparators
- · 5 DMA controller
  - Supported peripherals: Timer, UART, I2C, SPI, ADC and USB
- Up to 28 fast I/Os:
  - All mappable on 16 external interrupt vectors
  - Partial port can work on 5V
- · Debug mode
  - Serial wire debug (SWD)
- Up to 9 timers
  - 1 16-bit timer and 1 32-bit timer, with up to 4 IC/OC, usable for IR control decoding
  - 2 16-bit timer, with 1 IC/OC, 1 OCN, deadtime generation and emergency stop and modulator gate for IR control
  - 1 16-bit timer, with 1 IC/OC
  - 2 watchdog timers (independent and window type)
  - SysTick timer: 24-bit downcounter
- · Up to 6 Communication interfaces
  - 2 UARTs
  - 1 I2C
  - 1 SPI
  - 1 CAN
  - 1 USB
- Low cost peripheral component BOM cost
- 96-bit unique ID (UID)
- Packages LQFP48 and QFN32

For more information about the complete product, refer to Section 2.2 of the data sheet. The relevant information about the Cortex<sup>TM</sup>-M0, please refer to Cortex<sup>TM</sup>-M0 technical reference manual.

# **Specification**

Specification

#### 2.1 Device contrast

Table 1. Device features and peripheral counts

	Device	MM32W051/062/073PFB	MM32W051/062/073NTE	
Peripheral		WIWI32VVU31/002/0/3PFB	WIWI32 VVU3 1/U02/U7 3IN 1 E	
Flash memory -K Bytes		32/64/128	32/64/128	
SRAM	-K Bytes	4/8/8	4/8/8	
	General purpose			
Timers	(16 bit)	4	4	
	General purpose	1	1	
	(32 bit)	ı	1	
	Advanced control	1	1	
	UART	2	2	
Common	I2C	1	1	
interfaces	SPI	1	1	
	USB	0/1/1	0/1/1	
	CAN	0/1/1	0/1/1	
G	PIOs	28	22	
12-b	oit ADC	1	1	
(number	of channels)	10 channels	7 channels	
Com	parators	:	2	
Max CPI	J frequency	481	MHz	
Į.	\ES	YI	ES	
Operati	ng voltage	2.3V ~ 3.6V		
Pac	kages	LQFP48	QFP32	

#### 2.2 Summary

#### 2.2.1 ARM® Cortex<sup>TM</sup>-M0 and SRAM

The ARM® Cortex<sup>TM</sup>-M0 is a generation of ARM processors for embedded systems. It has been developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while delivering outstanding

computational performance and an advanced system response to interrupts.

The ARM® Cortex<sup>TM</sup>-M0 processors feature exceptional code-efficiency, delivering the high performance expected from an ARM core, with memory sizes usually associated with 8- and 16-bit devices.

The devices have embedded ARM core and are compatible with all ARM tools and software.

#### 2.2.2 Memory

128K Bytes of embedded Flash memory.

#### 2.2.3 SRAM

8K Bytes of embedded SRAM.

#### 2.2.4 Clocks and startup

When the system is powered up, the default clock is from PLL with the resource from HSE 48 MHz oscillator. An external 2  $\sim$  24 MHz clock can also be configured to monitor the system during power up phases.

Several prescalers allow the application to configure the frequency of the AHB and the APB domains. The maximum frequency of the AHB and the APB domains is 48MHzz.Refer to figure 3 for the clock drive block diagram.

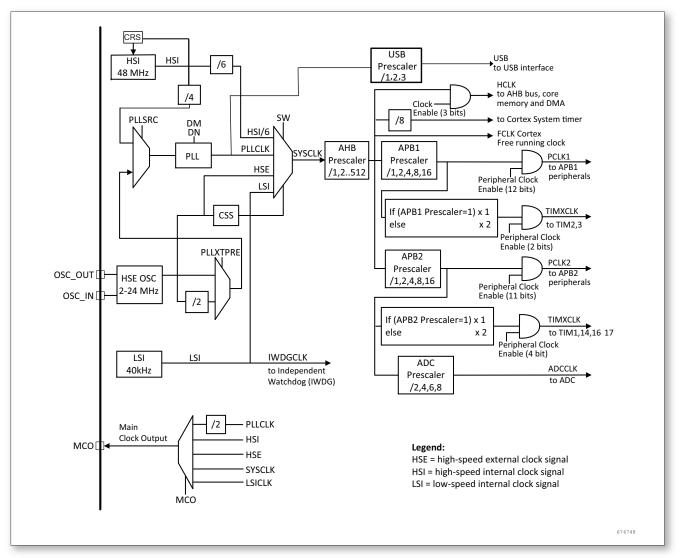


Figure 1. Clock tree

#### 2.2.5 Nested vectored interrupt controller (NVIC)

The device embeds a nested vectored interrupt controller and is able to handle up to 68 maskable interrupt channels (not including the 16 interrupt lines of Cortex<sup>™</sup>-M0) with 16 priority levels.

- Closely coupled NVIC gives low latency interrupt processing
- · Interrupt entry vector table address passed directly to the core
- · Closely coupled NVIC core interface
- · Allows early processing of interrupts
- · Processing of late arriving higher priority interrupts
- · Support for tail-chaining
- · Processor state automatically saved
- · Interrupt entry restored on interrupt exit with no instruction overhead

This hardware block provides flexible interrupt management features with minimal inter-

rupt latency.

#### 2.2.6 Extended interrupt/event controller (EXTI)

The extended interrupt/event controller consists of many edge detector lines are used to generate interrupt/event requests for waking up the system. Each line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the internal APB2 clock period. All GPIOs can be connected to the 16 external interrupt lines.

#### 2.2.7 Boot modes

At startup, the boot pin and boot selector option bit are used to select one of the three boot options:

- · Boot from User Flash memory
- · Boot from System Memory
- Boot from embedded SRAM

The boot loader is located in System Memory. It is used to reprogram the Flash memory by using UART1.

#### 2.2.8 Power supply schemes

- $V_{DD}$  = 2.3V ~ 3.6V: external power supply for I/Os and the internal regulator. Provided externally through  $V_{DD}$  pins.
- $V_{SSA}$ ,  $V_{DDA}$  = 2.3V  $\sim$  3.6V: external analog power supply for reset blocks, oscillators and PLL.  $V_{DDA}$  and  $V_{SSA}$  must be connected to  $V_{DD}$  and  $V_{SS}$ .

#### 2.2.9 Power supply supervisors

The device has integrated power-on reset (POR) and power-down reset (PDR) circuits. They are always active, and ensure proper operation above a threshold of 1.8V. The device remains in reset mode when the monitored supply voltage is below a specified threshold  $V_{POR/PDR}$ , without the need for an external reset circuit.

The device features an embedded programmable voltage detector (PVD) that monitors the  $V_{DD}/V_{DDA}$  power supply and compares it to the  $V_{PVD}$  threshold. An interrupt can be generated when VDD drops below the  $V_{PVD}$  threshold and/or when VDD is higher than the  $V_{PVD}$  threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

#### 2.2.10 Voltage regulator

The voltage regulator converts the external voltage to the internal digital logic and it is always enabled after reset.

#### 2.2.11 Low-power modes

The device support three low-power modes to achieve the best compromise between low power consumption, short startup time and available wakeup sources.

#### Sleep mode

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

#### Stop mode

Stop mode achieves very low power consumption while retaining the content of SRAM and registers. the HSI and the HSE crystal oscillators are disabled. The voltage regulator can also be put either in normal or in low power mode.

#### Standby mode

Standby mode achieves the lowest power consumption of the system. This mode turns off the voltage regulator in CPU deep sleep mode. The entire 1.5V power supply area is powered down. PLL HSI and HSE oscillators are also powered down. SRAM and register contents are missing.

#### 2.2.12 Direct memory access controller (DMA)

The 5-channel general-purpose DMAs manage memory-to-memory, peripheral-to-memory and memory-to-peripheral transfers. The DMA supports circular buffer management, removing the need for user code intervention when the controller reaches the end of the buffer.

Each channel is connected to dedicated hardware DMA requests, with support for software trigger on each channel. Configuration is made by software and transfer sizes between source and destination are independent.

DMA can be used with the main peripherals: UART、I2C、SPI、USB、CAN、ADC general-purpose and advanced-control timers TIMx.

#### 2.2.13 Timers and watchdogs

Medium capacity device include 1 advanced control 5 general-purpose timers 2 watchdog timers and 1 SysTick timer.

The following table compares the features of the different timers:

Table 2. Timer feature comparison

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/- compare channels	Complem -entary outputs
Advanced control	TIM1	16-bit	Up, down, up/down	integer from 1 to 65536	Yes	4	Yes
General	TIM2	32-bit	Up, down, up/down	integer from 1 to $2^{32} - 1$	Yes	4	No
purpose	TIM3	16-bit	Up, down, up/down	integer from 1 to 65536	Yes	4	No
basic	TIM14	16-bit	Up	integer from 1 to 65536	Yes	1	No
	TIM16 / TIM17	16-bit	Up	integer from 1 to 65536	Yes	1	Yes

#### Advanced-control timer (TIM1)

The advanced-control timer can be seen as a three-phase PWM multiplexed on six channels. It has complementary PWM outputs with programmable inserted dead times. It can also be seen as a complete general-purpose timer. The four independent channels can be used for:

- · Input capture
- · Output compare
- PWM generation (edge or center-aligned modes)
- One-pulse mode output

If configured as a standard 16-bit timer, it has the same features as the TIMx timer. If configured as the 16-bit PWM generator, it has full modulation capability (0  $\sim$  100%).

In debug mode, the counter can be frozen and the PWM output is disabled to cut off the switches controlled by these outputs.

Many features are shared with those of the standard timers which have the same architecture. The advanced control timer can therefore work together with the other timers via the Timer Link feature for synchronization or event chaining.

#### **General-purpose timers (TIMx)**

There are 5 synchronizable general-purpose timers (TIM2, TIM3).

#### **General-purpose timers 32-bit**

The timer is based on a 32-bit auto-reload up/downcounter and a 16-bit prescaler. The feature is 4 independent channels each for input capture/output compare, PWM or one-pulse mode output.

#### **General-purpose timers 16-bit**

#### TIM3

'The timer is based on a 16-bit auto-reload up/downcounter and a 16-bit prescaler. The feature is 4 independent channels each for input capture/output compare, PWM or one-pulse mode output.

The timer can work together or with the TIM1 advanced-control timer via the Timer Link feature for synchronization or event chaining. Their counter can be frozen in debug mode. Any of the general-purpose timers can be used to generate PWM outputs. They all have independent DMA request generation.

These timers are capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 3 hall-effect sensors.

#### **TIM14**

This timer is based on a 16-bit auto-reload upcounter and a 16-bit prescaler. TIM14 features one single channel for input capture/output compare, PWM or one-pulse mode output. Their counter can be frozen in debug mode.

#### **TIM16/TIM17**

Every timer is based on a 16-bit auto-reload upcounter and a 16-bit prescaler. They each have a single channel for input capture/output compare, PWM or one-pulse mode output. TIM16 and TIM17 have a complementary output with dead-time generation and independent DMA request generation. Their counters can be frozen in debug mode.

#### Independent watchdog (IWDG)

The independent watchdog is based on an 8-bit prescaler and 12-bit downcounter with user-defined refresh window. It is clocked from an independent 40 KHz internal oscillator and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free running timer for application timeout management. It is hardware or software configurable through the option bytes. The counter can be frozen in debug mode.

#### System window watchdog (WWDG)

The system window watchdog is based on a 7-bit downcounter that can be set as free running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the APB clock (PCLK). It has an early warning interrupt capability and the counter can be frozen in debug mode.

#### SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard down counter. It features:

- · A 24-bit down counter
- · Autoreload capability
- · Maskable system interrupt generation when the counter reaches 0
- · Programmable clock source

#### 2.2.14 Universal asynchronous receiver/transmitter (UART)

UART provides hardware management of the CTS, RTS.

Support LIN master-slave function.

All UART interface can be served by the DMA controller.

#### 2.2.15 I2C interface

The I2C interface can operate in multimaster or slave modes. It can support Standard mode, and Fast Mode.

It supports 7-bit and 10-bit addressing modes, multiple 7-bit slave addresses (two addresses, one with configurable mask).

#### 2.2.16 Serial peripheral interface (SPI)

The SPI interface, in slave or master mode, can be configured to 1  $\sim$  32 bits per frame.

All SPI interface can be served by the DMA controller.

#### 2.2.17 Universal serial bus (USB)

The microcontroller embeds a USB device peripheral compatible with the USB full-speed 12 Mbs. The USB interface implements a full-speed (12 Mbit/s) function interface. The dedicated 48 MHz clock is generated from the internal main PLL (the clock source must use a HSE crystal oscillator).

#### 2.2.18 Controller area network (CAN)

The CAN is compliant with specifications 2.0 A and B (active) with a bit rate up to 1 Mbit/s. It can receive and transmit standard frames with 11-bit identifiers as well as extended frames with 29-bit identifiers.

#### 2.2.19 General-purpose inputs/outputs (GPIO)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of

the GPIO pins are shared with digital or analog alternate functions. The I/O configuration can be locked if needed following a specific sequence in order to avoid spurious writing to the I/Os registers.

#### 2.2.20 Analog-to-digital converter (ADC)

The one 12-bit analog-to-digital converters is embedded into microcontrollers and the ADC shares up to 10 external channels, performing conversions in single-shot or scan modes. In scan mode, automatic conversion is performed on a selected group of analog inputs. The ADC can be served by the DMA controller.

The analog watchdog function allows very precise monitoring of all the way, multiple or all selected channels, and an interruption occurs when the monitored signal exceeds the preset threshold. The events generated by the general-purpose timers (TIMx) and the advanced-control timer (TIM1) can be internally connected to the ADC start trigger to allow the application to synchronize A/D conversion and timers.

#### 2.2.21 Serial single line SWD debug port (SW-DP)

Built-in ARM two-wire serial debug port (SW-DP).

An ARM SW-DP interface is provided to allow a serial wire debugging tool to be connected to the MCU.

#### 2.2.22 Comparator (COMP)

The devices embed 2 general purpose comparators. that can be used either as standalone devices (all terminal are available on I/Os) or combined with the timers. The comparators can be used for a variety of functions including:

- Wake-up from low-power mode triggered by an analog signal,
- · Analog signal conditioning,
- Cycle-by-cycle current control loop when combined with the PWM output from a timer.
- · Rail-to-rail comparators
- Each comparator has positive and configurable negative inputs used for flexible voltage
- · Selection:
  - Reusable I/O pins
  - Internal comparison voltage CRV selects the voltage divider value of AVDD or internal reference voltage
- · Programmable hysteresis
- Programmable speed/consumption
- The outputs can be redirected to an I/O or to timer inputs for triggering:
  - Capture events
  - OCref\_clr events (for cycle-by-cycle current control)
  - Break events for fast PWM shutdowns

#### 2.2.23 Bluetooth Low Energy Broadcast

The chip integrates the Bluetooth specification and RF transceiver and is compatible with the 2.4GHz ISM band defined by the International Telecommunications Union Radiocommunication Bureau.

After the chip is powered, the RF transceiver can be wirelessly transceived by simply building a simple peripheral component. It offers an excellent RF link budget of up to -80dBm with a shutdown current of less than 250uA for the entire chip.

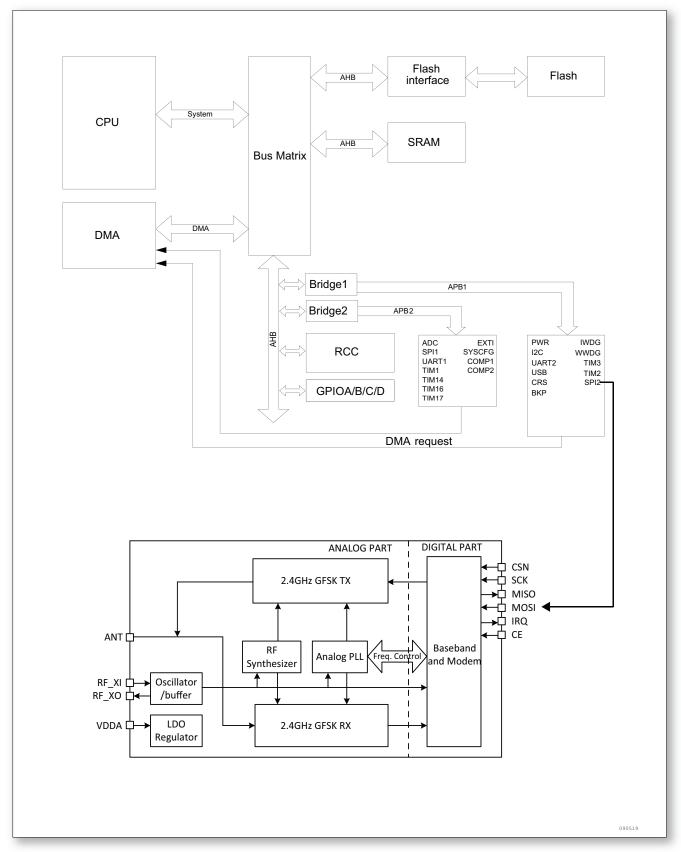


Figure 2. Block diagram

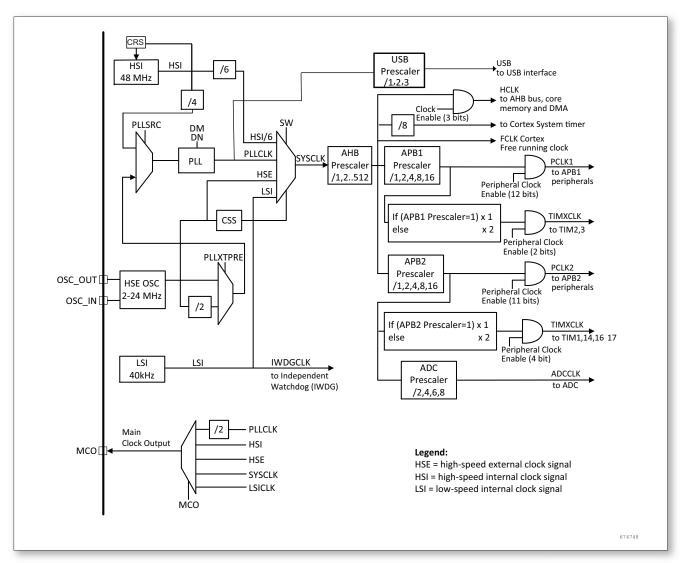


Figure 3. Clock tree

# Pin definition

Pin definition

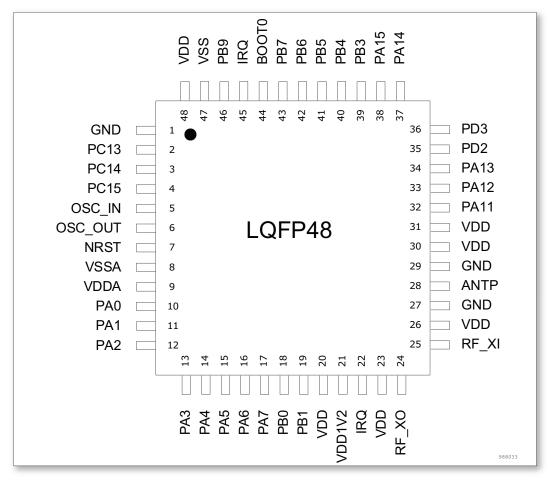


Figure 4. LQFP48 packet pinout

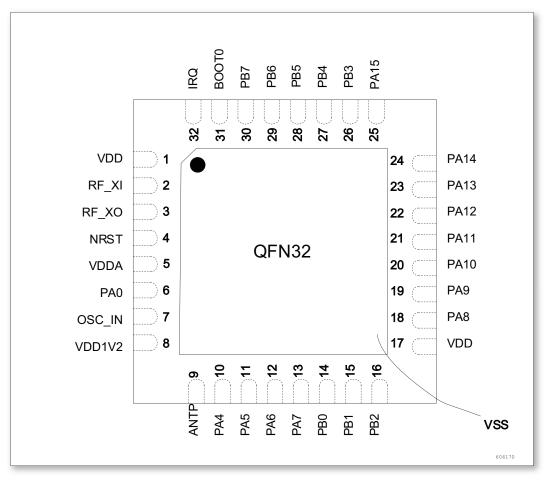


Figure 5. QFN32 packet pinout

Table 3. LQFP48 Pin definitions

Pin number	Pin name	<b>Type</b> (1)	I/O	Main	Alternate	Additional
LQFP48	Fill Hallie	Type	structure <sup>(2)</sup>	function	functions	functions
1	GND	S	-	NC	-	-
2	PC13	I/O	-	PC13	-	-
3	PC14	I/O	-	PC14	-	-
4	PC15	I/O	-	PC15	-	-
5	OSC_IN	I	-	OSC_IN	-	OSC_IN
6	OSC_OUT	0	-	OSC_OUT	-	OSC_OUT
7	NRST	I/O	-	NRST	-	-
8	VSSA	S	-	VSSA	-	-
9	VDDA	S	-	VDDA	-	-
10	PA0- WKUP	I/O	-	PA0	TIM2_CH1_ ETR/ UART2_CTS/ ADC_VIN[0]	WKUP/ COMP1_OUT
11	PA1	I/O	-	PA1	TIM2_CH2/ UART2_RTS/ ADC_VIN[1]	-

Time	Pin number	Pin name	<b>Type</b> <sup>(1)</sup>	I/O	Main	Alternate	Additional
12	LQFP48	Fill lialite	Type	structure <sup>(2)</sup>	function	functions	functions
ADC_VIN[2]   TIM2_CH4/   TIM2_CH4/   ADC_VIN[3]						TIM2_CH3/	
13	12	PA2	I/O	-	PA2	UART2_TX/	COMP2_OUT
13						ADC_VIN[2]	
ADC_VIN[3]						TIM2_CH4/	
14	13	PA3	I/O	-	PA3	UART2_RX/	-
14						ADC_VIN[3]	
ADC_VIN[4]   SPI1_SCK/   SPI1_SCK/   TIM2_CH1_ETR/   - ADC_VIN[5]   SPI1_MISO/   TIM3_CH1/   COMP1_OUT						SPI1_NSS/	
SPI1_SCK/   TIM2_CH1_ETR/   ADC_VIN[5]   SPI1_MISO/   TIM3_CH1/   COMP1_OUT	14	PA4	I/O	-	PA4	TIM14_CH1/	-
15						ADC_VIN[4]	
ADC_VIN[6]   SPI1_MISO/ TIM3_CH1/ TIM1_EKIN/ ADC_VIN[6]   SPI1_MOSI/ TIM3_CH2/ TIM1_CH1N/ TIM1_CH1N/ TIM1_CH1N/ TIM1_CH1/ TIM1_CH1/ ADC_VIN[6]   COMP2_OUT						SPI1_SCK/	
16	15	PA5	I/O	-	PA5	TIM2_CH1_ETR/	-
TIM3_CH1/ TIM1_BKIN/ ADC_VIN[6]   SPI1_MOSI/ TIM3_CH2/ TIM1_CH1N/ TIM1_CH1N/ TIM1_CH1N/ TIM1_CH1N/ TIM1_CH1/ ADC_VIN[7]   COMP2_OUT						ADC_VIN[5]	
16 PA6 I/O - PA6 TIM16_CH1/ TIM1_BKIN/ ADC_VIN[6]  17 PA7 I/O - PA7 TIM1_CH1N/ TIM1_CH1N/ TIM1_CH1N/ TIM1_CH1N/ TIM1_CH1/ ADC_VIN[7]  18 PB0 I/O - PB0 TIM1_CH2N/ - ADC_VIN[8]  19 PB1 I/O - PB1 TIM1_CH3N/ ADC_VIN[9]  20 VDD S - VDD TIM1_CH3N/ ADC_VIN[9]  20 VDD S - RF_VDDD12						SPI1_MISO/	
TIM1_BKIN/ ADC_VIN[6]  SPI1_MOSI/ TIM3_CH2/ TIM1_CH1N/ TIM14_CH1/ TIM17_CH1/ ADC_VIN[7]  18 PB0 I/O - PB0 TIM1_CH2N/ - ADC_VIN[8]  TIM3_CH3/ TIM1_CH2N/ - ADC_VIN[8]  TIM1_CH3N/ ADC_VIN[8]  TIM1_CH3N/ ADC_VIN[9]  20 VDD S - VDD 21 VDD1V2 S - RF_VDDD12 22 IRQ I/O - RF_IRQ 23 VDD S - RF_VDD 24 RF_XO O - RF_XO 25 RF_XI I - RF_XI						TIM3_CH1/	
ADC_VIN[6]   SPI1_MOSI/ TIM3_CH2/ TIM1_CH1N/ TIM14_CH1/ TIM1_CH1/ ADC_VIN[7]	16	PA6	I/O	-	PA6	TIM16_CH1/	COMP1_OUT
17						TIM1_BKIN/	
17						ADC_VIN[6]	
17						SPI1_MOSI/	
17 PA7 I/O - PA7 TIM14_CH1/ TIM17_CH1/ ADC_VIN[7]  18 PB0 I/O - PB0 TIM1_CH2N/ ADC_VIN[8]  19 PB1 I/O - PB1 TIM14_CH1/ TIM3_CH3/ TIM1_CH2N/ ADC_VIN[8]  20 VDD S - VDD 21 VDD1V2 S - RF_VDVDD12 22 IRQ I/O - RF_IRQ 23 VDD S - RF_VDDD 24 RF_XO O - RF_XO 25 RF_XI I - RF_XI						TIM3_CH2/	
TIM14_CH1/ TIM17_CH1/ ADC_VIN[7]  18 PB0 I/O - PB0 TIM1_CH2N/ - ADC_VIN[8]  19 PB1 I/O - PB1 TIM3_CH4/ TIM1_CH3N/ ADC_VIN[9]  20 VDD S - VDD 21 VDD1V2 S - RF_VDVDD1.2 22 IRQ I/O - RF_IRQ 23 VDD S - RF_VVDD 24 RF_XO O - RF_XO 25 RF_XI I - RF_XI	47	PA7	1/0	-	PA7	TIM1_CH1N/	COMP2_OUT
ADC_VIN[7]   TIM3_CH3/   - ADC_VIN[8]     TIM1_CH2N/   - ADC_VIN[8]     TIM1_CH2N/   - ADC_VIN[8]     TIM1_CH3N/   TIM1_CH3N/   ADC_VIN[9]	17		PA7   1/O			TIM14_CH1/	
TIM3_CH3/   TIM1_CH2N/   -   ADC_VIN[8]     TIM1_CH2N/   -   ADC_VIN[8]     TIM1_CH1/   TIM3_CH4/   TIM1_CH3N/   ADC_VIN[9]						TIM17_CH1/	
18 PB0 I/O - PB0 TIM1_CH2N/ - ADC_VIN[8]  19 PB1 I/O - PB1 TIM1_CH3N/ TIM3_CH4/ TIM1_CH3N/ ADC_VIN[9]  20 VDD S - VDD 21 VDD1V2 S - RF_VDVDD1.2 22 IRQ I/O - RF_IRQ 23 VDD S - RF_VDDD 24 RF_XO O - RF_XO 25 RF_XI I - RF_XI						ADC_VIN[7]	
ADC_VIN[8]  TIM14_CH1/ TIM3_CH4/ TIM1_CH3N/ ADC_VIN[9]  20 VDD S - VDD  21 VDD1V2 S - RF_VDVDD1.2  22 IRQ I/O - RF_IRQ  23 VDD S - RF_VVDD  24 RF_XO O - RF_XO  25 RF_XI I - RF_XI						TIM3_CH3/	
PB1 I/O - PB1 TIM14_CH1/ TIM3_CH4/ TIM1_CH3N/ ADC_VIN[9]  20 VDD S - VDD  21 VDD1V2 S - RF_V <sub>DVDD1.2</sub> 22 IRQ I/O - RF_IRQ  23 VDD S - RF_V <sub>DVDD</sub> 24 RF_XO O - RF_XO  25 RF_XI I - RF_XI	18	PB0	I/O	-	PB0	TIM1_CH2N/	-
19     PB1     I/O     -     PB1     TIM3_CH4/ TIM1_CH3N/ ADC_VIN[9]     -       20     VDD     S     -     VDD     -     -       21     VDD1V2     S     -     RF_V <sub>DVDD1.2</sub> -     -       22     IRQ     I/O     -     RF_IRQ     -     -       23     VDD     S     -     RF_V <sub>DD</sub> -     -       24     RF_XO     O     -     RF_XO     -     -       25     RF_XI     I     -     RF_XI     -     -						ADC_VIN[8]	
19 PB1 I/O - PB1 TIM1_CH3N/ ADC_VIN[9]  20 VDD S - VDD						TIM14_CH1/	
TIM1_CH3N/   ADC_VIN[9]	10	DD1	1/0		DD1	TIM3_CH4/	
20         VDD         S         -         VDD         -         -           21         VDD1V2         S         -         RF_V <sub>DVDD1.2</sub> -         -           22         IRQ         I/O         -         RF_IRQ         -         -           23         VDD         S         -         RF_V <sub>DD</sub> -         -           24         RF_XO         O         -         RF_XO         -         -           25         RF_XI         I         -         RF_XI         -         -	19	PDI	1/0	-	PDI	TIM1_CH3N/	-
21         VDD1V2         S         -         RF_VDVDD1.2         -         -         -           22         IRQ         I/O         -         RF_IRQ         -         -           23         VDD         S         -         RF_VDDD         -         -           24         RF_XO         O         -         RF_XO         -         -           25         RF_XI         I         -         RF_XI         -         -						ADC_VIN[9]	
22         IRQ         I/O         -         RF_IRQ         -         -           23         VDD         S         -         RF_V <sub>VDD</sub> -         -           24         RF_XO         O         -         RF_XO         -         -           25         RF_XI         I         -         RF_XI         -         -	20	VDD	S	-	VDD	-	-
23 VDD S - RF_V <sub>VDD</sub>	21	VDD1V2	S	-	RF_V <sub>DVDD1.2</sub>	-	-
24         RF_XO         O         -         RF_XO         -         -           25         RF_XI         I         -         RF_XI         -         -	22	IRQ	I/O	-	RF_IRQ	-	-
25 RF_XI I - RF_XI	23	VDD	S	-	RF_V <sub>VDD</sub>	-	-
	24	RF_XO	0	-	RF_XO	-	-
26 VDD S - RF_V <sub>VDD</sub>	25	RF_XI	ı	-	RF_XI	-	-
	26	VDD	S	-	RF_V <sub>VDD</sub>	-	-

Pin number	Pin name	<b>Type</b> <sup>(1)</sup>	I/O	Main	Alternate	Additional
LQFP48	structure <sup>(2)</sup> function		function	functions	functions	
27	GND	S	-	-	-	-
28	ANTP	-	-	ANTP	-	-
29	GND	S	-	-	-	-
30	VDD	S	-	RF_V <sub>VDD</sub>	-	-
31	VDD	S	-	RF_V <sub>VDD</sub>	-	-
					UART1_CTS/	
					TIM1_CH4/	
32	PA11	I/O	FT	PA11	CAN_RX/	COMP1_OUT
					I2C1_SCL/	
					USBDM	
					UART1_RTS/	
					TIM1_ETR/	
33	PA12	I/O	FT	PA12	CAN_TX/	COMP2_OUT
					I2C1_SDA/	
					USBDP	
34	PA13	I/O	FT	PA13	SWDIO	-
35	PD2	I/O	FT	PD2	-	-
36	PD3	I/O	FT	PD3	-	-
					SWDCLK/	
37	PA14	I/O	FT	PA14	UART2_TX	-
					SPI1_NSS/	
38	PA15	5 I/O	FT	PA15	UART2_RX/	-
					TIM2_CH1_ETR	
	550				SPI1_SCK/	
39	PB3	I/O	FT	PB3	TIM2_CH2	-
	55.4			55.4	SPI1_MISO/	
40	PB4	I/O	FT	PB4	TIM3_CH1	-
					SPI1_MOSI/	
41	PB5	I/O	FT	PB5	TIM3_CH2/	-
					TIM16_BKIN	
					UART1_TX/	
42	PB6	I/O	FT	PB6	I2C1_SCL/	_
					TIM16_CH1N	
					UART1_RX/	
43	PB7	I/O	FT	PB7	I2C1_SDA/	-
					TIM17_CH1N	
44	воото	ı	FT	воото	-	-
	1	1			1	<u>I</u>

Pin number	Pin name	Type <sup>(1)</sup>	I/O	Main	Alternate	Additional
LQFP48	Pin name		structure <sup>(2)</sup>	function	functions	functions
45	IRQ	I	FT	IRQ	-	-
46	PB9	I/O	FT	DDO	I2C1_SDA/	
46				PB9	TIM17_CH1	-
47	VSS	S	-	VSS	-	-
48	VDD	S	-	VDD	-	-

Table 4. QFN32 Pin definitions

Pin number	Pin name	<b>Type</b> <sup>(1)</sup>	I/O	Main	Alternate	Additional
QFN32	Pin name	Type	structure <sup>(2)</sup>	function	functions	functions
0	VSS	S	-	VSS	-	-
1	VDD	S	-	VDD	-	-
2	RF_XI	I	FT	RF_OSC_IN	-	RF_OSC_IN
3	RF_XO	0	FT	RF_OSC_OUT	-	RF_OSC_OUT
4	NRST	I/O	FT	NRST	-	-
5	VDDA	S	FT	VDDA		-
6	PA0- WKUP	I/O	-	PA0	UART2_CTS/ TIM2_CH1_ETR/ ADC_VIN[0]	WKUP/ COMP1_OUT
7	OSC_IN	I	FT	OSC_IN	-	OSC_IN
8	VDD1V2	S	-	RF_V <sub>VDD1V2</sub>	-	-
9	ANTP	-	-	ANTP	-	-
10	PA4	I/O	FT	PA4	SPI1_NSS/ TIM14_CH1/ ADC_VIN[4]	-
11	PA5	I/O	FT	PA5	SPI1_SCK/ TIM2_CH1_ETR/ ADC_VIN[5]	COMP1_OUT
12	PA6	I/O	FT	PA6	SPI1_MISO/ TIM3_CH1/ TIM1_BKIN/ TIM16_CH1/ ADC_VIN[6]	COMP2_OUT

Pin number	Pin name	<b>Type</b> <sup>(1)</sup>	I/O	Main	Alternate	Additional
QFN32	Filitialite	туре	structure <sup>(2)</sup>	function	functions	functions
					SPI1_MOSI/	
					TIM3_CH2/	
10	DAZ	1/0	ГТ	DA 7	TIM1_CH1N/	
13	PA7	I/O	FT	PA7	TIM14_CH1/	-
					TIM17_CH1/	
					ADC_VIN[7]	
					TIM3_CH3/	
14	PB0	I/O	FT	PB0	TIM1_CH2N/	-
					ADC_VIN[8]	
					TIM14_CH1/	
4-	55.4			55.4	TIM3_CH4/	
15	PB1	I/O	FT	PB1	TIM1_CH3N/	-
					ADC_VIN[9]	
16	PB2	I/O	FT	PB2	-	-
17	VDD	S	FT	RF_V <sub>VDD</sub>	-	-
40	540			D. 0	TIM1_CH1/	
18	PA8	I/O	FT	PA8	МСО	-
					UART1_TX/	
					TIM1_CH2/	
19	PA9	I/O	FT	PA9	UART1_RX/	-
					I2C1_SCL/	
					MCO	
					TIM17_BKIN/	
					UART1_RX/	
20	PA10	I/O	FT	PA10	TIM1_CH3/	-
					UART1_TX/	
					I2C1_SDA	
					UART1_CTS/	
					TIM1_CH4/	
21	PA11	I/O	FT	PA11	CAN_RX/	COMP1_OUT
					I2C1_SCL/	
					USBDM	
					UART1_RTS/	
					TIM1_ETR/	
22	PA12	I/O	FT	PA12	CAN_TX/	COMP2_OUT
					I2C_SDA/	
					USBDP	

Pin number	Pin name	<b>Type</b> <sup>(1)</sup>	I/O	Main	Alternate	Additional
QFN32	Fill Hallie	Type	structure <sup>(2)</sup>	function	functions	functions
23	PA13	I/O	-	PA13	SWDIO	-
24	DA 1.4	1/0		PA14	SWCLK/	
24	PA14	I/O	-	PA 14	UART2_TX	-
					TIM2_CH1_ETR/	
25	PA15	I/O	FT	PA15	SPI1_NSS/	-
					UART2_RX	
26	PB3	I/O	FT	PB3	TIM2_CH2/	
20	PBS	1/0	ГІ	FDJ	SPI1_SCK	-
27	PB4	I/O	FT	PB4	TIM3_CH1/	
21	FD4	1/0	ГІ	FD <del>4</del>	SPI1_MISO	-
					TIM3_CH2/	
28	PB5	I/O	FT	PB5	SPI1_MOSI/	-
					TIM16_CH1N	
					UART1_TX/	
29	PB6	I/O	FT	PB6	I2C_SCL/	-
					TIM16_CH1N	
					UART1_RX/	
30	PB7	I/O	FT	PB7	I2C_SDA/	-
					TIM17_CH1N	
31	воото	ı	FT	PB8	-	-
32	IRQ	I/O	FT	RF_IRQ	-	-

- 1. I = input, O = output, S = power supply, HiZ = high resistance.
- 2. FT: 5V tolerant, Input signal should be between VDD and 5V.

Table 5. Alternate functions for port A

Pin	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
Name			-					
PA0		UART2 CTS	TIM2_CH1					COMP1 OUT
PAU	-	UARIZ_CIS	_ETR	-	-	-	-	COMPT_OUT
PA4	SPI1_NSS	-	-	-	TIM14_CH1	-	-	-
PA5	CDI1 CCK	SPI1_SCK -	TIM2_CH1	-	1	-	-	
- FAG	SFII_SCK		_ETR					-
PA6	SPI1_MISO	TIM3_CH1	TIM1_BKIN	-	-	TIM16_CH1	1	COMP1_OUT
PA7	SPI1_MOSI	TIM3_CH2	TIM1_CH1N	-	TIM14_CH1	TIM17_CH1	1	COMP2_OUT
PA8	MCO	-	TIM1_CH1	-	CRS_SYNC	-	-	-
PA9	-	UART1_TX	TIM1_CH2	UART1_RX	I2C1_SCL	MCO	1	-
PA10	TIM17_BKIN	UART1_RX	TIM1_CH3	UART1_TX	I2C1_SDA	-	-	-
PA11	-	UART1_CTS	TIM1_CH4	-	CAN_RX	I2C1_SCL	-	COMP1_OUT

Pin	A.F.O.	A F.4	A.F.O.	A.F.2	A F.4	A 5 5	AFC	A F.7
Name	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PA12	-	UART1_RTS	TIM1_ETR	-	CAN_TX	I2C1_SDA	-	COMP2_OUT
PA13	SWDIO	-	-	-	-	-	-	-
PA14	SWDCLK	UART2_TX	-	-	-	-	-	-
DA15 CDI1 NCC		DIA NEC LIADTO DV	TIM2_CH1					
IO	PA15 SPI1_NSS	UART2_RX	_ETR	-	-	-	-	-

Table 6. Alternate functions for port B

Pin								
Name	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PB0	-	TIM3_CH3	TIM1_CH2N	-	-	-	-	-
PB1	TIM14_CH1	TIM3_CH4	TIM1_CH3N	-	-	-	-	-
PB2	-	-	-	-	-	-	-	-
PB3	SPI1_SCK	-	TIM2_CH2	-	-	-	-	-
PB4	SPI1_MISO	TIM3_CH1	-	-	-	TIM17_BKIN	-	-
PB5	SPI1_MOSI	TIM3_CH2	TIM16_BKIN	-	-	-	-	-
PB6	UART1_TX	I2C1_SCL	TIM16_CH1N	-	-	-	-	-
PB7	UART1_RX	I2C1_SDA	TIM17_CH1N	-	-	-	-	-
PB8	-	-	-	-	-	-	-	-
PB9	-	I2C1_SDA	TIM17_CH1	-	-		-	-
PB10	-	I2C1_SCL	TIM2_CH3	-	-		-	-
PB11	-	I2C1_SDA	TIM2_CH4	-	-	-	-	-
PB12	-	-	TIM1_BKIN	-	SPI2_MISO	-	-	-
PB13	-	-	TIM1_CH1N	-	SPI2_MOSI	I2C1_SCL	-	-
PB14	-	-	TIM1_CH2N	SPI2_SCK	-	I2C1_SDA	-	-
PB15	-	SPI2_NSS	TIM1_CH3N	-	-	-	-	-

Table 7. Alternate functions for port C and D

Pin	450	A F.4	450	450	A.F.4	A F.F.	A F.O.	A F-7
Name	AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7
PC13	-	-	-	-	-	-	-	-
PC14	-	-	-	-	-	-	-	-
PC15	-	-	-	-	-	-	-	-
PD0	CRS_SYNC	I2C1_SDA	-	-	-	-	-	-
PD1	-	I2C1_SCL	-	-	-	-	-	-
PD2	-	-	-	-	-	-	-	-
PD3	-	-	-	-	-	-	-	-

# **Memory mapping**

Memory mapping

Table 8. memory mapping

Bus	Boundaryaddress	Size	Peripheral	Notes
			Main flash memory, system	
	0x0000 0000 -0x0001 FFFF	128 KB	memory, or SRAM, depends on	
			the configuration of BOOT	
	0x0002 0000 -0x07FF FFFF	~ 128 MB	Reserved	
	0x0800 0000 -0x0801 FFFF	128 KB	Main Flash memory	
	0x0802 0000 -0x 1FFDFFFF	~ 256 MB	Reserved	
Flash	0x1FFE 0000 -0x1FFE 01FF	0.5 KB	Protect bytes	
	0x1FFE 0200 -0x1FFE 0FFF	3 KB	Reserved	
	0x1FFE 1000 -0x1FFE 1BFF	3 KB	Security space	
	0x1FFE 1C00 -0x1FFF F3FF	~ 256 MB	Reserved	
	0x1FFF F400 -0x1FFF F7FF	1 KB	System memory	
	0x1FFF F800 -0x1FFF F80F	16 B	Option bytes	
	0x1FFF F810 -0x1FFF FFFF	~2 KB	Reserved	
ODAM	0x2000 0000 -0x2000 1FFF	8 KB	SRAM	
SRAM	0x2000 2000 -0x2FFF FFFF	~ 512 MB	Reserved	
	0x4000 0000 -0x4000 03FF	1 KB	TIM2	
	0x4000 0400 -0x4000 07FF	1 KB	TIM3	
	0x4000 0800 -0x4000 0BFF	8 KB	Reserved	
	0x4000 2800 -0x4000 2BFF	1 KB	ВКР	
	0x4000 2C00 -0x4000 2FFF	1 KB	WWDG	
	0x4000 3000 -0x4000 33FF	1 KB	IWDG	
	0x4000 3400 -0x4000 37FF	1 KB	Reserved	
APB1	0x4000 3800 -0x4000 3BFF	1 KB	SPI2	
	0x4000 4000 -0x4000 43FF	1 KB	Reserved	
	0x4000 4400 -0x4000 47FF	1 KB	UART2	
	0x4000 4800 -0x4000 4BFF	3 KB	Reserved	
	0x4000 5400 -0x4000 57FF	1 KB	I2C	
	0x4000 5800 -0x4000 5BFF	1 KB	Reserved	
	0x4000 5C00 -0x4000 5FFF	1 KB	USB	
	0x4000 6000 -0x4000 63FF	1 KB	Reserved	
	0x4000 6400 -0x4000 67FF	1 KB	CAN	

Bus	Boundaryaddress	Size	Peripheral	Notes
	0x4000 6800 -0x4000 6BFF	1 KB	Reserved	
APB1	0x4000 6C00 -0x4000 6FFF	1 KB	CRS	
AFDI	0x4000 7000 -0x4000 73FF	1 KB	PWR	
	0x4000 7400 -0x4000 FFFF	35 KB	Reserved	
	0x4001 0000 -0x4001 03FF	1 KB	SYSCFG	
	0x4001 0400 -0x4001 07FF	1 KB	EXTI	
	0x4001 0800 -0x4001 23FF	7 KB	Reserved	
	0x4001 2400 -0x4001 27FF	1 KB	ADC	
	0x4001 2800 -0x4001 2BFF	1 KB	Reserved	
	0x4001 2C00 -0x4001 2FFF	1 KB	TIM1	
APB2	0x4001 3000 -0x4001 33FF	1 KB	SPI1	
	0x4001 3400 -0x4001 37FF	1 KB	DBGMCU	
	0x4001 3800 -0x4001 3BFF	1 KB	UART1	
	0x4001 3C00 -0x4001 3FFF	1 KB	COMP	
	0x4001 4000 -0x4001 43FF	1 KB	TIM14	
	0x4001 4400 -0x4001 47FF	1 KB	TIM16	
	0x4001 4800 -0x4001 4BFF	1 KB	TIM17	
	0x4001 4C00 -0x4001 7FFF	13 KB	Reserved	
	0x4002 0000 -0x4002 03FF	1 KB	DMA	
	0x4002 0400 -0x4002 0FFF	3 KB	Reserved	
	0x4002 1000 -0x4002 13FF	1 KB	RCC	
	0x4002 1400 -0x4002 1FFF	3 KB	Reserved	
	0x4002 2000 -0x4002 23FF	1 KB	Flash interface	
	0x4002 2400 -0x4002 5FFF	15 KB	Reserved	
AHB	0x4002 6000 -0x4002 63FF	1 KB	Reserved	
	0x4002 6400 -0x47FF FFFF	~ 128 MB	Reserved	
	0x4800 0000 -0x4800 03FF	1 KB	GPIOA	
	0x4800 0400 -0x4800 07FF	1 KB	GPIOB	
	0x4800 0800 -0x4800 0BFF	1 KB	GPIOC	
	0x4800 0C00 -0x4800 0FFF	1 KB	GPIOD	
	0x4800 1000 -0x5FFF FFFF	~ 384 MB	Reserved	

# Typical application circuit

Typical application circuit

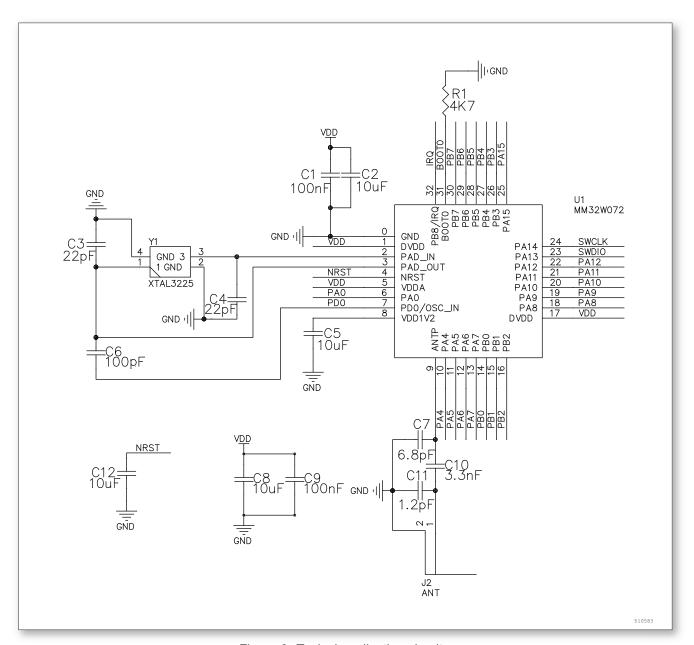


Figure 6. Typical application circuit

Note: This typical application circuit uses a radio frequency module and a control module to share a crystal oscillator solution. If the customer uses the internal clock of the control module, there is no need to add C6, and Pin7 can be used as GPIO.

Table 9. External component list

C1、C2、C8、C9	Decoupling filter capacitor		
C3、C4	Crystal loading capacitor		
C6	Shared clock capacitor		
C5	1.2V digital regulator decoupling capacitor		
C7、C11	RF Transformer / Matching Network		
Cr. Cil	Capacitor		
C10	Through resistance coupling capacitor		
R1	Start mode selection resistor		
Y1	16MHz crystal oscillator (clock source for		
r i	RF modules and control modules)		
J2	2.4G RF antenna		



# Absolute maximum rating operating conditions

Absolute maximum rating operating conditions

Table 10. Absolute maximum rating

Pin	Parameter	Value	Unit	
1、5、17	Control module DC-DC converter	-0.3 to +3.6	V	
	supply voltage input and output	-0.5 to 15.0	V	
2、3	RF module DC voltage crystal	-0.3 to +3.6	V	
	oscillator pin	0.0 to 10.0	v	
4	Control module reset pin	-0.3 to +5.5	V	
7	Control module DC voltage crystal	-0.3 to +5.5	V	
,	oscillator pin	-0.5 to +5.5	V	
8	RF module digital power supply	-0.3 to +1.5	V	
0	1.2V voltage output	-0.5 to +1.5	<b>v</b>	
9	RF Module 2.4G Antenna Pin		V	
31	Control module startup mode	-0.3 to +5.5	V	
	control pin	-0.5 to +5.5	V	
32	RF module interrupt control pin	-0.3 to +3.6	V	
6、10、11、				
12、13、14、				
15、16、18、	Control modulo DC voltage digital			
19、20、21、	Control module DC voltage digital	-0.3 to +5.5	V	
22、23、24、	input and output pin			
25、26、27、				
28、29、30				

# **Electrical characteristics**

Electrical characteristics

#### 7.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V<sub>SS</sub>.

#### 7.1.1 Minimum and maximum values

Unless otherwise specified, the minimum and maximum values are guaranteed with an ambient temperature at  $T_A = 25$ °C,  $V_{DD} = 3.3$ V.

#### 7.1.2 Typical values

Unless otherwise specified, typical data are based on  $T_A = 25^{\circ}C$  and  $V_{DD} = 3.3V$ . They are given only as design guidelines and are not tested.

#### 7.1.3 Typical curves

Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

#### 7.1.4 Loading capacitor

The load conditions used for pin parameter measurement are shown in the figure below.

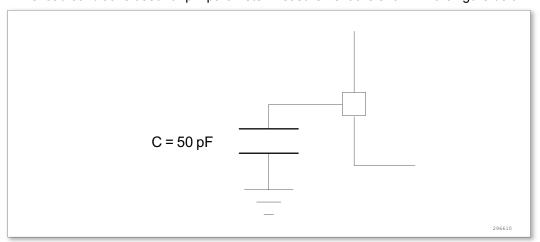


Figure 7. Pin loading conditions

#### 7.1.5 Pin input voltage

The input voltage measurement on a pin of the device is shown in the figure below.

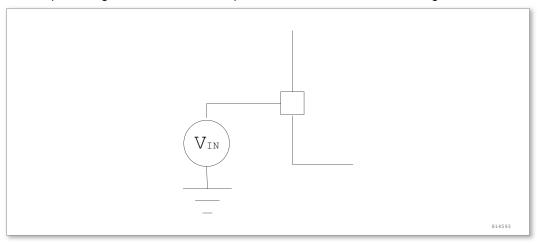


Figure 8. Pin input voltage

#### 7.1.6 Power supply scheme

The power supply design scheme is shown in the figure below.

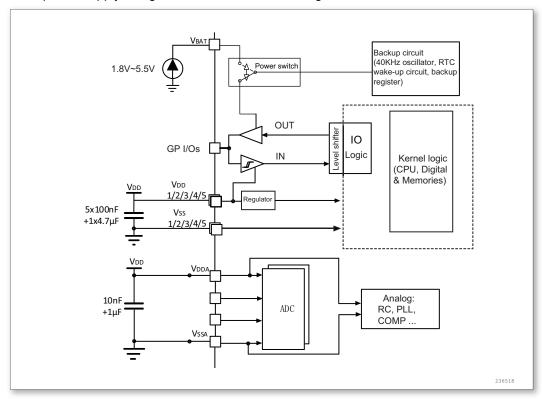


Figure 9. Power supply scheme

#### 7.1.7 Current consumption measurement

The measurement of the current consumption on the pin is shown in the figure below.

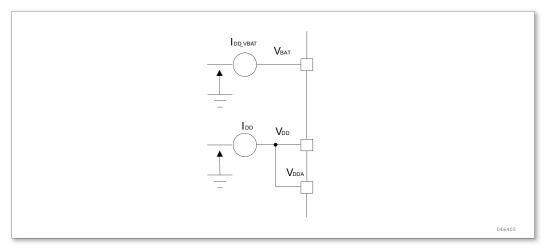


Figure 10. Current consumption measurement scheme

#### 7.2 RF general characteristics

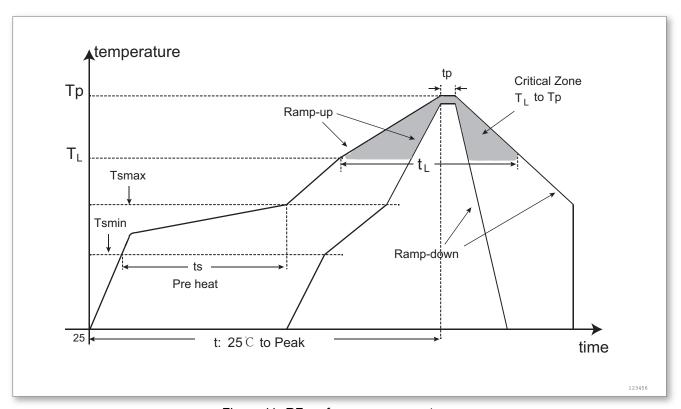


Figure 11. RF performance parameters

Table 11. RF general characteristics

Label	Parameter	Test Conditions	Minimum value	Typical value	Maximum	Unit
FREQ	Frequency change	$V_{DD} = 3.0V, T_A = 25^{\circ}C$	2400		2483.5	MHz
FC	Channel spacing	$V_{DD} = 3.0V, T_A = 25^{\circ}C$		2		MHz
RFch	RF channel center	$V_{DD} = 3.0V, T_A = 25^{\circ}C$	2402		2480	MHz

#### 7.3 RF transmitter characteristics

Table 12. RF Transmitter Characteristics Table

Labal	Parameter	Test Conditions	Minimum	Typical	Maximum	Unit
Label	Parameter	rest Conditions	value	value	Wiaximum	Onit
MOD	Modulation		GFSK			
ВТ	Bandwidth			0.5		
M <sub>index</sub>	Modulation index		0.45	0.5	0.55	
DB	Air transmission			1		Mhna
DR	index			ı		Mbps
P <sub>max</sub>	Maximum				+4	dBm
	transmission power				+4	UDIII
	6dB bandwidth					
P <sub>BW1M</sub>	modulated carrier		500			KHz
	(1Mbps)					
P <sub>SPUR</sub>	Spurious emission				-41	dBm
CE	Center frequency				1150	VU-
CF <sub>dev</sub>	offset				±150	KHz
Freq <sub>drift</sub>	Frequency drift				±50	KHz
IFrog	Initial carrier				130	VU-
IFreq <sub>drift</sub>	frequency drift				±20	KHz

#### 7.4 RF receiver characteristics

Table 13. RF receiver characteristics

Label	I Parameter Test Conditions		Minimum value	Typical value	Maximum	Unit
RX <sub>SENS</sub>	Receiving sensitivity	BER < 0.1%		-80		dBm

#### 7.5 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in Tables(Table 14. Table 15. Table 16) may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

50

Symbol Definition Min Unit Max External main supply  $V_{DD} - V_{SS}$ - 0.3 5.5 voltage(including  $V_{DDA}$  and  $V_{SSA}$ )<sup>(1)</sup> Input voltage on FT and FTf  $pins^{(2)}$  $V_{\text{SS}}$  - 0.3 5.5  $V_{\text{IN}}$ Input voltage on other pins(2)  $V_{SS}$  - 0.3 5.5 Variations between different V<sub>DD</sub>  $| \vartriangle V_{\mathsf{DDx}} |$ 50 power pins mV Variations between all the different

Table 14. Voltage characteristics

1. All main power (V<sub>DD</sub>, V<sub>DDA</sub>) and ground (V<sub>SS</sub>, V<sub>SSA</sub>) pins must always be connected to the external power supply, in the permitted range.

ground pins

2. V<sub>IN</sub> maximum must always be respected. Refer to Table below for maximum allowed injected current values.

Table 15. Current characteristics

 $|V_{\rm SSx} - V_{\rm SS}|$ 

Symbol	Definition	Max	Unit
I <sub>VDD</sub>	Total current into sum of all V <sub>DD</sub> /V <sub>DDA</sub> power	150	mA
	lines(source) <sup>(1)</sup>		
I <sub>VSS</sub>	Total current out of sum of all V <sub>SS</sub> ground lines(sink) <sup>(1)</sup>	150	mA
I <sub>IO</sub>	Output current sunk by any I/O and control pin	20	mA
I <sub>IO</sub>	Output current source by any I/O and control pin	-18	mA
I <sub>INJ(PIN)</sub> (2)(3)	Injected current on NRST pins	±5	mA
(2)(3)	Injected current on OSC_IN pin of HSE and OSC_IN pin	±5	mA
$I_{INJ(PIN)}^{(2)(3)}$	of LSE	15	IIIA
I <sub>INJ(PIN)</sub> (2)(3)	Injected current on other pins <sup>(4)</sup>	±5	mA
$\sum I_{\text{INJ(PIN)}}^{(2)}$	Total injected current(sum of all I/O and control pins)(4)	±25	mA

- 1. All main power( $V_{DD}$ ,  $V_{DDA}$ ) and ground( $V_{SS}$ ,  $V_{SSA}$ ) pins must always be connected to the external power supply, in the permitted range.
- 2.  $I_{INJ(PIN)}$  cannot exceed its limit, that is, to ensure that the  $V_{IN}$  does not exceed its maximum value. If  $V_{IN}$  does not guarantee that its maximum value is not exceeded, ensure that  $I_{INJ(PIN)}$  does not exceed its maximum value under external restrictions. When  $V_{IN} > V_{DD}$ , there is a forward injection current; when  $V_{IN} < V_{SS}$ , there is a reverse injection current.
- 3. Negative injection disturbs the analog performance of the device.
- 4. When several inputs are submitted to a current injection, the maximum I<sub>INJ(PIN)</sub> is the absolute sum of the positive and negative injected currents (instantaneous values).

Table 16. Thermal characteristics

Symbol	Symbol Definition		Unit
T <sub>STG</sub>	Storage temperature range	- 45 ~ + 150	°C
т.	Maximum junction	125	°C
IJ	temperature	125	

#### 7.6 Operating conditions

#### 7.6.1 General operating conditions

Table 17. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit
£	Internal AHB clock		0	00	
f <sub>HCLK</sub>	frequency		0	96	MHz
£	Internal APB1 clock		0	f.	IVITZ
f <sub>PCLK1</sub>	frequency		U	f <sub>HCLK</sub>	
f	Internal APB2 clock		0	f	
f <sub>PCLK2</sub>	frequency		U	f <sub>HCLK</sub>	
$V_{DD}^{(1)}$	Standard operating		2.0	5.5	V
	voltage		2.0	5.5	V
V	Analog operating	Must be the same voltage as V <sub>DD</sub> <sup>(1)</sup>	2.5	5.5	V
$V_{DDA}$	voltage	widst be the same voltage as v <sub>DD</sub>	2.5	5.5	<b>V</b>
\/	Backup part of working		1.8	5.5	V
$V_{BAT}$	voltage		1.0	5.5	V
	Power dissipation	LQFP64		203	
$P_D$	temperature:	LQFP48			mW
	$T_A = 85^{\circ}C^{(2)}$	LQFP32/QFN32			
	Ambient temperature:	Maximum power dissipation	-40	85	- °C
	T <sub>A</sub> =85°C	Low power dissipation <sup>(3)</sup>	-40	105	
	Ambient temperature:	Maximum power dissipation	-40	105	- °C
$T_A$	T <sub>x</sub> =105°C	Low power dissipation <sup>(3)</sup>	-40	125	

- 1. It is recommended to use the same power supply for  $V_{DD}$  and  $V_{DDA}$ , the maximum permissible difference between  $V_{DD}$  and  $V_{DDA}$  is 300mVduring power up and normal operation.
- 2. If  $T_A$  is low, higher  $P_D$  values are allowed as long as  $T_J$  does not exceed  $T_{Jmax}$  (See subsec 7.1).
- 3. In low power dissipation state,  $T_A$  can be extended to this range as long as  $T_J$  does not exceed  $T_{Jmax}$  (See subsec 7.1).

#### 7.6.2 Operating conditions at power-up/power-down

The parameters given in the table below are based on tests under normal operating conditions.

Table 18. Operating conditions at power-up/power-down

Symbol	Symbol Parameter		Min	Max	Unit
4	V <sub>VDD</sub> rise time rate	T - 27°C	100	∞	СЛ/
$t_{VDD}$	V <sub>VDD</sub> fall time rate	T <sub>A</sub> = 27°C	100	∞	μS/V

#### 7.6.3 Embedded reset and power control block characteristics

The parameters given in the table below are based on the ambient temperature and the  $V_{DD}$  supply voltage listed in Table 17.

Table 19. Embedded reset and power control block characteristics

Symbol Parameter		Conditions	Min	Тур	Max	Unit
		PLS[3: 0]=0000 (Rising edge)	1.813	1.819	1.831	V
		PLS[3: 0]=0000 (Falling edge)		1.705		V
		PLS[3: 0]=0001 (Rising edge)	2.112	2.116	2.124	V
		PLS[3: 0]=0001 (Falling edge)		2.0		V
		PLS[3: 0]=0010 (Rising edge)	2.411	2.414	2.421	V
		PLS[3: 0]=0010 (Falling edge)		2.297		V
		PLS[3: 0]=0011 (Rising edge)	2.711	2.714	2.719	V
		PLS[3: 0]=0011 (Falling edge)		2.597		V
		PLS[3: 0]=0100 (Rising edge)	3.011	3.013	3.018	V
		PLS[3: 0]=0100 (Falling edge)		2.895		V
V	La colonda de la constanta de	PLS[3: 0]=0101 (Rising edge)	3.311	3.313	3.317	V
$V_{PVD}$	Level selection of p	rogrammable voltage detectors PLS[3: 0]=0101 (Falling edge)		3.194		V
		PLS[3: 0]=0110 (Rising edge)	3.611	3.613	3.616	V
		PLS[3: 0]=0110 (Falling edge)		3.494		V
		PLS[3: 0]=0111 (Rising edge)	3.91	3.913	3.916	V
		PLS[3: 0]=0111 (Falling edge)		3.793		V
		PLS[3: 0]=1000 (Rising edge)	4.21	4.212	4.215	V
		PLS[3: 0]=1000 (Falling edge)		4.092		V
		PLS[3: 0]=1001 (Rising edge)	4.51	4.512	4.515	V
		PLS[3: 0]=1001 (Falling edge)		4.391		V
		PLS[3: 0]=1010 (Rising edge)	4.809	4.811	4.813	V
		PLS[3: 0]=1010 (Falling edge)		4.69		V
V <sub>PVDhyst</sub> (2)	PVD hysteresis			100		mV
\/	Power on/down	Falling edge	1.63 <sup>(1)</sup>	1.66	1.68	V
$V_{POR/PDR}$	reset threshold	Rising edge		1.75		V

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{PDRhys}^{(2)}$	PDR hysteresis			90.9		mV
T <sub>RSTTEMPO</sub> <sup>(2)</sup>	Reset duration			20		ms

- 1. The product behavior is guaranteed by design down to the minimum value V<sub>POR/PDR</sub>.
- 2. Guaranteed by design, not tested in production.

Note: The reset duration is measured from power-on (POR reset) to the time when the user application code reads the first instruction.

#### 7.6.4 Embedded internal reference voltage

The parameters given in the table below are based on the ambient temperature and the  $V_{DD}$  supply voltage listed in Table 17.

Table 20. Embedded internal reference voltage<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V	Internal reference voltage	$-40^{\circ} C < T_A < +105^{\circ} C$		1.2		V
$V_{REFINT}$	Internal reference voltage	$\text{-40}^{\circ}\text{C} < \text{T}_{\text{A}} < \text{+85}^{\circ}\text{C}$		1.2		V
T <sub>S_vrefint</sub> (1)	ADC sampling time when re	eading the internal reference vo	oltage <sup>d</sup> 0			μS

1. Shortest sampling time can be determined in the application by multiple iterations.

#### 7.6.5 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

All Run-mode current consumption measurements given in this section are performed with a reduced code.

#### **Maximum current consumption**

The MCU is placed under the following conditions:

- All I/O pins are in analog input mode, and are connected to a static level V<sub>DD</sub> or V<sub>SS</sub> (no load)
- · All peripherals are disabled except when explicitly mentioned
- The Flash memory access time is adjusted to the  $f_{HCLK}$  (0  $\sim$  24 MHz is 0 waiting period , 24  $\sim$  48 MHz is 1 waiting period, 48  $\sim$  72 MHz is 2 waiting period, 72  $\sim$  96 MHz is 3 waiting period ).
- The instruction prefetching function is on. When the peripherals are enabled:

 $f_{PCLK1} = f_{HCLK}$ .

Note: The instruction prefetching function must be set before setting the clock and bus divider.

The parameters given in the table (Table 21 \ Table 22 \ Table 23) are based on the ambient temperature and the  $V_{DD}$  supply voltage listed in Table 17.

Table 21. Typical current consumption in Run mode, code executing from Flash memory

Symbol	Parameter	Conditions	<b>f</b> <sub>HCLK</sub>	Тур	Unit	
<b>Gy</b> iii.50i	, aramotor	Conditions	•HCLK	All peripherals enabled <sup>(2)</sup>	All peripherals disabled	O.I.I.
			96MHz	26.23	15.2	
		Supply current in run mo <b>fe</b> ternal clock <sup>(2)</sup>	72MHz	20.52	12.19	
$I_{DD}$	Supply current in run m		<sup>)</sup> 48MHz	14.71	9.13	mA
			36MHz	11.76	7.58	
			24MHz	8.84	6.03	
	Supply current in run	External	ONAL I—	4.4	2.44	
I <sub>DD</sub>	mode	clock <sup>(2)</sup>	8MHz	4.1	3.14	

- 1. The typical value is tested at  $T_A = 25^{\circ}$ C.Guaranteed by design, not tested in production.
- 2. External clock is 8MHz, when  $f_{HCLK} > 8MHz$  enable PLL.

Table 22. Typical current consumption in Sleep mode, code executing from Flash memory or RAM

Symbol	Parameter	Conditions	<b>f</b> <sub>HCLK</sub>	Тур	Unit	
		Conditions	•HCLK	All peripherals enabled <sup>(2)</sup>	All peripherals disabled	
			96MHz	22.41	10.92	
			72MHz	17.57	8.96	
$I_{DD}$	Supply current in Sleer	r <del>lambu</del> nal clock <sup>(2</sup>	48MHz	12.68	6.96	mA
טטי	Supply current in Sieer	lly current in Sleep r <b>hate</b> rnal clock <sup>(2)</sup>	36MHz	10.29	5.95	
			24MHz	7.79	4.9	
			8MHz	3.46	2.8	

- 1. The typical value is tested at  $T_A$  = 25°C.From a comprehensive evaluation, it is tested in terms of  $V_{DDmax}$  and  $f_{HCLKmax}$  enable peripherals in production.
- 2. External clock is 8MHz, when  $f_{\mbox{\scriptsize HCLK}} > 8 \mbox{\scriptsize MHz}$  enable PLL.

Table 23. Maximum current consumption in Stop and Standby mode, code executing from Flash memory

Symbol	Parameter	Conditions	Max T <sub>A</sub> =25°C	Unit
I <sub>DD</sub>	Supply current in Stop mode	Enter the stop mode after reset, $V_{DD} = 3.3V$	402	

 $\mu A$ 

Symbol	Parameter	Conditions	Max	Unit
Зушьог	raiailletei	Conditions	<b>T</b> <sub>A</sub> <b>=25</b> °C	Oille
	Supply current in Standby	Enter the standby mode after reset,	0.4	
	mode	V <sub>DD</sub> = 3.3V	0.4	
	Supply current in the	Low speed oscillator and RTC are on,	0.2	
IDD_VBAT	backup area	$V_{DD}/V_{BAT} = 3.3V$	0.2	μА

#### 1. I/O status is analog input.

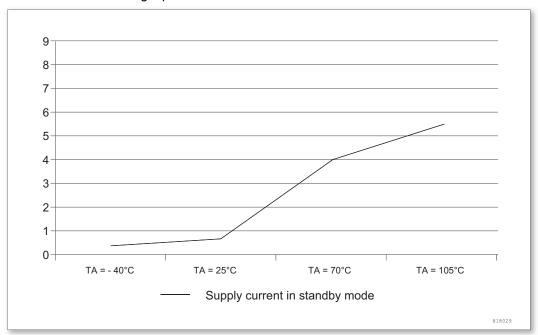


Figure 12. Typical current consumption in standby mode vs. temperature at  $V_{DD}$  = 3.3V

#### **Typical current consumption**

The MCU is placed under the following conditions:

- All I/O pins are in analog input configuration, and are connected to a static level V<sub>DD</sub> or V<sub>SS</sub> (no load).
- All the peripherals are closed, unless otherwise specified.
- The Flash memory access time is adjusted to the  $f_{HCLK}$  (0  $\sim$  24 MHz is 0 waiting period ,24  $\sim$  48 MHzis 1 waiting period, 48  $\sim$  72 MHzis 2 waiting period, 72  $\sim$  96 MHzis 3 waiting period ).
- $\bullet$  The ambient temperature and  $V_{DD}$  supply voltage conditions are summarized in Table 17
- The instruction prefetching function is on. When the peripherals are enabled:  $f_{PCLK1} = f_{HCLK}.$

Note: The instruction prefetch function must be set before the clock is set and the bus is divided.

#### On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in Table 24. The MCU is placed under the following conditions:

- All I/O pins are in analog input mode, and are connected to a static level V<sub>DD</sub> or V<sub>SS</sub> (no load).
- All peripherals are disabled except when explicitly mentioned.
- The given value is calculated by measuring the current consumption.
  - with all peripherals clocked OFF
  - with only one peripheral clocked on
- Ambient operating temperature and supply voltage conditions V<sub>DD</sub> summarized in Table 17.

Table 24. On-chip peripheral current consumption<sup>(1)</sup>

Perij	oheral	Typical consumption at 25 °C	Unit	Peri	pheral	Typical consumption at 25 °C	Unit
A DD4	TIM2	0.098		ADDO	GPIOA	0.045	4
APB1	TIM3		GPIOB	0.046	mA		
	TIM4	0.055			GPIOC	0.052	
	SPI2	0.133			GPIOD	0.046	
	UART2	0.077			ADC1	0.051	
APB1	UART3	0.078	mA	APB2	ADC2	0.052	mA
, ,, , , ,	I2C1	0.132	1 11/1	, u D2	TIM1	0.121	1117
	12C2	0.134	1		SPI1	0.122	
	USB	0.058	1		UART1	0.078	
	CAN	0.033	1				

<sup>1.</sup>  $f_{HCLK} = 96MHz$ ,  $f_{APB1} = f_{HCLK}/2$ ,  $f_{APB2} = f_{HCLK}$ , the prescale coefficient for each device is the default value.

#### 7.6.6 External clock source characteristics

#### High-speed external user clock generated from an external source

The characteristic parameters given in the following table are measured using a high-speed external clock source, ambient temperature and power supply voltage meet the conditions of General operating conditions.

Table 25. High-speed external user clock characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
£	User external clock source		2	0	24	MHz
THSE_ext	frequency <sup>(1)</sup>			0	24	IVITZ

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{HSEH}$	OSC_IN input pin high level voltage		0.7V <sub>DD</sub>		$V_{DD}$	V
V <sub>HSEL</sub>	OSC_IN input pin low level voltage		V <sub>SS</sub>		0.3V <sub>DD</sub>	V
t <sub>w(HSE)</sub>	OSC_IN high or low time <sup>(1)</sup>		16			
$t_{r(HSE)}$ $t_{f(HSE)}$	OSC_IN rise or fall time <sup>(1)</sup>				20	nS
C <sub>in(HSE)</sub>	OSC_IN input capacitance <sup>(1)</sup>			5		pF
DuCy <sub>(HSE)</sub>	Duty cycle		45		55	%
Ι <sub>L</sub>	OSC_IN input leakage current	$V_{SS} \le V_{IN} \le V_{DD}$			±1	uA

1. Guaranteed by design, not tested in production.

#### Low-speed external user clock characteristics

The characteristic parameters given in the following table are measured using a low-speed external clock source, ambient temperature and power supply voltage meet the conditions of General operating conditions.

Table 26. Low-speed external user clock characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
	User external clock source		10	22.700	200	KHz
$f_{LSE\_ext}$	frequency <sup>(1)</sup>		16	32.768	200	KIIZ
V	OSC_IN input pin high level				1.2	V
$V_{LSEH}$	voltage				1.2	V
· · ·	OSC_IN input pin low level		0.25			V
$V_{LSEL}$	voltage		0.25			V
$t_{w(LSE)}$	OSC_IN high or low time <sup>(1)</sup>			15259		nS
$t_{r(LSE)}$	OSC_IN rise time <sup>(1)</sup>			30		nS
$t_{f(LSE)}$	OSC_IN fall time <sup>(1)</sup>			30		nS
$C_{in(LSE)}$	OSC_IN input capacitance <sup>(1)</sup>			5		pF
DuCy <sub>(LSE)</sub>	Duty cycle			50		%
IL	OSC_IN input leakage current	$V_{SS} \le V_{IN} \le V_{DD}$		0.03		uA

1. Guaranteed by design, not tested in production.

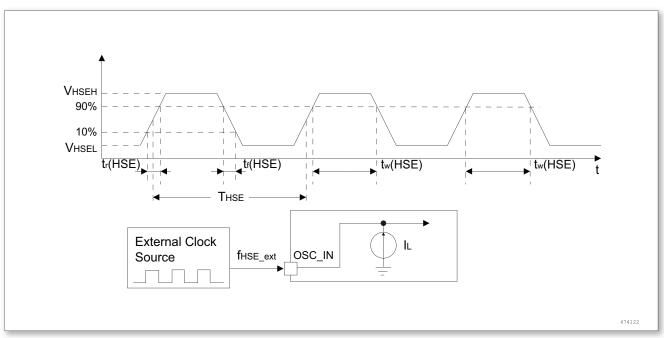


Figure 13. High-speed external clock source AC timing diagram

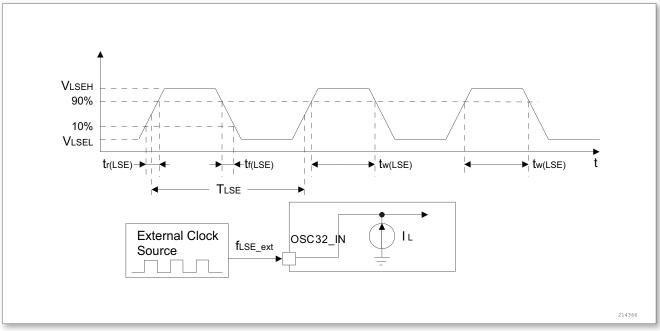


Figure 14. Low-speed external clock source AC timing diagram

# High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with an 2 to 24 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in the table below. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteris-

tics (frequency, package, accuracy...).

Table 27. HSE oscillator characteristics<sup>(1)(2)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>OSC_IN</sub>	Oscillator frequency		2	8	24	MHz
R <sub>F</sub>	Feedback resistor	$R_S = 30\Omega$		1000		kΩ
$\begin{array}{c} C_{L1} \\ C_{L2} ^{(3)} \end{array}$	The proposed load capacitance corresponds to the crystal serial impedance $\left(R_{S}\right)^{(4)}$	$V_{DD} = 3.3V$ $V_{IN} = V_{SS}$ 30pF load		30		pF
l <sub>2</sub>	HSE current consumption	Startup			1	mA
g <sub>m</sub>	Oscillator transconductance	V <sub>DD</sub> is stabilized	25			mA/V
t <sub>SU(HSE)</sub> <sup>(5)</sup>	Startup time	$R_S = 30\Omega$		2		mS

- Resonator characteristics given by the crystal/ceramic resonator manufacturer characteristics Parameter.
- 2. Guaranteed by design, not tested in production.
- 3. For  $C_{L1}$  and  $C_{L2}$ , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 25 pF range (Typ.) , designed for high-frequency applications, and selected to match the requirements of the crystal or resonator.  $C_{L1}$  and  $C_{L2}$  are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of  $C_{L1}$  and  $C_{L2}$ . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing  $C_{L1}$  and  $C_{L2}$ .
- 4. The relatively low value of the RF resistance can be used to avoid problems arising from the use of wet conditions to provide protection, this environment resulting in leakage and bias conditions have changed. However, if the MCU is applied in bad wet conditions, the design needs to take this parameter into account.
- t<sub>SU(HSE)</sub> is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.

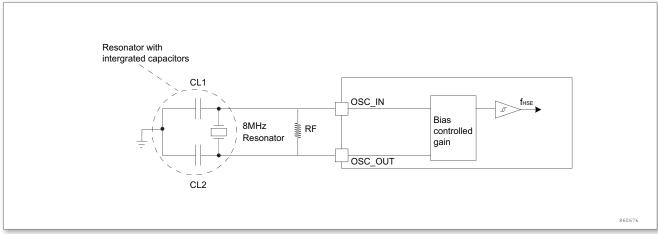


Figure 15. Typical application with an 8 MHz crystal

# Low-speed external clock generated from a crystal/ceramic resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in the table below. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy). Note: For C<sub>L1</sub> and C<sub>L2</sub>, it is recommended to use a high quality ceramic capacitor between 5pF and 15pF and select the crystal or resonator that meets the requirements. Usually C<sub>L1</sub> and C<sub>L2</sub> have the same parameters. Crystal manufacturers typically give the parameters of the load capacitance in a serial combination of  $C_{L1}$  and  $C_{L2}$ . The load capacitance  $C_L$  is calculated by:  $C_L = C_{L1} \times C_{L2}/(C_{L1} + C_{L2})$ + C<sub>stray</sub>, where C<sub>stray</sub> is the capacitance of the pin and the capacitance associated with the PCB or PCB. Its typical value is between 2pF and 7pF. WARNING: To avoid exceeding the maximum value of  $C_{L1}$  and  $C_{L2}$  (15pF), it is highly recommended to use a resonator with a load capacitance of C<sub>L</sub> ≤ 7pF. A resonator with a load capacitance of 12.5pF cannot be used. For example, if a resonator with a load capacitance of C<sub>L</sub> = 6pF is selected and  $C_{\text{stray}} = 2pF$ , then  $C_{L1} = C_{L2} = 8pF$ .

Table 28. LSI oscillator characteristics (f<sub>LSE</sub>=32.768KHz) (1)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
R <sub>F</sub>	Feedback resistor			25		ΜΩ
	The proposed load					
$C_{L1}$	capacitance corresponds to	D = 200			4	, nF
$C_{L2}\ ^{(2)}$	the crystal serial impedance	$R_S = 30\Omega$			4	pF
	(R <sub>S</sub> ) <sup>(3)</sup>					
	LCE ourrent consumption	V <sub>DD</sub> = 3.3V		0.00		
l <sub>2</sub>	LSE current consumption	V <sub>IN</sub> = V <sub>SS</sub>		0.08		μA
g <sub>m</sub>	Oscillator transconductance			0.5		μA/V
t <sub>SU(HSE)</sub> (4)	Startup time	V <sub>DD</sub> is stabilized		1	4	S

- 1. Guaranteed by design, not tested in production.
- 2. See the note and warning paragraphs above this table.
- 3. Selecting a high quality oscillator with a small RS value (such as MSIV-TIN 32.768KHz) optimizes current consumption. Please consult the crystal manufacturer for details.
- 4. t<sub>SU(HSE)</sub> is the start-up time, which is measured from the time the software enables HSE until a stable 8MHz oscillation is obtained. This value is measured on a standard crystal resonator, which may vary greatly depending on the crystal manufacturer.

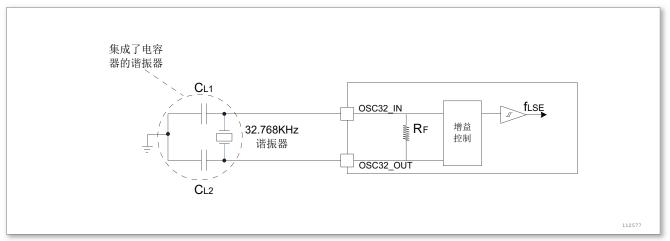


Figure 16. Typical application with a 32.768 kHz crystal

#### 7.6.7 Internal clock source characteristics

The characteristic parameters given in the table below are measured using ambient temperature and supply voltage in accordance with general operating conditions.

#### High-speed internal (HSI) oscillator

Table 29. HSI oscillator characteristics<sup>(1)(2)</sup>

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>HSI</sub>	Frequency		40	48	64	MHz
ACC <sub>HSI</sub>	Accuracy of the HSI oscillator	T <sub>A</sub> = -40°C ~ 105°C	-5		5	%
ACC <sub>HSI</sub>	Accuracy of the HSI oscillator	T <sub>A</sub> = -10°C ~ 85°C	-3		3	%
ACC <sub>HSI</sub>	Accuracy of the HSI oscillator	T <sub>A</sub> = 0°C ~ 70°C	-2		2	%
ACC <sub>HSI</sub>	Accuracy of the HSI oscillator	T <sub>A</sub> = 25	-1		1	%
t <sub>SU(HSI)</sub>	HSI oscillator startup time				2	μS
I <sub>DD(HSI)</sub>	HSI oscillator power consumption			81	200	μА

- 1.  $V_{DD}$  = 3.3V,  $T_A$  = -40°C  $\sim$  105°C, unless otherwise specified.
- 2. Guaranteed by design, not tested in production.

#### Low-speed internal (LSI) oscillator

Table 30. LSI oscillator characteristics(1)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
f <sub>LSI</sub> <sup>(2)</sup>	Frequency		31	40	75	KHz
t <sub>SU(LSI)</sub> (2)	LSI oscillator startup time				1	μS

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
I <sub>DD(LSI)</sub> (3)	LSI oscillator power			1 1	1.7	μA
	consumption			1.1	1.7	μΛ

- 1.  $V_{DD}$  = 3.3V,  $T_A$  = -40°C  $\sim$  105°C, Unless otherwise stated
- 2. Comprehensive assessment, not tested in production.
- 3. Guaranteed by design, not tested in production.

#### Wake-up times from low power mode

The wake-up times listed in the table below are measured during the wake-up phase of the internal clock HSI. The clock source used when waking up depends on the current operating mode:

- · Stop or Standby mode: The clock source is the oscillator
- · Sleep mode: The clock source is the clock used when entering sleep mode

All times are measured using ambient temperature and supply voltage in accordance with common operating conditions.

Table 31. Low-power mode wakeup timings

Symbol	Parameter	Conditions	Max	Unit
t <sub>WUSLEEP</sub> (1)	Wakeup from Sleep mode	HSI clock wakeup	4.2	
t <sub>wustop</sub> (1)	Wakeup from Stop mode (The regulator is in run mode)	HSI clock wakeup = 2μS	6.3	μS
t <sub>wustdby</sub> (1)	Wakeup from Standby mode	HSI clock wakeup = 2µS  The regulator wakes up  from the off mode = 38µS	47	mS

1. The wake-up time is measured from the start of the wake-up event to the user program to read the first instruction.

#### 7.6.8 PLL characteristics

The parameters listed in the table below are measured using ambient temperature and supply voltage in accordance with common operating conditions.

Table 32. PLL characteristics<sup>(1)</sup>

Symbol	Parameter	Min	Тур	Max	Unit
<u> </u>	PLL input clock <sup>(2)</sup>	2		24	MHz
$f_{PLL\_IN}$	PLL input clock duty cycle	40		60	%
f <sub>PLL_OUT</sub>	PLL multiplier output clock	40		200	MHz
t <sub>LOCK</sub>	PLL lock time			100	μS

- 1. Guaranteed by design, not tested in production.
- 2. Take care to use the appropriate multiplier factors to obtain PLL input clock values compatible with the range defined by  $f_{\text{PLL\_OUT}}$ .

#### 7.6.9 Memory characteristics

#### Flash memory

The characteristics are given at  $T_A$  = -  $40^{\circ}$ C  $\sim 105^{\circ}$ Cunless otherwise specified.

Table 33. Flash memory characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
4	O hit was are remined times	T <sub>A</sub> = -40°C ~	4			0
t <sub>prog</sub>	8-bit programming time	125°C	4			μS
+	Dago (E12K butos) araso timo	T <sub>A</sub> = -40°C ~		4	5	m <sup>C</sup>
t <sub>ERASE</sub>	Page (512K bytes) erase time	125°C		4	5	mS
+	Mass erase time	T <sub>A</sub> = -40°C ~	20		40	mS
t <sub>ME</sub>	Mass erase time	125°C	20		40	IIIO
		Read mode, f <sub>HCLK</sub> =		5	6	mA
		48MHz		5	0	ША
		Write mode,f <sub>HCLK</sub> =			7	mA
$I_{DD}$	Supply current	48MHz			ı	ША
		Erase mode, f <sub>HCLK</sub>			2	mA
		= 48MHz				111/4
$I_{SB}$	Standby current			1@25°C	50@125°C	μA
$I_{DEP}$	Deep Standby current			0.5	15@125°C	μΑ

Table 34. Flash memory endurance and data retention (1)(2)

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
	Endurance					
	(Annotation:	T - 400C 050C				
NEND	Erase	$T_A = -40^{\circ}\text{C} \sim 85^{\circ}\text{C}$	10			K cycle
	number of	$T_A = -40^{\circ}C \sim 105^{\circ}C$				
	times)					

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
	Data	1 K cycle <sup>(2)</sup> at T <sub>A</sub> = 85°C	30			
$t_{RET}$	retention	1 K cycle <sup>(1)(2)</sup> at T <sub>A</sub> = 105°C	10			Year
		10 K cycle <sup>(1)(2)</sup> at T <sub>A</sub> = 55°C	20			

- 1. Guaranteed by design, not tested in production.
- 2. Cycle tests are carried out in the whole temperature range.

#### 7.6.10 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

#### Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports), the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- Electrostatic discharge (ESD) (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- FTB: A Burst of Fast Transient voltage (positive and negative) is applied to VDD and VSS through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 1000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in the following table. They are based on the EMS levels and classes defined in application note.

Table 35. EMS characteristics

Symbol	Parameter	Conditions	Level/Class
	Fast transientvoltage burst		
	limits to be applied through	$V_{DD} = 3.3V, T_A = +25^{\circ}C,$	
$V_{EFT}$	100 pF on $V_{DD}$ and $V_{SS}$	f <sub>HCLK</sub> =48MHz.Conformingto	TBD
	pinsto induce a functional	IEC 1000-4-4	
	disturbance		

#### Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

#### Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical Data corruption (for example control registers)

#### **Prequalification trials**

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors.

#### 7.6.11 Absolute Maximum (Electrical Sensitivity)

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

#### Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts  $\times$  (n+1) supply pins). This test conforms to the JESD22-A114/C101 standard.

#### Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- · A supply overvoltage is applied to each power supply pin
- A current injection is applied to each input, output and configurable I/O pin

These tests are compliant with EIA/JESD78A IC latch-up standard.

Table 36. ESD characteristics

Symbol	Parameter	Conditions	Туре	Max	Unit
V <sub>ESD(HBM)</sub>	Electrostatic discharge voltage (Human body model)	$T_A = +25^{\circ}C$ , Conforming to JESD22-A114		2000	V
$V_{\text{ESD}(\text{CDM})}$	Electrostatic discharge voltage (Charging device model)	$T_A = +25$ °C, Conforming to JESD22-C101		500	

		T <sub>A</sub> =		
I <sub>LU</sub>	Latch-up current	+25°C,Conforming to	200	mA
		JESD78A		

#### 7.6.12 I/O port characteristics

#### General input/output characteristics

Unless otherwise specified, the parameters given in Table 14 are derived from tests.

Table 37. I/O static characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{IL}$	Low level input voltage	CMOS Port	-0.5		1.1	V
$V_{IH}$	High level input voltage	CMOS Port	2.08			V
V <sub>hys</sub>	Schmitt trigger hysteresis <sup>(1)</sup>		500	700	800	mV
I <sub>lkg</sub>	Input leakage current <sup>(2)</sup>				1	μА
R <sub>PU</sub>	Weak pull-up equivalent resistor <sup>(3)</sup>	V <sub>IN</sub> = V <sub>SS</sub>	30	50	100	kΩ
$R_{PD}$	Weak pull-down equivalent resistor <sup>(3)</sup>	V <sub>IN</sub> = V <sub>DD</sub>	30	50	100	
C <sub>IO</sub>	I/O pin capacitance				5	pF

- Schmitt Trigger switching hysteresis voltage level. Data based on design simulation only. Not tested in production.
- 2. The leakage could be higher than the maximum value, if negative current is injected on adjacent pins.
- Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This PMOS/NMOS contribution to the series resistance is minimal (10% order).

All I/Os are CMOS (no software configuration required). Their characteristics cover more than the strict CMOS-technology.

- For V<sub>IH</sub>:
  - If  $V_{DD}$  is between [2.50V $\sim$  3.08V]; use CMOS features.
  - If V<sub>DD</sub> is between [3.08V∼ 3.60V]; include CMOS.
- For V<sub>IL</sub>:
  - Use CMOS features.

#### **Output driving current**

The GPIOs (general purpose input/outputs) can sink or source up to ±20mA.

n the user application, the number of I/O pins which can drive current must be limited to

respect the absolute maximum rating specified in 7.5:

- The sum of the currents obtained from V<sub>DD</sub> for all I/O ports, plus the maximum operating current that the MCU obtains on V<sub>DD</sub>, cannot exceed the absolute maximum rating I<sub>VDD</sub>.
- The sum of the currents drawn by all I/O ports and flowing out of V<sub>SS</sub>, plus the maximum operating current of the MCU flowing out on V<sub>SS</sub>, cannot exceed the absolute maximum rating I<sub>VSS</sub>.

#### **Output voltage levels**

Unless otherwise stated, the parameters listed in the table below are measured using the ambient temperature and  $V_{DD}$  supply voltage in accordance with the condition Table 14. All I/O ports are CMOS compatible.

Table 38. Output voltage characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
V <sub>OL</sub> <sup>(1)</sup>	Output low level voltage for an I/O pin,when 8 pins absorb current	CMOS Port, $I_{IO}$ = +8mA 2.7V < $V_{DD}$ < 3.6V		0.4	
V <sub>OH</sub> <sup>(2)</sup>	Output high level voltage for an I/O pin,when 8 pins output current	CMOS Port, $I_{IO}$ = +8mA 2.7V < $V_{DD}$ < 3.6V	0.8V <sub>DD</sub>		
V <sub>OL</sub> <sup>(1)(3)</sup>	Output low level voltage for an I/O pin,when 8 pins absorb current	$I_{IO}$ = +20mA 2.7V < $V_{DD}$ < 3.6V		0.4	V
$V_{OH}^{(2)(3)}$	Output high level voltage for an I/O pin,when 8 pins output current		$0.8V_{DD}$		
V <sub>OL</sub> <sup>(2)(3)</sup>	Output low level voltage for an I/O pin,when 8 pins absorb current	$I_{IO} = +6mA$ $2V < V_{DD} < 2.7V$		TBD	
V <sub>OH</sub> <sup>(2)(3)</sup>	Output high level voltage for an I/O pin,when 8 pins output current	$I_{IO} = +6mA$ $2V < V_{DD} < 2.7V$	TBD		

- 1. The current absorbed by the chip  $I_{IO}$  must always follow the absolute maximum ratings given in the table, and the sum of  $I_{IO}$  (all I/O feet and control pins) must not exceed  $I_{VSS}$ .
- 2. The current output  $I_{IO}$  of the chip must always follow the absolute maximum rating given in the table, and the sum of  $I_{IO}$  (all I/O pins and control pins) must not exceed  $I_{VDD}$ .
- 3. Data based on characterization results. Not tested in production.

#### Input/output AC characteristics

The definitions and values of the input and output AC characteristics are given in figure 17 and Table 39, respectively.

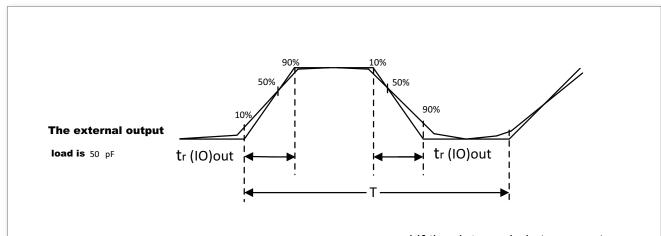
Unless otherwise stated, the parameters listed in Table 39 are measured using the ambient temperature and supply voltage in accordance with the condition Table 14.

Table 39. I/O AC characteristics<sup>(1)</sup>

OSPEEDRy [1:0] value (1)	Symbol	Parameter	er Conditions		Max	Unit
01	f <sub>max(IO)out</sub>	Maximum	$C_L = 50pF,$		10	MHz
	1	frequency <sup>(2)</sup>	V <sub>DD</sub> = 2V~3.6V		25 <sup>(3)</sup>	
11 (50MHz)  S  (1:0] value (1)  fmax  fmax  fmax  fmax  fmax  fmax  fmax	t <sub>f(IO)out</sub>	Output fall time	C <sub>L</sub> = 50pF,		25 <sup>(3)</sup>	nS
	t <sub>r(IO)out</sub>	Output rise time	V <sub>DD</sub> = 2V ~ 3.6V		25(0)	
10	f <sub>max(IO)out</sub>	Maximum frequency <sup>(2)</sup>	$C_L$ = 50pF, $V_{DD}$ = 2V ~ 3.6V		20	MHz
(20MHz)	t <sub>f(IO)out</sub>	Output fall time	C <sub>L</sub> = 50pF,		125 <sup>(3)</sup>	
	t <sub>r(IO)out</sub>	Output rise time	V <sub>DD</sub> = 2V ∼ 3.6V		125 <sup>(3)</sup>	——nS
			$C_L$ = 30pF, $V_{DD}$ = 2.7V ~ 3.6V		50	110
	f <sub>max(IO)out</sub> Maximum frequency <sup>(2)</sup>	$C_L$ = 50pF, $V_{DD}$ = 2.7V ~ 3.6V		30		
			$C_L$ = 50pF, $V_{DD}$ = 2V ~ 2.7V		20	
			$C_L$ = 30pF, $V_{DD}$ = 2.7V $\sim$ 3.6V		5	
	$\mathbf{t}_{f(IO)out}$	Output fall time	$C_L$ = 50pF, $V_{DD}$ = 2.7V ~ 3.6V		8	
			$C_L$ = 50pF, $V_{DD}$ = 2V ~ 2.7V		12	nS
			$C_L$ = 30pF, $V_{DD}$ = 2.7V $\sim$ 3.6V		5	
	$t_{r(IO)out}$	Output rise time	$C_L = 50pF,$ $V_{DD} = 2.7V \sim 3.6V$		8	

OSPEEDRy [1:0] value (1)	Symbol	Parameter	Conditions	Min	Max	Unit
			$C_L$ = 50pF, $V_{DD}$ = 2V ~ 2.7V		12	
	t <sub>EXTIPW</sub>	Pulse width of external signals detected by the EXTI controller		10		nS

- 1. The speed of the I/O port can be configured via MODEx[1:0]. See the description of the GPIO Port Configuration Register in this chip reference manual.
- 2. The maximum frequency is defined in figure 17.



Maximum frequency is achieved if ( $(t_r + t_f) \le 2/3$ )T, and if the duty cycle is (45 ~ 55%) when loaded by  $C_L$ (see the i/O AC characteristics definition)

86830

Figure 17. I/O AC characteristics

#### 7.6.13 NRST pin characteristics

The NRST pin input driver uses the CMOS technology. It is connected to a permanent pullup resistor,  $R_{\text{PU}}$ .

Unless otherwise stated, the parameters listed in the table below are measured using the ambient temperature and V<sub>DD</sub> supply voltage in accordance with the condition Table 14.

Table 40. NRST pin characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
$V_{IL(NRST)}^{(1)}$	NRST input low level voltage		-0.5		0.8	V
V <sub>IH(NRST)</sub> <sup>(1)</sup>	NRST input high level voltage		2		$V_{DD}$	V

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V	NRST Schmitt trigger voltage			0.21/		V
$V_{hys(NRST)}$	hysteresis			0.2V <sub>DD</sub>		V
	Weak pull-up equivalent	\/ - \/		15		kΩ
$R_{PU}$	resistor <sup>(2)</sup>	$V_{IN} = V_{SS}$		15		K77
$V_{F(NRST)}^{(1)}$	V <sub>F(NRST)</sub> <sup>(1)</sup> NRST input filtered pulse				100	ns
$V_{NF(NRST)}{}^{(1)}$	NRST input not filtered pulse		300			

- 1. Data based on design simulation only. Not tested in production.
- 2. The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is minimal (10% order).

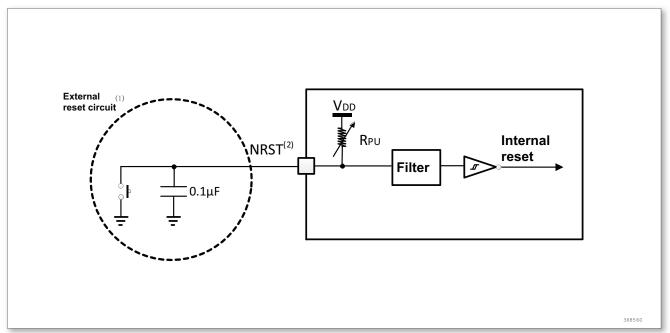


Figure 18. Recommended NRST pin protection

- 1. The reset network is to prevent parasitic reset
- 2. The user must ensure that the potential of the NRST pin is below the maximum  $V_{\text{IL}(\text{NRST})}$  listed in Table 40, otherwise the MCU cannot be reset.

#### 7.6.14 Timer characteristics

The parameters given in the following tables are guaranteed by design.

For details on the characteristics of the I/O multiplexing function pins (output compare, input capture, external clock, PWM output), see subsubsec 7.6.12.

Table 41. TIMx<sup>(1)</sup> characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
t	Timer resolution time		1		t <sub>TIMxCLK</sub>
ι <sub>res(TIM)</sub>	Timer resolution time				

Symbol	Parameter	Conditions	Min	Max	Unit
		f <sub>TIMxCLK</sub> =	10.4		nS
		96MHz	10.4		110
$f_{EXT}$	Timer external clock		0	f <sub>TIMxCLK</sub> /2	MHz
IEXT	frequency on CH1 to CH4	$f_{TIMxCLK} =$	0	48	IVII IZ
		96MHz		40	
Res <sub>TIM</sub>	Timer resolution			16	Bit
4	16-bit timer		1	65536	t <sub>TIMxCLK</sub>
t <sub>COUNTER</sub>	maximum period	$f_{TIMxCLK} =$	0.0104	682	μS
		96MHz	0.0104	002	μΟ
+	The maximum possible count			65536 × 65536	t <sub>TIMxCLK</sub>
t <sub>MAX_COUNT</sub>	The maximum possible count	f <sub>TIMxCLK</sub> =		44.7	S
		96MHz		44.7	3

1. TIMx is a generic name, representing TIM4.

#### 7.6.15 Communication interfaces

#### **I2C** interface characteristics

Unless otherwise specified, the parameters given in Table 42 are derived from tests performed under the ambient temperature, f<sub>PCLK1</sub>frequency and supply voltage conditions summarized in Table 17: General operating conditions.

The I2C interface conforms to the standard I2C communication protocol, but has the following limitations: SDA and SCL are not true pins. When configured as open-drain output, the PMOS transistor between the pin and  $V_{DD}$  Was closed but still exists.

The I2C I/Os characteristics are listed in Table 42, the alternate function characteristics of I/Os (SDA and SCL) refer to subsubsec 7.6.12.

Table 42. I2C characteristics

0	Danamatan	Standa	rd I2C <sup>(1)</sup>	Fast I2	<b>C</b> (1)(2)	11:4
Symbol	Parameter	Min	Max	Min	Max	Unit
t <sub>w(SCLL)</sub>	SCL clock fall time	4.7		1.3		μS
t <sub>w(SCLH)</sub>	SCL clock rise time	4.0		0.6		μS
t <sub>su(SDA)</sub>	SDA setup time	250		100		
t <sub>h(SDA)</sub>	SDA data hold time	0(3)		0 <sup>(4)</sup>	900(3)	ns
$t_{r(SDA)} t_{r(SDL)}$	SDA and SCL rise time		1000	2.0+0.1C <sub>b</sub>	300	115
$t_{\text{f(SDA)}} \; t_{\text{f(SDL)}}$	SDA and SCL fall time		300		300	
t <sub>h(STA)</sub>	Start condition hold time	4.0		0.6		
t <sub>su(STA)</sub>	Start condition setup time	4.7		0.6		1
t <sub>su(STO)</sub>	Stop condition setup time	4.0		0.6		μS

Symbol	Doromotor	Standard I2C(1)		Fast I2C (1)(2)		Unit	
	Parameter	Min	Max	Min	Max	Oilit	
•	Time from Stop condition to	4.7		1.3			
Ūw(STO:STA)	Start condition	4.7		1.3			
C <sub>b</sub>	Capacitive load of each bus		400		400	pF	

- 1. Guaranteed by design, not tested in production.
- 2. f<sub>PCLK1</sub>must be at least 3MHz to achieve standard mode I2C frequencies. It must be at least 12MHz to achieve fast mode I2C frequencies.
- 3. The maximum Data hold time has only to be met if the interface does not stretch the low period of SCL signal.
- 4. In order to span the undefined area of the falling edge of SCL, it must ensure that the SDA signal has a hold time of at least 300nS.

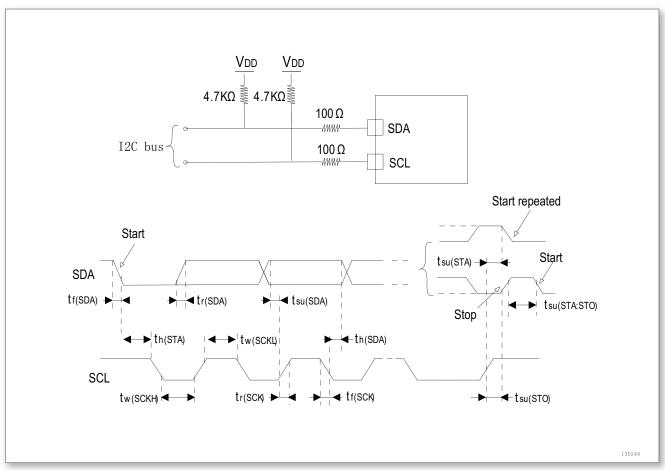


Figure 19. I2C bus AC waveform and measurement circuit<sup>(1)</sup>

1. Measurement point is set to the CMOS level:  $0.3V_{DD}$  and  $0.7V_{DD}$ .

#### **SPI** characteristics

Unless otherwise specified, the parameters given in Table 43 are derived from tests performed under the ambient temperature,  $f_{PCLKx}$  frequency and  $V_{DD}$  supply voltage conditions summarized in Table 17.

Refer to subsubsec 7.6.12 for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO).

Table 43. SPI characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Max	Unit
f 1/4	CDI plack fraguancy	Master mode	0	36	MHz
$f_{SCK}1/t_{c(SCK)}$	SPI clock frequency	Slave mode	0	18	IVITIZ
$t_{r(SCK)}$	SPI clock rise and fall	Load capacitance: C = 30pF		8	
$t_{\text{f}(\text{SCK})}$	time	Load capacitance. C – 30pr		0	
$t_{su(NSS)}^{(2)}$	NSS setup time	Slave mode	4t <sub>PCLK</sub>		
$t_{h(NSS)}^{(2)}$	NSS hold time	Slave mode	73		
$t_{\text{w(SCKH)}}^{(2)}$	SCK high and low time	Master mode, f <sub>PCLK</sub> = 36MHz,	50	60	ns
$t_{w(\text{SCKL})}{}^{(2)}$	SCK flight and low time	prescale coefficient = 4	50	00	
<b>4</b> (2)	Data input setup time,	SPI1	1		
$t_{su(MI)}^{(2)}$	Master mode	SPI1	1		
+ (2)	Data input setup time,		1		
$t_{su(SI)}^{(2)}$	Slave mode		ı		

Symbol	Parameter	Conditions	Min	Max	Unit
+ (2)	Data input hold time,	SPI1	1		
$t_{h(MI)}^{(2)}$	Master mode	JPII	ı		
t <sub>h(SI)</sub> (2)	Data input hold time,		3		
اh(SI) ک	Slave mode		3		
		Slave mode, f <sub>PCLK</sub> = 36MHz,	0	55	
$t_{a(SO)}^{(2)(3)}$	Data output access time	prescale coefficient = 4	U	33	
		Slave mode, f <sub>PCLK</sub> = 24MHz		4t <sub>PCLK</sub>	ns
t <sub>dis(SO)</sub> (2)(4)	Data output disable time	Slave mode	10		
$t_{v(SO)}^{(2)(1)(1$	Data output valid time	Slave mode (after enable edge)		25	
t <sub>v(MO)</sub> (2)(1)	Data output valid time	Master mode (after enable		3	
v(MO) ` ′ ′ ′	Data output valid time	edge)		3	
t <sub>h(SO)</sub> (2)	Data output hold time	Slave mode (after enable edge)			
<b>4</b> (2)	Data output hold time	Master mode (after enable	4		
t <sub>h(MO)</sub> (2)		edge)	4		

- 1. Remapping SPI1 characteristics needs to be further determined.
- 2. Data based on characterization results. Not tested in production.
- 3. Min time is for the minimum time to drive the output and the max time is for the maximum time to validate the data.
- 4. Min time is for the minimum time to invalidate the output and the max time is for the maximum time to put the data in Hi-Z.

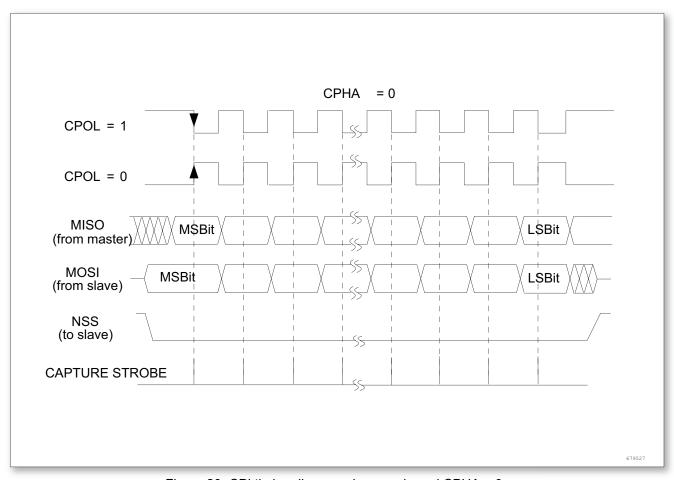


Figure 20. SPI timing diagram-slave mode and CPHA = 0

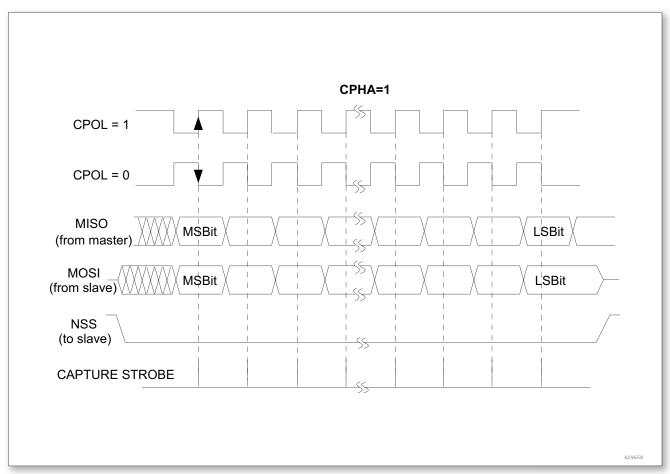


Figure 21. SPI timing diagram-slave mode and CPHA =  $\mathbf{1}^{(1)}$ 

1. Measurement points are done at CMOS levels:  $0.3\mbox{V}_{\mbox{\scriptsize DD}}$  and  $0.7\mbox{V}_{\mbox{\scriptsize DD}}.$ 

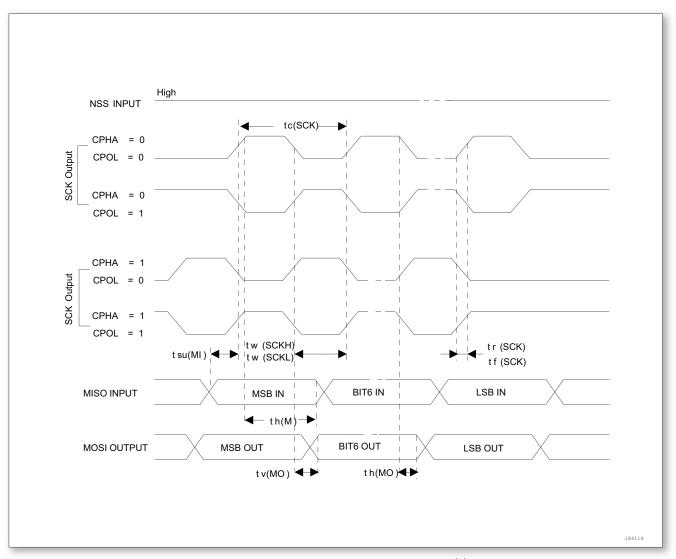


Figure 22. SPI timing diagram-master mode<sup>(1)</sup>

1. Measurement points are done at CMOS levels:  $0.3V_{DD}$  and  $0.7V_{DD}$ .

#### **USB** characteristics

Table 44. USB startup time

Symbol	Parameter	Max	Unit
t <sub>START</sub> (1)	USB transceiver startup time	1	μ\$

1. Guaranteed by design. Not tested in production.

Table 45. USB DC electrical characteristics

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Max <sup>(1)</sup>	Unit
		Input levels			
$V_{DD}$	USB operating voltage <sup>(2)</sup>		3.0(3)	3.6	
$V_{DI}^{(4)}$	Differential input sensitivity	I(USBDP, USBDM)	0.2		V

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Max <sup>(1)</sup>	Unit
V <sub>CM</sub> <sup>(4)</sup>	Differential common	Includes V <sub>DI</sub> range	0.8 2.5		
▼ CM · ·	mode range	molades v <sub>D</sub> range	0.0	2.0	
V <sub>SE</sub> <sup>(4)</sup>	Single ended receiver		1.3	2	
V SE V	threshold		1.3	2	
		Output levels			
$V_{OL}$	Static output level low	$R_L$ of $1.5 k\Omega$ to $3.6 V^{(5)}$		0.3	V
V <sub>OH</sub>	Static output level high	$R_L$ of 15k $\Omega$ to $V_{SS}$ $^{(5)}$	2.8	3.6	<b>V</b>

- 1. All the voltages are measured from the local ground potential.
- 2. To be compliant with the USB 2.0 full-speed electrical specification, USBDP (D +) pin has a built-in  $1.5k\Omega$  resistor connected to the  $V_{DD}$ , no need to external connect.
- 3. The USB functionality is ensured down to 2.7V but not the full USB electrical characteristics which are degraded in the 2.7V  $\sim$  3.6V  $V_{DD}$  voltage range.
- 4. Guaranteed by design. Not tested in production.
- 5.  $R_L$  is the load connected on the USB drivers

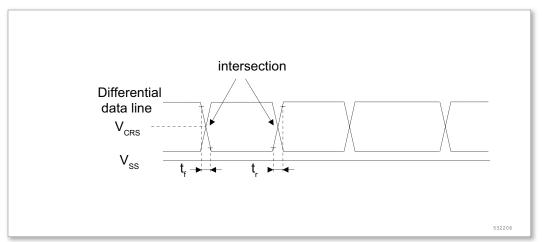


Figure 23. USB timing diagram: definition of data signal rise and fall time

Table 46. USB Full-speed electrical characteristics<sup>(1)</sup>

Symbol	Parameter	Conditions	Min	Max	Unit
t <sub>r</sub>	Rise time <sup>(2)</sup>	C <sub>L</sub> <= 50pF	7.041	23.13	ns
$t_f$	Fall time <sup>(2)</sup>	C <sub>L</sub> <= 50pF	6.866	26.76	ns
t <sub>rfm</sub>	Rise/fall time matching	t <sub>r</sub> /t <sub>f</sub>	96.52	125.1	%
$V_{CRS}$	Output signal crossover		1.391	2.967	V
V CRS	voltage		1.591	2.307	V

- 1. Guaranteed by design. Not tested in production.
- 2. Measured from 10% to 90% of the data signal. For more detailed information, please refer to USB Specification Section 7 (version 2.0).

#### 7.6.16 CAN (controller area network) interface

Refer to subsubsec 7.6.12 for more details on the input/output alternate function characteristics (CAN\_TX and CAN\_RX).

#### 7.6.17 12-bit ADC characteristics

Unless otherwise specified, The parameters in the table below are measured using the ambient temperature,  $f_{PCLK2}$  frequency and  $V_{DDA}$  supply voltage in accordance with the conditions of Table 17.

Note: It is recommended to perform a calibration after each power-up

Table 47. ADC characteristics

Symbol	Parameter	Conditions	Min	Туре	Max	Unit	
$V_{DDA}$	Supply voltage		2.5	5	5.5	V	
$V_{REF+}$	Positive reference			$V_{DDA}$		V	
	voltage						
$f_{ADC}^{(1)(3)}$	ADC clock				15	MHz	
	frequency						
<b>f</b> S <sup>(1)(3)</sup>	Sampling rate				1	MHz	
$f_{TRIG}^{}^{(1)}$	External trigger	$f_{ADC} = 15MHz$			833	KHz	
TRIG	frequency				18	1/f <sub>ADC</sub>	
$V_{AIN}^{(2)}$	Conversion voltage		0 (V <sub>SSA</sub> or				
			V <sub>REF</sub> -connected		$V_{REF^+}$	V	
	range		to ground)				
R <sub>AIN</sub> <sup>(1)</sup>	External sample		See For	kΩ			
KAIN	and hold capactor		366 1 011	K3Z			
R <sub>ADC</sub> <sup>(1)</sup>	Sampling switch			0.75			
NADC Y	resistance				0.75	kΩ	
$C_{ADC}^{(1)}$	Internal sample and			10		, F	
CADC	hold capacitor			10		pF	
ts <sup>(1)</sup>	Sampling time	$f_{ADC} = 15MHz$	0.1		16	μ\$	
us` /	Sampling time		1.5		239.5	1/f <sub>ADC</sub>	
t <sub>STAB</sub> (1)	Stabilization time			1		μ\$	
	Total conversion	f <sub>ADC</sub> = 15MHz	1		17.44	μ\$	
$t_{conv}^{(1)}$	time (including		15 ~ 253 (	sampling t <sub>S+</sub> ) s	tepwise	1 /F	
	Sampling time)		арр	proximation 13.5	5	1/f <sub>ADC</sub>	

- 1. Guaranteed by design. Not tested in production.
- 2. In this series of products,  $V_{REF+}$  is internally connected to  $_{DDA}$ ,  $V_{REF-}$  is internally connected to  $_{SSA}$ .
- 3.  $f_{ADC}$  Maximum support 15MHz,  $f_{S}$  Maximum support 1MHz ( $f_{PCLK2}$  = 60MHz, ADC Prescaler = 4,  $f_{ADC}$  = 15MHz, TS = 1.5)

#### Formula 1: Maximum RAIN Formula

$$R_{AIN} < \frac{T_S}{f_{ADC} \times C_{ADC} \times (N+3) \times ln(2)} - R_{ADC}$$

The above formula (Equation 1) is used to determine the maximum external impedance so that the error can be less than 1/4 LSB. Where N = 12 (representing 12-bit resolution) . Ln(2) = 0.69314718.

Table 48. Maximum  $R_{AIN}$  at  $f_{ADC} = 15MHz^{(1)}$ 

T <sub>S</sub> (cycles)	<b>t</b> <sub>S</sub> (μ <b>s</b> )	$\mathbf{R}_{AIN}$ max ( $k\Omega$ )
1.5	0.1	0.2
7.5	0.5	4.1
13.5	0.9	7.9
28.5	1.9	17.5
41.5	2.8	25.9
55.5	3.7	34.8
71.5	4.8	NA
239.5	16.0	NA

1. Guaranteed by design. Not tested in production.

Table 49. ADC Accuracy - Limit Test Conditions(1)(2)

Symbol	Parameter	Test Conditions	Туре	Max	Unit
ET	Comprehensive error		8	10	
EO	Offset error	$f_{PCLK2} = 60MHz, f_{ADC} =$	3	3	
EG	Gain error	15MHz, $R_{AIN}$ < 10K $\Omega$ , $V_{DDA}$	1	1	LSB
ED	Differential linearity error	= 5V,T <sub>A</sub> = 25°C	6.5	7	
EL	Integral linearity error		8	8	

1. ADC Accuracy vs. Negative Injection Current: Injecting negative current on any of the standard (non-robust) analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to standard analog pins which may potentially inject negative current.

Any positive injection current within the limits specified for  $I_{INJ(PIN)}$  and  $\Sigma I_{INJ(PIN)}$  in subsubsec 7.6.13 does not affect the ADC accuracy.

2. Guaranteed based on test during characterization. Not tested in production.

ET = Total unadjusted error: The maximum deviation between the actual and ideal transmission curves.

EO = Offset error: The deviation between the first actual conversion and the first ideal conversion.

EG = Gain error: The deviation between the last ideal transition and the last actual transition.

ED = Differential linearity error: The maximum deviation between the actual step and the ideal value.

EL = Integral linearity error: The maximum deviation between any actual conversion and the associated line of the endpoint.

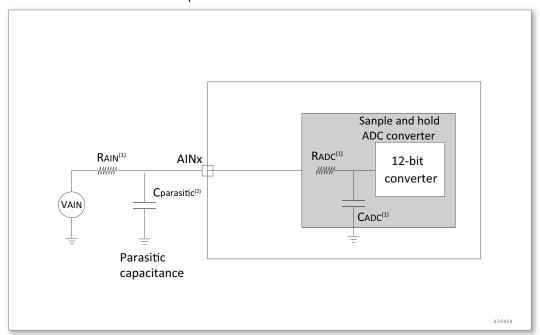


Figure 24. Typical connection diagram using the ADC

- 1. See Table 49 for the values of  $R_{\text{AIN}}$ ,  $R_{\text{ADC}}$  and  $C_{\text{ADC}}$ .
- 2. C<sub>parasitic</sub> represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 7pF). A high C<sub>parasitic</sub> value will downgrade conversion accuracy. To remedy this, f<sub>ADC</sub> should be reduced.

#### PCB design recommendations

The power supply must be connected as shown below. The 10nFcapacitor in the figure must be a ceramic capacitor (good quality), and they should be as close as possible to the MCU chip.

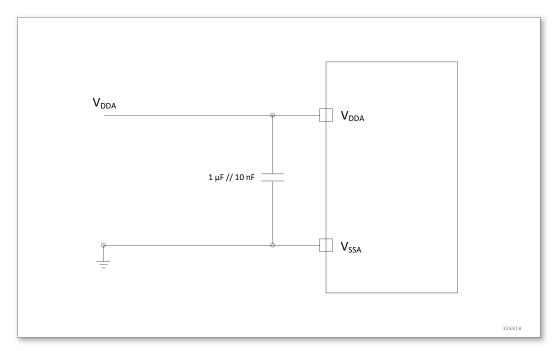


Figure 25. Power supply and reference power supply decoupling circuit

#### 7.6.18 Temperature sensor characteristics

Table 50. Temperature sensor characteristics  $^{(3)}(4)$ 

Symbol	Parameter	Min	Туре	Max	Unit
T <sub>L</sub> <sup>(1)</sup>	V <sub>SENSE</sub> linearity with respect to				0.0
	temperature		±5		°C
Avg_Slope <sup>(1)</sup>	Average slope	4.571	4.801	5.984	mV/°C
$V_{25}^{(1)}$	Voltage at 25°C	1.433	1.451	1.467	V
t <sub>start</sub> (2)	Setup time			10	μS
T <sub>S_temp</sub> (2)	ADC sampling time when	40			0
	reading temperature	10	10		μ\$

- 1. Guaranteed based on test during characterization. Not tested in production.
- 2. Guaranteed by design. Not tested in production.
- 3. The shortest Sampling time can be determined by the application through multiple iterations.
- 4.  $V_{DD} = 3.3V$ .

#### 7.6.19 Comparator characteristics

Table 51. Comparator characteristics

Symbol	Parameter	Register configuration	Min	Туре	Max	Unit
HYST	Hysteresis	00		0		mV
HYST	Hysteresis	01		15		mV
HYST	Hysteresis	10		30		mV

Symbol	Parameter	Register configuration	Min	Туре	Max	Unit
HYST	Hysteresis	11		90		mV
OFFSET	Offset voltage	00	0.091	0.213	0.358	mV
OFFSET	Offset voltage	01	3.23	7.51	12.08	mV
OFFSET	Offset voltage	10	9.79	15	20.8	mV
OFFSET	Offset voltage	11	34.25	47.4	62.22	mV
DELAY <sup>(1)</sup>	Propagation delay	00		80		nS
DELAY <sup>(1)</sup>	Propagation delay	01		51		nS
DELAY <sup>(1)</sup>	Propagation delay	10		26		nS
DELAY <sup>(1)</sup>	Propagation delay	11		9		nS
$I_q^{(2)}$	Operating current mean	00		4.5		uA
$I_q^{(2)}$	Operating current mean	01		4.4		uA
$I_q^{(2)}$	Operating current mean	10		4.4		uA
$I_q^{(2)}$	Operating current mean	11		4.4		uA

- 1. The output flips 50% of the time and the time difference between the input and the flip.
- 2. Total current consumption, operating current.

# 8

### PCB design recommendations

PCB design recommendations

#### 8.1 Power supply design recommendations

The power supply must be connected as shown below. The 10nF capacitors in the figure must be ceramic dielectric capacitors (good quality) and they should be as close as possible to the MCU chip.

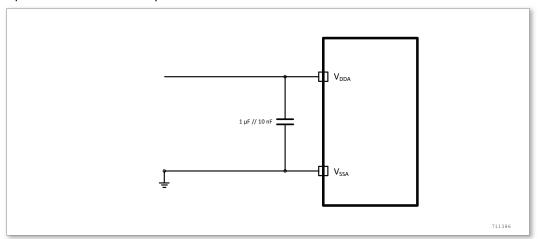


Figure 26. Power supply and reference power supply

#### 8.2 PCB considerations

Bluetooth works in the 2.4G wireless frequency band, and should avoid the influence of various factors on the wireless transceiver. Pay attention to the following points:

- The product enclosure surrounding the Bluetooth module avoids the use of metal. When
  using a part of the metal casing, try to keep the module antenna part away from the metal
  part.
- The internal metal wire or metal screw of the product should be as far as possible from the antenna part of the module.
- The antenna part of the module should be placed around the carrier PCB. It is not
  allowed to be placed in the board, and the carrier board under the antenna is slotted.
   In the direction parallel to the antenna, copper or wiring is not allowed. It is also a good
  choice to directly expose the antenna part directly to the carrier board.
- Place a large GND under the module as far as possible, and extend the trace as far as possible to the outside.
- · It is recommended to use insulating materials for isolation at the module mounting posi-

tion on the substrate, for example by placing a single screen print (TopOverLay) at this position.

- The wiring of the power supply line and ground line is directly related to the performance
  of the product, and the noise interference is minimized. When wiring, try to widen the
  ground wire, power cable width, ground wire > power wire > signal wire, usually the
  signal line width is 0.2 0.3mm, the power line width is 1.2 2.5mm, and the large-area
  copper layer is used for grounding. The unused space on the PCB is paved.
- Power supply plus two decoupling filter capacitors: If using LDO power supply, the values are 1uF and 0.1uF respectively for filtering; if using button battery power, the values are 10uF and 10uF respectively for voltage regulation.
- The trace between the chip ANT and the antenna should not be too long. The line width should consider the impedance matching requirements.

#### 8.3 2.4G RF antenna design

Small antenna sizes can cause large changes due to performance. Therefore, it is highly recommended to make an accurate reference design for optimum performance. When drawing a PCB antenna, draw the antenna with reference to the dimensions given in the figure below.

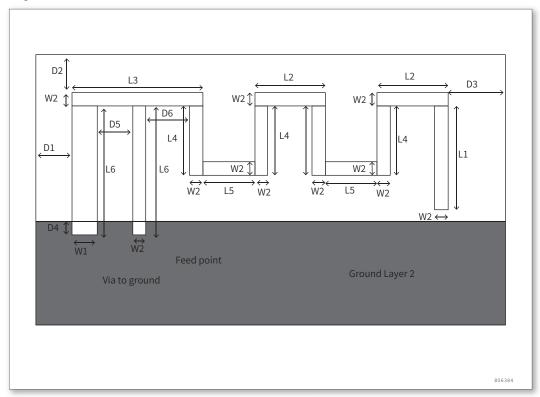


Figure 27. Antenna size

Table 52. Antenna size

No.	Typical value(mm)	
L1	3.94	
L2	2.70	

No.	Typical value(mm)
L3	5.00
L4	2.64
L5	2.00
L6	4.90
W1	0.90
W2	0.50
D1	0.50
D2	0.30
D3	0.30
D4	0.50
D5	1.40
D6	1.70

9

## **Package information**

Package information

#### 9.1 LQFP48 Package information

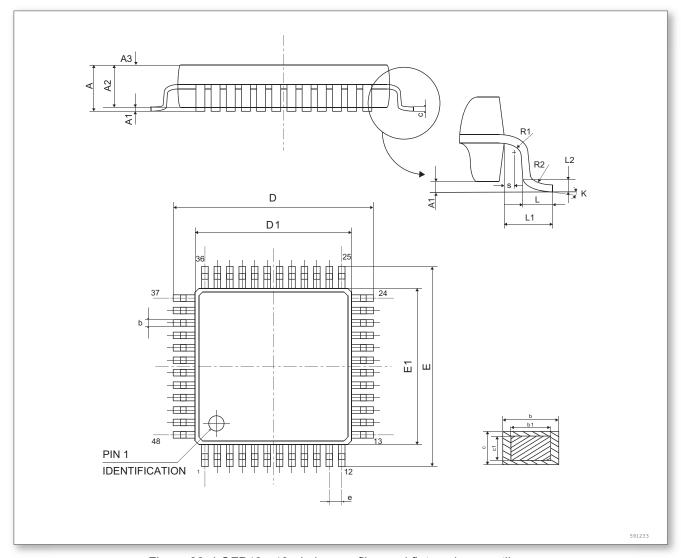


Figure 28. LQFP48 - 48-pin low-profile quad flat package outline

- 1. Drawing is not to scale.
- 2. Dimensions are expressed in millimeters.

Table 53. LQFP48 mechanical data

0	Millimeters		
Symbol	Min	Тур	Max
A			1.60
A1	0.05		0.15
A2	1.35	1.40	1.45
A3	0.59	0.64	0.69
b	0.18		0.27
b1	0.17	0.20	0.23
С	0.13		0.18
c1	0.12	0.127	0.134
D	8.80	9.00	9.20
D1	6.90	7.00	7.10
E	8.80	9.00	9.20
E1	6.90	7.00	7.10
е		0.50	
L	0.45	0.60	0.75
L1		1.00REF	
L2		0.25BSC	
R1	0.08		
R2	0.08		0.20
S	0.20		
N		Number of pins = 48	

#### 9.2 QFN32 Package information

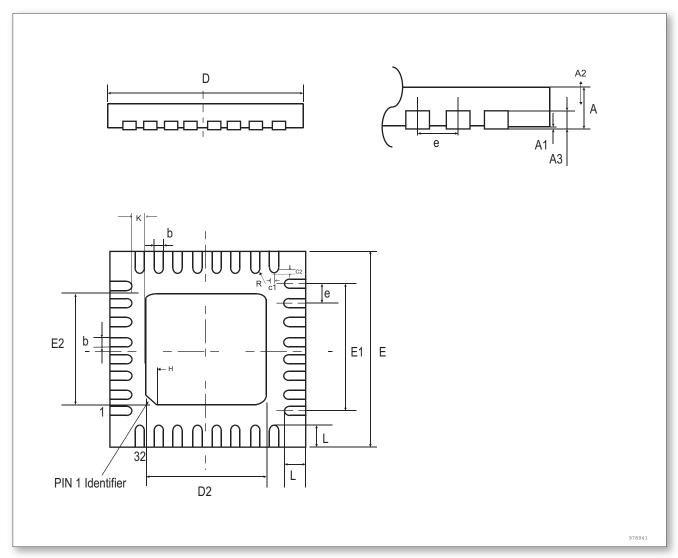


Figure 29. QFN32 - 32-pin quad flat no-leads package outline

- 1. Drawing is not to scale.
- 2. Dimensions are expressed in millimeters.

Table 54. QFN32 mechanical data

Symbol		Millimeters			
	Min	Тур	Max		
А	0.7	0.75	0.80		
A1	0.00	0.02	0.05		
A2	0.50	0.55	0.60		
A3		0.20REF			
b	0.20	0.25	0.30		
D	4.90	5.00	5.10		
Е	4.90	5.00	5.10		
D2	3.40	3.50	3.60		

Comple al	Millimeters				
Symbol	Min	Тур	Max		
E2	3.40	3.50	3.60		
е		0.5			
Н	0.30REF				
K		0.35REF			
L	0.35	0.40	0.45		
R	0.09				
c1		0.08			
c2		0.08			
N	Number of pins = 32				

# 10 Ordering information

Ordering information

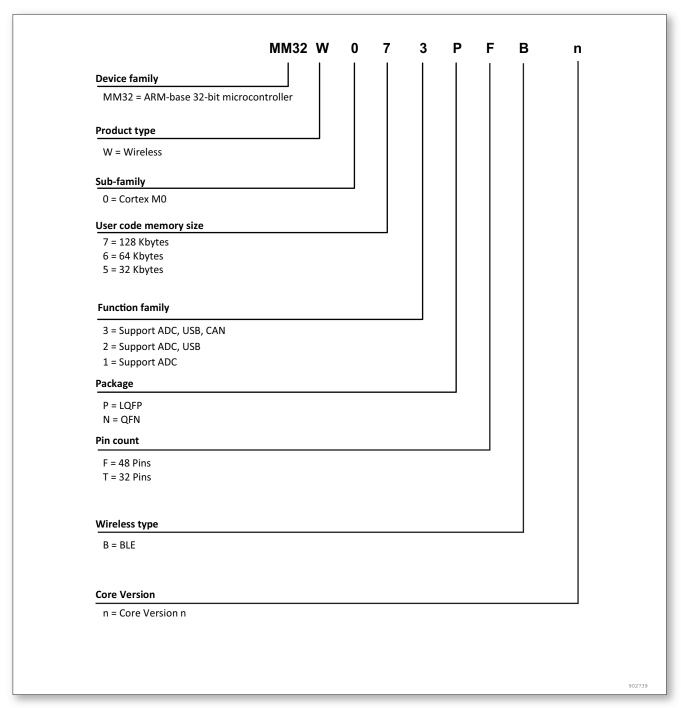


Figure 30. Ordering information scheme

# 1 Revision history Revision history

Table 55. Document revision history

Revision	Changes	Date
Rev1.82	Modify the package parameters.	2019/3/11
Rev1.81	Modify ADC electrical parameters.	2019/1/7
Rev1.8	Change parameters	2018/12/3
	Revise the capacitance value of the matching	
David 7	crystal in the application circuit diagram and the	2040/2/22
Rev1.7	capacitance value in the reset circuit, and the	2018/2/23
	power consumption parameters of Table 10.	
Rev1.2	Modify the typical application circuit diagram.	2017/6/21
Rev1.1	Modify the working voltage and clock map.	2017/6/13
Rev1.0	Initial release.	2017/5/31