

CHAPTER 7

7.4, 7.6, 7.8, 7.10, 7.14, 7.16

7.4: (a) use filter coeffs: $H(z) = \frac{1}{3} + \frac{1}{3}z^{-1} + \frac{1}{3}z^{-2}$.

(b) use positive powers to extract poles and zeros

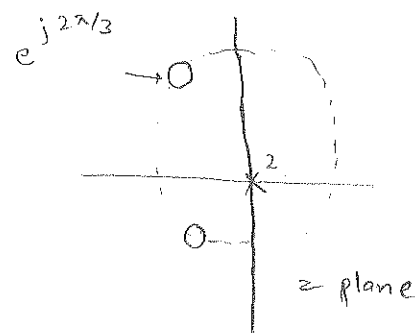
$$H(z) = \frac{1}{z^2} \left(\frac{1}{3}z^2 + \frac{1}{3}z + \frac{1}{3} \right)$$

Two poles at $z=0$

Zeros at

$$z = \frac{-1 \pm \sqrt{1-4}}{2} = -\frac{1}{2} \pm j\frac{\sqrt{3}}{2}$$

zeros: $1 \angle \pm j2\pi/3$



(c) $H(\omega) = H(e^{j\omega}) = H(z) \big|_{z=e^{j\omega}}$

$$= \frac{1}{3} + \frac{1}{3}e^{-j\omega} + \frac{1}{3}e^{-j2\omega} = \frac{1}{3}e^{-j\omega}(e^{j\omega} + 1 + e^{-j\omega})$$

$$= e^{-j\omega} \left(\frac{1 + 2\cos\omega}{3} \right)$$

ANOTHER FORMULA

$$H(\omega) = e^{-j\omega} \left(\frac{\sin(3\omega/2)}{3\sin(\omega/2)} \right)$$

(e) Use Linearity & Frequency response at $\hat{\omega} = 0$, $\hat{\omega} = \pi/4$ and $\hat{\omega} = 2\pi/3$. These are marked on the plots of frequency response.

$$Y[n] = 4H(0) + |H(\pi/4)| \cos\left(\frac{\pi}{4}n - \frac{\pi}{4} + \angle H(\pi/4)\right) - 3|H(2\pi/3)| \cos\left(2\pi/3 n + \angle H(2\pi/3)\right)$$

$$H(0) = \frac{1}{3} + \frac{1}{3} + \frac{1}{3} = 1.$$

$$H(\pi/4) = e^{-j\pi/4} (1 + 2\sqrt{2}/2)/3 = \frac{1+\sqrt{2}}{3} e^{-j\pi/4} = 0.8047e^{-j\pi/4}$$

$$H(2\pi/3) = 0 \text{ because } H(z) = 0 \text{ at } z = e^{\pm j2\pi/3}.$$

$$\therefore Y[n] = 4 + 0.8047 \cos(\pi/4 n - \pi/2)$$

7.6 (a) $Y_1(z) = H_1(z) X(z)$

$$\begin{aligned} Y(z) &= H_2(z) Y_1(z) = H_2(z) (H_1(z) X(z)) \\ &= \underbrace{(H_2(z) H_1(z))}_{H(z)} X(z) \end{aligned}$$

because $H(z) = \frac{Y(z)}{X(z)}$

(b) since $H_2(z) H_1(z) = H_1(z) H_2(z)$ because $H_1(z)$ and $H_2(z)$ are scalar functions

$$\Rightarrow Y(z) = H_1(z) \underbrace{H_2(z)}_{\text{means that } H_2(z) \text{ is applied first}} X(z)$$

(c) $H_1(z) = \frac{1}{3} (1 + z^{-1} + z^{-2})$ by using filter coeffs

$$\begin{aligned} H(z) &= H_2(z) H_1(z) \\ &= \frac{1}{3} (1 + z^{-1} + z^{-2}) \cdot \frac{1}{3} (1 + z^{-1} + z^{-2}) \\ &= \frac{1}{4} (1 + 2z^{-1} + 3z^{-2} + 2z^{-3} + z^{-4}) \end{aligned}$$

(d) convert to difference equation (i.e. filter coeffs)

$$y[n] = \frac{1}{9} (x[n] + 2x[n-1] + 3x[n-2] + 2x[n-3] + x[n-4])$$

(e) Find the poles and zeros of $H_2(z)$, then "double" them because $H_1(z) = H_2(z)$.

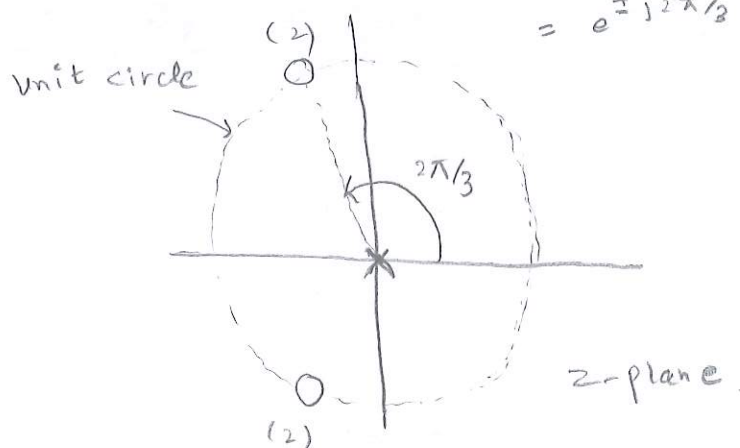
$$H_2(z) = \frac{1}{3} z^{-2} (z^2 + z + 1)$$

$\frac{1}{z^2}$ contributes two poles at $z=0$

zeros are:

$$\frac{-1 \pm \sqrt{1-4}}{2} = \frac{-1 \pm j\sqrt{3}}{2}$$

$$= e^{\pm j2\pi/3}$$



$$(f) H(e^{j\hat{\omega}}) = H_1(e^{j\hat{\omega}}) H_2(e^{j\hat{\omega}})$$

$$= \frac{1}{9} (1 + e^{-j\hat{\omega}} + e^{-j2\hat{\omega}})^2$$

$$= \frac{1}{9} e^{-j2\hat{\omega}} (e^{j\hat{\omega}} + 1 + e^{-j\hat{\omega}})^2$$

$$= \frac{1}{9} e^{-j2\hat{\omega}} (1 + 2\cos(\hat{\omega}))^2$$

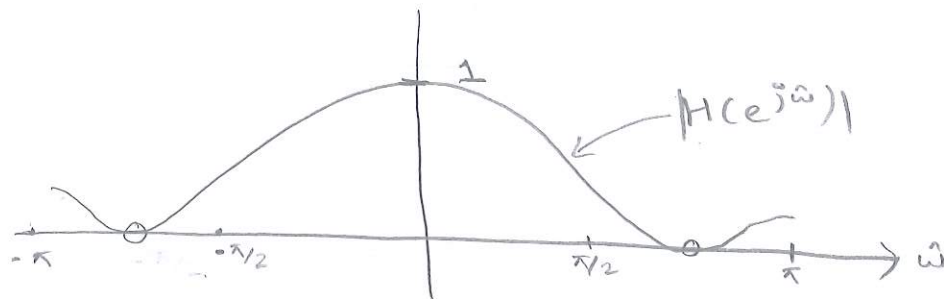
$$|H(e^{j\hat{\omega}})| = \frac{1}{9} (1 + 2\cos(\hat{\omega}))^2$$

At $\hat{\omega} = 0$, $|H| = \frac{1}{9} (3)^2 = 1$

At $\hat{\omega} = \pi/2$, $|H| = \frac{1}{9} (1)^2 = 1/9$

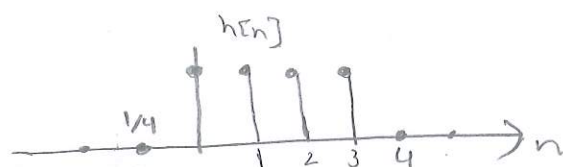
At $\hat{\omega} = 2\pi/3$, $|H| = 0$ because there is a zero on the unit circle.

At $\hat{\omega} = \pi$, $|H| = \frac{1}{9} (1-2)^2 = 1/9$



Problem 7.8:

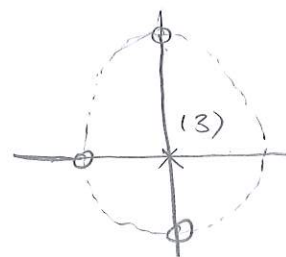
(a) $h[n] = \frac{1}{4} \{ \delta[n] + \delta[n-1] + \delta[n-2] + \delta[n-3] \}$



(b) $H(z) = \frac{1}{4} (1 + z^{-1} + z^{-2} + z^{-3})$ by using $h[n]$.

(c) Poles and zeros:

$$H(z) = \frac{1}{4} \frac{z^3 + z^2 + z + 1}{z^3} \rightarrow \boxed{\text{3 poles at } z=0}$$



$$z^3 + z^2 + z + 1 = \frac{z^4 - 1}{z - 1} \leftarrow \text{zeros at } z$$

(d)
$$H(z) = \frac{1}{4} \frac{1 - z^{-4}}{1 - z^{-1}} = \frac{1}{4} \frac{e^{-j2\omega} (e^{j2\omega} - e^{-j2\omega})}{e^{-j\omega/2} (e^{j\omega/2} - e^{-j\omega/2})}$$

$$= \frac{1}{4} e^{-j3\omega/2} \frac{\sin(2\omega)}{\sin(\omega/2)}$$

At $\omega = 0$, $H(e^{j\omega}) = \frac{1}{4} e^{j0} \cdot 4 = 1$

At $\omega = \pi/2, \pi, -\pi/2$, $H(e^{j\omega}) = 0$ because $\sin(2\omega) = 0$

(f) Evaluate $H(e^{j\omega})$ at $\omega = 0$, $\omega = 0.2\pi$ and $\omega = 0.5\pi$.
These are marked on the frequency response plots.

$$H(e^{j0}) = 1 \quad H(e^{j0.2\pi}) = 0.771 e^{-j0.3\pi}$$

$$H(e^{j0.5\pi}) = 0$$

$$\begin{aligned} \neq Y[n] &= 5 + 4(0.771) \cos(0.2\pi n - 0.3\pi) + 0 \\ &= 5 + 3.084 \cos(0.2\pi n - 0.3\pi) \end{aligned}$$

→ ANGLE = -54°
or -0.94 rads.

Problem 7.10:

(a) Convert $H(z)$ to a difference equation

$$Y[n] = X[n] - 3X[n-2] + 2X[n-3] + 4X[n-6]$$

The most delay is 6 samples, so the term $4X[n-6]$ in $X[n]$ is delayed to $168[n-10]$.

The least amount of delay is $2\delta[n]$ experiencing no delay. Thus the output starts at $n=0$ and ends at $n=10$

$$\Rightarrow y[n] = 0 \quad \text{for } n < 0 \quad \& \quad n > 10$$

$$N_1 = 0 \quad \text{and} \quad N_2 = 10$$

$$(b) \quad X(z) = z + z^{-1} - 2z^{-2} + 4z^{-4}$$

$$Y(z) = H(z) X(z)$$

$$= (1 - 3z^{-2} + 2z^{-3} + 4z^{-6})(z + z^{-1} - 2z^{-2} + 4z^{-4})$$

$$\begin{aligned} = & z + z^{-1} - 2z^{-2} + 4z^{-4} - 3z^{-2} - 3z^{-3} + 6z^{-4} - 12z^{-6} + \\ & 4z^{-3} + 2z^{-4} - 4z^{-5} + 8z^{-7} + 8z^{-6} + \\ & 4z^{-7} - 8z^{-8} + 16z^{-10} \end{aligned}$$

Combine terms with common exponents

$$Y(z) = 2 + z^{-1} - 8z^{-2} + z^{-3} + 12z^{-4} - 4z^{-5} - 4z^{-6} \\ + 12z^{-7} - 8z^{-8} + 16z^{-10}$$

Invert:

$$y[n] = 2\delta[n] + \delta[n-1] - 8\delta[n-2] + \delta[n-3] + \\ 12\delta[n-4] - 4\delta[n-5] - 4\delta[n-6] + 12\delta[n-7] \\ - 8\delta[n-8] + 16\delta[n-10]$$

Problem 7.14:

$$H(z) = 1 - 2z^{-2} - 4z^{-4}$$

$$h[n] = \delta[n] - 2\delta[n-2] - 4\delta[n-4]$$

$$x[n] = 20e^{j0n} + 20\cos(\pi/2n + \pi/4) - 20\delta[n] \\ \downarrow \qquad \qquad \qquad \downarrow \qquad \qquad \qquad \downarrow \\ H(e^{j0}) \cdot 20 \qquad \text{Need } H(e^{j\pi/2}) \qquad -20h[n]$$

$$H(e^{j\omega}) = 1 - 2e^{-j2\omega} - 4e^{-j4\omega}$$

$$H(e^{j0}) = 1 - 2 - 4 = -5$$

$$H(e^{j\pi/2}) = 1 - 2e^{-j\pi} - 4e^{-j2\pi}$$

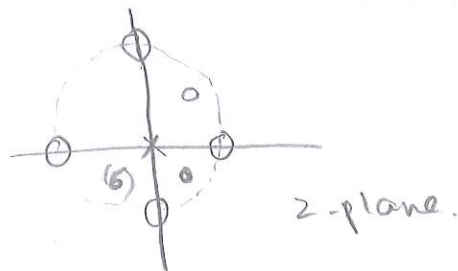
$$= 1 + 2 - 4 = -1$$

$$\therefore y[n] = -100 - 20\cos(\pi/2n + \pi/4) - 20\delta[n] + \\ 40\delta[n-2] + 80\delta[n-4]$$

Problem 7.16:

(a) $H(z)$ has 6 zeros & 6 poles at $z=0$

The zeros are: $z = \pm 1, \pm j, 0.8 e^{\pm j\pi/4}$



(b) $w[n] = x[n] - x[n-4]$

$$\Rightarrow H_1(z) = 1 - z^{-4} = (1 + z^{-2})(1 - z^{-2})$$

To get $H_2(z)$ divide:

$$H_2(z) = \frac{H(z)}{H_1(z)} = \frac{(1 - 0.8 e^{-j\pi/4} z^{-1})}{(1 - 0.8 e^{+j\pi/4} z^{-1})}$$

$$H_2(z) = 1 - 1.6 \cos \pi/4 z^{-1} + 0.64 z^{-2}$$

(c) $Y[n] = x[n] - (0.8\sqrt{2}) x[n-1] + 0.64 x[n-2]$
 $= 1.1314$