

# MIS Homework 4

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## Problem A5.5

Solution:

**a**

```
const_basis = fourier_basis_const(n)
# Check the norm of the constant basis vector
norm(const_basis)
# Check the inner product between the constant basis vector
dot(const_basis, const_basis)
sin_basis = fourier_basis_sin(1, n)
# Check the norm of the sine basis vector
norm(sin_basis)
# Check the inner product of the constant basis vector with the sine basis
vector
dot(const_basis, sin_basis)
# Check the inner product between the sine basis vector
dot(sin_basis, sin_basis)
# Check the norm of the cosine basis vector
cos_basis = fourier_basis_cos(1, n)
norm(cos_basis)
# Check the inner product between the cosine basis vector
dot(cos_basis, cos_basis)
# Check the inner product of the sine basis vector with the cosine basis
vector
dot(sin_basis, cos_basis)
```

**b**

Using Plots

```
n = 50
d = fourier_basis_const(n)
s1 = fourier_basis_sin(1, n)
c1 = fourier_basis_cos(1, n)
```

```

s2 = fourier_basis_sin(2,n)
c2 = fourier_basis_cos(2,n)
a = fourier_basis_alt(n)
plot(1 : n, d, label = "d")
plot!(1 : n, s1, label = "s1")
plot!(1 : n, c1, label = "c1")
plot!(1 : n, s2, label = "s2")
plot!(1 : n, c2, label = "c2")
plot!(1 : n, a, label = "a")

```

**c**

n=6

```

coefficient_d = dot(d, x)
coefficient_s1 = dot(s1, x)
coefficient_c1 = dot(c1, x)
coefficient_s2 = dot(s2, x)
coefficient_c2 = dot(c2, x)
coefficient_a = dot(a, x)
x' = coefficient_d*d + coefficient_s1*s1 + coefficient_c1*c1 + coefficient_c2*
c2 + coefficient_s2*s2 + coefficient_a*a
Then, we can have that x'=[1,2,3,4,5,6].

```

## Problem 6.18

Solution:

6.18

An  $m \times n$  Vandermonde matrix are linearly dependent when an  $n$ -vector  $C$  exists,

$$V_C = \sum_{i=1}^n C_i V_i = 0, \text{ all entries of } C$$

are not 0. The entries of  $V_C$  are

$$p(t) = C_1 + C_2 t + \dots + C_n t^{n-1} \text{ when } t_1, \dots, t_m.$$

Using the fact from hint, when  $V_C = 0$ , then  $C = 0$ , so the columns of  $V$  are linearly independent.

## Problem 6.22

Solution:

**a**

Adding A and B costs  $mn$  flops, adding  $x$  and  $y$  costs  $n$  flops, and the matrix-vector multiplying costs  $2mn$  flops, so the total is  $3mn+n$  flops and the approximate flop is  $3mn$  flops.

**b**

Each matrix-vector multiplication costs  $2mn$  flops, so the total is  $8mn$  flops. Three vector additions costs  $3m$  flops, so the total is  $8mn+3m$  flops and the approximate flop is  $8mn$  flops.

**c**

The first method requires fewer flops.

## Problem A6.2

Solution:

```
mis4.jl > ...
1 function Vandermonde(n,t::AbstractArray)
2     m=length(t)
3     tmp=map(tuple,t[i]^(j-1) for i=1:m,j=1:n)
4     return tmp
5 end
6 # Vandermonde(4,rand(1:10,6))
7 println("Vandermonde :",Vandermonde(4,rand(1:10,6)))
```

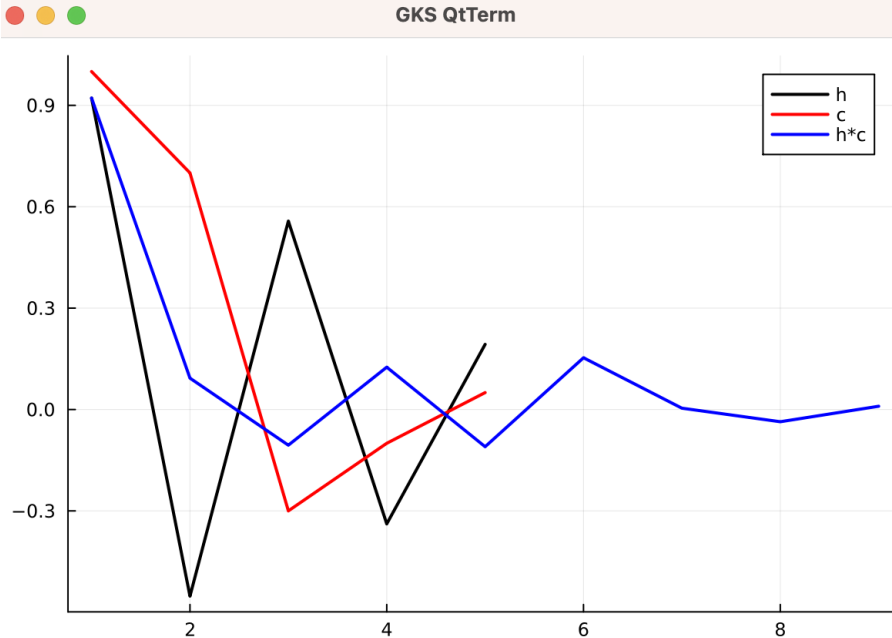
6×4 Matrix{Tuple{Int64}}:

(1,)	(1,)	(1,)	(1,)
(1,)	(3,)	(9,)	(27,)
(1,)	(2,)	(4,)	(8,)
(1,)	(4,)	(16,)	(64,)
(1,)	(4,)	(16,)	(64,)
(1,)	(2,)	(4,)	(8,)

## Problem A7.1

Solution:

```
include("channel_equalization_data.jl")
using Plots
import Plots:plot
a=length(h)
b=length(c)
result=zeros(a+b-1)
for i in 1:(a+b-1)
    for j in 1:a
        if i+1-j>=1 && i+1-j<=b
            result[i]+=h[j]*c[i+1-j]
        end
    end
end
display(result)
y1=plot(1:1:a,h,label="h",color="black",linewidth=2,linestyle=:solid)
y1=plot!(1:1:b,c,label="c",color="red",linewidth=2,linestyle=:solid)
y1=plot!(1:1:a+b-1,result,label="h*c",color="blue",linewidth=2,linestyle=:solid)
```



Chanel  $c$  is the system impulse response, and the plot of  $c$  shows that  $c$  waves are random. However, adding  $h$ , which had been bouncing up and down about line 0, and results in a longer and average serious impulse response,  $h*c$  is more equalized and stabled.

Then we plot  $s$ ,  $y$ , and  $y$ -tilde

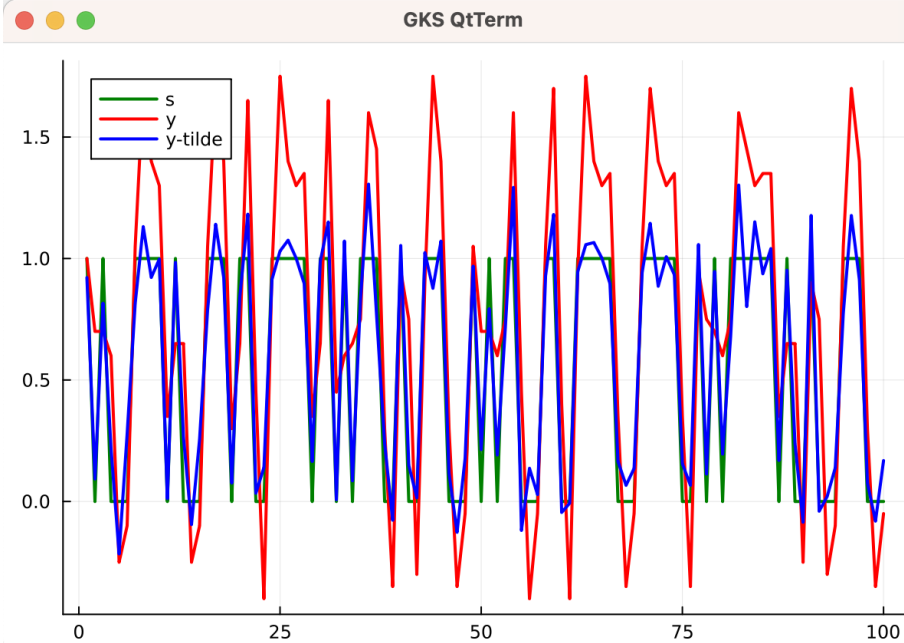
```

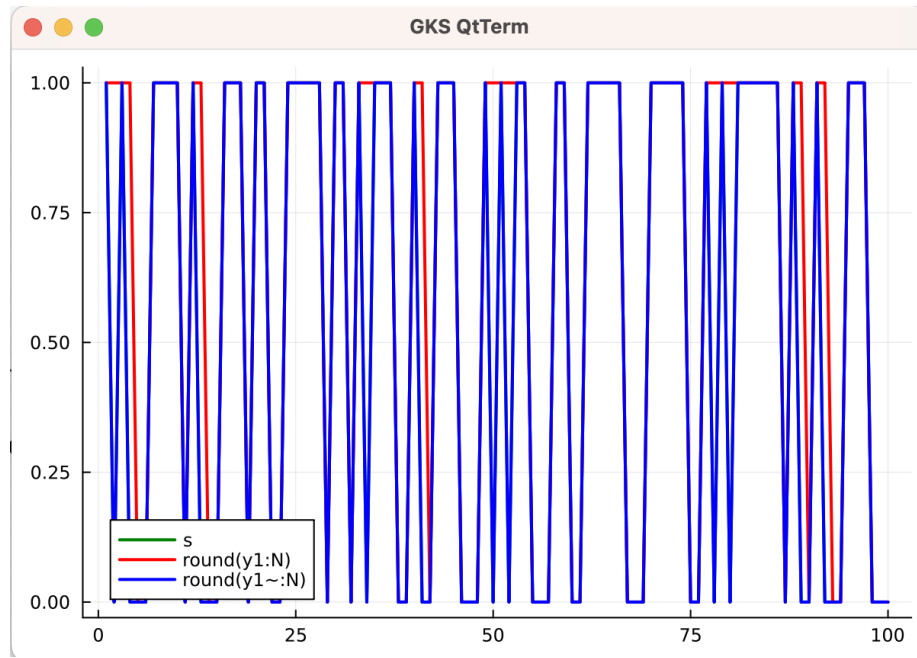
n=length(c)
m=length(result)
l=length(s)
y=zeros(l+n-1)
yt=zeros(m+l-1)
for i in 1:l+n-1
    for j in 1:n
        if i+1-j>=1 && i+1-j<=l
            y[i]+=c[j]*s[i+1-j]
        end
    end
end
for i in 1:l+m-1
    for j in 1:m
        if i+1-j>=1 && i+1-j<=l
            yt[i]+=result[j]*s[i+1-j]
        end
    end
end

y2=plot(1:1:100,s[1:100],label="s",color="green",linewidth=2,linestyle=:solid)
y2=plot!(1:1:100,y[1:100],label="y",color="red",linewidth=2,linestyle=:solid)
y2=plot!(1:1:100,yt[1:100],label="y-tilde",color="blue",linewidth=2,linestyle=:solid)

y=1*(y.>0.5)
yt=1*(yt.>0.5)
y3=plot(1:1:100,s[1:100],label="s",color="green",linewidth=2,linestyle=:solid)
y3=plot!(1:1:100,y[1:100],label="round(y1~N)",color="red",linewidth=2,linestyle=:solid)
y3=plot!(1:1:100,yt[1:100],label="round(y1~N)",color="blue",linewidth=2,linestyle=:solid)

```





Plots  $S = \text{round}(y1 : N)$  and  $S^{eq} = \text{round}(y1 : N)$

It is clear that  $s = \text{round}(y1 : N)$  is worse estimate of  $s^{eq} = \text{round}(y1 : N)$

```

BER1=0
BER2=0
for i in 1:length(s)
    if y[i]!=s[i]
        BER1+=1
    end
end
for i in 1:length(s)
    if yt[i]!=s[i]
        BER2+=1
    end
end
println("BER for s^: ",BER1/length(y))
println("BER for s^eq: ",BER2/length(y))

```

```

julia> println("BER for s^: ",BER1/length(y))
BER for s^: 0.11553784860557768

```

```

[julia> println("BER for s^eq: ",BER2/length(y))
BER for s^eq: 0.0

```

As we can see from the figure, the BER for  $s$  is 0.11553784860557768, and for  $s^{eq}$  is 0.

### Problem A7.3

Solution:

a

```
using WAV
x,f=wavread("audio_filtering_original.wav");
x=vec(x)
wavplay(x,f)

h_smooth = 1 / 44 * ones(44);
output = conv(h_smooth, x);
wavplay(output, f);
```

The music is more peace and softer after convolving by  $h^{smooth}$ .

b

Since  $k$  is the number of samples in 0.25s,  $k=0.25*(441000/10)=11025$ , which means  $h_{echo}=(1,0,\dots,0,0.5)$ .

```
h_echo=zeros(11025)
h_echo[1]=1
h_echo[11025]=0.5
output1=conv(h_echo,x)
wavplay(output1,f)
h_echo2=conv(h_echo,h_echo)
output2=conv(h_echo2,x)
wavplay(output2,f)
```

We could hear two music, the first one is just like the original one, and the second one is weaker than the original music.



### Problem 7.8

Solution:

$$Ax + S = 0$$

$$1^T(Ax + S) = 0$$

$$(1^T A)x + (1^T S) = 0$$

$A$  is the incidence matrix, and  $1^T A$  is the row vector of column sums of  $A$ , so  $1^T A = 0$ .

Thus we can have  $1^T S = 0$ .