

LAB1 Report

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In lab 1 We explored the basic configuration of e² studio, ultrasound module and the basic signal processing flow of the wireless transmitted signal.

1 Hardware configuration

2 Error Source Analysis

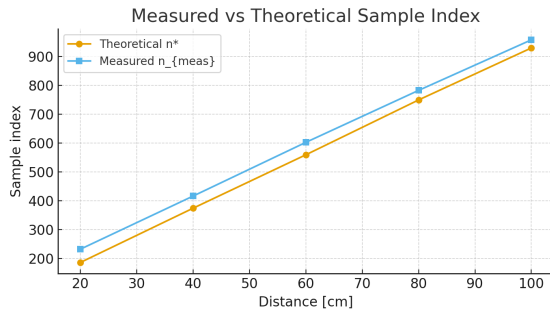
We inspect the relative distance and try to determine the model of the error.

Pair (cm)	n_{theory}^*	n_{meas}	$\Delta n_{\text{theory}}^*$	Δn_{meas}	$\frac{\Delta n_{\text{meas}}}{\Delta n_{\text{theory}}^*}$	Relative Error (%)
20→40	184.97→369.94	231→412	184.97	181	0.979	-2.1
40→60	369.94→554.91	412→598	184.97	186	1.006	+0.6
60→80	554.91→739.88	598→773	184.97	175	0.946	-5.4
80→100	739.88→924.86	773→953	184.97	180	0.973	-2.7

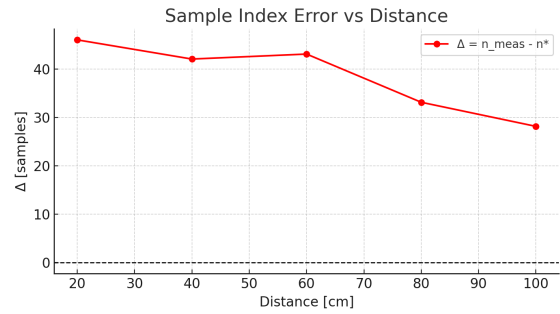
Table 1: Relative Distance Comparison (Measured vs. Theoretical)

We retrieve back n_{meas} by aligning the starting sample by the maximum Tx data

We plot out the



(a) Measured Peak and the theoretical sample (T=25)



(b) Error vs. distance

Figure 1: Comparing robust and non-robust design for linear precoder

3 Signal Processing

some discussion below:

- **Correct "Detection" threshold.** The physically correct measure of the time of flight (TOF) would be the *wavefront* of the received signal.

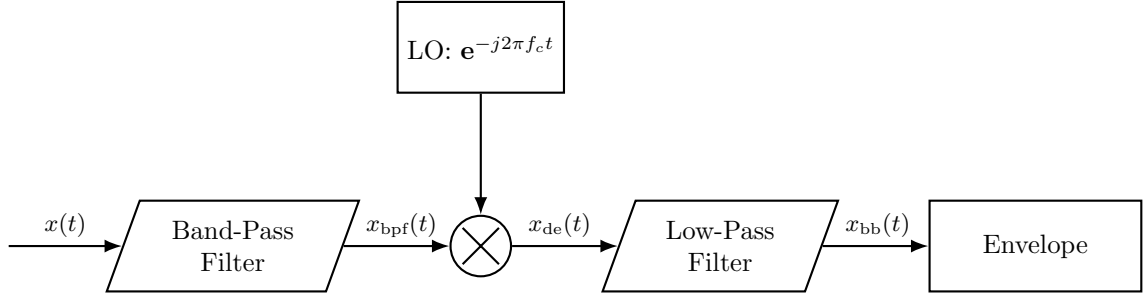


Figure 2: block diagram: signal \rightarrow BPF \rightarrow demod (mixer) \rightarrow LPF \rightarrow envelope.

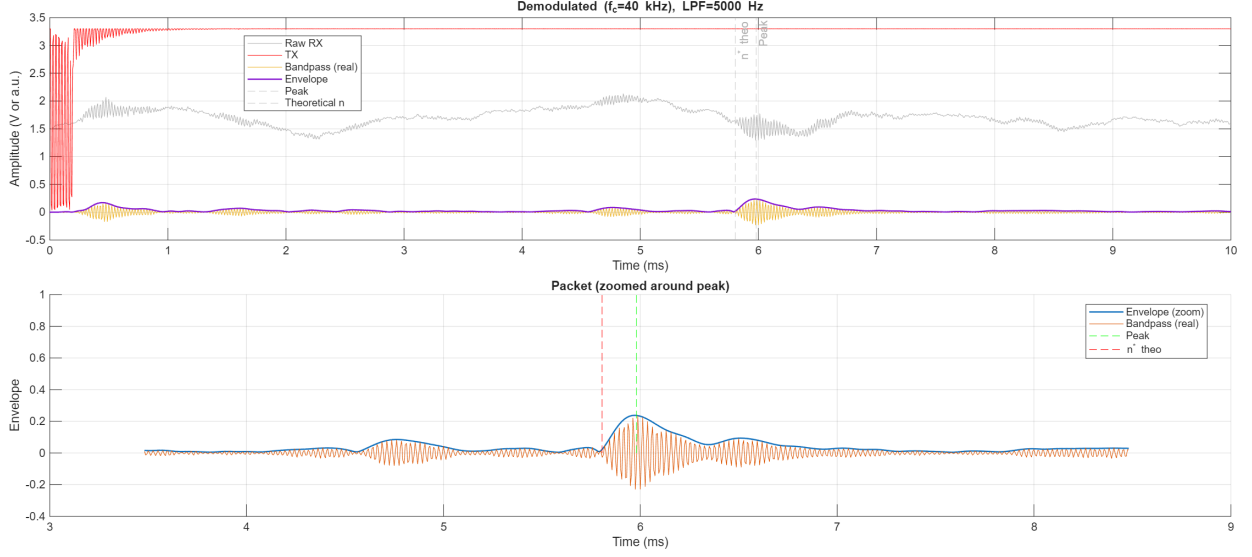


Figure 3: Time domain signal before and after filtering & demodulation

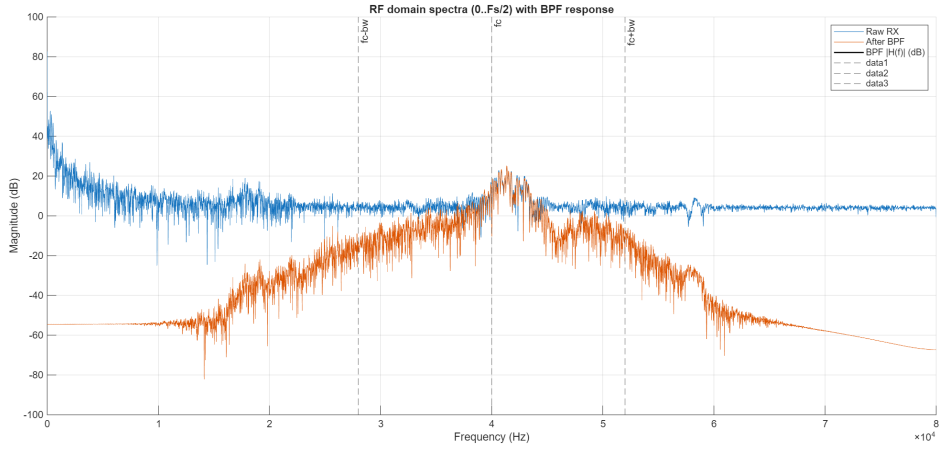


Figure 4: bpf

Bandpass Filtering

Given the sampled received signal $x(t)$, we first apply a bandpass filter to isolate the signal components near the carrier frequency. The bandpass-filtered signal is

$$x_{\text{bpf}}(t) = x(t) * h_{\text{bpf}}(t) \quad (1)$$

where $*$ denotes convolution and $h_{\text{bpf}}(t)$ is the impulse response of the bandpass filter.

Filter Design. We choose a 6th-order IIR Butterworth bandpass filter with half-power frequencies

$$f_{\text{bp},1} = f_c - 2f_w, \quad f_{\text{bp},2} = f_c + 2f_w,$$

where the carrier frequency $f_c = 40$ kHz and the signal bandwidth is

$$f_w = \frac{1}{T_{\text{burst}}} \approx 5 \text{ kHz}.$$

To ensure the filter design remains within valid frequency bounds, we compute:

$$\begin{aligned} \text{bp_bw} &= \max(2f_w, 12 \text{ kHz}), \\ \text{bp_f1} &= \max(10, f_c - \text{bp_bw}), \\ \text{bp_f2} &= \min\left(\frac{F_s}{2} - 10, f_c + \text{bp_bw}\right), \end{aligned}$$

where F_s is the sampling frequency.

MATLAB Implementation. The filter is implemented and applied using MATLAB as follows:

```
1 bp_bw = max(2*fw, 12e3); % Bandwidth selection
2 bp_f1 = max(10, fc - bp_bw); % Lower cutoff frequency
3 bp_f2 = min(Fs/2-10, fc + bp_bw); % Upper cutoff frequency
4
5 dbp = designfilt('bandpassiir','FilterOrder',6, ...
6     'HalfPowerFrequency1', bp_f1, ...
7     'HalfPowerFrequency2', bp_f2, ...
8     'SampleRate', Fs);
9
10 rx_bp = filtfilt(dbp, rx_dc); % Zero-phase filtering
```

This produces the zero-phase bandpass-filtered signal $x_{\text{bpf}}(t)$.

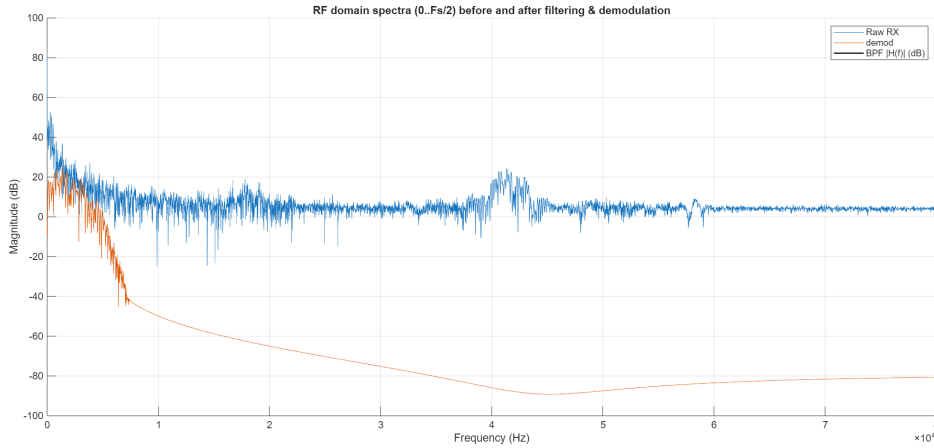


Figure 5: demod

References

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