

Analysis and Design of High Speed Residue to Binary Reverse Converter for the Moduli Set $\{2n+1, 2n, 2n-1\}$

K. VijayaVardhan^{1,‡}, Gonuguntla Sailakshmi^{1,†},

¹Dept. Electronics and Communication Engineering,
Vignan's Foundation for Science, Technology and Research,
Guntur, Andhra Pradesh-522213, India.

[‡]kvvardhan405@gmail.com, [†]gssrc147@gmail.com,

Sarada Musala^{1,♀}, Avireni Srinivasulu^{2,*}, *SM-IEEE*

²Dept. Electronics and Communication Engineering,
JECRC University,
Jaipur-303905, Rajasthan (State), India.

[♀]sarada.marasu@gmail.com, ^{*}avireni@jecrcu.edu.in

Abstract— Residue Number System (RNS) has the specific feature to perform addition, subtraction independently with carry-free propagation. In RNS, conversion is of two types: Forward Conversion and Reverse Conversion. Chinese Remainder Theorem (CRT) and Mixed Radix Conversion (MRC) are the two extensively used techniques adapted for reverse conversion and in this case, CRT is proved faster than MRC. This paper proposes a residue to binary reverse converter for three moduli set $\{2n+1, 2n, 2n-1\}$ by modifying the CRT. It is simplified in order to design a reverse converter that does not require modular operations. The operations required in the existing designs are $\text{mod}(2n) \text{ mod}(2n-1)$ and $\text{mod}(2n-1)$ operations. The proposed design provides better results in terms of power, area, delay and number of cells over the existing CRT design and does not require mod operations which are however necessary in the existing CRT technique. This design has been simulated using NC Launch - Encounter tool in Cadence.

Keywords - Residue Number System; Moduli Set; Forward Conversion; Reverse Conversion; Mixed Radix Conversion and Chinese Remainder Theorem.

I. INTRODUCTION

A non-weighted number system, Residue Number System provides the single step multiplication, carry-free addition and borrow-free subtraction [1]-[6]. Data conversion and moduli selection are the primary factors that determines the performance of RNS hardware and may limit the use of RNS in DSP applications [7]. RNS is used in digital signal processing applications i.e., digital filtering and convolutions [8].



Figure 1. General structure of RNS processor

The RNS operation is performed in three stages and the structure is shown in Fig. 1 [9]. They are (i). Forward Conversion, (ii). Modulo Channels, (iii). Reverse Conversion. Reverse conversion is of two types: Chinese Remainder Theorem (CRT), Mixed Radix Conversion (MRC). The CRT based conversion is parallel process where as MRC is a

sequential process. MRC is a weighted system, used to implement the magnitude of comparison which is problematic in RNS [7]. In CRT, the limitations are the complex and delayed modulo (mod)-M operations [10]-[19]. The existing designs depend on $\text{mod}(2n+1) \text{ mod}(2n-1)$, $\text{mod}(2n) \text{ mod}(2n-1)$ and $\text{mod}(2n-1)$ operations for the three moduli set [20]-[27]. An efficient reverse converter is proposed in this paper which does not require the mod operations for the moduli set.

The remaining sections of the paper are designed as follows. In section II, the conventional design, Proposed design, Simulated and RTL schematic results are presented in section III and IV respectively. Finally, the conclusion is drawn in section V.

II. CONVENTIONAL DESIGN

The reverse conversion is the process of converting the residue representations to binary/decimal conventional notations.

A. Mixed Radix Conversion

In the mixed radix conversion, each of the weight determines a distinct radix. The radices are m_n, m_{n-1}, \dots, m_1 , residue representation $(r_n, r_{n-1}, r_{n-2}, \dots, r_1)$ and number X can be expressed as [9]

$$X = (Z_n, Z_{n-1}, Z_{n-2}, \dots, Z_1)$$

$$X = Z_n m_{n-1} m_{n-2} \dots m_1 + \dots + Z_3 m_2 m_1 + Z_2 m_1 + Z_1 \quad (1)$$

$$Z_1 = r_1$$

$$Z_2 = \|m_1^{-1} \|_{m_2} (r_2 - z_1) \|_{m_2} \text{ and so on}$$

$$Z_n = \| (m_1 m_2 \dots m_{n-1})^{-1} \|_{m_n} (r_n - (r_{n-1} m_{n-2} \dots Z_2 m_1 + Z_1)) \|_{m_n}$$

B. Chinese Remainder Theorem

In Chinese remainder theorem, $(r_1, r_2, \text{ and } r_3)$ are the residue representations of X with prime moduli m_1, m_2, m_3 and represented as [9]

$$|X|_M = \left| \sum_{i=1}^N r_i \|M_i^{-1} \|_{m_i} M_i \right|_M \quad (2)$$

where M is the dynamic range

$$M = m_1 \cdot m_2 \cdot m_3$$

The simplified expression of equation (2) is

$$|X|_M = \left| \frac{M}{2}(r_1 + r_3) + \frac{m_2 m_3}{2} - m_1 m_3 r_2 + \frac{m_1 m_2}{2} r_3 \right|_M \quad (3)$$

The term $(r_1 + r_3)$ can be either even or odd in the above equation (3).

If $(r_1 + r_3)$ is odd, then

$$\left| \frac{M}{2}(r_1 + r_3) \right|_M = \frac{M}{2}$$

The equation can be rewritten as

$$|X|_M = \left| \frac{M}{2} + \frac{m_2 m_3}{2} - m_1 m_3 r_2 + \frac{m_1 m_2}{2} r_3 \right|_M \quad (4)$$

If $(r_1 + r_3)$ is even, then

$$\left| \frac{M}{2}(r_1 + r_3) \right|_M = 0$$

The equation can be rewritten as

$$|X|_M = \left| \frac{m_2 m_3}{2} - m_1 m_3 r_2 + \frac{m_1 m_2}{2} r_3 \right|_M \quad (5)$$

According to [2], the simplified equations of (4) and (5) are as follows that uses mod $(2n)$ mod $(2n-1)$ operations in the process and the design is shown in Fig. 2.

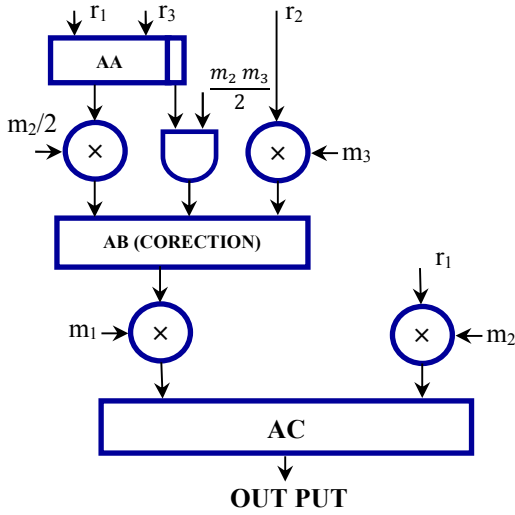


Figure 2. Structure of existing reverse converter [2]

If $(r_1 + r_3)$ is odd, then

$$X = -m_2 r_1 + m_1 \left| \frac{m_2 m_3}{2} + \frac{m_2}{2} \left(\frac{r_1 + r_3}{2} \right) - m_3 r_2 \right|_{m_2 m_3} \quad (6)$$

If $(r_1 + r_3)$ is even, then

$$X = -m_2 r_1 + m_1 \left| \frac{m_2}{2} \left(\frac{r_1 + r_3}{2} \right) - m_3 r_2 \right|_{m_2 m_3} \quad (7)$$

In adder AA, the inputs r_1 and r_3 are added and multiplied with $m_2/2$ which is one of the input to adder AB. The remaining inputs are one from multiplier $m_3(-m_3 r_2)$ and $m_2 m_3/2$. The output is multiplied with m_1 and provided as one of the input to adder AC. The output from the multiplier $m_2(m_2 r_1)$ is the other input to adder AC in which subtraction takes place to provide desired result.

According to [1], the simplified equations of (6) and (7) are as follows based on lemma given $|a m_1|_{m_1 m_2} = m_1 |a|_{m_2}$ and $m_3 = m_2 - 1$ in the existing structure to perform mod $2n-1$ operation and structure is shown in Fig. 3 [1].

If $(r_1 + r_3)$ is odd,

$$X = m_2(r_2 - r_1) + r_2 + (m_1 m_2 \left| \frac{m_3}{2} + \left(\frac{r_1 + r_3}{2} - r_2 \right) \right|_{m_3}) + m_3 \quad (8)$$

If $(r_1 + r_3)$ is even,

$$X = m_2(r_2 - r_1) + r_2 (m_1 m_2 \left| \left(\frac{r_1 + r_3}{2} - r_2 \right) \right|_{m_3}) + m_3 \quad (9)$$

In Adder A, the residue r_1 is subtracted from r_2 and multiplied with m_2 . The 3:2 carry save adder is used to implement Adder B to compute $((r_1 + r_3)/2 - r_2)$ with carry propagate adder. To compute mod $-m_3$ with the addition of $m_3/2$ and m_3 , one corrective addition or subtraction is required and which is implemented by Adder C with the selection line from the comparator that compares the output.

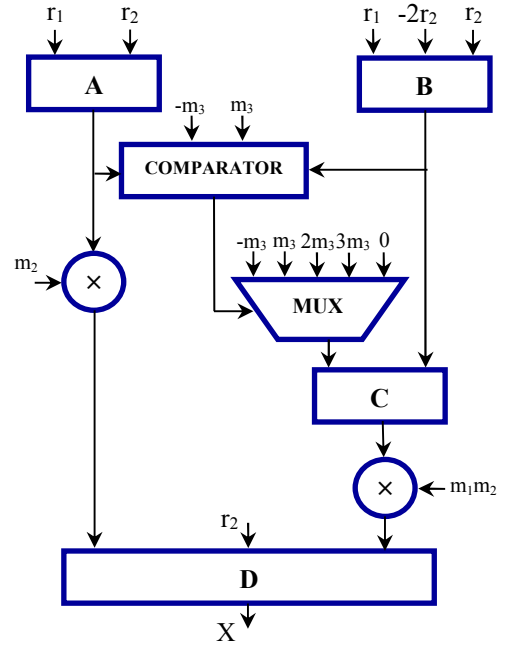


Figure 3. Structure of existing reverse converter [1]

The result is multiplied with $m_1 m_2$ which are added with inputs r_2 and one from the multiplier m_2 at Adder D. The adder D performs mod-M operation to the output. The output of adder D is the desired decimal equivalent of residues. The drawbacks in the existing design are more of hardware, power and area due to mod $(2n-1)$ operation. A design has been proposed to install reduce power, hardware and area.

III. PROPOSED DESIGN

The existing design is based on mod $(2n-1)$ operation which has limitation in the design that requires more hardware, power and area.

The equations (8) and (9) can be modified, which does not require the mod operations in the conversion process by

applying the mod operation. Thus, it reduces the complexity, power and area. The structure of proposed reverse converter is shown in Fig.4. The modified equations are as follows:

If $(r_1 + r_3)$ is odd,

$$X = m_2(r_2 - r_1) + r_2 + m_1m_2\left(\frac{m_3}{2} + \left(\frac{r_1+r_3}{2} - r_2\right)\right) \quad (10)$$

If $(r_1 + r_3)$ is even,

$$X = m_2(r_2 - r_1) + r_2 + m_1m_2\left(\frac{r_1+r_3}{2} - r_2\right) \quad (11)$$

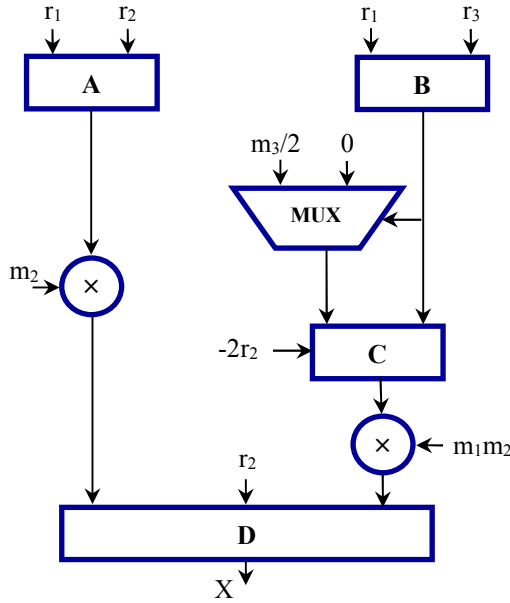


Figure 4. Structure of proposed reverse converter

In the proposed design, the residue r_2 is subtracted from r_1 at Adder A and is multiplied with m_2 . Adder B is the ripple carry adder which has inputs r_1 and r_3 . The LSB of output is used as selection input to multiplexer that provides either $m_3/2$ or 0 as output. The result is added with output of adder B and input $2r_2$. This result is multiplied with m_1m_2 which is given as one of the inputs of adder D. The other inputs are r_2 and output of multiplier m_2 that produces the output which is desired decimal equivalent. The examples are shown below for both cases in the proposed design.

IV. SIMULATION AND RTL SCHEMATIC RESULTS

The proposed design is simulated in NC launch simulation analysis environment (Sim-vision) using Cadence. The RTL

schematic of existing and proposed reverse converters are shown in Fig. 5, 6 and 7. The number of components in the proposed design is less than the existing design form Fig. 5 and 6. Table.1 shows the comparison of conventional reverse converters[1], [2], [3] in VHDL and implementing them on Xilinx Spartan 3 FPGA with Xilinx ISE 10.0.03 and the proposed design in Encounter tool for the moduli set $\{2n+1, 2n, 2n-1\}$. It is observed that from Table. 1, the proposed design has low power, delay and area than conventional designs.

The proposed design has been implemented in Cadence Encounter Digital Implementation RTL to GDSII system. The cell area for proposed reverse converter is shown in Table.1.

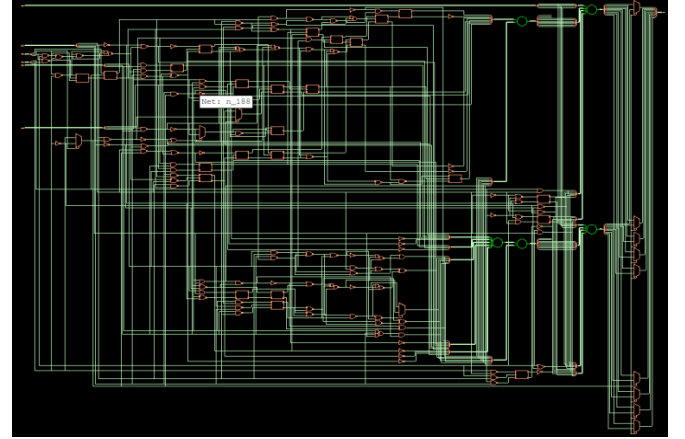


Figure 5. RTL Schematic of existing reverse converter [2].

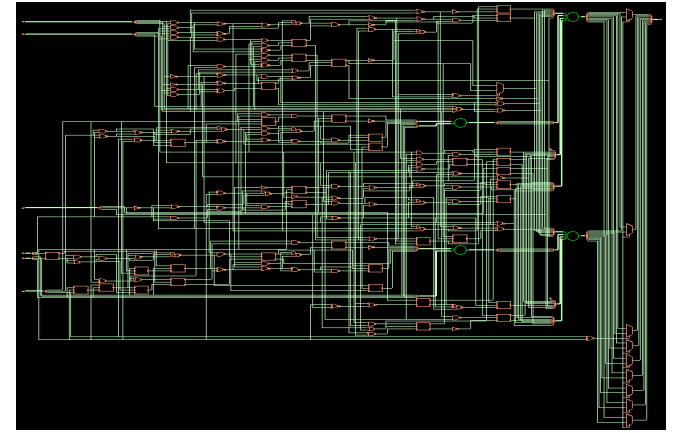


Figure 6. RTL Schematic of existing reverse converter [1].

TABLE 1. COMPARISON OF CONVENTIONAL REVERSE CONVERTERS AND PROPOSED REVERSE CONVERTER IN VHDL XILINX ISE.

n	Proposed			[1]			[2]			[3]		
	Area (μm^2)	Delay (ps)	Power (mW)	Area (slices)	Delay (ps)	Power (mW)	Area (slices)	Delay (ps)	Power (mW)	Area (slices)	Delay (ps)	Power (mW)
4	2669.586	3935	0.207	28	19857	15	49	31865	16	41	28575	15
21	4938.772	6767	0.428	90	36262	19	121	44718	20	126	44999	22
129	12264.807	9721	1.848	101	37599	19	136	42329	22	141	41312	24
813	15335.550	12256	2.426	251	45543	33	328	53556	37	305	55609	42
32769	30964.021	17800	5.940	177	39256	25	218	44896	28	253	47955	34
1321123	49469.469	22862	11.187	589	57261	65	749	65295	68	735	71930	95

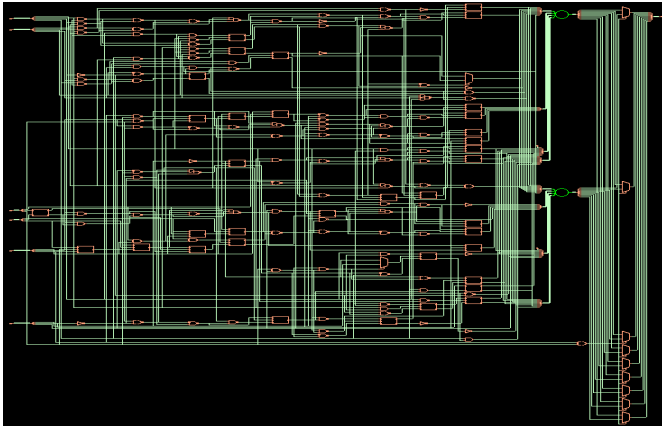


Figure 7. RTL Schematic of proposed reverse converter.

V. CONCLUSION

The proposed reverse converter does not require mod operations. This design has been proved advantageous in terms of power, cells, high speed and cell area over the existing ones. Table 1 demonstrates that the proposed structure provides low power, high speed and less area for implementing 90 nm technology. This RNS reverse conversion is used in the filtering and convolution process of DSP applications to minimize the faults and complexity.

REFERENCES

- [1] K. A. Gbolagade, G. R. Voicu, and S. D. Cotofana, "An Efficient FPGA Design of Residue to Binary Converter for the Moduli Set $\{2n+1, 2n, 2n-1\}$ ", IEEE Transactions on Very Large Scale Integration Systems, Aug. 2011, vol. 19, no. 8, pp. 1500-1503. DOI:10.1109/TVLSI.2010.2050608.
- [2] K. A. Gbolagade and S. D. Cotofana, "An Efficient RNS to Binary Converter Using the Moduli Set $\{2n+1, 2n, 2n-1\}$ ", in *proc. of 23rd Conference of DCI System*, Grenoble, France, Nov. 2008.
- [3] Yuke Wang, M. N. S. Swamy, and M. Omair Ahmad, "Residue-to-Binary Number Converters for Three Moduli Sets", IEEE Transactions of Circuits Systems II: Express Briefs, Feb. 1999, vol. 46, no.2, pp. 180-183. DOI:10.1109/82.752949.
- [4] T. Senthil Kumar and G. Prakash, "A Novel FPGA Design of Modified Residue to Binary Converter for Three Moduli Set", in *proc. of 2013 International Conference on Emerging Trends in VLSI, Embedded System, Nano Electronics and Tele Communication Systems*, Jan. 2013.
- [5] S. Akhter, R. Gaurav and S. Khan "Analysis and Design of Residue Number System Based Building Blocks," in *proc. of 5th International Conference on Signal Processing and Integrated Networks (SPIN)*, 2018, pp. 441-445. DOI: 10.1109/SPIN.2018.8474204.
- [6] B. Raghuvaiah and Omprakash, "Implementation of Hamming Coding in Residue Number System", in *proc. of 2018 IEEE International Conference on Current Trends Toward Converging Technologies (ICCTCT)*, Mar. 2018, pp. 1-5. DOI: 10.1109/ICCTCT.2018.8551122.
- [7] Wei Wang, M. N. S. Swamy, M. O. Ahmad, Yuke Wang, "A Study of the Residue-to-Binary Converters for the Three-Moduli Set", IEEE Transactions of Circuits Systems I: Regular Papers, Feb. 2003, vol. 50, no. 2, pp. 235-243. DOI: 10.1109/TCSI.2002.808191.
- [8] R. Conway and J. Nelson, "Improved RNS FIR Filter Architectures", IEEE Transactions of Circuits Systems II: Express Briefs, Jan. 2004, vol. 51, no. 1, pp. 26-28. DOI:10.1109/TCSII.2003.821524.
- [9] A. Omondi and B. Premkumar, "Residue number system: Theory and Implementation", Imperial College Press, 2007.
- [10] A. Prem Kumar, "An RNS to Binary Converter in $\{2n+1, 2n, 2n-1\}$ Moduli Set", IEEE Transactions of Circuits Systems II: Express Briefs, Jul. 1992, vol. 39, no.7, pp. 480-482. DOI: 10.1109/82.160172.
- [11] A. Prem kumar, "An RNS to Binary Converter in a Three Moduli Set With Common Factors", IEEE Transactions of Circuits Systems II, Express Briefs, Apr. 1995, vol. 42, no. 4, pp. 298-301. DOI:10.1109/82.378047.
- [12] P. V. Lakshmi, M. Sarada, A. Srinivasulu and D. Pal, "Three Novel Single-Stage Full Swing 3-Input XOR", International Journal of Electronics, vol. 105, no. 8. pp. 1416-1432, 2018. DOI: 10.1080/00207217.2018.1460767.
- [13] K. Gbolagade and S. Cotofana, "A Residue to Binary Converter for the $\{2n+2, 2n+1, 2n\}$ Moduli Set", in *proc. of 42nd Asilomar Conference on Signals, Systems and Computers (ACSSC)*, Oct. 2008, pp. 1785-1789. DOI: 10.1109/ACSSC.2008.5074734.
- [14] M. Sarada, A. Srinivasulu and D. Pal, "Novel Low-Supply, Differential XOR/ XNOR with Rail-to-Rail Swing, for Hamming-Code Generation", International Journal of Electronics Letters, vol. 6, no. 3. pp. 272-287, 2018, DOI:10.1080/21681724.2017.1357761.
- [15] B. Vinnakota and V. Rao, "Fast Conversion Techniques for Binary-Residue Number Systems," IEEE Trans. Circuits Systems I, Regular Papers, Dec. 1994, vol. 41, no. 12, pp. 927 – 929.
- [16] M. Akkal and P. Siy, "A New Mixed Radix Conversion Algorithm MRC- II", Journal of Systems and Architectures, 2007, vol. 53, pp. 577-586. <https://doi.org/10.1016/j.sysarc.2006.12.006>.
- [17] Avireni Srinivasulu and K. Sivadasan, "Optical Exclusive-OR gate", Journal of Microwaves, Optoelectronics and Electromagnetic Applications, vol. 3, no. 1, pp. 20-25, Apr 2003.
- [18] Y. Wang, "Residue - to - Binary Converters Based on New CHINESE Remainder Theorems", IEEE Transactions on Circuits Systems II: Express Briefs, Mar. 2000, vol. 47, no.3, pp. 197-205. DOI:10.1109/82.826745.
- [19] J. K. Saini, Avireni Srinivasulu and R. Kumawat, "High-Performance Low-Power 5:2 Compressor With 30 CNTFETs Using 32 nm Technology", Int. J. of Sen., Wir. Comm. and Con., vol. 9, issue. 4, pp. 462-467, 2019, DOI: 10.2174/2210327909666190206144601.
- [20] E. Vassalos, and D. Bakalis, "Residue-to-Binary Converter for the New RNS Moduli Set $\{22n-2, 2n-1, 2n+1\}$ ", in *proc. of Panhellenic Conference on Electronics & Telecommunications (PACET)*, 8-9 Nov. 2019, Volos, Greece, DOI: 10.1109/PACET48583.2019.8956249.
- [21] K. Vijayavardhan, M. Sarada, K. M. Santhoshini, A. Srinivasulu, "A Critical Look at Modular Adders using Residue Number System", Solid State Electronics Letters, vol. 1, no. 2. pp. 84-91, 2019, doi : 10.1016/j.ssel.2019.11.001.
- [22] E. K. Bankas, K. A. Gbolagade, "An Efficient VLSI Design of Residue to Binary Converter Circuit for a New Moduli Set $\{2^n, 2^{2n-1}-1, 2^{2n-1}+1\}$ ", in *proc. of 4th International Conference on Integrated Circuits and Microsystems*, 25-27 Oct. 2019, Beijing, China.
- [23] J. K. Saini, A. Srinivasulu and R. Kumawat, "A Low Power - High Speed CNTFETs Based Full Adder Cell With Overflow Detection", Micro and Nanosystems, vol. 11, no. 1, pp. 80-87, 2019, doi : 10.2174/1876402911666190211154634.
- [24] M.V.N. Madhavi Latha, R. Ramesh Rachh, P. V. Ananda Mohan, "Residue-to-Binary converters for the seven moduli set $\{2^{n-5}-1, 2^{n-3}-1, 2^{n-2}+1, 2^{n-1}-1, 2^{n-1}+1, 2^n, 2^n+1\}$ for n even", in *proc. of Asia Pacific Conference on Postgraduate Research in Microelectronics and Electronics*, 11-14 Nov. 2019, Bangkok, Thailand.
- [25] S. Zahiruddin, A. Srinivasulu, M. Sarada "A New Current Mode Multiplier Using Single CCCII Without Passive Components", in *proc. of the Int. Con. on 'Emerging Trends for Smart Grid Automation and Industry 4.0'*, BIT, Mesra, Ranchi, India, Dec. 05-07, 2019.
- [26] M.V.N. M. Latha, R. R. Rachh, P. V. A. Mohan, "An efficient residue-to-binary converter for the moduli set $\{2^{n-1}-1, 2^{n-k}, 2^n-1\}$ ", in *proc. of Asia Pacific Conference on Postgraduate Research in Microelectronics and Electronics*, 31 Oct.-2 Nov. 2017, Kuala Lumpur, Malaysia,
- [27] J. K. Saini, A. Srinivasulu, R. Kumawat, "Real-time Error Detectable Multiplier Circuit Using CNFET", Journal of Advanced Research in Dynamical and Control Systems, vol. 11, Issue. 07-SI, pp. 150-156, 2019.
- [28] A. Hiasat, "A Reverse Converter and Sign Detectors for an Extended RNS Five-Moduli Set", IEEE Transactions on Circuits and Systems I: Regular Papers, vol. 64, issue. 1, pp. 111-121, Jan. 2017. DOI: 10.1109/TCSI.2016.2612723.