

HW 2
Solutions

1) a) $h[n] = \frac{1}{4} \delta[n] + \frac{1}{4} \delta[n-1] + \frac{1}{4} \delta[n-2] + \frac{1}{4} \delta[n-3]$

b) $H(\hat{\omega}) = \frac{1}{4} (1 + e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} + e^{-j3\hat{\omega}})$
 $= \frac{1}{4} e^{-j\frac{3}{2}\hat{\omega}} (e^{j\frac{3}{2}\hat{\omega}} + e^{j\frac{1}{2}\hat{\omega}} + e^{-j\frac{1}{2}\hat{\omega}} + e^{-j\frac{3}{2}\hat{\omega}})$
 $= \frac{1}{2} e^{-j\frac{3}{2}\hat{\omega}} (\cos(\frac{1}{2}\hat{\omega}) + \cos(\frac{3}{2}\hat{\omega}))$

c) $x[n] = 5 + 4\cos(0.2\pi n) + 3\cos(0.5\pi n + \pi/4)$
 $\quad \quad \quad \uparrow \quad \quad \quad \uparrow \quad \quad \quad \uparrow$
 $\quad \quad \quad \text{Need } H(0) \quad \quad \quad \text{Need } H(0.2\pi) \quad \quad \quad H(0.5\pi)$

$H(0) = 1$

$H(0.2\pi) = 0.769 e^{-j0.3\pi}$

ANGLE = $-54^\circ = -0.942 \text{ rad}$

$H(0.5\pi) = 0$

$\Rightarrow y[n] = 5 + \underbrace{4(0.769)}_{3.078} \cos(0.2\pi n - 0.3\pi)$

d) $x_1[n] = 0$ for $n < 0$

$y_1[n] = \frac{1}{4} (x_1[n] + x_1[n-1] + x_1[n-2] + x_1[n-3])$

Since $x_1[n] = x[n]$ for $n \geq 0$, the filtered outputs will be the same when $n-3 \geq 0$

$\Rightarrow n \geq 3$ is the region where $y_1[n] = y[n]$

Here's a table of the first few values:

n	-1	0	1	2	3	4...
$y[n]$	4.029	6.809	7.927	7.927	6.809	5...
$y_1[n]$	0	2.480	4.329	5.338	6.809	5...

② a) $\delta[n+5]$

$X(z) = z^5$

b) $\delta[n-5]$

$X(z) = z^{-5}$

c) $\delta[n-1]$

$X(z) = z^{-1}$

d) $X(z) = 2 + 4z^{-1} + 6z^{-2} + 4z^{-3} + 2z^{-4}$

$x[n] = 2\delta[n] + 4\delta[n-1] + 6\delta[n-2] + 4\delta[n-3] + 2\delta[n-4]$

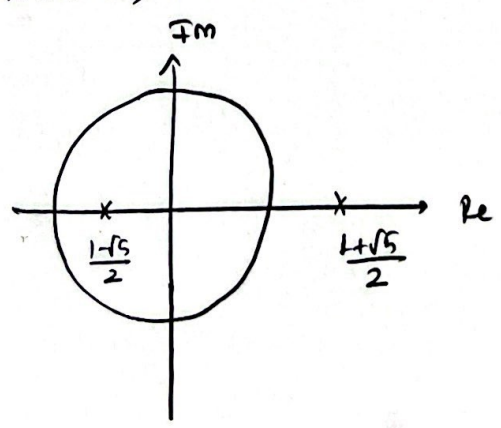
e) $X(z) = 1 - 2z^{-1} + 3z^{-3} - z^{-5}$

$x[n] = \delta[n] - 2\delta[n-1] + 3\delta[n-3] - \delta[n-5]$

③

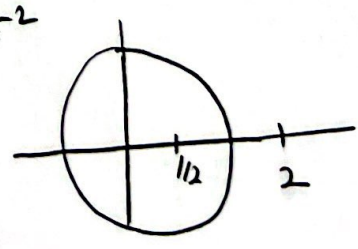
$H(z) = \frac{Y(z)}{X(z)} = \frac{z^{-1}}{1 - z^{-1} - z^{-2}} \rightarrow \text{for } y[n] = y[n-1] + y[n-2] + x[n-1]$

Poles of $H(z)$ are at $z = (1/2) \pm (\sqrt{5}/2)$. $H(z)$ has a zero at $z=0$.
 $h[n]$ is causal, the ROC for $H(z)$ has to be $|z| > (1/2) + (\sqrt{5}/2)$



③ $H(z) = \frac{Y(z)}{X(z)} = \frac{1}{z - \frac{5}{2} + z^{-1}} = \frac{z^{-1}}{1 - \frac{5}{2}z^{-1} + z^{-2}}$ for $y[n-1] - \frac{5}{2}y[n] + y[n+1] = x[n]$

$H(z) = \frac{-2/3}{1 - \frac{1}{2}z^{-1}} + \frac{2/3}{1 - 2z^{-1}}$
 $a=1/2 \quad z=2$
zeros $\rightarrow 1/2, 2$



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a) $h[n] = \delta[n-2] \Rightarrow$ filter is a delay by 2

$$y[n] = u[n-3] - u[n-6]$$

To find $x[n]$ we need to "un-delay" $y[n]$.

$$\Rightarrow x[n] = u[n-1] - u[n-4]$$

b) First difference FIR $\Rightarrow h[n] = \delta[n] - \delta[n-1]$

The first-difference filter has a nonzero output at n when $x[n]$ & $x[n-1]$ are not equal.

If $y[n] = \delta[n] - \delta[n-4]$, then the input $x[n]$ changes value at $n=0$ and $n=4$. At $n=0$, it pumps up by one; at $n=4$, it pumps down.

$$\Rightarrow x[n] = u[n] - u[n-4]$$

\uparrow pump up by one
 \nwarrow pump down

c) 4 pt averager : $y[n] = \frac{1}{4} (x[n] + x[n-1] + x[n-2] + x[n-3])$

$$\text{If } y[n] = -5\delta[n] - 5\delta[n-2]$$

$$y[0] = -5 = \frac{1}{4} (x[0] + x[-1] + x[-2] + x[-3])$$

** If we assume $x[n] = 0$ for $n < 0$, then $x[0] = -20$

$$y[1] = 0 = \frac{1}{4} (x[1] + x[0] + x[-1] + x[-2]) = \frac{1}{4} x[1] - 5$$

$$\Rightarrow x[1] = 20$$

$$\begin{aligned}
 y[2] = -5 &= \frac{1}{4} (x[2] + x[1] + x[0] + x[-1]) \\
 &= \frac{1}{4} (x[2] + 20 - 20 + 0) = \frac{1}{4} x[2]
 \end{aligned}
 \left. \vphantom{\begin{aligned} y[2] = -5 \\ = \frac{1}{4} x[2] \end{aligned}} \right\} x[2] = -20$$

$$y[3] = 0 = \frac{1}{4} (x[3] + x[2] + x[1] + x[0]) \Rightarrow x[3] = -20$$

$$\Rightarrow x[n] = \begin{cases} 0 & \text{for } n < 0 \\ -20 & \text{for } n \text{ even} \\ 20 & \text{for } n \text{ odd} \end{cases}$$

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a) $b_0 = b_3 = 0, b_1 = b_2$

$$H(e^{j\omega}) = b_1 e^{-j\omega} + b_2 e^{-j2\omega} = 2b_1 e^{-j3\omega/2} \cos(\omega/2)$$

$$|H(e^{j\omega})| = 2|b_1| |\cos(\omega/2)|$$

b) $b_1 = b_2 = 0, b_0 = b_3$

$$H(e^{j\omega}) = b_0 + b_3 e^{-j3\omega} = 2b_0 e^{-j3\omega/2} \cos(3\omega/2)$$

$$|H(e^{j\omega})| = 2|b_0| |\cos(3\omega/2)|$$

c) $b_0 = b_1 = b_2 = b_3$

$$H(e^{j\omega}) = b_0 + b_1 e^{-j\omega} + b_2 e^{-j2\omega} + b_3 e^{-j3\omega} = 2b_0 e^{-j3\omega/2} \cos(\omega) \cos(\omega/2)$$

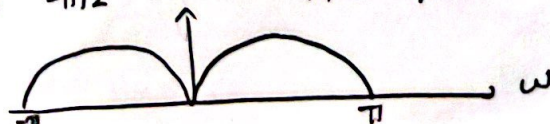
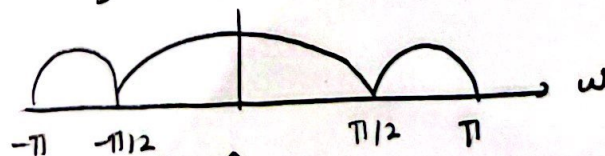
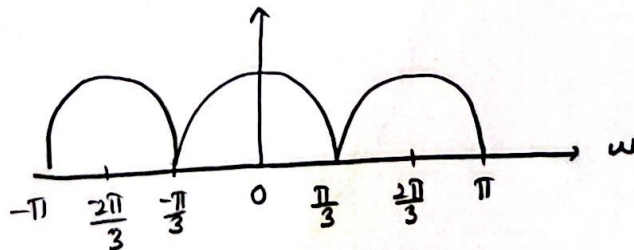
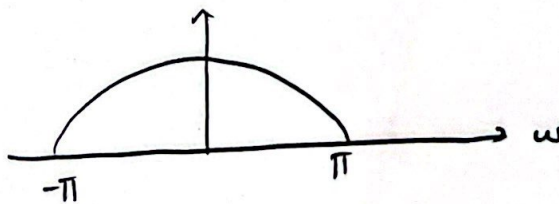
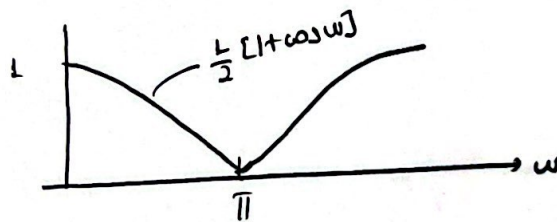
$$|H(e^{j\omega})| = 2|b_0| |\cos(\omega)| |\cos(\omega/2)|$$

d) $b_0 = -b_1 = b_2 = -b_3$

$$H(e^{j\omega}) = b_0 + b_1 e^{-j\omega} + b_2 e^{-j2\omega} + b_3 e^{-j3\omega} = -2b_0 e^{-j3\omega/2} \sin(\omega) \sin(\omega/2)$$

$$|H(e^{j\omega})| = 2|b_0| |\sin(\omega)| |\sin(\omega/2)|$$

Graphs:



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(5)

$$(6) \quad h(t) = e^{-0.1(t-2)} (u(t-2) - u(t-12))$$

a) The system is stable because $\int_{-\infty}^{+\infty} |h(t)| dt < \infty$

$$\int_{-\infty}^{+\infty} |h(t)| dt = \int_2^{12} |e^{-0.1(t-2)}| dt < \int_2^{12} dt = 10 < \infty$$

b) The system is causal because $h(t) = 0$ for $t < 0$

A plot of $h(t)$ starts at $t=2$

c) $x(t) = f(t-2)$

$$\Rightarrow y(t) = f(t-2) * h(t) \\ = h(t-2)$$

$$(7) \quad H(e^{j\omega}) = \frac{1}{2\pi} (H_L(e^{j\omega}) * \{2\pi\delta(\omega - \pi/2) + 2\pi\delta(\omega + \pi/2)\})$$

$$H_L(e^{j\omega}) = \begin{cases} 1, & |\omega| < \omega_c \\ 0, & \omega_c < |\omega| < \pi \end{cases}$$

using properties of Fourier transform, we obtain

$$h[n] = h_L[n] [2\cos(\pi n/2)]$$

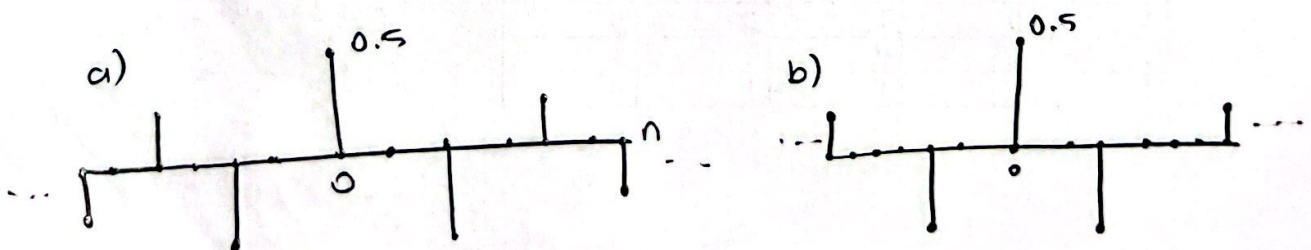
where

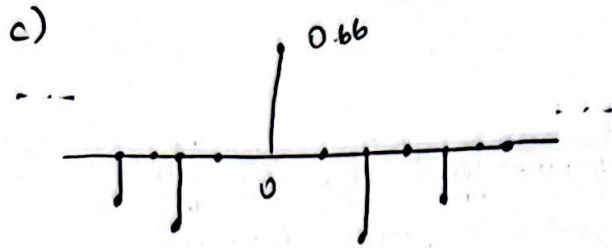
$$h_L[n] = \frac{\sin(\omega_c n)}{\pi n}$$

a) when $\omega_c = \pi/5$, $h[n] = 2 \frac{\sin(\pi n/5)}{\pi n} \cos(\pi n/2)$

b) when $\omega_c = \pi/4$, $h[n] = 2 \frac{\sin(\pi n/4)}{\pi n} \cos(\pi n/2)$

c) when $\omega_c = \pi/3$, $h[n] = 2 \frac{\sin(\pi n/3)}{\pi n} \cos(\pi n/2)$





- 8)
- a) Linear, stable
 - b) memoryless, linear, causal, stable
 - c) Linear
 - d) Linear, causal, stable
 - e) Time invariant, linear, causal, stable
 - f) Linear, stable
 - g) Time invariant, linear, causal

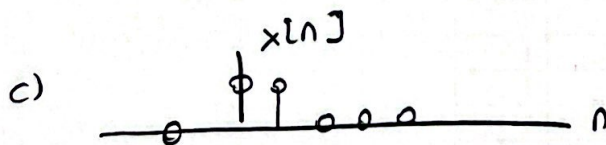
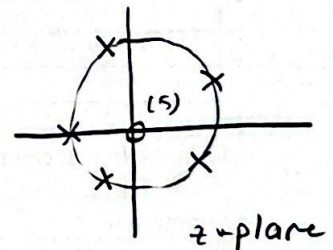
9) a) $H(z) = \frac{1}{1+z^{-5}}$

b) FIVE POLES - find roots of

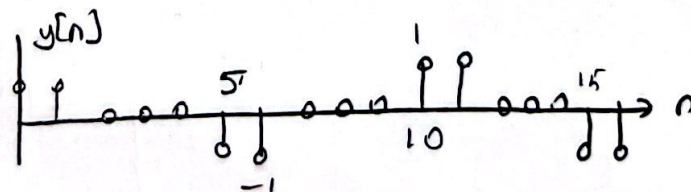
$$z^5 + 1 = 0$$

$$z = e^{j\pi/5}, e^{j3\pi/5}, e^{j\pi}, e^{-j\pi/5}, e^{-j3\pi/5}$$

$\nearrow 36^\circ$ $\nearrow 108^\circ$



n	0	1	2	3	4	5	6	7	8	9	10	11	12	
y[n]	0	1	1	0	0	0	-1	-1	0	0	0	1	1	0



d) PERIOD = 10

which can be determined from the plot above

10) Using the properties of Fourier Transform, we obtain

$$Y(j\omega) = X_1(j\omega) X_2(j\omega)$$

Therefore $Y(j\omega) = 0$ for $|\omega| > 1000\pi$. This implies that the Nyquist rate for $y(t)$ is $2 \times 1000\pi = 2000\pi$.
Therefore the sampling period T can at most be $2\pi / (2000\pi) = 10^{-3}$ sec. Therefore we have to use $T < 10^{-3}$ sec in order to be able to recover $y(t)$ from $y_p(t)$.