

**EEE 431: Telecommunications 1**

**MIDTERM**

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Section: \_\_\_\_\_

Prob. 1: \_\_\_\_\_ / 21

Prob. 2: \_\_\_\_\_ / 21

Prob. 3: \_\_\_\_\_ / 32

Prob. 4: \_\_\_\_\_ / 26

**Total: \_\_\_\_\_ / 100**

**Problem 1.** Consider the following signal:  $s(t) = \cos(2000\pi t) + \sin(1000\pi t)$ . Suppose  $s(t)$  is sampled at every 0.25 millisecond for  $-\infty < t < \infty$ . Let  $X$  denote a random variable corresponding to these samples.

- (a) Can we reconstruct  $s(t)$  from these samples? Why or why not?
- (b) Find the probability mass function (PMF) of  $X$ .
- (c) Calculate the entropy of  $X$ .
- (d) Perform Huffman coding of  $X$ . List all the codewords and calculate the average codeword length.

**Problem 2.** A source has a probability density function (PDF) as expressed below:

$$f_X(x) = \begin{cases} k e^{(x-1)^3} , & \text{if } 0 \leq x \leq 2 \\ k e x/2 , & \text{if } 2 < x \leq 4 \\ 0 , & \text{otherwise} \end{cases}$$

where  $k$  is a suitable constant (no need to find it). This source is quantized by using the following 2-level quantizer:

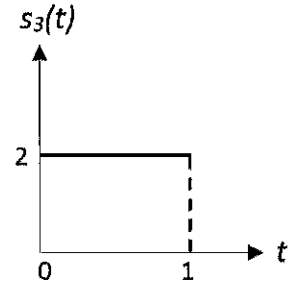
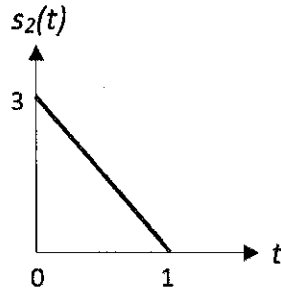
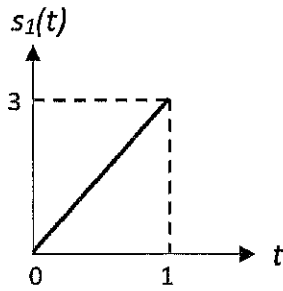
$$Q(x) = \begin{cases} 1 , & \text{if } x \in [0, 2] \\ 3 , & \text{if } x \in (2, 4] \end{cases}$$

- (a) Obtain the mean squared error distortion  $E\{(X - Q(X))^2\}$  in its simplest form.
- (b) Calculate the following ratio of probabilities:  $P(Q(X) = 1)/P(Q(X) = 3)$ . Simplify as much as possible.
- (c) Suppose that the source produces 20000 samples (outputs) per second, and we would like to transmit the source outputs by using uniform PCM with 1 bit per sample, as specified by the quantizer above. What is the minimum bandwidth required to transmit this PCM signal?
- (d) Propose a compander to improve the SQNR of this system. Namely, plot or define a function  $g(x)$  corresponding to the “compressor” at the transmitter. Specify the domain and range of function  $g(x)$ . Why do you think the proposed compander will increase SQNR? (You do not need to calculate the SQNR. Your compander does not need to be optimal.)

**Problem 3.** Consider a wide sense stationary (WSS) Gaussian random process  $X(t)$  with zero mean and autocorrelation function  $R_X(\tau) = 2 - |\tau|$  if  $|\tau| < 2$  and  $R_X(\tau) = 0$  otherwise. This process is input to a system which generates the output process  $Y(t)$  given by  $Y(t) = X(t) - 3X(t - 2)$ .

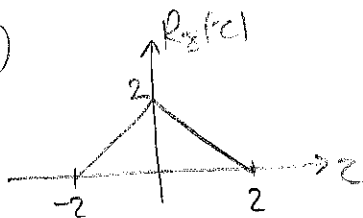
- (a) Determine the power spectral density of the input process  $X(t)$ .
- (b) Determine the power spectral density of the output process  $Y(t)$ .
- (c) Determine the autocorrelation function of  $Y(t)$ .
- (d) Calculate the average powers of  $X(t)$  and  $Y(t)$ .
- (e) Specify the joint probability distribution of  $Y(0)$  and  $Y(1)$  (i.e., name the distribution and calculate all the necessary parameters).
- (f) Calculate the following expectation:  $E[(2Y(0) - 3Y(1))^2]$ .

**Problem 4. (a)** For the following signals, find a set of orthonormal basis functions (both write down their mathematical expressions and plot them), and represent each signal as a vector in the corresponding signal space.



**(b)** Find a signal  $x(t)$  such that it has the same energy as  $s_3(t)$  above, and the angle between  $x(t)$  and  $s_3(t)$  is equal to  $\pi/3$  (i.e., 60 degrees). Both write down the mathematical expression of  $x(t)$  and plot it.

32 ③



③ ①  $S_x(f) = F\{R_x(\tau)\}$   
 $= \underline{4 \text{sinc}^2(2f)}$

$\begin{cases} \Lambda(\tau) \leftrightarrow \text{sinc}^2 f \\ 2\Lambda(\frac{\tau}{2}) \leftrightarrow 4 \text{sinc}^2(2f) \end{cases}$

⑦ ⑥  $S_y(f) = S_x(f) |H(f)|^2$   
 $= 4 \text{sinc}^2(2f) ((1 - 3 \cos(4\pi f))^2 + 9 \sin^2(4\pi f))$   
 $= \underline{4 \text{sinc}^2(2f) (10 - 6 \cos(4\pi f))}$

$Y(f) = X(f) \underbrace{(1 - 3e^{-j2\pi f^2})}_{H(f)}$

⑥ ③  $R_y(\tau) = E[Y(t+\tau)Y(t)] = E[(X(t+\tau) - 3X(t+\tau-2))(X(t) - 3X(t-2))]$   
 $= \underline{10R_x(\tau) - 3R_x(\tau-2) - 3R_x(\tau+2)}$

③ ④  $P_x = R_x(0) = \underline{2}$   $P_y = R_y(0) = 10R_x(0) - 3R_x(-2) - 3R_x(2) = \underline{20}$

⑦ ②  $\begin{bmatrix} y(0) \\ y(1) \end{bmatrix} \sim N\left(\begin{bmatrix} 0 \\ 0 \end{bmatrix}, \begin{bmatrix} 20 & 7 \\ 7 & 20 \end{bmatrix}\right)$   $E[y(1)y(0)] = R_y(1) = 10R_x(1) - 3R_x(-1) - 3R_x(3) = \underline{7}$

④ ④  $4E[y^2(0)] - 12E[y(0)y(1)] + 3E[y^2(1)]$   
 $= 4 \cdot 20 - 12 \cdot 7 + 3 \cdot 20 = \underline{206}$

② ① ②  $2f_{\max} = 2000 \text{ Hz}$   $f_{\text{samp}} = \frac{1}{0.25 \text{ ms}} = 4000 \text{ Hz}$   $f_{\text{samp}} > 2f_{\max} \rightarrow \text{Yes } \checkmark$

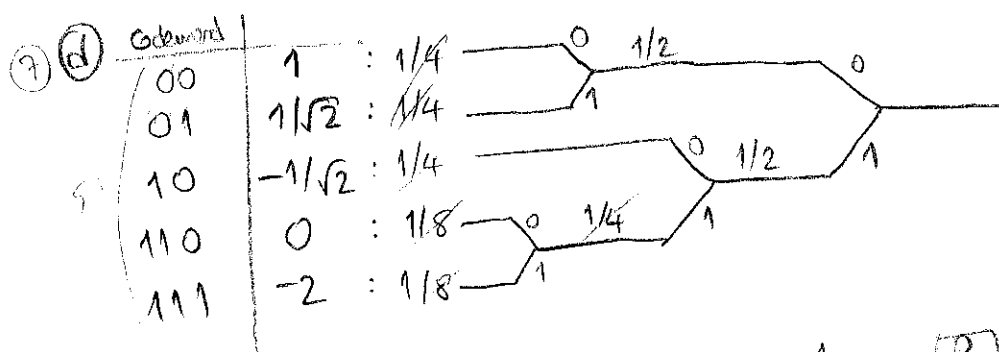
⑦ ⑥

t (ms)	0	0.25	0.5	0.75	1	1.25	1.5	1.75
s(t)	1	$\frac{1}{\sqrt{2}}$	0	$\frac{1}{\sqrt{2}}$	1	$-\frac{1}{\sqrt{2}}$	-2	$-\frac{1}{\sqrt{2}}$

← 1 period

$X = \{-2, -\frac{1}{\sqrt{2}}, 0, \frac{1}{\sqrt{2}}, 1\}$  w/ probs  $\{\frac{1}{8}, \frac{1}{4}, \frac{1}{8}, \frac{1}{4}, \frac{1}{4}\}$ , respectively.

② ②  $H(X) = 2 \cdot \frac{1}{8} \log_2 8 + 3 \cdot \frac{1}{4} \log_2 4 = \frac{3}{4} + \frac{6}{4} = \underline{\frac{9}{4} \text{ bits/sample}}$



Avg. Code word length =  $2 \cdot \frac{1}{4} \cdot 3 + 3 \cdot \frac{1}{8} \cdot 2 = \underline{\frac{9}{4} \text{ bits/sample}}$

(2)  $f_X(x) = \begin{cases} ke^{(x-1)^3}, & 0 \leq x \leq 2 \\ ke^{x/2}, & 2 < x \leq 4 \\ 0, & \text{o.w.} \end{cases}$

$Q(x) = \begin{cases} 1, & 0 \leq x \leq 2 \\ 3, & 2 < x \leq 4 \end{cases}$

(12) (9)  $E[(X-Q(X))^2] = \int_0^2 (x-1)^2 ke^{(x-1)^3} dx + \int_2^4 (x-3)^2 \frac{ke^x}{2} dx$

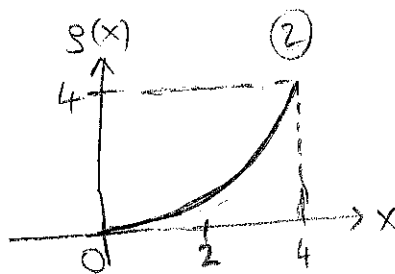
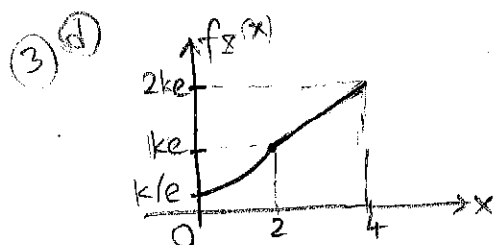
$u = (x-1)^3 \rightarrow du = 3(x-1)^2 dx$

$= \frac{k}{3} e^u \Big|_{-1}^1 + \frac{ke}{2} \left( \frac{(x-3)^4}{4} \Big|_2^4 + (x-3)^3 \Big|_2^4 \right)$

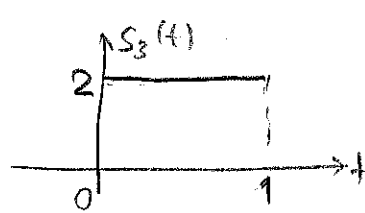
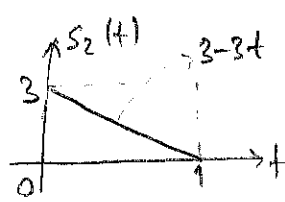
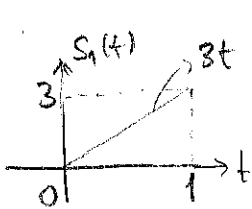
$= \frac{k}{3} (e - e^{-1}) + \frac{ke}{2} \left( \frac{1}{4} - \frac{1}{4} + 1 + 1 \right) = \frac{k}{3} (e - e^{-1}) + ke$

(5) (6)  $P(Q(X)=1) = 1 - P(Q(X)=3)$   
 $P(Q(X)=3) = P(X \in (2, 4]) = \int_2^4 ke^{\frac{x}{2}} dx = \frac{ke}{4} (16 - 4) = 3ke$   
 So  $\frac{P(Q(X)=1)}{P(Q(X)=3)} = \frac{1 - 3ke}{3ke}$

(1) (c) 20000 bits/sec  $\rightarrow$  10000 Hz needed.



(1) More prob. values mapped to larger regions



(13) (a)

$$\gamma_1(t) = \frac{S_1(t)}{\sqrt{E_1}} \quad E_1 = \int_0^1 (3t)^2 dt = 9 \left[ \frac{t^3}{3} \right]_0^1 = 3$$

(4)

$$\gamma_1(t) = \frac{S_1(t)}{\sqrt{3}} = \sqrt{3}t$$

(3)

$$S_{21} = \int_0^1 S_2(t) \gamma_1(t) dt = \int_0^1 (3-3t) \sqrt{3}t dt = \left( 3\sqrt{3} \frac{t^2}{2} - 3\sqrt{3} \frac{t^3}{3} \right) \Big|_0^1 = \frac{\sqrt{3}}{2}$$

$$d_2(t) = S_2(t) - S_{21} \gamma_1(t) = (3-3t) - \frac{\sqrt{3}}{2} \sqrt{3}t = 3 - \frac{3t}{2}$$

$$\int_0^1 \left(3 - \frac{3t}{2}\right)^2 dt = 9 - \frac{27}{2} + \frac{27}{4} \left[ \frac{t^3}{3} \right]_0^1 = \frac{9}{4}$$

$$\gamma_2(t) = \frac{d_2(t)}{\sqrt{9/4}} = \frac{2}{3} \left(3 - \frac{3t}{2}\right) = 2 - 3t$$

Since  $S_3(t) = \frac{2}{3} S_1(t) + \frac{2}{3} S_2(t)$ , no need for another basis fn.

(6)

$$\underline{S}_1 = \begin{bmatrix} \sqrt{3} \\ 0 \end{bmatrix} \quad \underline{S}_2 = \begin{bmatrix} \sqrt{3}/2 \\ 3/2 \end{bmatrix} \quad \underline{S}_3 = \begin{bmatrix} \sqrt{3} \\ 1 \end{bmatrix}$$

$$S_{22} = \int_0^1 S_2(t) \gamma_2(t) dt = \int_0^1 (3-3t)(2-3t) dt = \frac{3}{2}$$

(7) (b)

$x(t) \rightarrow \text{energy } 4, \quad \cos \theta = \frac{1}{2}$

$$\frac{1}{2} = \frac{\langle x(t), S_3(t) \rangle}{\sqrt{4 \cdot 4}} \Rightarrow \langle x(t), S_3(t) \rangle = 2 \Rightarrow x_1 \sqrt{3} + x_2 1 = 2$$

$\swarrow$  any  $x_1, x_2$  satisfying this.

Let  $x(t) = x_1 \gamma_1(t) + x_2 \gamma_2(t)$

e.g.  $x_1 = 0, x_2 = 2$

Then,  $x(t) = 2 \gamma_2(t)$