



Low-Power, 8-Channel, 24-Bit Analog Front-End for Biopotential Measurements

Check for Samples: [ADS1294](#), [ADS1294R](#), [ADS1296](#), [ADS1296R](#), [ADS1298](#), [ADS1298R](#)

FEATURES

- Eight Low-Noise PGAs and Eight High-Resolution ADCs (ADS1298, ADS1298R)
- Low Power: 0.75mW/channel
- Input-Referred Noise: $4\mu\text{V}_{\text{PP}}$ (150Hz BW, $G = 6$)
- Input Bias Current: 200pA
- Data Rate: 250SPS to 32kSPS
- CMRR: –115dB
- Programmable Gain: 1, 2, 3, 4, 6, 8, or 12
- Supports AAMI EC11, EC13, IEC60601-1, IEC60601-2-27, and IEC60601-2-51 Standards
- Unipolar or Bipolar Supplies:
 $\text{AVDD} = 2.7\text{V to } 5.25\text{V}$, $\text{DVDD} = 1.65\text{V to } 3.6\text{V}$
- Built-In Right Leg Drive Amplifier, Lead-Off Detection, WCT, PACE Detection, Test Signals
- Integrated Respiration Impedance Measurement (ADS1294R/6R/8R only)
- Digital PACE Detection Capability
- Built-In Oscillator and Reference
- Flexible Power-Down, Standby Modes
- SPI™-Compatible Serial Interface
- Operating Temperature Range:
 $-40^{\circ}\text{C to } +85^{\circ}\text{C}$

APPLICATIONS

- Medical Instrumentation (ECG, EMG and EEG): Patient monitoring; Holter, event, stress, and vital signs including ECG, AED, telemedicine Bispectral index (BIS), Evoked audio potential (EAP), Sleep study monitor
- High-Precision, Simultaneous, Multichannel Signal Acquisition

DESCRIPTION

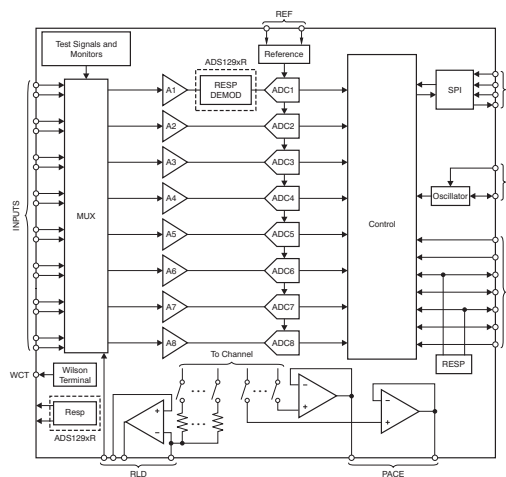
The ADS1294/6/8/4R/6R/8R are a family of multichannel, simultaneous sampling, 24-bit, delta-sigma ($\Delta\Sigma$) analog-to-digital converters (ADCs) with built-in programmable gain amplifiers (PGAs), internal reference, and an onboard oscillator. The ADS1294/6/8/4R/6R/8R incorporate all of the features that are commonly required in medical electrocardiogram (ECG) and electroencephalogram (EEG) applications.

With its high levels of integration and exceptional performance, the ADS1294/6/8/4R/6R/8R family enables the development of scalable medical instrumentation systems at significantly reduced size, power, and overall cost.

The ADS1294/6/8/4R/6R/8R have a flexible input multiplexer per channel that can be independently connected to the internally-generated signals for test, temperature, and lead-off detection. Additionally, any configuration of input channels can be selected for derivation of the right leg drive (RLD) output signal. The ADS1294/6/8/4R/6R/8R operate at data rates as high as 32kSPS, thereby allowing the implementation of software PACE detection. Lead-off detection can be implemented internal to the device, either with a pull-up/pull-down resistor or an excitation current sink/source. Three integrated amplifiers generate the Wilson Central Terminal (WCT) and the Goldberger Central Terminals (GCT) required for a standard 12-lead ECG. The ADS1294R/6R/8R versions include a fully-integrated, respiration impedance measurement function.

Multiple ADS1294/6/8/4R/6R/8R devices can be cascaded in high channel count systems in a daisy-chain configuration.

Package options include a tiny 8mm × 8mm, 64-ball BGA and a TQFP-64. The ADS1294/6/8 BGA version is specified over the commercial temperature range of $0^{\circ}\text{C to } +70^{\circ}\text{C}$. The ADS1294R/6R/8R BGA and ADS1294/6/8 TQFP versions are specified over the industrial temperature range of $-40^{\circ}\text{C to } +85^{\circ}\text{C}$.



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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

Table 1. FAMILY AND ORDERING INFORMATION⁽¹⁾

| PRODUCT | PACKAGE OPTION | NUMBER OF CHANNELS | ADC RESOLUTION | MAXIMUM SAMPLE RATE (kSPS) | OPERATING TEMPERATURE RANGE | RESPIRATION CIRCUITRY |
|----------|----------------|--------------------|----------------|----------------------------|-----------------------------|-----------------------|
| ADS1194 | BGA | 4 | 16 | 8 | 0°C to +70°C | No |
| | TQFP | 4 | 16 | 8 | 0°C to +70°C | No |
| ADS1196 | BGA | 6 | 16 | 8 | 0°C to +70°C | No |
| | TQFP | 6 | 16 | 8 | 0°C to +70°C | No |
| ADS1198 | BGA | 8 | 16 | 8 | 0°C to +70°C | No |
| | TQFP | 8 | 16 | 8 | 0°C to +70°C | No |
| ADS1294 | BGA | 4 | 24 | 32 | 0°C to +70°C | External |
| ADS1294R | BGA | 4 | 24 | 32 | –40°C to +85°C | Yes |
| ADS1294 | TQFP | 4 | 24 | 32 | –40°C to +85°C | External |
| ADS1296 | BGA | 6 | 24 | 32 | 0°C to +70°C | External |
| ADS1296R | BGA | 6 | 24 | 32 | –40°C to +85°C | Yes |
| ADS1296 | TQFP | 6 | 24 | 32 | –40°C to +85°C | External |
| ADS1298 | BGA | 8 | 24 | 32 | 0°C to +70°C | External |
| ADS1298R | BGA | 8 | 24 | 32 | –40°C to +85°C | Yes |
| ADS1298 | TQFP | 8 | 24 | 32 | –40°C to +85°C | External |

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or visit the device product folder at www.ti.com.

ABSOLUTE MAXIMUM RATINGS⁽¹⁾

Over operating free-air temperature range, unless otherwise noted.

| | | ADS1294, ADS1296, ADS1298 ADS1294R, ADS1296R, ADS1298R | UNIT |
|--|---|---|------|
| AVDD to AVSS | | –0.3 to +5.5 | V |
| DVDD to DGND | | –0.3 to +3.9 | V |
| AVSS to DGND | | –3 to +0.2 | V |
| V _{REF} input to AVSS | | AVSS – 0.3 to AVDD + 0.3 | V |
| Analog input to AVSS | | AVSS – 0.3 to AVDD + 0.3 | V |
| Digital input voltage to DGND | | –0.3 to DVDD + 0.3 | V |
| Digital output voltage to DGND | | –0.3 to DVDD + 0.3 | V |
| Input current (momentary) | | 100 | mA |
| Input current (continuous) | | 10 | mA |
| Operating temperature range | Commerical Grade: ADS1294, ADS1296, ADS1298 | 0 to +70 | °C |
| | Industrial grade: ADS1294I, ADS1296I, ADS1298I, ADS1294RI, ADS1296RI, ADS1298RI | –40 to +85 | °C |
| ESD ratings | Human body model (HBM) JEDEC standard 22, test method A114-C.01, all pins | ±2000 | V |
| | Charged device model (CDM) JEDEC standard 22, test method C101, all pins | ±500 | V |
| Storage temperature range | | –60 to +150 | °C |
| Maximum junction temperature (T _J) | | +150 | °C |

(1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.

ELECTRICAL CHARACTERISTICS

Minimum/maximum specifications apply for all commercial grade (0°C to +70°C) devices and from –40°C to +85°C for industrial grades devices. Typical specifications are at +25°C. All specifications at DVDD = 1.8V, AVDD – AVSS = 3V⁽¹⁾, V_{REF} = 2.4V, external f_{CLK} = 2.048MHz, data rate = 500SPS, High-Resolution mode, and gain = 6, unless otherwise noted.

| PARAMETER | TEST CONDITIONS | ADS1294, ADS1296, ADS1298 ADS1294R, ADS1296R, ADS1298R | | | UNIT | |
|--|--|--|-----|-----|------------------|------------------|
| | | MIN | TYP | MAX | | |
| ANALOG INPUTS | | | | | | |
| Full-scale differential input voltage (A _{INP} – A _{INN}) | | ±V _{REF} /GAIN | | | V | |
| Input common-mode range | | See the Input Common-Mode Range subsection of the PGA Settings and Input Range section | | | | |
| Input capacitance | | 20 | | | pF | |
| Input bias current | T _A = +25°C, input = 1.5V | ±200 | | | pA | |
| | T _A = 0°C to +70°C, input = 1.5V | ±1 | | | nA | |
| | T _A = –40°C to +85°C, input = 1.5V | ±1.2 | | | nA | |
| DC input impedance | No lead-off | 1000 | | | MΩ | |
| | Current source lead-off detection | 500 | | | MΩ | |
| | Pull-up resistor lead-off detection | 10 | | | MΩ | |
| PGA PERFORMANCE | | | | | | |
| Gain settings | | 1, 2, 3, 4, 6, 8, 12 | | | | |
| Bandwidth | | See Table 16 | | | | |
| ADC PERFORMANCE | | | | | | |
| Resolution | Data rates up to 8kSPS, no missing codes | 24 | | | Bits | |
| | 16kSPS data rate | 19 | | | Bits | |
| | 32kSPS data rate | 17 | | | Bits | |
| Data rate | f _{CLK} = 2.048MHz, High-Resolution mode | 500 | | | 32000 | SPS |
| | f _{CLK} = 2.048MHz, Low-Power mode | 250 | | | 16000 | SPS |
| CHANNEL PERFORMANCE | | | | | | |
| DC Performance | | | | | | |
| Input-referred noise | Gain = 6 ⁽²⁾ , 10 seconds of data | 5 | | | μV _{PP} | |
| | Gain = 6, 256 points, 0.5 seconds of data | 4 | | | 7 | μV _{PP} |
| | Gain settings other than 6, data rates other than 500SPS | See Noise Measurements section | | | | |
| Integral nonlinearity | Full-scale with gain = 6, best fit | 8 | | | ppm | |
| | Full-scale with gain = 6, best fit, ADS1294R/6R/8R channel 1 | 40 | | | ppm | |
| | –20dBFS with gain = 6, best fit, ADS1294R/6R/8R channel 1 | 8 | | | ppm | |
| Offset error | | ±500 | | | μV | |
| Offset error drift | | 2 | | | μV/°C | |
| Gain error | Excluding voltage reference error | ±0.2 | | | ±0.5 | % of FS |
| Gain drift | Excluding voltage reference drift | 5 | | | ppm/°C | |
| Gain match between channels | | 0.3 | | | % of FS | |

- (1) Performance is applicable for 5V operation as well. Production testing for limits is performed at 3V.
(2) Noise data measured in a 10-second interval. Test not performed in production. Input-referred noise is calculated with input shorted (without electrode resistance) over a 10-second interval.

ELECTRICAL CHARACTERISTICS (continued)

Minimum/maximum specifications apply for all commercial grade (0°C to +70°C) devices and from –40°C to +85°C for industrial grades devices. Typical specifications are at +25°C. All specifications at DVDD = 1.8V, AVDD – AVSS = 3V⁽¹⁾, V_{REF} = 2.4V, external f_{CLK} = 2.048MHz, data rate = 500SPS, High-Resolution mode, and gain = 6, unless otherwise noted.

| PARAMETER | TEST CONDITIONS | ADS1294, ADS1296, ADS1298 ADS1294R, ADS1296R, ADS1298R | | | UNIT |
|---|--|---|----------------------|------------|-------------------|
| | | MIN | TYP | MAX | |
| CHANNEL PERFORMANCE (continued) | | | | | |
| AC Performance | | | | | |
| Common-mode rejection ratio (CMRR) | f _{CM} = 50Hz, 60Hz ⁽³⁾ | −105 | −115 | | dB |
| Power-supply rejection ratio (PSRR) | f _{PS} = 50Hz, 60Hz | | 90 | | dB |
| Crosstalk | f _{IN} = 50Hz, 60Hz | | −126 | | dB |
| Signal-to-noise ratio (SNR) | f _{IN} = 10Hz input, gain = 6 | | 112 | | dB |
| Total harmonic distortion (THD) | 10Hz, −0.5dBFs | | −98 | | dB |
| | ADS1294R/6R/8R channel 1, 10Hz, −0.5dBFs | | −70 | | dB |
| | 100Hz, −0.5dBFs ⁽⁴⁾ | | −100 | | dB |
| | ADS1294R/6R/8R channel 1, 100Hz, −0.5dBFs ⁽⁴⁾ | | −68 | | dB |
| | ADS1294R/6R/8R channel 1, 100Hz, −20dBFs ⁽⁴⁾ | | −86 | | dB |
| DIGITAL FILTER | | | | | |
| −3dB bandwidth | | | 0.262f _{DR} | | Hz |
| Digital filter settling | Full setting | | 4 | | Conversions |
| RIGHT LEG DRIVE (RLD) AMPLIFIER AND PACE AMPLIFIERS | | | | | |
| RLD integrated noise | BW = 150Hz | | 7 | | μV _{RMS} |
| PACE integrated noise | BW = 8kHz | | 20 | | μV _{RMS} |
| PACE amplifier crosstalk | Crosstalk between PACE amplifiers | | 60 | | dB |
| Gain bandwidth product | 50kΩ 10pF load, gain = 1 | | 100 | | kHz |
| Slew rate | 50kΩ 10pF load, gain = 1 | | 0.25 | | V/μs |
| PACE and RLD amplifier drive strength | Short-circuit to GND (AVDD = 3V) | | 270 | | μA |
| | Short-circuit to supply (AVDD = 3V) | | 550 | | μA |
| | Short-circuit to GND (AVDD = 5V) | | 490 | | μA |
| | Short-circuit to supply (AVDD = 5V) | | 810 | | μA |
| PACE and RLD current | Peak swing (AVSS + 0.3V to AVDD + 0.3V) at AVDD = 3V | | 50 | | μA |
| | Peak swing (AVSS + 0.3V to AVDD + 0.3V) at AVDD = 5V | | 75 | | μA |
| PACE amplifier output resistance | | | 100 | | Ω |
| Total harmonic distortion | f _{IN} = 100Hz, gain = 1 | | −70 | | dB |
| Common-mode input range | | AVSS + 0.7 | | AVDD − 0.3 | V |
| Common-mode resistor matching | Internal 200kΩ resistor matching | | 0.1 | | % |
| Short-circuit current | | | ±0.25 | | mA |
| Quiescent power consumption | Either RLD or PACE amplifier | | 20 | | μA |
| WILSON CENTRAL TERMINAL (WCT) AMPLIFIER | | | | | |
| Integrated noise | BW = 150Hz | | See Table 6 | | nV/√Hz |
| Gain bandwidth product | | | See Table 6 | | kHz |
| Slew rate | | | See Table 6 | | V/s |
| Total harmonic distortion | f _{IN} = 100Hz | | 90 | | dB |
| Common-mode input range | | AVSS + 0.3 | | AVDD − 0.3 | V |
| Short-circuit current | Through internal 30kΩ resistor | | ±0.25 | | mA |
| Quiescent power consumption | | | See Table 6 | | μA |

(3) CMRR is measured with a common-mode signal of AVSS + 0.3V to AVDD – 0.3V. The values indicated are the maximum of the eight channels.

(4) Harmonics above the second harmonic are attenuated by the digital filter.

ELECTRICAL CHARACTERISTICS (continued)

Minimum/maximum specifications apply for all commercial grade (0°C to +70°C) devices and from –40°C to +85°C for industrial grades devices. Typical specifications are at +25°C. All specifications at DVDD = 1.8V, AVDD – AVSS = 3V⁽¹⁾, V_{REF} = 2.4V, external f_{CLK} = 2.048MHz, data rate = 500SPS, High-Resolution mode, and gain = 6, unless otherwise noted.

| PARAMETER | TEST CONDITIONS | ADS1294, ADS1296, ADS1298 ADS1294R, ADS1296R, ADS1298R | | | UNIT |
|---|--|---|-----|-------|------------------|
| | | MIN | TYP | MAX | |
| LEAD-OFF DETECT | | | | | |
| Frequency | See the Register Map section for settings | 0, f _{DR} /4 | | | kHz |
| Current | See the Register Map section for settings | 6, 12, 18, 24 | | | nA |
| Current accuracy | | ±20 | | | % |
| Comparator threshold accuracy | | ±30 | | | mV |
| RESPIRATION (ADS1294R/6R/8R Only) | | | | | |
| Frequency | Internal source | 32, 64 | | | kHz |
| | External source | 32 | | 64 | kHz |
| Phase shift | See the Register Map section for settings | 22.5 | 90 | 157.5 | Degrees |
| Impedance range | I _{RESP} = 30μA | 10 | | | kΩ |
| Impedance measurement noise | 0.05Hz to 2Hz brick wall filter, 32kHz modulation clock, phase = 112.5, using I _{RESP} = 30μA with 2kΩ baseline load, gain = 4 test condition | 20 | | | mΩ _{pp} |
| Modulator current | internal reference, signal path = 82kΩ, baseline = 2.21kΩ | 29 | | | μA |
| EXTERNAL REFERENCE | | | | | |
| Reference input voltage | 3V supply V _{REF} = (VREFP – VREFN) | 2.5 | | | V |
| | 5V supply V _{REF} = (VREFP – VREFN) | 4.0 | | | V |
| Negative input (VREFN) | | AVSS | | | V |
| Positive input (VREFP) | | AVSS + 2.5 | | | V |
| Input impedance | | 10 | | | kΩ |
| INTERNAL REFERENCE | | | | | |
| Output voltage | Register bit CONFIG3.VREF_4V = 0, AVDD ≥ 2.7V | 2.4 | | | V |
| | Register bit CONFIG3.VREF_4V = 1, AVDD ≥ 4.4V | 4.0 | | | V |
| V _{REF} accuracy | | ±0.2 | | | % |
| Internal reference drift | T _A = +25°C | 35 | | | ppm/°C |
| | Commerical grade, 0°C to +70°C | 35 | | | ppm |
| | Industrial grade, –40°C to +85°C | 45 | | | ppm |
| Start-up time | | 150 | | | ms |
| SYSTEM MONITORS | | | | | |
| Analog supply reading error | | 2 | | | % |
| Digital supply reading error | | 2 | | | % |
| Device wake up | From power-up to $\overline{\text{DRDY}}$ low | 150 | | | ms |
| | STANDBY mode | 9 | | | ms |
| Temperature sensor reading, voltage | T _A = +25°C | 145 | | | mV |
| Temperature sensor reading, coefficient | | 490 | | | μV/°C |
| Test Signal | | | | | |
| Signal frequency | See Register Map section for settings | f _{CLK} /2 ²¹ , f _{CLK} /2 ²⁰ | | | Hz |
| Signal voltage | See Register Map section for settings | ±1, ±2 | | | mV |
| Accuracy | | ±2 | | | % |

ELECTRICAL CHARACTERISTICS (continued)

Minimum/maximum specifications apply for all commercial grade (0°C to +70°C) devices and from –40°C to +85°C for industrial grades devices. Typical specifications are at +25°C. All specifications at DVDD = 1.8V, AVDD – AVSS = 3V⁽¹⁾, V_{REF} = 2.4V, external f_{CLK} = 2.048MHz, data rate = 500SPS, High-Resolution mode, and gain = 6, unless otherwise noted.

| PARAMETER | TEST CONDITIONS | ADS1294, ADS1296, ADS1298 ADS1294R, ADS1296R, ADS1298R | | | UNIT |
|--|---|---|------------|------|------|
| | | MIN | TYP | MAX | |
| CLOCK | | | | | |
| Internal oscillator clock frequency | Nominal frequency | 2.048 | | | MHz |
| Internal clock accuracy | T _A = +25°C | ±0.5 | | | % |
| | 0°C ≤ T _A ≤ +70°C | ±2 | | | % |
| | −40°C ≤ T _A ≤ +85°C, ADS1298I industrial grade version only | ±2.5 | | | % |
| Internal oscillator start-up time | | 20 | | | μs |
| Internal oscillator power consumption | | 120 | | | μW |
| External clock input frequency | CLKSEL pin = 0 | 1.94 | 2.048 | 2.25 | MHz |
| DIGITAL INPUT/OUTPUT (DVDD = 1.65V to 3.6V) | | | | | |
| Logic level | | | | | |
| V _{IH} | | 0.8DVDD | DVDD + 0.1 | | V |
| V _{IL} | | −0.1 | 0.2DVDD | | V |
| V _{OH} | I _{OH} = −500μA | DVDD − 0.4 | | | V |
| V _{OL} | I _{OL} = +500μA | 0.4 | | | V |
| Input current (I _{IN}) | 0V < V _{DigitalInput} < DVDD | −10 | +10 | | μA |
| POWER-SUPPLY REQUIREMENTS | | | | | |
| Analog supply (AVDD − AVSS) | | 2.7 | 3.0 | 5.25 | V |
| Digital supply (DVDD) | | 1.65 | 1.8 | 3.6 | V |
| AVDD − DVDD | | −2.1 | 3.6 | | V |
| SUPPLY CURRENT (RLD, WCT, and PACE Amplifiers Turned Off) | | | | | |
| High-Resolution mode (ADS1298) | | | | | |
| I _{AVDD} | AVDD − AVSS = 3V | 2.75 | | | mA |
| | AVDD − AVSS = 5V | 3.1 | | | mA |
| I _{DVDD} | DVDD = 3.0V | 0.5 | | | mA |
| | DVDD = 1.8V | 0.3 | | | mA |
| Low-Power mode (ADS1298) | | | | | |
| I _{AVDD} | AVDD − AVSS = 3V | 1.8 | | | mA |
| | AVDD − AVSS = 5V | 2.1 | | | mA |
| I _{DVDD} | DVDD = 3.0V | 0.5 | | | mA |
| | DVDD = 1.8V | 0.3 | | | mA |
| POWER DISSIPATION (Analog Supply = 3V, RLD, WCT, and PACE Amplifiers Turned Off) | | | | | |
| Quiescent power dissipation | | | | | |
| ADS1298/8R | High-Resolution mode | 8.8 9.5 | | | mW |
| | Low-Power mode (250SPS) | 6.0 7.0 | | | mW |
| ADS1296/6R | High-Resolution mode | 7.2 7.9 | | | mW |
| | Low-Power mode | 5.3 6.6 | | | mW |
| ADS1294/4R | High-Resolution mode | 5.4 6 | | | mW |
| | Low-Power mode | 4.1 4.4 | | | mW |
| Power-down | | 10 | | | μW |
| Standby mode | | 2 | | | mW |
| Quiescent channel power | PGA + ADC | 818 | | | μW |

ELECTRICAL CHARACTERISTICS (continued)

Minimum/maximum specifications apply for all commercial grade (0°C to +70°C) devices and from –40°C to +85°C for industrial grades devices. Typical specifications are at +25°C. All specifications at DVDD = 1.8V, AVDD – AVSS = 3V⁽¹⁾, V_{REF} = 2.4V, external f_{CLK} = 2.048MHz, data rate = 500SPS, High-Resolution mode, and gain = 6, unless otherwise noted.

| PARAMETER | TEST CONDITIONS | ADS1294, ADS1296, ADS1298 ADS1294R, ADS1296R, ADS1298R | | | UNIT |
|--|----------------------|---|-----|------|------|
| | | MIN | TYP | MAX | |
| POWER DISSIPATION (Analog Supply = 5V, RLD, WCT, and PACE Amplifiers Turned Off) | | | | | |
| Quiescent power dissipation | | | | | |
| ADS1298/8R | High-Resolution mode | 17.5 | | | mW |
| | Low-Power mode | 12.5 | | | mW |
| ADS1296/6R | High-Resolution mode | 14.1 | | | mW |
| | Low-Power mode | 10 | | | mW |
| ADS1294/4R | High-Resolution mode | 10.1 | | | mW |
| | Low-Power mode | 8.3 | | | mW |
| Power-down | | 20 | | | μW |
| Standby mode | | 4 | | | mW |
| Quiescent channel power | PGA + ADC | 1.5 | | | mW |
| TEMPERATURE | | | | | |
| Specified temperature range | | 0 | | +70 | °C |
| Operating temperature range | | 0 | | +70 | °C |
| Specified temperature range (industrial grade only) | | −40 | | +85 | °C |
| Operating temperature range (industrial grade only) | | −40 | | +85 | °C |
| Storage temperature range | | −60 | | +150 | °C |

THERMAL INFORMATION

| THERMAL METRIC ⁽¹⁾ | | ADS1294/6/8 | ADS1294/6/8/ 4R/6R/8R | UNITS |
|-------------------------------|--|-------------|--------------------------|-------|
| | | PAG | ZXG | |
| | | 64 PINS | 64 PINS | |
| θ _{JA} | Junction-to-ambient thermal resistance | 35 | 48 | °C/W |
| θ _{JCtop} | Junction-to-case (top) thermal resistance | 31 | 8 | |
| θ _{JB} | Junction-to-board thermal resistance | 26 | 25 | |
| ψ _{JT} | Junction-to-top characterization parameter | 0.1 | 0.5 | |
| ψ _{JB} | Junction-to-board characterization parameter | — | 22 | |
| θ _{JCbot} | Junction-to-case (bottom) thermal resistance | — | — | |

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

NOISE MEASUREMENTS

NOTE

The ADS1294R/6R/8R channel performance differs from the ADS1294/6/8 in regards to respiration circuitry found on channel one. Unless otherwise noted, ADS129x refers to all specifications and functional descriptions of the ADS1294, ADS1296, ADS1298, ADS1294R, ADS1296R, and ADS1298R.

The ADS129x noise performance can be optimized by adjusting the data rate and PGA setting. As the averaging is increased by reducing the data rate, the noise drops correspondingly. Increasing the PGA value reduces the input-referred noise, which is particularly useful when measuring low-level biopotential signals. [Table 2](#) and [Table 3](#) summarize the noise performance of the ADS129x in the High-Resolution (HR) mode and Low-Power (LP) mode, respectively, with a 3V analog power supply. [Table 4](#) and [Table 5](#) summarize the noise performance of the ADS129x in the HR mode and LP mode, respectively, with a 5V analog power supply. The data are representative of typical noise performance at $T_A = +25^\circ\text{C}$. The data shown are the result of averaging the readings from multiple devices and are measured with the inputs shorted together. A minimum of 1000 consecutive readings are used to calculate the RMS and peak-to-peak noise for each reading. For the two highest data rates, the noise is limited by quantization noise of the ADC and does not have a gaussian distribution. Thus, the ratio between rms noise and peak-to-peak noise is approximately 10. For the lower data rates, the ratio is approximately 6.6.

[Table 2](#) to [Table 5](#) show measurements taken with an internal reference. The data are also representative of the ADS129x noise performance when using a low-noise external reference such as the [REF5025](#).

**Table 2. Input-Referred Noise ($\mu\text{V}_{\text{RMS}}/\mu\text{V}_{\text{PP}}$) in High-Resolution Mode
3V Analog Supply and 2.4V Reference⁽¹⁾**

| DR BITS OF CONFIG1 REGISTER | OUTPUT DATA RATE (SPS) | -3dB BANDWIDTH (Hz) | PGA GAIN = 1 | PGA GAIN = 2 | PGA GAIN = 3 | PGA GAIN = 4 | PGA GAIN = 6 | PGA GAIN = 8 | PGA GAIN = 12 |
|-----------------------------------|------------------------------|---------------------------|-----------------|-----------------|-----------------|-----------------|-----------------|-----------------|------------------|
| 000 | 32000 | 8398 | 335/3553 | 168/1701 | 112/1100 | 85/823 | 58/529 | 42.5/378 | 28.6/248 |
| 001 | 16000 | 4193 | 56/613 | 28/295 | 18.8/188 | 14.3/143 | 9.7/94 | 7.4/69 | 5.2/44.3 |
| 010 | 8000 | 2096 | 12.4/111 | 6.5/54 | 4.5/37.9 | 3.5/29.7 | 2.6/21.7 | 2.2/17.8 | 1.8/13.8 |
| 011 | 4000 | 1048 | 6.1/44.8 | 3.2/23.3 | 2.4/17.1 | 1.9/14.0 | 1.5/11.1 | 1.3/9.7 | 1.2/8.5 |
| 100 | 2000 | 524 | 4.1/27.8 | 2.2/15.4 | 1.6/11.0 | 1.3/9.1 | 1.1/7.3 | 1.0/6.5 | 0.9/6.0 |
| 101 | 1000 | 262 | 2.9/19.0 | 1.6/10.1 | 1.2/7.5 | 1.0/6.2 | 0.8/5.0 | 0.7/4.6 | 0.6/4.1 |
| 110 | 500 | 131 | 2.1/12.5 | 1.1/6.8 | 0.9/5.1 | 0.7/4.3 | 0.6/3.5 | 0.5/3.1 | 0.5/2.9 |

(1) At least 1000 consecutive readings were used to calculate the RMS and peak-to-peak noise values in this table.

**Table 3. Input-Referred Noise ($\mu\text{V}_{\text{RMS}}/\mu\text{V}_{\text{PP}}$) in Low-Power Mode
3V Analog Supply and 2.4V Reference⁽¹⁾**

| DR BITS OF CONFIG1 REGISTER | OUTPUT DATA RATE (SPS) | -3dB BANDWIDTH (Hz) | PGA GAIN = 1 | PGA GAIN = 2 | PGA GAIN = 3 | PGA GAIN = 4 | PGA GAIN = 6 | PGA GAIN = 8 | PGA GAIN = 12 |
|-----------------------------------|------------------------------|---------------------------|-----------------|-----------------|-----------------|-----------------|-----------------|-----------------|------------------|
| 000 | 16000 | 4193 | 333/3481 | 166/1836 | 111/1168 | 84/834 | 56/576 | 42/450 | 28/284 |
| 001 | 8000 | 2096 | 56/554 | 28/272 | 19/177 | 14.3/133 | 9.7/85 | 7.4/64 | 5.0/42.4 |
| 010 | 4000 | 1048 | 12.5/99 | 6.5/51 | 4.5/35.0 | 3.4/25.9 | 2.4/18.8 | 2.0/14.5 | 1.5/11.3 |
| 011 | 2000 | 524 | 6.1/41.8 | 3.2/22.2 | 2.3/15.9 | 1.8/12.1 | 1.4/9.3 | 1.2/7.8 | 1.0/6.7 |
| 100 | 1000 | 262 | 4.1/26.3 | 2.2/14.6 | 1.6/9.9 | 1.3/8.1 | 1.0/6.2 | 0.8/5.4 | 0.7/4.7 |
| 101 | 500 | 131 | 3.0/17.9 | 1.6/9.8 | 1.1/6.8 | 0.9/5.7 | 0.7/4.2 | 0.6/3.6 | 0.5/3.4 |
| 110 | 250 | 65 | 2.1/11.9 | 1.1/6.3 | 0.8/4.6 | 0.7/4.0 | 0.5/3.0 | 0.5/2.6 | 0.4/2.4 |

(1) At least 1000 consecutive readings were used to calculate the RMS and peak-to-peak noise values in this table.

**Table 4. Input-Referred Noise ($\mu V_{RMS}/\mu V_{PP}$) in High-Resolution Mode
5V Analog Supply and 4V Reference⁽¹⁾**

| DR BITS OF CONFIG1 REGISTER | OUTPUT DATA RATE (SPS) | –3dB BANDWIDTH (Hz) | PGA GAIN = 1 | PGA GAIN = 2 | PGA GAIN = 3 | PGA GAIN = 4 | PGA GAIN = 6 | PGA GAIN = 8 | PGA GAIN = 12 |
|-----------------------------------|------------------------------|---------------------------|-----------------|-----------------|-----------------|-----------------|-----------------|-----------------|------------------|
| 000 | 32000 | 8398 | 521/5388 | 260/2900 | 173/1946 | 130/1403 | 87/917 | 65/692 | 44/483 |
| 001 | 16000 | 4193 | 86/1252 | 43/633 | 29/402 | 22/298 | 15/206 | 11/141 | 7/91 |
| 010 | 8000 | 2096 | 17/207 | 9/112 | 6/71 | 4/57 | 3/36 | 3/29 | 2/18 |
| 011 | 4000 | 1048 | 6.4/48.2 | 3.4/25.9 | 2.4/17.7 | 1.9/15.4 | 1.5/11.2 | 1.3/9.6 | 1.1/8.2 |
| 100 | 2000 | 524 | 4.2/29.9 | 2.3/15.9 | 1.6/11.1 | 1.3/9.3 | 1.0/7.5 | 0.9/6.6 | 0.8/5.8 |
| 101 | 1000 | 262 | 2.9/18.8 | 1.6/10.4 | 1.1/7.8 | 0.9/6.1 | 0.7/4.9 | 0.6/4.7 | 0.6/3.9 |
| 110 | 500 | 131 | 2.0/12.8 | 1.1/7.2 | 0.8/5.2 | 0.7/4.0 | 0.5/3.3 | 0.5/3.3 | 0.4/2.7 |

(1) At least 1000 consecutive readings were used to calculate the RMS and peak-to-peak noise values in this table.

**Table 5. Input-Referred Noise ($\mu V_{RMS}/\mu V_{PP}$) in Low-Power Mode
5V Analog Supply and 4V Reference⁽¹⁾**

| DR BITS OF CONFIG1 REGISTER | OUTPUT DATA RATE (SPS) | –3dB BANDWIDTH (Hz) | PGA GAIN = 1 | PGA GAIN = 2 | PGA GAIN = 3 | PGA GAIN = 4 | PGA GAIN = 6 | PGA GAIN = 8 | PGA GAIN = 12 |
|-----------------------------------|------------------------------|---------------------------|-----------------|-----------------|-----------------|-----------------|-----------------|-----------------|------------------|
| 000 | 16000 | 4193 | 526/5985 | 263/2953 | 175/1918 | 132/1410 | 88/896 | 66/681 | 44/458 |
| 001 | 8000 | 2096 | 88/1201 | 44/619 | 29/411 | 22/280 | 15/191 | 11/139 | 7/83 |
| 010 | 4000 | 1048 | 17/208 | 9/103 | 6/62 | 4/52 | 3/37 | 2/25 | 2/16 |
| 011 | 2000 | 524 | 6.0/41.1 | 3.3/23.3 | 2.2/15.5 | 1.8/12.3 | 1.3/9.8 | 1.1/7.8 | 0.9/6.5 |
| 100 | 1000 | 262 | 4.1/27.1 | 2.3/14.8 | 1.5/10.1 | 1.2/8.1 | 0.9/6.0 | 0.8/5.4 | 0.7/4.4 |
| 101 | 500 | 131 | 2.9/17.4 | 1.6/9.6 | 1.1/6.6 | 0.9/5.9 | 0.7/4.3 | 0.6/3.4 | 0.5/3.2 |
| 110 | 250 | 65 | 2.1/11.9 | 1.1/6.6 | 0.8/4.6 | 0.6/3.7 | 0.5/3.0 | 0.4/2.5 | 0.4/2.2 |

(1) At least 1000 consecutive readings were used to calculate the RMS and peak-to-peak noise values in this table.

Table 6. Typical WCT Performance

| PARAMETER | ANY ONE (A, B, or C) | ANY TWO (A+B, A+C, or B+C) | ALL THREE (A+B+C) | UNIT |
|------------------|-------------------------|-------------------------------|----------------------|-------------------|
| Integrated noise | 540 | 382 | 312 | nV _{RMS} |
| Power | 53 | 59 | 65 | μW |
| –3dB BW | 30 | 59 | 89 | kHz |
| Slew rate | BW limited | BW limited | BW limited | — |

PIN CONFIGURATIONS

ZXG PACKAGE BGA-64 (TOP VIEW, SOLDER BUMPS ON BOTTOM SIDE)

| H | G | F | E | D | C | B | A | |
|-------|-----------|-----------------|-----------------|------|--------|--------|-------|---|
| IN1P | IN2P | IN3P | IN4P | IN5P | IN6P | IN7P | IN8P | 1 |
| IN1N | IN2N | IN3N | IN4N | IN5N | IN6N | IN7N | IN8N | 2 |
| VREFP | VCAP4 | TESTN_PACE_OUT2 | TESTP_PACE_OUT1 | WCT | RLDINV | RLDOUT | RLDIN | 3 |
| VREFN | RESP_MODP | RESP_MODN | RESV1 | AVSS | RLDREF | AVDD | AVDD | 4 |
| VCAP1 | PWDN | GPIO1 | GPIO4 | AVSS | AVSS | AVSS | AVSS | 5 |
| VCAP2 | RESET | DAISY_IN | GPIO3 | DRDY | AVDD | AVDD | AVDD | 6 |
| DGND | START | CS | GPIO2 | DGND | DGND | VCAP3 | AVDD1 | 7 |
| DIN | CLK | SCLK | DOUT | DVDD | DVDD | CLKSEL | AVSS1 | 8 |

Table 7. BGA PIN ASSIGNMENTS

| NAME | TERMINAL | FUNCTION | DESCRIPTION |
|---------------------|----------|--------------|--|
| IN8P ⁽¹⁾ | 1A | Analog input | Differential analog positive input 8 (ADS1298/8R) |
| IN7P ⁽¹⁾ | 1B | Analog input | Differential analog positive input 7 (ADS1298/8R) |
| IN6P ⁽¹⁾ | 1C | Analog input | Differential analog positive input 6 (ADS1296/8/6R/8R) |
| IN5P ⁽¹⁾ | 1D | Analog input | Differential analog positive input 5 (ADS1296/8/6R/8R) |
| IN4P ⁽¹⁾ | 1E | Analog input | Differential analog positive input 4 |
| IN3P ⁽¹⁾ | 1F | Analog input | Differential analog positive input 3 |
| IN2P ⁽¹⁾ | 1G | Analog input | Differential analog positive input 2 |
| IN1P ⁽¹⁾ | 1H | Analog input | Differential analog positive input 1 |
| IN8N ⁽¹⁾ | 2A | Analog input | Differential analog negative input 8 (ADS1298/8R) |
| IN7N ⁽¹⁾ | 2B | Analog input | Differential analog negative input 7 (ADS1298/8R) |
| IN6N ⁽¹⁾ | 2C | Analog input | Differential analog negative input 6 (ADS1296/8/6R/8R) |
| IN5N ⁽¹⁾ | 2D | Analog input | Differential analog negative input 5 (ADS1296/8/6R/8R) |
| IN4N ⁽¹⁾ | 2E | Analog input | Differential analog negative input 4 |
| IN3N ⁽¹⁾ | 2F | Analog input | Differential analog negative input 3 |
| IN2N ⁽¹⁾ | 2G | Analog input | Differential analog negative input 2 |
| IN1N ⁽¹⁾ | 2H | Analog input | Differential analog negative input 1 |

(1) Connect unused terminals to AVDD.

Table 7. BGA PIN ASSIGNMENTS (continued)

| NAME | TERMINAL | FUNCTION | DESCRIPTION |
|--------------------------------|----------|----------------------------|--|
| RLDIN ⁽²⁾ | 3A | Analog input | Right leg drive input to MUX |
| RLDOUT | 3B | Analog output | Right leg drive output |
| RLDINV | 3C | Analog input/output | Right leg drive inverting input |
| WCT | 3D | Analog output | Wilson Central Terminal output |
| TESTP_PACE_OUT1 ⁽²⁾ | 3E | Analog input/buffer output | Internal test signal/single-ended buffer output based on register settings |
| TESTN_PACE_OUT2 ⁽²⁾ | 3F | Analog input/output | Internal test signal/single-ended buffer output based on register settings |
| VCAP4 | 3G | Analog output | Analog bypass capacitor |
| VREFP | 3H | Analog input/output | Positive reference voltage |
| AVDD | 4A | Supply | Analog supply |
| AVDD | 4B | Supply | Analog supply |
| RLDREF | 4C | Analog input | Right leg drive noninverting input |
| AVSS | 4D | Supply | Analog ground |
| RESV1 | 4E | Digital input | Reserved for future use; must tie to logic low (DGND). |
| RESP_MODN | 4F | Analog output | ADS1294R/6R/8R: modulation clock for respiration measurement, negative side. ADS1294/6/8: leave floating. |
| RESP_MODP | 4G | Analog output | ADS1294R/6R/8R: modulation clock for respiration measurement, positive side. ADS1294/6/8: leave floating. |
| VREFN | 4H | Analog input | Negative reference voltage |
| AVSS | 5A | Supply | Analog ground |
| AVSS | 5B | Supply | Analog ground |
| AVSS | 5C | Supply | Analog ground |
| AVSS | 5D | Supply | Analog ground |
| GPIO4 | 5E | Digital input/output | GPIO4 in normal mode |
| GPIO1 | 5F | Digital input/output | General purpose input/output pin |
| PWDN | 5G | Digital input | Power-down; active low |
| VCAP1 | 5H | Analog input/output | Analog bypass capacitor |
| AVDD | 6A | Supply | Analog supply |
| AVDD | 6B | Supply | Analog supply |
| AVDD | 6C | Supply | Analog supply |
| DRDY | 6D | Digital output | Data ready; active low |
| GPIO3 | 6E | Digital input/output | GPIO3 in normal mode |
| DAISY_IN | 6F | Digital input | Daisy-chain input; if not used, short to DGND. |
| RESET | 6G | Digital input | System reset; active low |
| VCAP2 | 6H | — | Analog bypass capacitor |
| AVDD1 | 7A | Supply | Analog supply for charge pump |
| VCAP3 | 7B | — | Analog bypass capacitor; internally generated AVDD + 1.9V. |
| DGND | 7C | Supply | Digital ground |
| DGND | 7D | Supply | Digital ground |
| GPIO2 | 7E | Digital input/output | General-purpose input/output pin |
| CS | 7F | Digital input | SPI chip select; active low |
| START | 7G | Digital input | Start conversion |
| DGND | 7H | Supply | Digital ground |
| AVSS1 | 8A | Supply | Analog ground for charge pump |
| CLKSEL | 8B | Digital input | Master clock select |
| DVDD | 8C | Supply | Digital power supply |
| DVDD | 8D | Supply | Digital power supply |
| DOUT | 8E | Digital output | SPI data out |
| SCLK | 8F | Digital input | SPI clock |
| CLK | 8G | Digital input/output | External Master clock input or internal clock output. |
| DIN | 8H | Digital input | SPI data in |

(2) Connect unused terminals to AVDD.

**PAG PACKAGE
TQFP-64
(TOP VIEW)**

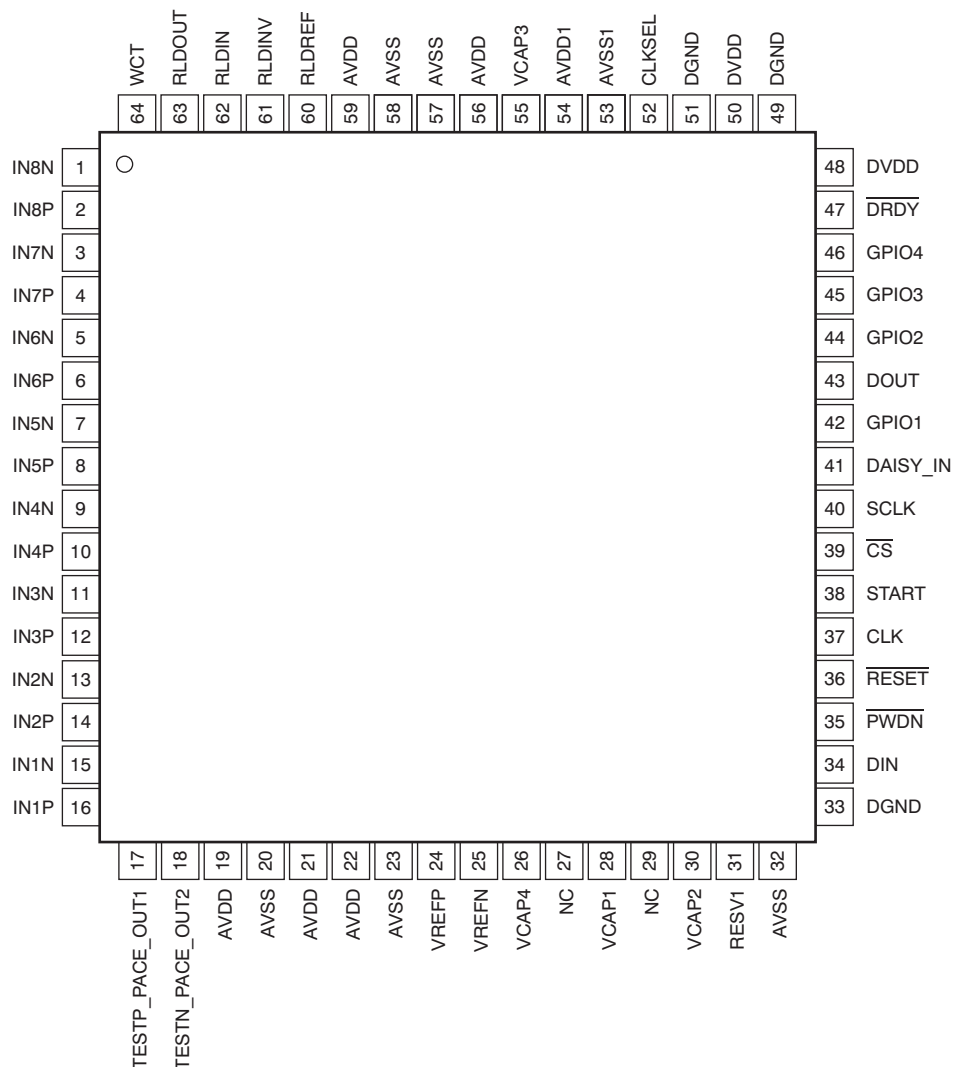


Table 8. PAG PIN ASSIGNMENTS

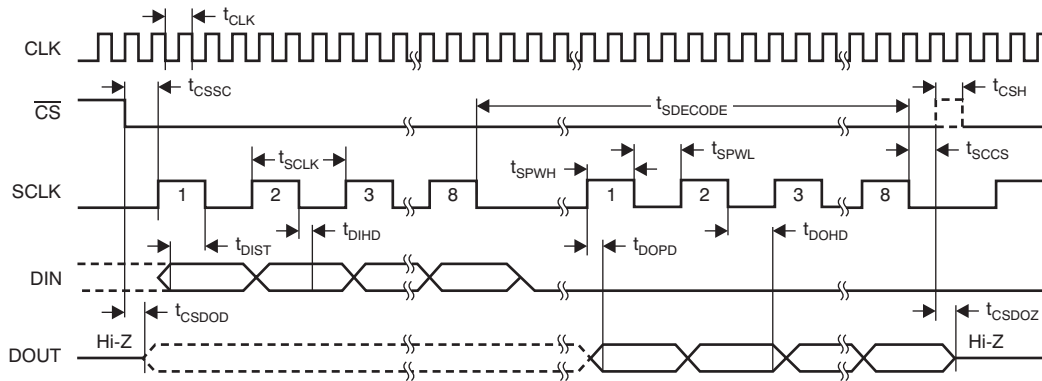
| NAME | PIN | FUNCTION | DESCRIPTION |
|--------------------------------|-----|----------------------------|--|
| IN8N ⁽¹⁾ | 1 | Analog input | Differential analog negative input 8 (ADS1298) |
| IN8P ⁽¹⁾ | 2 | Analog input | Differential analog positive input 8 (ADS1298) |
| IN7N ⁽¹⁾ | 3 | Analog input | Differential analog negative input 7 (ADS1298) |
| IN7P ⁽¹⁾ | 4 | Analog input | Differential analog positive input 7 (ADS1298) |
| IN6N ⁽¹⁾ | 5 | Analog input | Differential analog negative input 6 (ADS1296/8) |
| IN6P ⁽¹⁾ | 6 | Analog input | Differential analog positive input 6 (ADS1296/8) |
| IN5N ⁽¹⁾ | 7 | Analog input | Differential analog negative input 5 (ADS1296/8) |
| IN5P ⁽¹⁾ | 8 | Analog input | Differential analog positive input 5 (ADS1296/8) |
| IN4N ⁽¹⁾ | 9 | Analog input | Differential analog negative input 4 |
| IN4P ⁽¹⁾ | 10 | Analog input | Differential analog positive input 4 |
| IN3N ⁽¹⁾ | 11 | Analog input | Differential analog negative input 3 |
| IN3P ⁽¹⁾ | 12 | Analog input | Differential analog positive input 3 |
| IN2N ⁽¹⁾ | 13 | Analog input | Differential analog negative input 2 |
| IN2P ⁽¹⁾ | 14 | Analog input | Differential analog positive input 2 |
| IN1N ⁽¹⁾ | 15 | Analog input | Differential analog negative input 1 |
| IN1P ⁽¹⁾ | 16 | Analog input | Differential analog positive input 1 |
| TESTP_PACE_OUT1 ⁽¹⁾ | 17 | Analog input/buffer output | Internal test signal/single-ended buffer output based on register settings |
| TESTN_PACE_OUT2 ⁽¹⁾ | 18 | Analog input/output | Internal test signal/single-ended buffer output based on register settings |
| AVDD | 19 | Supply | Analog supply |
| AVSS | 20 | Supply | Analog ground |
| AVDD | 21 | Supply | Analog supply |
| AVDD | 22 | Supply | Analog supply |
| AVSS | 23 | Supply | Analog ground |
| VREFP | 24 | Analog input/output | Positive reference voltage |
| VREFN | 25 | Analog input | Negative reference voltage |
| VCAP4 | 26 | Analog output | Analog bypass capacitor |
| NC | 27 | — | No connection, can be connected to AVDD or AVSS with a 10 kΩ resistor |
| VCAP1 | 28 | — | Analog bypass capacitor |
| NC | 29 | — | No connection, can be connected to AVDD or AVSS with a 10 kΩ resistor |
| VCAP2 | 30 | — | Analog bypass capacitor |
| RESV1 | 31 | Digital input | Reserved for future use; must tie to logic low (DGND). |
| AVSS | 32 | Supply | Analog ground |
| DGND | 33 | Supply | Digital ground |
| DIN | 34 | Digital input | SPI data in |
| $\overline{\text{PWDN}}$ | 35 | Digital input | Power-down; active low |
| $\overline{\text{RESET}}$ | 36 | Digital input | System reset; active low |
| CLK | 37 | Digital input/output | External Master clock input or internal clock output. |
| START | 38 | Digital input | Start conversion |
| $\overline{\text{CS}}$ | 39 | Digital input | SPI chip select; active low |
| SCLK | 40 | Digital input | SPI clock |
| DAISY_IN | 41 | Digital input | Daisy-chain input; if not used, short to DGND. |
| GPIO1 | 42 | Digital input/output | General-purpose input/output pin |
| DOUT | 43 | Digital output | SPI data out |
| GPIO2 | 44 | Digital input/output | General-purpose input/output pin |
| GPIO3 | 45 | Digital input/output | General-purpose input/output pin |
| GPIO4 | 46 | Digital input/output | General-purpose input/output pin |
| $\overline{\text{DRDY}}$ | 47 | Digital output | Data ready; active low |
| DVDD | 48 | Supply | Digital power supply |
| DGND | 49 | Supply | Digital ground |
| DVDD | 50 | Supply | Digital power supply |

(1) Connect unused terminals to AVDD.

Table 8. PAG PIN ASSIGNMENTS (continued)

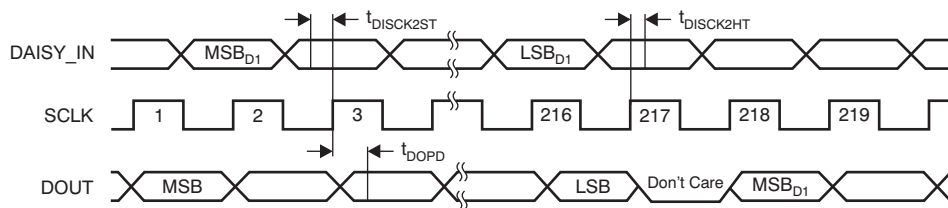
| NAME | PIN | FUNCTION | DESCRIPTION |
|----------------------|-----|---------------------|--|
| DGND | 51 | Supply | Digital ground |
| CLKSEL | 52 | Digital input | Master clock select |
| AVSS1 | 53 | Supply | Analog ground |
| AVDD1 | 54 | Supply | Analog supply |
| VCAP3 | 55 | Analog | Analog bypass capacitor; internally generated AVDD + 1.9V. |
| AVDD | 56 | Supply | Analog supply |
| AVSS | 57 | Supply | Analog ground |
| AVSS | 58 | Supply | Analog ground |
| AVDD | 59 | Supply | Analog supply |
| RLDREF | 60 | Analog input | Right leg drive noninverting input |
| RLDINV | 61 | Analog input/output | Right leg drive inverting input |
| RLDIN ⁽¹⁾ | 62 | Analog input | Right leg drive input to MUX |
| RLDOUT | 63 | Analog output | Right leg drive output |
| WCT | 64 | Analog output | Wilson Central Terminal output |

TIMING CHARACTERISTICS



NOTE: SPI settings are CPOL = 0 and CPHA = 1.

Figure 1. Serial Interface Timing



NOTE: Daisy-chain timing shown for eight-channel ADS1298, ADS1298R, and ADS1298L.

Figure 2. Daisy-Chain Interface Timing

TIMING REQUIREMENTS FOR [Figure 1](#) AND [Figure 2](#)

Specifications apply from –40°C to +85°C, unless otherwise noted. Load on D_{OUT} = 20pF || 100kΩ.

| PARAMETER | DESCRIPTION | 2.7V ≤ DVDD ≤ 3.6V | | | 1.65V ≤ DVDD ≤ 2V | | | UNIT |
|-----------------------|--|--------------------|-----|-----|-------------------|-----|-----|-------------------|
| | | MIN | TYP | MAX | MIN | TYP | MAX | |
| t _{CLK} | Master clock period | 414 | | 514 | 414 | | 514 | ns |
| t _{CSSC} | \overline{CS} low to first SCLK, setup time | 6 | | | 17 | | | ns |
| t _{SCLK} | SCLK period | 50 | | | 66.6 | | | ns |
| t _{SPWH, L} | SCLK pulse width, high and low | 15 | | | 25 | | | ns |
| t _{DIST} | DIN valid to SCLK falling edge: setup time | 10 | | | 10 | | | ns |
| t _{DIHD} | Valid DIN after SCLK falling edge: hold time | 10 | | | 11 | | | ns |
| t _{DOHD} | SCLK falling edge to invalid DOUT: hold time | 10 | | | 10 | | | ns |
| t _{DOPD} | SCLK rising edge to DOUT valid: setup time | | | 17 | | | 32 | ns |
| t _{CSH} | \overline{CS} high pulse | 2 | | | 2 | | | t _{CLKs} |
| t _{CSDOD} | \overline{CS} low to DOUT driven | 10 | | | 20 | | | ns |
| t _{SCCS} | Eighth SCLK falling edge to \overline{CS} high | 4 | | | 4 | | | t _{CLKs} |
| t _{SDECODE} | Command decode time | 4 | | | 4 | | | t _{CLKs} |
| t _{CSDOZ} | \overline{CS} high to DOUT Hi-Z | | | 10 | | | 20 | ns |
| t _{DISCK2ST} | DAISY_IN valid to SCLK rising edge: setup time | 10 | | | 10 | | | ns |
| t _{DISCK2HT} | DAISY_IN valid after SCLK rising edge: hold time | 10 | | | 10 | | | ns |

TYPICAL CHARACTERISTICS

All plots at $T_A = +25^\circ\text{C}$, $AVDD = 3\text{V}$, $AVSS = 0\text{V}$, $DVDD = 1.8\text{V}$, internal $VREFP = 2.4\text{V}$, $VREFN = AVSS$, external clock = 2.048MHz , data rate = 500SPS , High-Resolution mode, and gain = 6, unless otherwise noted.

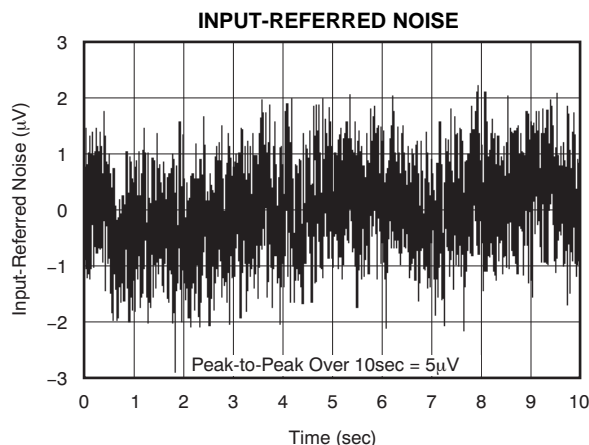


Figure 3.

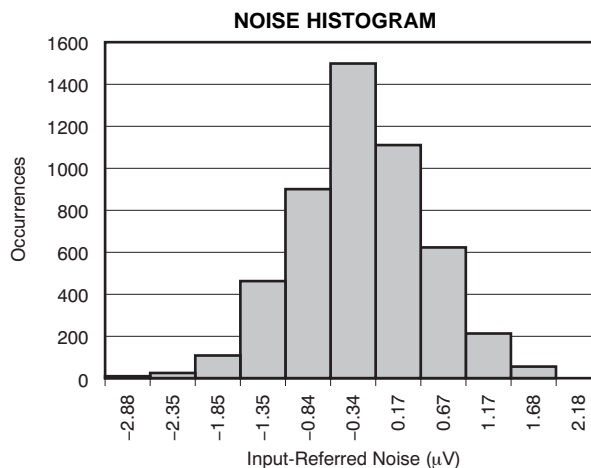


Figure 4.

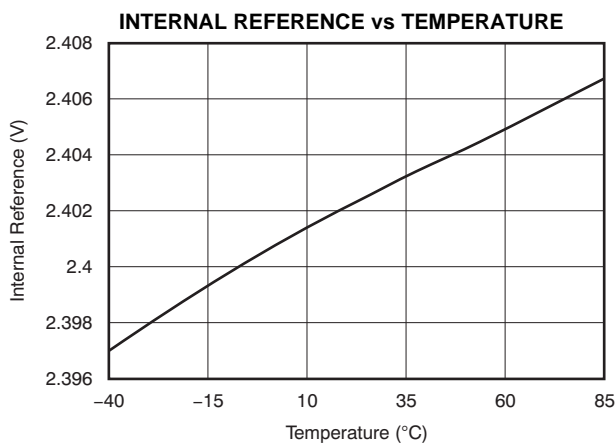


Figure 5.

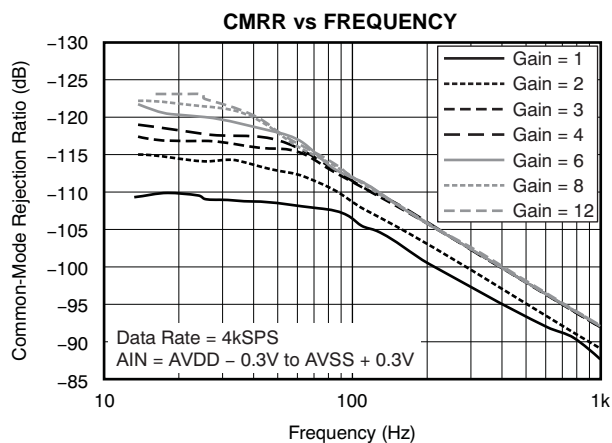


Figure 6.

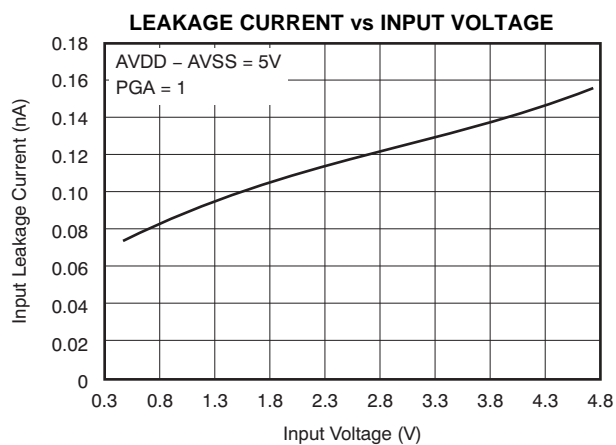


Figure 7.

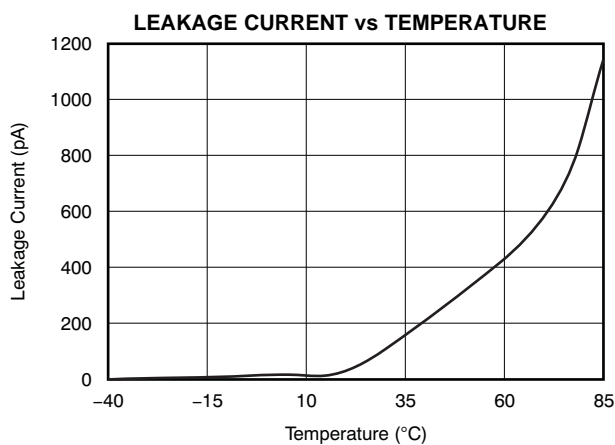


Figure 8.

TYPICAL CHARACTERISTICS (continued)

All plots at $T_A = +25^\circ\text{C}$, $AV_{DD} = 3\text{V}$, $AV_{SS} = 0\text{V}$, $DV_{DD} = 1.8\text{V}$, internal $V_{REFP} = 2.4\text{V}$, $V_{REFN} = AV_{SS}$, external clock = 2.048MHz, data rate = 500SPS, High-Resolution mode, and gain = 6, unless otherwise noted.

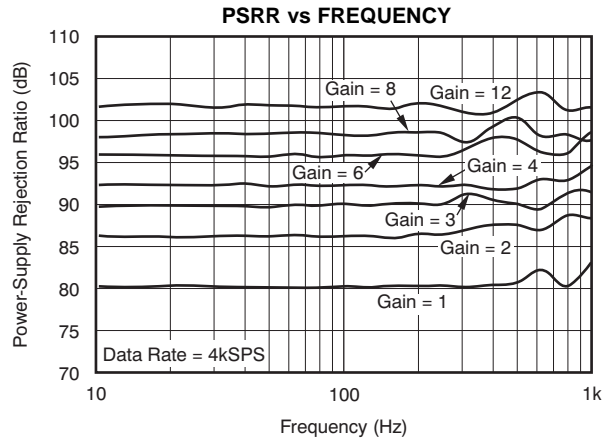


Figure 9.

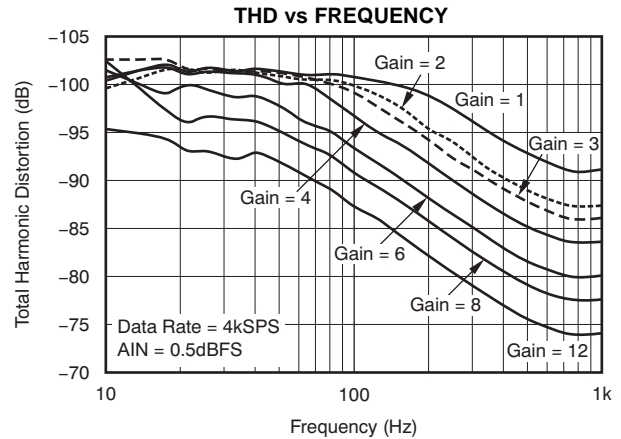


Figure 10.

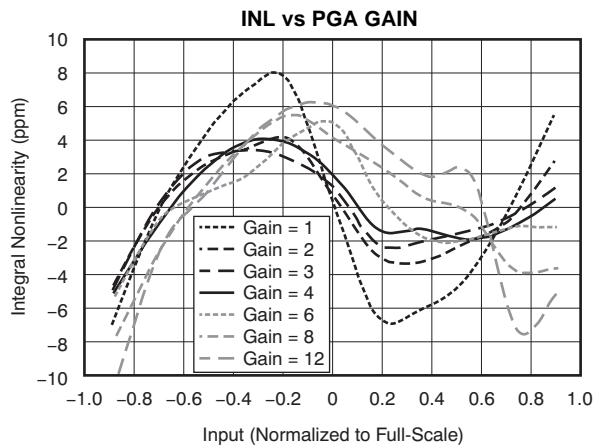


Figure 11.

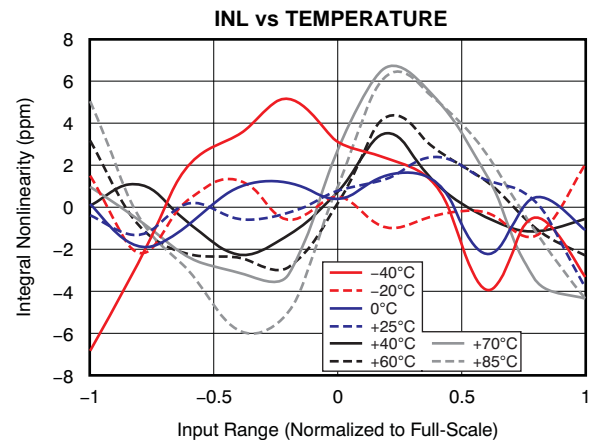


Figure 12.

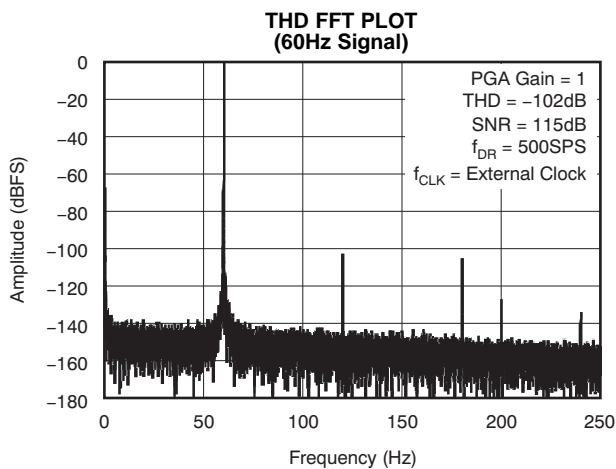


Figure 13.

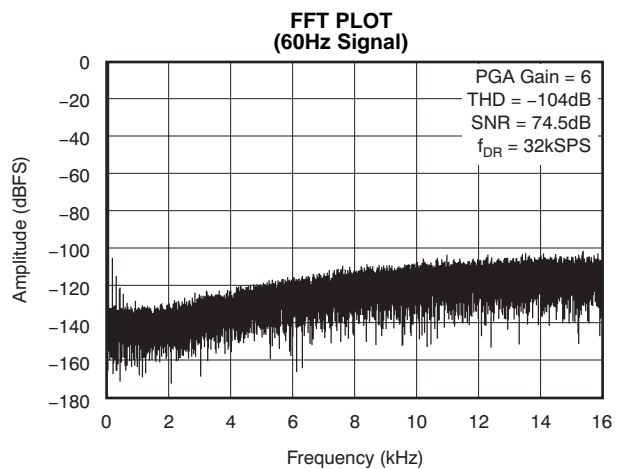


Figure 14.

TYPICAL CHARACTERISTICS (continued)

All plots at $T_A = +25^\circ\text{C}$, $AVDD = 3\text{V}$, $AVSS = 0\text{V}$, $DVDD = 1.8\text{V}$, internal $VREFP = 2.4\text{V}$, $VREFN = AVSS$, external clock = 2.048MHz, data rate = 500SPS, High-Resolution mode, and gain = 6, unless otherwise noted.

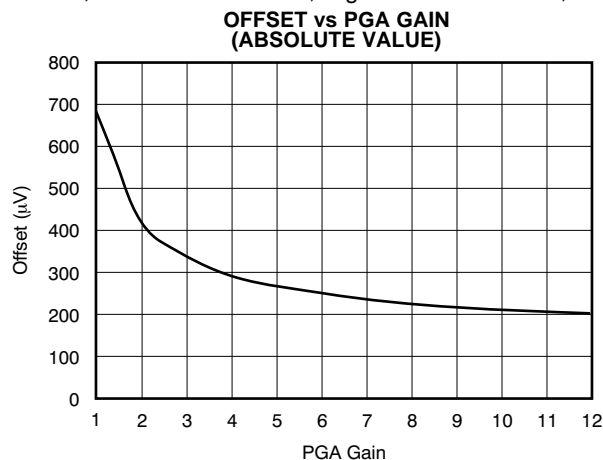


Figure 15.

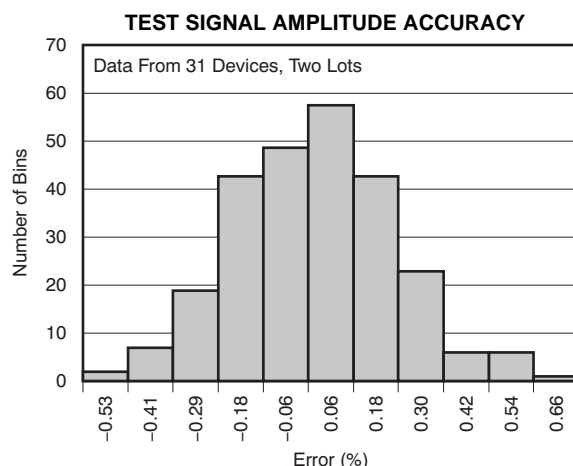


Figure 16.

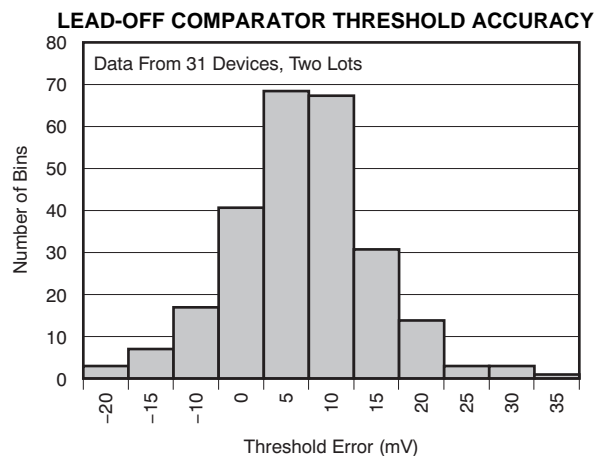


Figure 17.

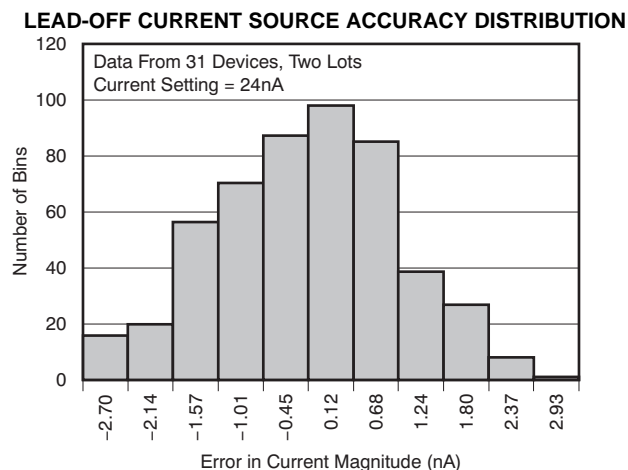


Figure 18.

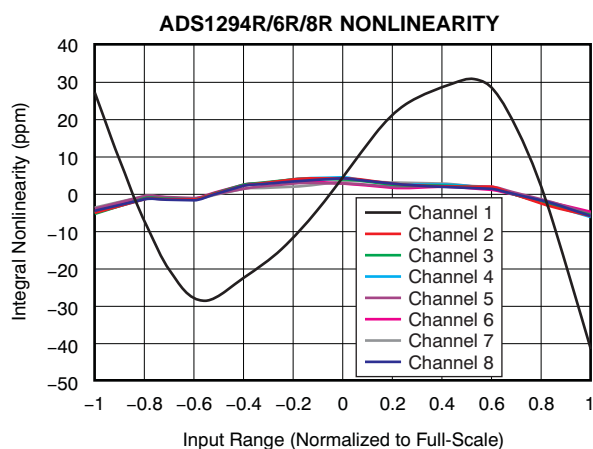


Figure 19.

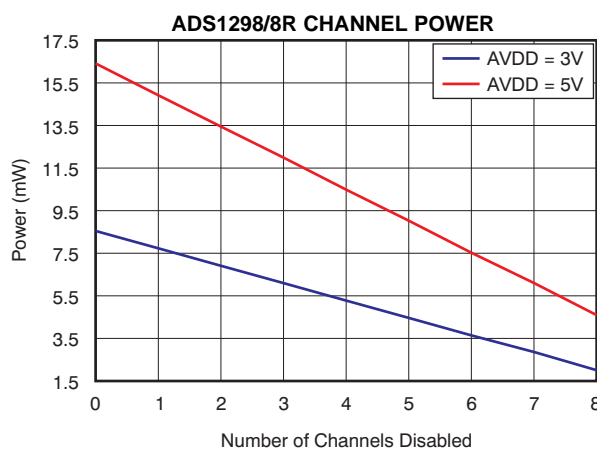


Figure 20.

TYPICAL CHARACTERISTICS (continued)

All plots at $T_A = +25^\circ\text{C}$, $AVDD = 3\text{V}$, $AVSS = 0\text{V}$, $DVDD = 1.8\text{V}$, internal $V_{REFP} = 2.4\text{V}$, $V_{REFN} = AVSS$, external clock = 2.048MHz, data rate = 500SPS, High-Resolution mode, and gain = 6, unless otherwise noted.

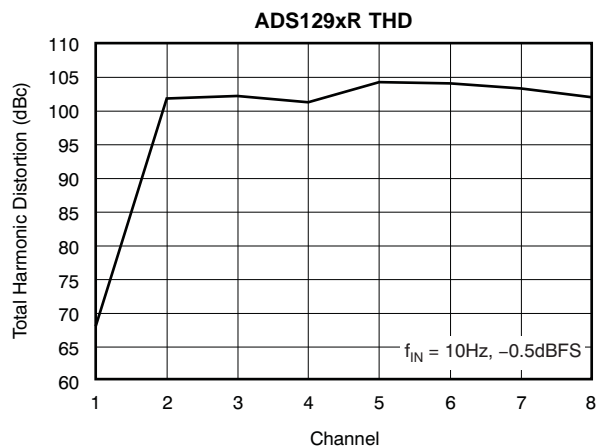


Figure 21.

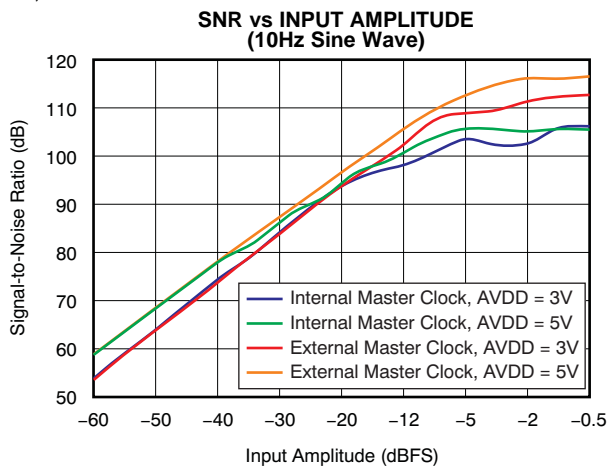


Figure 22.

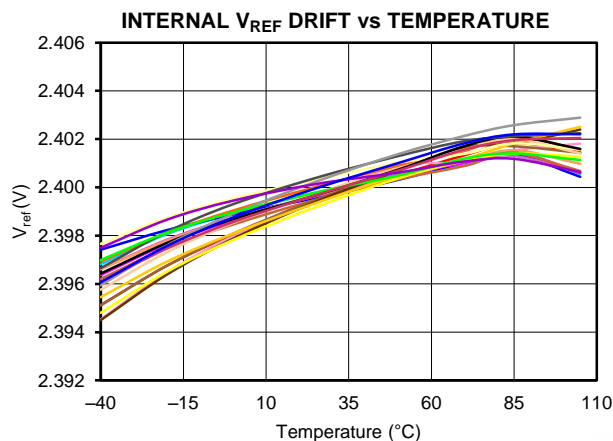


Figure 23.

OVERVIEW

NOTE

The ADS1294R/6R/8R channel performance differs from the ADS1294/6/8 in regards to respiration circuitry found on channel one. Unless otherwise noted, ADS129x refers to all specifications and functional descriptions of the ADS1294, ADS1296, ADS1298, ADS1294R, ADS1296R, and ADS1298R.

The ADS129x are low-power, multichannel, simultaneously-sampling, 24-bit delta-sigma ($\Delta\Sigma$) analog-to-digital converters (ADCs) with integrated programmable gain amplifiers (PGAs). These devices integrate various ECG-specific functions that make them well-suited for scalable electrocardiogram (ECG), electroencephalography (EEG), and electromyography (EMG) applications. The devices can also be used in high-performance, multichannel data acquisition systems by powering down the ECG-specific circuitry.

The ADS129x have a highly programmable multiplexer that allows for temperature, supply, input short, and RLD measurements. Additionally, the multiplexer allows any of the input electrodes to be programmed as the patient reference drive. The PGA gain can be chosen from one of seven settings (1, 2, 3, 4, 6, 8, and 12). The ADCs in the device offer data rates from 250SPS to 32kSPS. Communication to the device is accomplished using an SPI-compatible interface. The device provides four GPIO pins for general use. Multiple devices can be synchronized using the START pin.

The internal reference can be programmed to either 2.4V or 4V. The internal oscillator generates a 2.048MHz clock. The versatile right leg drive (RLD) block allows the user to choose the average of any combination of electrodes to generate the patient drive signal. Lead-off detection can be accomplished either by using a pull-up/pull-down resistor or a current source/sink. An internal ac lead-off detection feature is also available. The device supports both hardware PACE detection and software PACE detection. The Wilson Central Terminal (WCT) block can be used to generate the WCT point of the standard 12-lead ECG.

Additionally, the ADS1294R, ADS1296R, and ADS1298R provide options for an internal respiration modulator and a demodulator circuit in the signal path of channel 1.

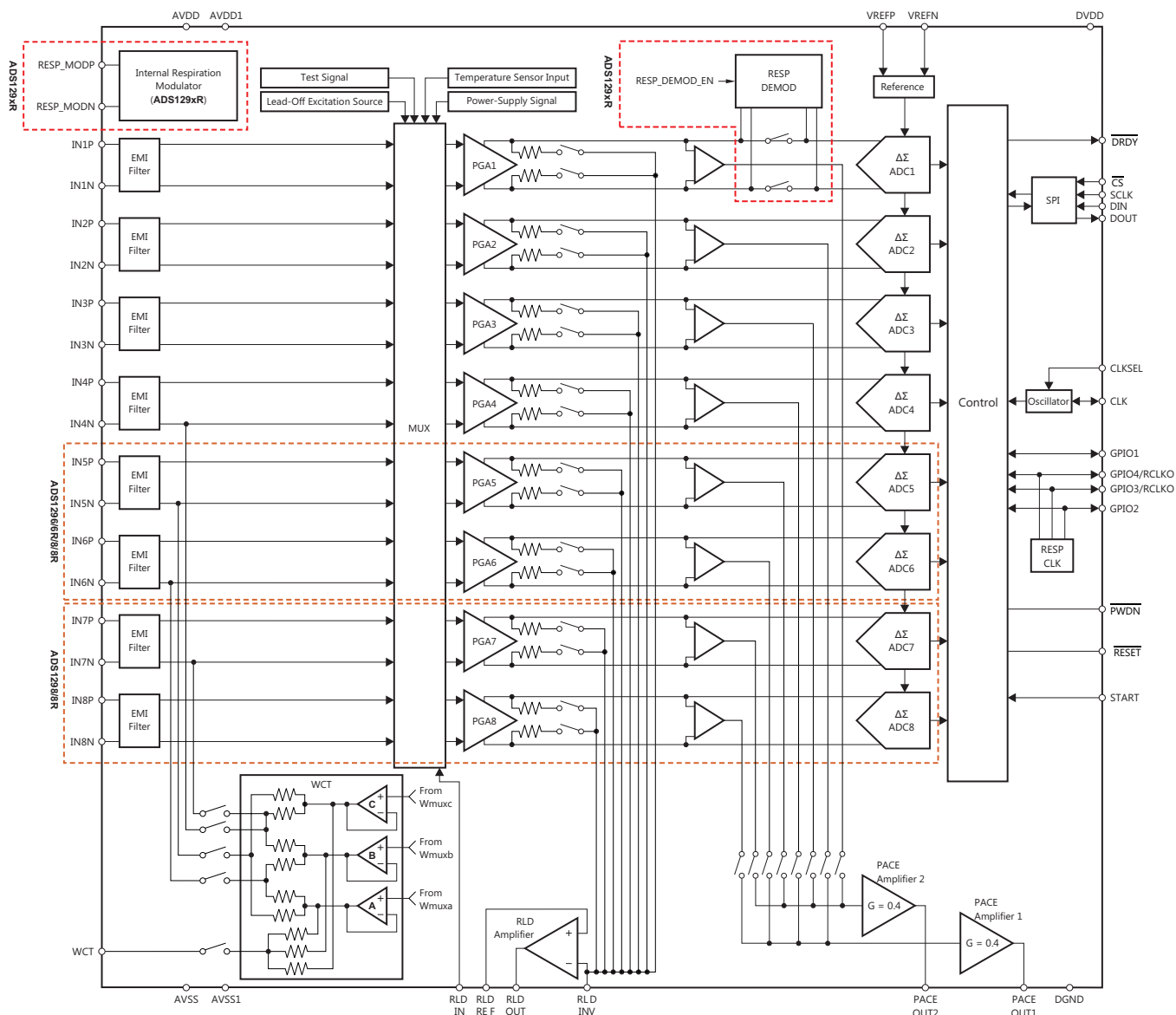


Figure 24. Functional Block Diagram

REGISTER MAP

Table 9 lists the various ADS129x registers.

Table 9. Register Assignments

| ADDRESS | REGISTER | RESET VALUE (Hex) | BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|--|---------------------------|-------------------------|----------------------|----------------------|----------------------|----------------------|-------------|------------|---------------|------------|
| Device Settings (Read-Only Registers) | | | | | | | | | | |
| 00h | ID | xx | DEV_ID7 | DEV_ID6 | DEV_ID5 | 1 | 0 | DEV_ID2 | DEV_ID1 | DEV_ID0 |
| Global Settings Across Channels | | | | | | | | | | |
| 01h | CONFIG1 | 06 | HR | DAISY_EN | CLK_EN | 0 | 0 | DR2 | DR1 | DR0 |
| 02h | CONFIG2 | 40 | 0 | 0 | WCT_CHOP | INT_TEST | 0 | TEST_AMP | TEST_FREQ1 | TEST_FREQ0 |
| 03h | CONFIG3 | 40 | PD_REFBUF | 1 | VREF_4V | RLD_MEAS | RLDREF_INT | PD_RLD | RLD_LOFF_SENS | RLD_STAT |
| 04h | LOFF | 00 | COMP_TH2 | COMP_TH1 | COMP_TH0 | VLEAD_OFF_EN | ILEAD_OFF1 | ILEAD_OFF0 | FLEAD_OFF1 | FLEAD_OFF0 |
| Channel-Specific Settings | | | | | | | | | | |
| 05h | CH1SET | 00 | PD1 | GAIN12 | GAIN11 | GAIN10 | 0 | MUXn2 | MUXn1 | MUXn0 |
| 06h | CH2SET | 00 | PD2 | GAIN22 | GAIN21 | GAIN20 | 0 | MUX22 | MUX21 | MUX20 |
| 07h | CH3SET | 00 | PD3 | GAIN32 | GAIN31 | GAIN30 | 0 | MUX32 | MUX31 | MUX30 |
| 08h | CH4SET | 00 | PD4 | GAIN42 | GAIN41 | GAIN40 | 0 | MUX42 | MUX41 | MUX40 |
| 09h | CH5SET ⁽¹⁾ | 00 | PD5 | GAIN52 | GAIN51 | GAIN50 | 0 | MUX52 | MUX51 | MUX50 |
| 0Ah | CH6SET ⁽¹⁾ | 00 | PD6 | GAIN62 | GAIN61 | GAIN60 | 0 | MUX62 | MUX61 | MUX60 |
| 0Bh | CH7SET ⁽¹⁾ | 00 | PD7 | GAIN72 | GAIN71 | GAIN70 | 0 | MUX72 | MUX71 | MUX70 |
| 0Ch | CH8SET ⁽¹⁾ | 00 | PD8 | GAIN82 | GAIN81 | GAIN80 | 0 | MUX82 | MUX81 | MUX80 |
| 0Dh | RLD_SENSP ⁽²⁾ | 00 | RLD8P ⁽¹⁾ | RLD7P ⁽¹⁾ | RLD6P ⁽¹⁾ | RLD5P ⁽¹⁾ | RLD4P | RLD3P | RLD2P | RLD1P |
| 0Eh | RLD_SENSN ⁽²⁾ | 00 | RLD8N ⁽¹⁾ | RLD7N ⁽¹⁾ | RLD6N ⁽¹⁾ | RLD5N ⁽¹⁾ | RLD4N | RLD3N | RLD2N | RLD1N |
| 0Fh | LOFF_SENSP ⁽²⁾ | 00 | LOFF8P | LOFF7P | LOFF6P | LOFF5P | LOFF4P | LOFF3P | LOFF2P | LOFF1P |
| 10h | LOFF_SENSN ⁽²⁾ | 00 | LOFF8N | LOFF7N | LOFF6N | LOFF5N | LOFF4N | LOFF3N | LOFF2N | LOFF1N |
| 11h | LOFF_FLIP | 00 | LOFF_FLIP8 | LOFF_FLIP7 | LOFF_FLIP6 | LOFF_FLIP5 | LOFF_FLIP4 | LOFF_FLIP3 | LOFF_FLIP2 | LOFF_FLIP1 |
| Lead-Off Status Registers (Read-Only Registers) | | | | | | | | | | |
| 12h | LOFF_STATP | 00 | IN8P_OFF | IN7P_OFF | IN6P_OFF | IN5P_OFF | IN4P_OFF | IN3P_OFF | IN2P_OFF | IN1P_OFF |
| 13h | LOFF_STATN | 00 | IN8N_OFF | IN7N_OFF | IN6N_OFF | IN5N_OFF | IN4N_OFF | IN3N_OFF | IN2N_OFF | IN1N_OFF |
| GPIO and OTHER Registers | | | | | | | | | | |
| 14h | GPIO | 0F | GPIO4 | GPIO3 | GPIO2 | GPIO1 | GPIOC4 | GPIOC3 | GPIOC2 | GPIOC1 |
| 15h | PACE | 00 | 0 | 0 | 0 | PACEE1 | PACEE0 | PACEO1 | PACEO0 | PD_PACE |
| 16h | RESP | 00 | RESP_DEMOD_EN1 | RESP_MOD_EN1 | 1 | RESP_PH2 | RESP_PH1 | RESP_PH0 | RESP_CTRL1 | RESP_CTRL0 |
| 17h | CONFIG4 | 00 | RESP_FREQ2 | RESP_FREQ1 | RESP_FREQ0 | 0 | SINGLE_SHOT | WCT_TO_RLD | PD_LOFF_COMP | 0 |
| 18h | WCT1 | 00 | aVF_CH6 | aVL_CH5 | aVR_CH7 | avR_CH4 | PD_WCTA | WCTA2 | WCTA1 | WCTA0 |
| 19h | WCT2 | 00 | PD_WCTC | PD_WCTB | WCTB2 | WCTB1 | WCTB0 | WCTC2 | WCTC1 | WCTC0 |

- (1) CH5SET and CH6SET are not available for the ADS1294/4R. CH7SET and CH8SET registers are not available for the ADS1294/4R and ADS1296/6R.
- (2) The RLD_SENSP, PACE_SENSP, LOFF_SENSP, LOFF_SENSN, and LOFF_FLIP registers bits[5:4] are not available for the ADS1294/4R. Bits[7:6] are not available for the ADS1294/6/4R/6R.

User Register Description

ID: ID Control Register (Factory-Programmed, Read-Only)

Address = 00h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|---------|---------|---------|-------|-------|---------|---------|---------|
| DEV_ID7 | DEV_ID6 | DEV_ID5 | 1 | 0 | DEV_ID2 | DEV_ID1 | DEV_ID0 |

The ID Control Register is programmed during device manufacture to indicate device characteristics.

Bits[7:5] DEV_ID[7:5]: Device family identification

These bits indicate the device family.

000 = Reserved

011 = Reserved

100 = ADS129x device family

101 = Reserved

110 = ADS129xR device family

111 = Reserved

Bit 4 This bit reads high.

Bit 3 This bit reads low.

Bits[2:0] DEV_ID[2:0]: Channel number identification

These bits indicates number of channels.

000 = 4-channel ADS1294 or ADS1294R

001 = 6-channel ADS1296 or ADS1296R

010 = 8-channel ADS1298 or ADS1298R

011 = Reserved

111 = Reserved

CONFIG1: Configuration Register 1

Address = 01h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|-------|----------|--------|-------|-------|-------|-------|-------|
| HR | DAISY_EN | CLK_EN | 0 | 0 | DR2 | DR1 | DR0 |

Bit 7 HR: High-Resolution/Low-Power mode

This bit determines whether the device runs in Low-Power or High-Resolution mode.

0 = Low-Power mode (default)

1 = High-Resolution mode

Bit 6 DAISY_EN: Daisy-chain/multiple readback mode

This bit determines which mode is enabled.

0 = Daisy-chain mode (default)

1 = Multiple readback mode

Bit 5 CLK_EN: CLK connection⁽¹⁾

This bit determines if the internal oscillator signal is connected to the CLK pin when the CLKSEL pin = 1.

0 = Oscillator clock output disabled (default)

1 = Oscillator clock output enabled

Bits[4:3] Must always be set to '0'

Bits[2:0] DR[2:0]: Output data rate

For High-Resolution mode, $f_{MOD} = f_{CLK}/4$. For low power mode, $f_{MOD} = f_{CLK}/8$.

These bits determine the output data rate of the device.

(1) Additional power is consumed when driving external devices.

| BIT | DATA RATE | HIGH-RESOLUTION MODE ⁽¹⁾ | LOW-POWER MODE ⁽²⁾ |
|---------------|----------------|-------------------------------------|-------------------------------|
| 000 | $f_{MOD}/16$ | 32kSPS | 16kSPS |
| 001 | $f_{MOD}/32$ | 16kSPS | 8kSPS |
| 010 | $f_{MOD}/64$ | 8kSPS | 4kSPS |
| 011 | $f_{MOD}/128$ | 4kSPS | 2kSPS |
| 100 | $f_{MOD}/256$ | 2kSPS | 1kSPS |
| 101 | $f_{MOD}/512$ | 1kSPS | 500SPS |
| 110 (default) | $f_{MOD}/1024$ | 500SPS | 250SPS |
| 111 | Do not use | n/a | n/a |

(1) Additional power is consumed when driving external devices.

(2) $f_{CLK} = 2.048\text{MHz}$.

CONFIG2: Configuration Register 2

Address = 02h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|-------|-------|----------|----------|-------|----------|------------|------------|
| 0 | 0 | WCT_CHOP | INT_TEST | 0 | TEST_AMP | TEST_FREQ1 | TEST_FREQ0 |

Configuration Register 2 configures the test signal generation. See the [Input Multiplexer](#) section for more details.

Bits[7:6] **Must always be set to '0'**

Bit 5 **WCT_CHOP: WCT chopping scheme**

This bit determines whether the chopping frequency of WCT amplifiers is variable or fixed.

0 = Chopping frequency varies, see [Table 17](#)

1 = Chopping frequency constant at $f_{MOD}/16$

Bit 4 **INT_TEST: TEST source**

This bit determines the source for the Test signal.

0 = Test signals are driven externally (default)

1 = Test signals are generated internally

Bit 3 **Must always be set to '0'**

Bit 2 **TEST_AMP: Test signal amplitude**

These bits determine the calibration signal amplitude.

0 = $1 \times -(V_{REFP} - V_{REFN})/2400$ V (default)

1 = $2 \times -(V_{REFP} - V_{REFN})/2400$ V

Bits[1:0] **TEST_FREQ[1:0]: Test signal frequency**

These bits determine the calibration signal frequency.

00 = Pulsed at $f_{CLK}/2^{21}$ (default)

01 = Pulsed at $f_{CLK}/2^{20}$

10 = Not used

11 = At dc

CONFIG3: Configuration Register 3

Address = 03h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|--------------------------------|-------|---------|----------|------------|-----------------------------|---------------|----------|
| $\overline{\text{PD_REFBUF}}$ | 1 | VREF_4V | RLD_MEAS | RLDREF_INT | $\overline{\text{PD_RLD}}$ | RLD_LOFF_SENS | RLD_STAT |

Configuration Register 3 configures multi-reference and RLD operation.

Bit 7 **$\overline{\text{PD_REFBUF}}$: Power-down reference buffer**

This bit determines the power-down reference buffer state.
 0 = Power-down internal reference buffer (default)
 1 = Enable internal reference buffer

Bit 6 **Must always be set to '1'**

Default is '1' at power-up.

Bit 5 **VREF_4V: Reference voltage**

This bit determines the reference voltage, VREFP.
 0 = VREFP is set to 2.4V (default)
 1 = VREFP is set to 4V (use only with a 5V analog supply)

Bit 4 **RLD_MEAS: RLD measurement**

This bit enables RLD measurement. The RLD signal may be measured with any channel.
 0 = Open (default)
 1 = RLD_IN signal is routed to the channel that has the MUX_Setting 010 (V_{REF})

Bit 3 **RLDREF_INT: RLDREF signal**

This bit determines the RLDREF signal source.
 0 = RLDREF signal fed externally (default)
 1 = RLDREF signal ($\text{AVDD} - \text{AVSS}$)/2 generated internally

Bit 2 **$\overline{\text{PD_RLD}}$: RLD buffer power**

This bit determines the RLD buffer power state.
 0 = RLD buffer is powered down (default)
 1 = RLD buffer is enabled

Bit 1 **RLD_LOFF_SENS: RLD sense function**

This bit enables the RLD sense function.
 0 = RLD sense is disabled (default)
 1 = RLD sense is enabled

Bit 0 **RLD_STAT: RLD lead-off status**

This bit determines the RLD status.
 0 = RLD is connected (default)
 1 = RLD is not connected

LOFF: Lead-Off Control Register

Address = 04h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|----------|----------|----------|--------------|------------|------------|------------|------------|
| COMP_TH2 | COMP_TH1 | COMP_TH0 | VLEAD_OFF_EN | ILEAD_OFF1 | ILEAD_OFF0 | FLEAD_OFF1 | FLEAD_OFF0 |

The Lead-Off Control Register configures the Lead-Off detection operation.

Bits[7:5] COMP_TH[2:0]: Lead-off comparator threshold

These bits determine the lead-off comparator threshold level setting. See the [Lead-Off Detection](#) subsection of the [ECG-Specific Functions](#) section for a detailed description.

Comparator positive side

000 = 95% (default)
001 = 92.5%
010 = 90%
011 = 87.5%
100 = 85%
101 = 80%
110 = 75%
111 = 70%

Comparator negative side

000 = 5% (default)
001 = 7.5%
010 = 10%
011 = 12.5%
100 = 15%
101 = 20%
110 = 25%
111 = 30%

Bit 4 VLEAD_OFF_EN: Lead-off detection mode

This bit determines the lead-off detection mode.
0 = Current source mode lead-off (default)
1 = Pull-up/pull-down resistor mode lead-off

Bits[3:2] ILEAD_OFF[1:0]: Lead-off current magnitude

These bits determine the magnitude of current for the current lead-off mode.
00 = 6nA (default)
01 = 12nA
10 = 18nA
11 = 24nA

Bits[1:0] FLEAD_OFF[1:0]: Lead-off frequency

These bits determine the frequency of lead-off detect for each channel.
00 = When any bits of the LOFF_SENSP or LOFF_SENSN registers are turned on, make sure that FLEAD[1:0] are either set to 01 or 11 (default)
01 = AC lead-off detection at $f_{DR}/4$
10 = Do not use
11 = DC lead-off detection turned on

CHnSET: Individual Channel Settings (n = 1 : 8)

Address = 05h to 0Ch

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|-------|-------|-------|-------|-------|-------|-------|-------|
| PD | GAIN2 | GAIN1 | GAIN0 | 0 | MUXn2 | MUXn1 | MUXn0 |

The CH[1:8]SET Control Register configures the power mode, PGA gain, and multiplexer settings channels. See the [Input Multiplexer](#) section for details. CH[2:8]SET are similar to CH1SET, corresponding to the respective channels.

Bit 7 PD: Power-down

This bit determines the channel power mode for the corresponding channel.

0 = Normal operation (default)

1 = Channel power-down.

When powering down a channel, it is recommended that the channel be set to input short by setting the appropriate MUXn[2:0] = 001 of the CHnSET register.

Bits[6:4] GAIN[2:0]: PGA gain

These bits determine the PGA gain setting.

000 = 6 (default)

001 = 1

010 = 2

011 = 3

100 = 4

101 = 8

110 = 12

Bit 3 Always write '0'

Bits[2:0] MUXn[2:0]: Channel input

These bits determine the channel input selection.

000 = Normal electrode input (default)

001 = Input shorted (for offset or noise measurements)

010 = Used in conjunction with RLD_MEAS bit for RLD measurements. See the [Right Leg Drive \(RLD DC Bias Circuit\)](#) subsection of the [ECG-Specific Functions](#) section for more details.

011 = MVDD for supply measurement

100 = Temperature sensor

101 = Test signal

110 = RLD_DRP (positive electrode is the driver)

111 = RLD_DRN (negative electrode is the driver)

RLD_SENSP

Address = 0Dh

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|-------|-------|-------|-------|-------|-------|-------|-------|
| RLD8P | RLD7P | RLD6P | RLD5P | RLD4P | RLD3P | RLD2P | RLD1P |

This register controls the selection of the positive signals from each channel for right leg drive derivation. See the [Right Leg Drive \(RLD DC Bias Circuit\)](#) subsection of the [ECG-Specific Functions](#) section for details.

Note that registers bits[5:4] are not available for the ADS1294/4R. Bits[7:6] are not available for the ADS1294/6/4R/6R.

RLD_SENSN

Address = 0Eh

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|-------|-------|-------|-------|-------|-------|-------|-------|
| RLD8N | RLD7N | RLD6N | RLD5N | RLD4N | RLD3N | RLD2N | RLD1N |

This register controls the selection of the negative signals from each channel for right leg drive derivation. See the [Right Leg Drive \(RLD DC Bias Circuit\)](#) subsection of the [ECG-Specific Functions](#) section for details.

Note that registers bits[5:4] are not available for the ADS1294/4R. Bits[7:6] are not available for the ADS1294/6/4R/6R.

LOFF_SENSP

Address = 0Fh

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|--------|--------|--------|--------|--------|--------|--------|--------|
| LOFF8P | LOFF7P | LOFF6P | LOFF5P | LOFF4P | LOFF3P | LOFF2P | LOFF1P |

This register selects the positive side from each channel for lead-off detection. See the [Lead-Off Detection](#) subsection of the [ECG-Specific Functions](#) section for details. Note that the LOFF_STATP register bits are only valid if the corresponding LOFF_SENSP bits are set to '1'.

Note that registers bits[5:4] are not available for the ADS1294/4R. Bits[7:6] are not available for the ADS1294/6/4R/6R.

LOFF_SENSN

Address = 10h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|--------|--------|--------|--------|--------|--------|--------|--------|
| LOFF8N | LOFF7N | LOFF6N | LOFF5N | LOFF4N | LOFF3N | LOFF2N | LOFF1N |

This register selects the negative side from each channel for lead-off detection. See the [Lead-Off Detection](#) subsection of the [ECG-Specific Functions](#) section for details. Note that the LOFF_STATN register bits are only valid if the corresponding LOFF_SENSN bits are set to '1'.

Note that registers bits[5:4] are not available for the ADS1294/4R. Bits[7:6] are not available for the ADS1294/6/4R/6R.

LOFF_FLIP

Address = 11h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|------------|------------|------------|------------|------------|------------|------------|------------|
| LOFF_FLIP8 | LOFF_FLIP7 | LOFF_FLIP6 | LOFF_FLIP5 | LOFF_FLIP4 | LOFF_FLIP3 | LOFF_FLIP2 | LOFF_FLIP1 |

This register controls the direction of the current used for lead-off derivation. See the [Lead-Off Detection](#) subsection of the [ECG-Specific Functions](#) section for details.

LOFF_STATP (Read-Only Register)

Address = 12h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|----------|----------|----------|----------|----------|----------|----------|----------|
| IN8P_OFF | IN7P_OFF | IN6P_OFF | IN5P_OFF | IN4P_OFF | IN3P_OFF | IN2P_OFF | IN1P_OFF |

This register stores the status of whether the positive electrode on each channel is on or off. See the [Lead-Off Detection](#) subsection of the [ECG-Specific Functions](#) section for details. Ignore the LOFF_STATP values if the corresponding LOFF_SENSP bits are not set to '1'.

'0' is lead-on (default) and '1' is lead-off. When the LOFF_SENSEP bits are '0', the LOFF_STATP bits should be ignored.

LOFF_STATN (Read-Only Register)

Address = 13h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|----------|----------|----------|----------|----------|----------|----------|----------|
| IN8N_OFF | IN7N_OFF | IN6N_OFF | IN5N_OFF | IN4N_OFF | IN3N_OFF | IN2N_OFF | IN1N_OFF |

This register stores the status of whether the negative electrode on each channel is on or off. See the [Lead-Off Detection](#) subsection of the [ECG-Specific Functions](#) section for details. Ignore the LOFF_STATN values if the corresponding LOFF_SENSE bits are not set to '1'.

'0' is lead-on (default) and '1' is lead-off. When the LOFF_SENSE bits are '0', the LOFF_STATP bits should be ignored.

GPIO: General-Purpose I/O Register

Address = 14h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|--------|--------|--------|--------|--------|--------|--------|--------|
| GPIOD4 | GPIOD3 | GPIOD2 | GPIOD1 | GPIOC4 | GPIOC3 | GPIOC2 | GPIOC1 |

The General-Purpose I/O Register controls the action of the three GPIO pins. Note that when RESP_CTRL[1:0] is in mode 01 and 11, the GPIO2, GPIO3, and GPIO4 pins are not available for use.

Bits[7:4] GPIOD[4:1]: GPIO data

These bits are used to read and write data to the GPIO ports.

When reading the register, the data returned correspond to the state of the GPIO external pins, whether they are programmed as inputs or as outputs. As outputs, a write to the GPIOD sets the output value. As inputs, a write to the GPIOD has no effect. GPIO is not available in certain respiration modes.

Bits[3:0] GPIOC[4:1]: GPIO control (corresponding GPIOD)

These bits determine if the corresponding GPIOD pin is an input or output.

0 = Output

1 = Input (default)

PACE: PACE Detect Register

Address = 15h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|-------|-------|-------|--------|--------|--------|--------|------------------------------|
| 0 | 0 | 0 | PACEE1 | PACEE0 | PACEO1 | PACEO0 | $\overline{\text{PD_PACE}}$ |

This register provides the PACE controls that configure the channel signal used to feed the external PACE detect circuitry. See the [PACE Detect](#) subsection of the [ECG-Specific Functions](#) section for details.

Bits[7:5] Must always be set to '0'

Bits[4:3] PACEE[1:0]: PACE even channels

These bits control the selection of the even number channels available on TEST_PACE_OUT1. Note that only one channel may be selected at any time.

00 = Channel 2 (default)

01 = Channel 4

10 = Channel 6, ADS1296/8/6R/8R only

11 = Channel 8, ADS1298/8R only

Bits[2:1] PACEO[1:0]: PACE odd channels

These bits control the selection of the odd number channels available on TEST_PACE_OUT2. Note that only one channel may be selected at any time.

00 = Channel 1 (default)

01 = Channel 3

10 = Channel 5, ADS1296/8/6R/8R only

11 = Channel 7, ADS1298/8R only

Bit 0 $\overline{\text{PD_PACE}}$: PACE detect buffer

This bit is used to enable/disable the PACE detect buffer.

0 = PACE detect buffer turned off (default)

1 = PACE detect buffer turned on

RESP: Respiration Control Register

Address = 16h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|----------------|--------------|-------|----------|----------|----------|------------|------------|
| RESP_DEMOD_EN1 | RESP_MOD_EN1 | 1 | RESP_PH2 | RESP_PH1 | RESP_PH0 | RESP_CTRL1 | RESP_CTRL0 |

This register provides the controls for the respiration circuitry; see the [Respiration](#) section for details.

- Bit 7** **RESP_DEMOD_EN1: Enables respiration demodulation circuitry (ADS1294R/6R/8R only, for ADS1294/6/8 always write '0')**
This bit enables/disables the demodulation circuitry on channel 1.
0 = RESP demodulation circuitry turned off (default)
1 = RESP demodulation circuitry turned on
- Bit 6** **RESP_MOD_EN1: Enables respiration modulation circuitry (ADS1294R/6R/8R only, for ADS1294/6/8 always write '0')**
This bit enables/disables the modulation circuitry on channel 1.
0 = RESP modulation circuitry turned off (default)
1 = RESP modulation circuitry turned on
- Bit 5** **Reserved**
Must always be set to '1' for ADS1294R/6R/8R. This bit does not have any effect for the ADS1294/6/8.
- Bits[4:2]** **RESP_PH[2:0]: Respiration phase⁽¹⁾**
These bits control the phase of the respiration demodulation control signal. (GPIO4 is out-of-phase with GPIO3 by the phase determined by the RESP_PH bits.)
000 = 22.5° (default)
001 = 45°
010 = 67.5°
011 = 90°
100 = 112.5°
101 = 135°
110 = 157.5°
111 = N/A
- Bits[1:0]** **RESP_CTRL[1:0]: Respiration control**
These bits set the mode of the respiration circuitry.
00 = No respiration (default)
01 = External respiration
10 = Internal respiration with internal signals
11 = Internal respiration with user-generated signals

(1) RESP_PH[2:0] phase control bits only for internal respiration (RESP_CTRL = 10) and external respiration (RESP_CTRL = 01) modes when the CONFIG4.RESP_FREQ[2:0] register bits are 000b or 001b.

CONFIG4: Configuration Register 4

Address = 17h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|------------|------------|------------|-------|-------------|------------|--------------|-------|
| RESP_FREQ2 | RESP_FREQ1 | RESP_FREQ0 | 0 | SINGLE_SHOT | WCT_TO_RLD | PD_LOFF_COMP | 0 |

Bits[7:5] RESP_FREQ[2:0]: Respiration modulation frequency

These bits control the respiration control frequency when RESP_CTRL[1:0] = 10 or RESP_CTRL[1:0] = 10⁽¹⁾.

000 = 64kHz modulation clock

001 = 32kHz modulation clock

010 = 16kHz square wave on GPIO3 and GPIO04. Output on GPIO4 is 180 degree out of phase with GPIO3.

011 = 8kHz square wave on GPIO3 and GPIO04. Output on GPIO4 is 180 degree out of phase with GPIO3.

100 = 4kHz square wave on GPIO3 and GPIO04. Output on GPIO4 is 180 degree out of phase with GPIO3.

101 = 2kHz square wave on GPIO3 and GPIO04. Output on GPIO4 is 180 degree out of phase with GPIO3.

110 = 1kHz square wave on GPIO3 and GPIO04. Output on GPIO4 is 180 degree out of phase with GPIO3.

111 = 500Hz square wave on GPIO3 and GPIO04. Output on GPIO4 is 180 degree out of phase with GPIO3.

Modes 000 and 001 are modulation frequencies in internal and external respiration modes. In internal respiration mode, the control signals appear at the RESP_MODP and RESP_MODN terminals. All other bit settings generate square waves as described above on GPIO4 and GPIO3.

(1) These frequencies assume $f_{CLK} = 2.048\text{MHz}$.

Bit 4 Must always be set to '0'

Bit 3 SINGLE_SHOT: Single-shot conversion

This bit sets the conversion mode.

0 = Continuous conversion mode (default)

1 = Single-shot mode

Bit 2 WCT_TO_RLD: Connects the WCT to the RLD

This bit connects WCT to RLD.

0 = WCT to RLD connection off (default)

1 = WCT to RLD connection on

Bit 1 PD_LOFF_COMP: Lead-off comparator power-down

This bit powers down the lead-off comparators.

0 = Lead-off comparators disabled (default)

1 = Lead-off comparators enabled

Bit 0 Must always be set to '0'

WCT1: Wilson Central Terminal and Augmented Lead Control Register

Address = 18h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|---------|---------|---------|---------|------------------------------|-------|-------|-------|
| aVF_CH6 | aVL_CH5 | aVR_CH7 | aVR_CH4 | $\overline{\text{PD_WCTA}}$ | WCTA2 | WCTA1 | WCTA0 |

The WCT1 control register configures the device WCT circuit channel selection and the augmented leads.

Bit 7 **aVF_CH6: Enable (WCTA + WCTB)/2 to the negative input of channel 6 (ADS1296/8/6R/8R)**
 0 = Disabled (default)
 1 = Enabled

Bit 6 **aVL_CH5: Enable (WCTA + WCTC)/2 to the negative input of channel 5 (ADS1296/8/6R/8R)**
 0 = Disabled (default)
 1 = Enabled

Bit 5 **aVR_CH7: Enable (WCTB + WCTC)/2 to the negative input of channel 7 (ADS1298/8R)**
 0 = Disabled (default)
 1 = Enabled

Bit 4 **aVR_CH4: Enable (WCTB + WCTC)/2 to the negative input of channel 4**
 0 = Disabled (default)
 1 = Enabled

Bit 3 **$\overline{\text{PD_WCTA}}$: Power-down WCTA**
 0 = Powered down (default)
 1 = Powered on

Bits[2:0] **WCTA[2:0]: WCT amplifier A channel selection; typically connected to RA electrode.**

These bits select one of the eight electrode inputs of channels 1 to 4.

000 = Channel 1 positive input connected to WCTA amplifier (default)
 001 = Channel 1 negative input connected to WCTA amplifier
 010 = Channel 2 positive input connected to WCTA amplifier
 011 = Channel 2 negative input connected to WCTA amplifier
 100 = Channel 3 positive input connected to WCTA amplifier
 101 = Channel 3 negative input connected to WCTA amplifier
 110 = Channel 4 positive input connected to WCTA amplifier
 111 = Channel 4 negative input connected to WCTA amplifier

WCT2: Wilson Central Terminal Control Register

Address = 19h

| BIT 7 | BIT 6 | BIT 5 | BIT 4 | BIT 3 | BIT 2 | BIT 1 | BIT 0 |
|------------------------------|------------------------------|-------|-------|-------|-------|-------|-------|
| $\overline{\text{PD_WCTC}}$ | $\overline{\text{PD_WCTB}}$ | WCTB2 | WCTB1 | WCTB0 | WCTC2 | WCTC1 | WCTC0 |

The WCT2 configuration register configures the device WCT circuit channel selection.

Bit 7 $\overline{\text{PD_WCTC}}$: Power-down WCTC

0 = Powered down (default)
1 = Powered on

Bit 6 $\overline{\text{PD_WCTB}}$: Power-down WCTB

0 = Powered down (default)
1 = Powered on

Bits[5:3] WCTB[2:0]: WCT amplifier B channel selection; typically connected to LA electrode.

These bits select one of the eight electrode inputs of channels 1 to 4.

000 = Channel 1 positive input connected to WCTB amplifier (default)
001 = Channel 1 negative input connected to WCTB amplifier
010 = Channel 2 positive input connected to WCTB amplifier
011 = Channel 2 negative input connected to WCTB amplifier
100 = Channel 3 positive input connected to WCTB amplifier
101 = Channel 3 negative input connected to WCTB amplifier
110 = Channel 4 positive input connected to WCTB amplifier
111 = Channel 4 negative input connected to WCTB amplifier

Bits[2:0] WCTC[2:0]: WCT amplifier C channel selection; typically connected to LL electrode.

These bits select one of the eight electrode inputs of channels 1 to 4.

000 = Channel 1 positive input connected to WCTC amplifier (default)
001 = Channel 1 negative input connected to WCTC amplifier
010 = Channel 2 positive input connected to WCTC amplifier
011 = Channel 2 negative input connected to WCTC amplifier
100 = Channel 3 positive input connected to WCTC amplifier
101 = Channel 3 negative input connected to WCTC amplifier
110 = Channel 4 positive input connected to WCTC amplifier
111 = Channel 4 negative input connected to WCTC amplifier

THEORY OF OPERATION

This section discusses the details of the ADS129x internal functional elements. The analog blocks are reviewed first, followed by the digital interface. Blocks implementing ECG-specific functions are covered in the end.

Throughout this document, f_{CLK} denotes the frequency of the signal at the CLK pin, t_{CLK} denotes the period of the signal at the CLK pin, f_{DR} denotes the output data rate, t_{DR} denotes the time period of the output data, and f_{MOD} denotes the frequency at which the modulator samples the input.

SPI INTERFACE

The SPI-compatible serial interface consists of four signals: \overline{CS} , SCLK, DIN, and DOUT. The interface reads conversion data, reads and writes registers, and controls the ADS129x operation. The \overline{DRDY} output is used as a status signal to indicate when data are ready. \overline{DRDY} goes low when new data are available.

Chip Select (\overline{CS})

Chip select (\overline{CS}) selects the ADS129x devices for SPI communication. While \overline{CS} is low the serial interface is active \overline{CS} must remain low for the entire duration of the serial communication. After the serial communication is finished, always wait four or more t_{CLK} cycles before taking \overline{CS} high. When \overline{CS} is taken high, the serial interface is reset, SCLK and DIN are ignored, and DOUT enters a high-impedance state. \overline{DRDY} asserts when data conversion is complete, regardless of whether \overline{CS} is high or low.

While ADS129x is selected the device will attempt to decode and execute commands every eight serial clocks. If the devices ceases to execute serial commands, it is possible extra clock pulses were presented and placed the serial interface in an unknown state. Take \overline{CS} high and back low to reset the serial interface to a known state.

Serial Clock (SCLK)

SCLK is the serial peripheral interface (SPI) serial clock. It is used to shift in commands and shift out data from the device. The serial clock (SCLK) features a Schmitt-triggered input and clocks data on the DIN and DOUT pins into and out of the ADS129x. Even though the input has hysteresis, it is recommended to keep SCLK as clean as possible to prevent glitches from accidentally forcing a clock event. The absolute maximum limit for SCLK is specified in the [Serial Interface Timing](#) table.

While ADS129x is selected ($\overline{CS} = \text{LOW}$), the device attempts to decode and execute commands every eight serial clocks. It is therefore recommended that multiples of 8 SCLKs be presented every serial transfer to keep the interface in a normal operating mode. If the interface ceases to function because of extra serial clocks, it can be reset by toggling \overline{CS} high and back to low.

For a single device, the minimum speed needed for the SCLK depends on the number of channels, number of bits of resolution, and output data rate. (For multiple cascaded devices, see the [Cascade Mode](#) subsection of the [Multiple Device Configuration](#) section.)

$$t_{SCLK} < (t_{DR} - 4t_{CLK}) / (N_{BITS} \times N_{CHANNELS} + 24) \quad (1)$$

For example, if the ADS1298 is used in a 500SPS mode (eight channels, 24-bit resolution), the minimum SCLK speed is 110kHz.

Data retrieval can be done either by putting the device in RDATA mode or by issuing a RDATA command for data on demand. The above SCLK rate limitation applies to RDATA mode. For the RDATA command, the limitation applies if data must be read between two consecutive \overline{DRDY} signals. The above calculation assumes that there are no other commands issued between data captures.

SCLK Clocking Methods

As shown in [Figure 25](#), there are two different SCLK clocking methods to satisfy the decode timing specification shown in [Figure 1](#) for multiple byte commands.

For SCLK speeds that meet the $t_{SDECODE}$ timing requirement shown in [Figure 1](#), SCLK is transmitted in a continuous stream when \overline{CS} is low. This method is not to be confused with a free-running SCLK, where SCLK operates when \overline{CS} is high. Free-running SCLK operation is not supported by this device.

For faster SCLK speeds that do not meet the $t_{SDECODE}$ timing requirement, SCLK is transmitted in 8-bit bursts with a delay between bursts. The absolute maximum SCLK limit is specified in the [Serial Interface Timing](#) table. [Figure 25](#) shows the difference between the two SCLK clocking methods for this device.

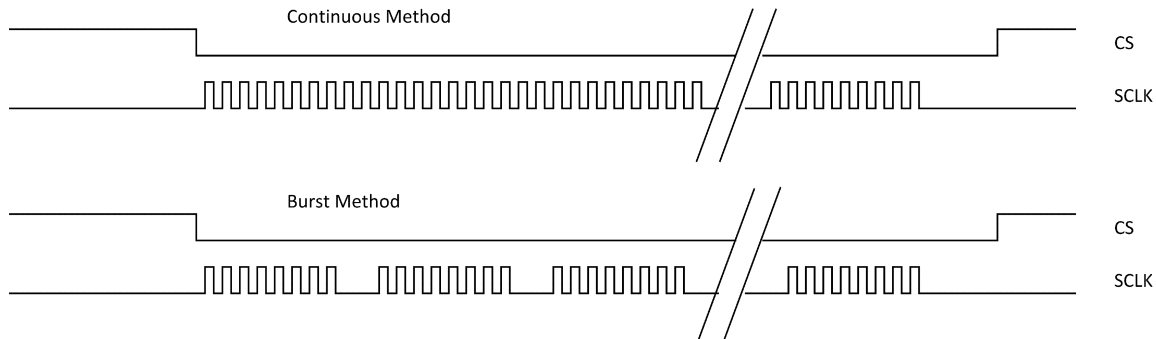


Figure 25. SCLK Clocking Methods

Data Input (DIN)

The data input pin (DIN) is used along with SCLK to communicate with the ADS129x (opcode commands and register data). The device latches data on DIN on the falling edge of SCLK.

Data Output (DOUT)

The data output pin (DOUT) is used with SCLK to read conversion and register data from the ADS129x. Data on DOUT are shifted out on the rising edge of SCLK. DOUT goes to a high-impedance state when \overline{CS} is high. In read data continuous mode (see the [SPI Command Definitions](#) section for more details), the DOUT output line also indicates when new data are available. This feature can be used to minimize the number of connections between the device and the system controller.

[Figure 26](#) shows the data output protocol for ADS1298.

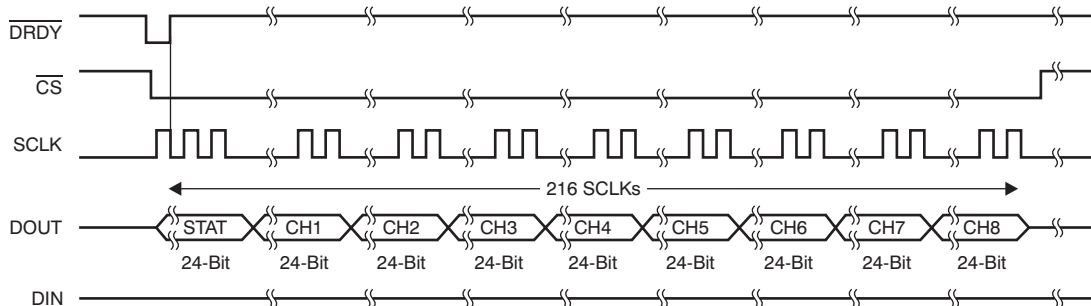


Figure 26. SPI Bus Data Output for the ADS1298 (Eight Channels)

SPI COMMAND DEFINITIONS

The ADS129x provide flexible configuration control. The opcode commands, summarized in [Table 10](#), control and configure the operation of the ADS129x. The opcode commands are stand-alone, except for the register read and register write operations that require a second command byte plus data. CS can be taken high or held low between opcode commands but must stay low for the entire command operation (especially for multi-byte commands). System opcode commands and the RDATA command are decoded by the ADS129x on the seventh falling edge of SCLK. The register read/write opcodes are decoded on the eighth SCLK falling edge. Be sure to follow SPI timing requirements when pulling CS high after issuing a command.

Table 10. Opcode Command Definitions

| COMMAND | DESCRIPTION | FIRST BYTE | SECOND BYTE |
|-------------------------------|--|--|----------------------------------|
| System Commands | | | |
| WAKEUP | Wake-up from standby mode | 0000 0010 (02h) | |
| STANDBY | Enter standby mode | 0000 0100 (04h) | |
| RESET | Reset the device | 0000 0110 (06h) | |
| START | Start/restart (synchronize) conversions | 0000 1000 (08h) | |
| STOP | Stop conversion | 0000 1010 (0Ah) | |
| Data Read Commands | | | |
| RDATA | Enable Read Data Continuous mode. This mode is the default mode at power-up. ⁽¹⁾ | 0001 0000 (10h) | |
| SDATA | Stop Read Data Continuously mode | 0001 0001 (11h) | |
| RDATA | Read data by command; supports multiple read back. | 0001 0010 (12h) | |
| Register Read Commands | | | |
| RREG | Read <i>n nnnn</i> registers starting at address <i>r rrrr</i> | 001 <i>r rrrr</i> (2xh) ⁽²⁾ | 000 <i>n nnnn</i> ⁽²⁾ |
| WREG | Write <i>n nnnn</i> registers starting at address <i>r rrrr</i> | 010 <i>r rrrr</i> (4xh) ⁽²⁾ | 000 <i>n nnnn</i> ⁽²⁾ |

(1) When in RDATA mode, the RREG command is ignored.

(2) *n nnnn* = number of registers to be read/written – 1. For example, to read/write three registers, set *n nnnn* = 0 (0010). *r rrrr* = starting register address for read/write opcodes.

WAKEUP: Exit STANDBY Mode

This opcode exits the low-power standby mode; see the [STANDBY: Enter STANDBY Mode](#) subsection of the [SPI Command Definitions](#) section. Time is required when exiting standby mode (see the [Electrical Characteristics](#) for details). **There are no restrictions on the SCLK rate for this command and it can be issued any time.** Any subsequent command must be sent after 4 t_{CLK} cycles.

STANDBY: Enter STANDBY Mode

This opcode command enters the low-power standby mode. All parts of the circuit are shut down except for the reference section. The standby mode power consumption is specified in the [Electrical Characteristics](#). **There are no restrictions on the SCLK rate for this command and it can be issued any time.** Send a WAKEUP command to return device to normal operation. Serial interface is active, thus register read/write commands are permitted while in this mode.

RESET: Reset Registers to Default Values

This command resets the digital filter cycle and returns all register settings to the respective default values. See the [Reset \(RESET\)](#) subsection of the [SPI Interface](#) section for more details. **There are no restrictions on the SCLK rate for this command and it can be issued any time.** 18 t_{CLK} cycles are required to execute the RESET command. Avoid sending any commands during this time.

START: Start Conversions

This opcode starts data conversions. Tie the START pin low to control conversions by command. If conversions are in progress this command has no effect. The STOP opcode command is used to stop conversions. If the START command is immediately followed by a STOP command, there must be a gap of 4 t_{CLK} cycles between the two commands. When the START opcode is sent to the device, keep the START pin low until the STOP command is issued. (See the [START](#) subsection of the [SPI Interface](#) section for more details.) **There are no restrictions on the SCLK rate for this command and it can be issued any time.**

STOP: Stop Conversions

This opcode stops conversions. Tie the START pin low to control conversions by command. When the STOP command is sent, the conversion in progress completes and further conversions are stopped. If conversions are already stopped, this command has no effect. **There are no restrictions on the SCLK rate for this command and it can be issued any time.**

RDATAC: Read Data Continuous

This opcode enables the output of conversion data on each \overline{DRDY} without the need to issue subsequent read data opcodes. This mode places the conversion data in the output register and may be shifted out directly. The read data continuous mode is the default mode of the device and the device defaults to this mode on power-up and reset.

RDATAC mode is cancelled by the Stop Read Data Continuous command. If the device is in RDATAC mode, a SDATAC command must be issued before any other commands can be sent to the device. There is no restriction on the SCLK rate for this command. However, the subsequent data retrieval SCLKs or the SDATAC opcode command should wait at least 4 t_{CLK} cycles. The timing for RDATAC is shown in [Figure 27](#). As [Figure 27](#) shows, there is a *keep out* zone of 4 t_{CLK} cycles around the \overline{DRDY} pulse when this command cannot be issued. If no data are retrieved from the device, DOUT and \overline{DRDY} behave similarly in this mode. To retrieve data from the device after RDATAC command is issued, make sure either the START pin is high or the START command is issued. [Figure 27](#) shows the recommended way to use the RDATAC command. RDATAC is ideally suited for applications such as data loggers or recorders where registers are set once and do not need to be re-configured.

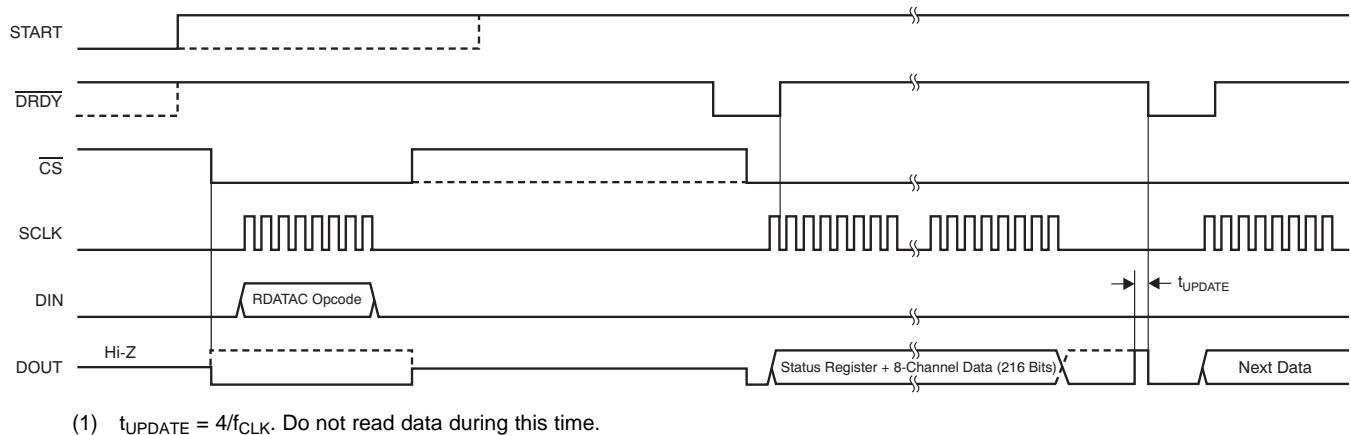


Figure 27. RDATAC Usage

SDATAC: Stop Read Data Continuous

This opcode cancels the Read Data Continuous mode. There is no restriction on the SCLK rate for this command, but the subsequent command must wait for 4 t_{CLK} cycles.

RDATA: Read Data

Issue this command after $\overline{\text{DRDY}}$ goes low to read the conversion result (in Stop Read Data Continuous mode). There is no restriction on the SCLK rate for this command, and there is no wait time needed for the subsequent commands or data retrieval SCLKs. To retrieve data from the device after RDATA command is issued, make sure either the START pin is high or the START command is issued. When reading data with the RDATA command, the read operation can overlap the occurrence of the next $\overline{\text{DRDY}}$ without data corruption. Figure 28 shows the recommended way to use the RDATA command. RDATA is best suited for ECG- and EEG-type systems where register settings must be read or changed often between conversion cycles.

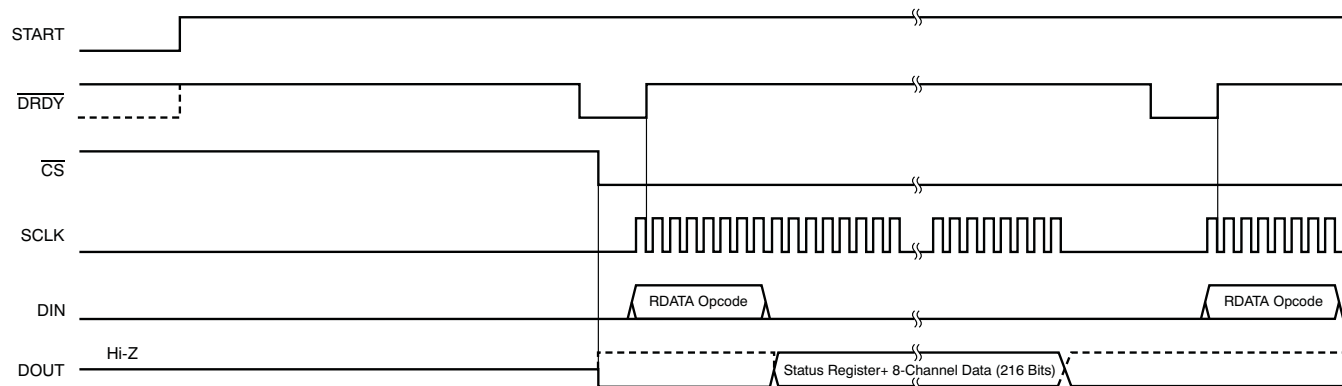


Figure 28. RDATA Usage

Sending Multi-Byte Commands

The ADS129x serial interface decodes commands in bytes and requires $4 t_{\text{CLK}}$ cycles to decode and execute. Therefore, when sending multi-byte commands, a $4 t_{\text{CLK}}$ period must separate the end of one byte (or opcode) and the next.

Assume CLK is 2.048MHz, then t_{SDECODE} ($4 t_{\text{CLK}}$) is 1.96 μs . When SCLK is 16MHz, one byte can be transferred in 500ns. This byte transfer time does not meet the t_{SDECODE} specification; therefore, a delay must be inserted so the end of the second byte arrives 1.46 μs later. If SCLK is 4MHz, one byte is transferred in 2 μs . Because this transfer time exceeds the t_{SDECODE} specification, the processor can send subsequent bytes without delay. In this later scenario, the serial port can be programmed to cease single-byte transfer per cycle to multiple bytes.

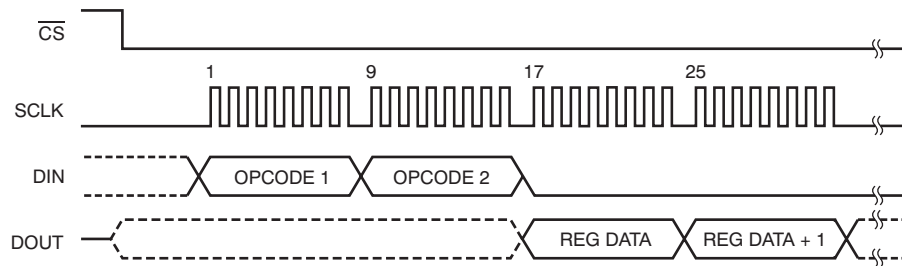
RREG: Read From Register

This opcode reads register data. The Register Read command is a two-byte opcode followed by the output of the register data. The first byte contains the command opcode and the register address. The second byte of the opcode specifies the number of registers to read – 1.

First opcode byte: 001r rrrr, where r rrrr is the starting register address.

Second opcode byte: 000n nnnn, where n nnnn is the number of registers to read – 1.

The 17th SCLK rising edge of the operation clocks out the MSB of the first register, as shown in Figure 29. When the device is in read data continuous mode, it is necessary to issue a SDATAC command before a RREG command can be issued. An RREG command can be issued any time. However, because this command is a multi-byte command, there are restrictions on the SCLK rate depending on the way the SCLKs are issued. See the [Serial Clock \(SCLK\)](#) subsection of the [SPI Interface](#) section for more details. Note that $\overline{\text{CS}}$ must be low for the entire command.



**Figure 29. RREG Command Example: Read Two Registers Starting from Register 00h (ID Register)
(OPCODE 1 = 0010 0000, OPCODE 2 = 0000 0001)**

WREG: Write to Register

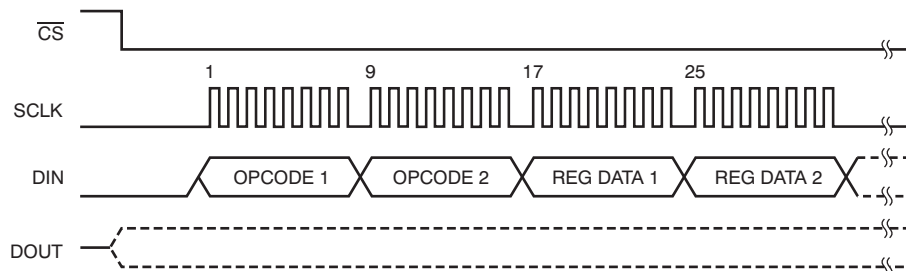
This opcode writes register data. The Register Write command is a two-byte opcode followed by the input of the register data. The first byte contains the command opcode and the register address.

The second byte of the opcode specifies the number of registers to write – 1.

First opcode byte: 010r rrrr, where r rrrr is the starting register address.

Second opcode byte: 000n nnnn, where n nnnn is the number of registers to write – 1.

After the opcode bytes, the register data follows (in MSB-first format), as shown in Figure 30. WREG command can be issued any time. However, because this command is a multi-byte command, there are restrictions on the SCLK rate depending on the way the SCLKs are issued. See the [Serial Clock \(SCLK\)](#) subsection of the [SPI Interface](#) section for more details. Note that $\overline{\text{CS}}$ must be low for the entire command.



**Figure 30. WREG Command Example: Write Two Registers Starting from 00h (ID Register)
(OPCODE 1 = 0100 0000, OPCODE 2 = 0000 0001)**

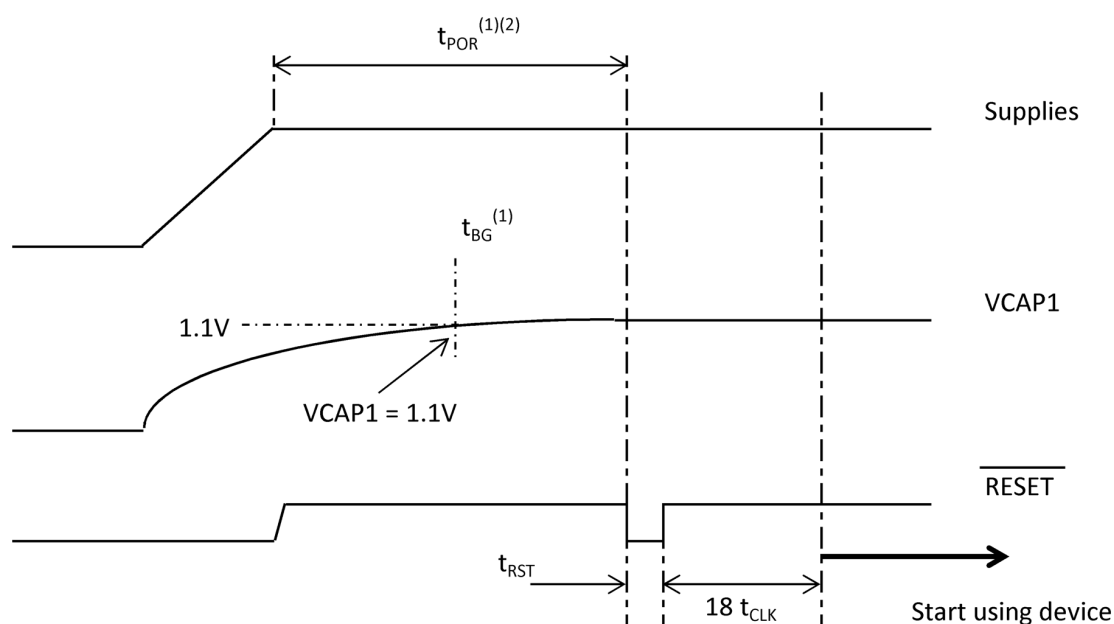
POWER-UP SEQUENCING

Before device power-up, all digital and analog inputs must be low. At the time of power-up, all of these signals should remain low until the power supplies have stabilized, as shown in [Figure 31](#).

Allow for the supply voltages to reach their final value and then begin supplying the master clock signal to the CLK pin. Wait for time t_{POR} , then transmit a reset pulse using either the RESET pin or RESET command to initialize the digital portion of the chip. The RESET should be issued after t_{POR} or after the VCAP1 voltage is greater than 1.1 V, whichever time is longer.

- t_{POR} is described in [Table 11](#).
- The VCAP1 pin charge time is set by the RC time constant; see [Figure 53](#).

After releasing RESET, the configuration registers must be programmed, see the [CONFIG1: Configuration Register 1](#) subsection of the [Register Map](#) section for details. The power-up sequence timing is shown in [Table 11](#).



- (1) Timing to reset pulse is t_{POR} or after t_{BG} , whichever is longer.
(2) When using an external clock, t_{POR} timing does not start until CLK is valid.

Figure 31. Power-Up Timing Diagram

Table 11. Power-Up Sequence Timing

| SYMBOL | DESCRIPTION | MIN | TYP | MAX | UNIT |
|-----------|---------------------------------|----------|-----|-----|-----------|
| t_{POR} | Wait after power-up until RESET | 2^{18} | | | t_{CLK} |
| t_{RST} | Reset low width | 2 | | | t_{CLK} |

DATA ACQUISITION

This section describes the data acquisition process in relation to the START and $\overline{\text{DRDY}}$ pins, settled data, and data readback.

START

The START pin must be set high for at least $2 t_{\text{CLK}}$ s or the START command sent to begin conversions. When START is low or if the START command has not been sent, the device does not issue a $\overline{\text{DRDY}}$ signal (conversions are halted).

When using the START opcode to begin conversions, hold the START pin low. The ADS129x feature two modes to control conversion: continuous mode and single-shot mode. The mode is selected by SINGLE_SHOT (bit 3 of the CONFIG4 register). In multiple device configurations the START pin is used to synchronize devices (see the [Multiple Device Configuration](#) subsection of the [SPI Interface](#) section for more details).

Settling Time

The settling time (t_{SETTLE}) is the time it takes for the converter to output fully settled data when START signal is pulled high.

START Pin Low-to-High Transition or START Command Sent

When START is pulled high or when the START command is sent, the device ADCs are actively converting the input signals and $\overline{\text{DRDY}}$ is pulled high. The next falling edge of $\overline{\text{DRDY}}$ indicates that data is ready. [Figure 32](#) shows the timing diagram and [Table 12](#) shows the settling time for different data rates as a function of t_{CLK} . The settling time depends on f_{CLK} and the decimation ratio (controlled by the DR[2:0] bits in the CONFIG1 register).

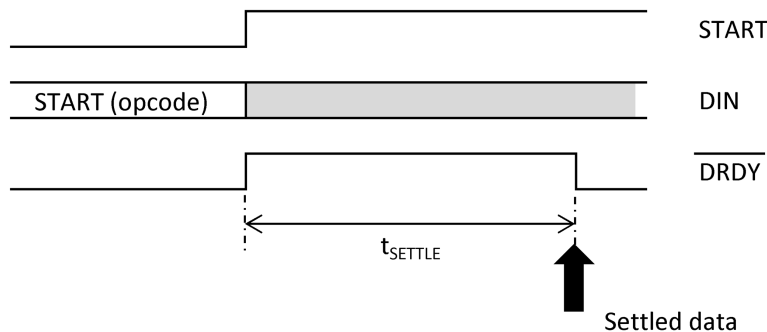


Figure 32. Settling Time For Initial Conversion

Table 12. Settling Times for Different Data Rates

| DR[2:0] | HIGH-RESOLUTION MODE | LOW-POWER MODE | UNIT |
|---------|----------------------|----------------|------------------|
| 000 | 296 | 584 | t_{CLK} |
| 001 | 584 | 1160 | t_{CLK} |
| 010 | 1160 | 2312 | t_{CLK} |
| 011 | 2312 | 4616 | t_{CLK} |
| 100 | 4616 | 9224 | t_{CLK} |
| 101 | 9224 | 18440 | t_{CLK} |
| 110 | 18440 | 36872 | t_{CLK} |

START High and Input Signal Step

Note that when START is held high and there is a step change in the input signal, it takes $3 \times t_{DR}$ for the filter to settle to the new value. Settled data is available on the fourth \overline{DRDY} pulse. This time must be considered when trying to measure narrow PACE pulses for PACE detection. Note that data is available to read at each \overline{DRDY} high-to-low transition, but can be ignored.

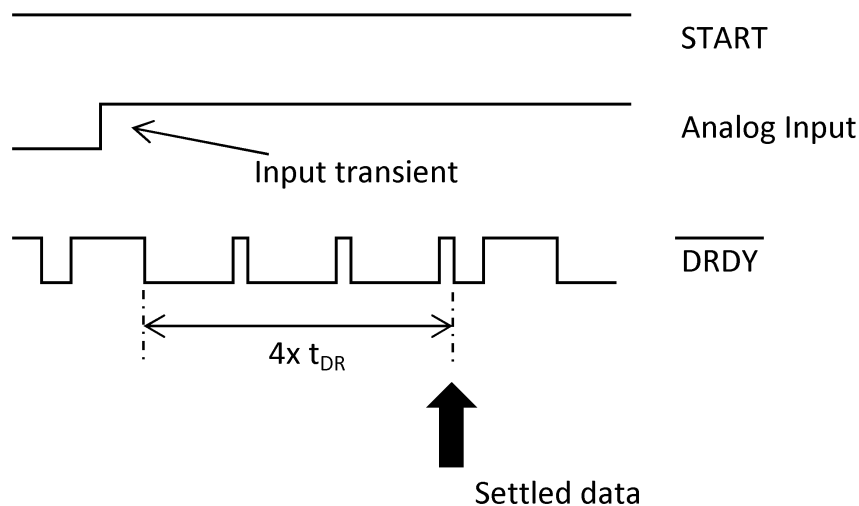


Figure 33. Settling Time For Input Transient

Data Ready (\overline{DRDY})

\overline{DRDY} is an output. When it transitions low, new conversion data are ready. The \overline{CS} signal has no effect on the data ready signal. Regardless of the status of the \overline{CS} signal, a rising edge on SCLK pulls \overline{DRDY} high. Hence, when using multiple devices in the SPI bus, it is recommended that SCLK be gated with \overline{CS} . The behavior of \overline{DRDY} is determined by whether the device is in RDATA mode or the RDATA command is being used to read data on demand. (See the [RDATA: Read Data Continuous](#) and [RDATA: Read Data](#) subsections of the [SPI Command Definitions](#) section for further details).

When reading data with the RDATA command, the read operation can overlap the occurrence of the next \overline{DRDY} without data corruption.

The START pin or the START command is used to place the device either in normal data capture mode or pulse data capture mode.

Figure 34 shows the relationship between $\overline{\text{DRDY}}$, DOUT, and SCLK during data retrieval (in the case of an ADS129x with a selected data rate that gives 24-bit resolution). DOUT is latched out at the rising edge of SCLK. $\overline{\text{DRDY}}$ is pulled high at the falling edge of SCLK. Note that $\overline{\text{DRDY}}$ goes high on the first falling edge SCLK regardless of whether data are being retrieved from the device or a command is being sent through the DIN pin. The data starts from the MSB of the STATUS word and then proceeds to the ADC channel data in sequential order (i.e. Channel 1, Channel 2, ..., Channel X). Channels that are powered down will still have a position in the data stream, however the data is not valid and should be ignored.

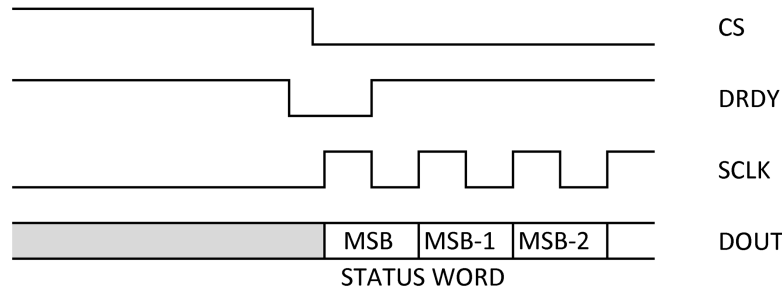


Figure 34. $\overline{\text{DRDY}}$ with Data Retrieval ($\overline{\text{CS}} = 0$)

One consideration with the SCLK signal is that the $\overline{\text{DRDY}}$ signal is cleared on the first SCLK falling edge, regardless of the state of $\overline{\text{CS}}$. No data are clocked out, but the $\overline{\text{DRDY}}$ signal is cleared. This condition should be taken in consideration if the SPI bus is used to communicate with other devices on the same bus. Figure 35 shows a timing diagram for this multiplexing.

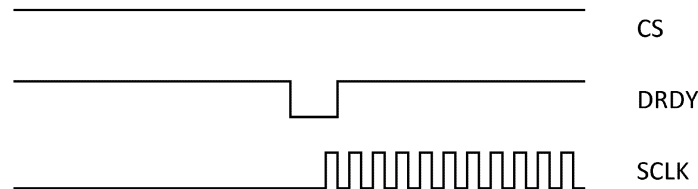


Figure 35. $\overline{\text{DRDY}}$ and SCLK Behavior for SPI Bus Multiplexing

Data Retrieval

Data retrieval can be accomplished in one of two methods. The read data continuous command (see the [RDATAC: Read Data Continuous](#) section) can be used to set the device in a mode to read the data continuously without sending opcodes. The read data command (see the [RDATAC: Read Data](#) section) can be used to read just one data output from the device (see the [SPI Command Definitions](#) section for more details). The conversion data are read by shifting the data out on DOUT. The MSB of the data on DOUT is clocked out on the first SCLK rising edge. $\overline{\text{DRDY}}$ returns to high on the first SCLK falling edge. DIN should remain low for the entire read operation.

STATUS Word

The ADS129x data readback is preceded by a status word that provides information on the state of the ADC. The STATUS word is 24 bits long and contains the values for LOFF_STATP, LOFF_STATN, and part of the GPIO registers. The content alignment is shown in Figure 36.



Figure 36. STATUS Word Content

Readback Length

The number of bits in the data output depends on the number of channels and the number of bits per channel. The data format for each channel data is two's complement and MSB first. For the ADS129x with 32- and 64-kSPS data rates, the number of data bits is: 24 status bits + 16 bits per channel × 8 channels = 152 bits. For all other data rates, the number of data bits is: 24 status bits + 24 bits per channel × 8 channels = 216 bits. When channels are powered down using the user register setting, the corresponding channel output is set to '0'. However, the sequence of channel outputs remains the same. The ADS1294 outputs four channels of data and the ADS1296 outputs six channels of data.

The ADS129x also provide a multiple readback feature. Data can be read out multiple times by simply providing more SCLKs, in which case the MSB data byte repeats after reading the last byte. The DAISY_IN bit in the CONFIG1 register must be set to '1' for multiple readbacks.

Data Format

The ADS129x output 24 bits of data per channel in binary two's complement format, MSB first. The LSB has a weight of $V_{REF}/(2^{23} - 1)$. A positive full-scale input produces an output code of 7FFFFFFh and the negative full-scale input produces an output code of 800000h. The output clips at these codes for signals exceeding full-scale. Table 13 summarizes the ideal output codes for different input signals. Note that for DR[2:0] = 000 and 001, the device has only 17 and 19 bits of resolution, respectively. The last seven (in 17-bit mode) or five (in 19-bit mode) bits can be ignored.

Table 13. Ideal Output Code versus Input Signal^{(1) (2)}

| INPUT SIGNAL, V_{IN} ($A_{INP} - A_{INN}$) | IDEAL OUTPUT CODE ⁽³⁾ |
|---|----------------------------------|
| $\geq V_{REF}$ | 7FFFFFFh |
| $+V_{REF}/(2^{23} - 1)$ | 000001h |
| 0 | 000000h |
| $-V_{REF}/(2^{23} - 1)$ | FFFFFFh |
| $\leq -V_{REF}/(2^{23} - 1)$ | 800000h |

(1) Only valid for 24-bit resolution data rates.

(2) Assumes gain = 1.

(3) Excludes effects of noise, linearity, offset, and gain error.

Single-Shot Mode

The single-shot mode is enabled by setting the SINGLE_SHOT bit in CONFIG4 register to '1'. In single-shot mode, the ADS129x perform a single conversion when the START pin is taken high or when the START opcode command is sent. As seen in Figure 39, when a conversion is complete, DRDY goes low and further conversions are stopped. Regardless of whether the conversion data are read or not, DRDY remains low. To begin a new conversion, take the START pin low and then back high for at least 2 t_{CLK} s, or transmit the START opcode again. Note that when switching from continuous mode to pulse mode, make sure the START signal is pulsed or issue a STOP command followed by a START command.

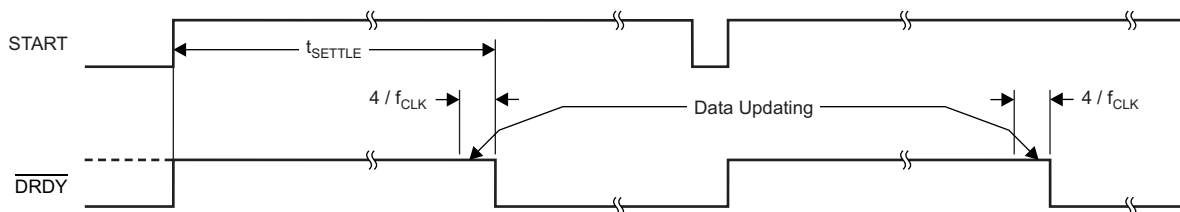
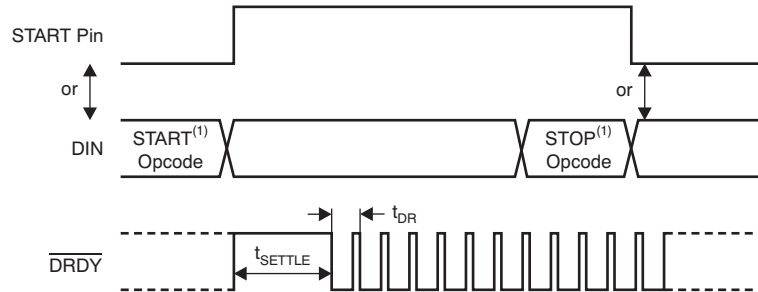


Figure 37. \overline{DRDY} with No Data Retrieval in Single-Shot Mode

This conversion mode is provided for applications that require non-standard or non-continuous data rates. Issuing a START command or toggling the START pin high resets the digital filter, effectively dropping the data rate by a factor of four. This mode leaves the system more susceptible to aliasing effects, thus requiring more complex analog or digital filtering. Loading on the host processor increases because it must toggle the START pin or send a START command to initiate a new conversion cycle.

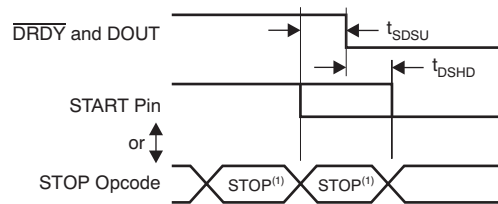
Continuous Mode

Conversions begin when the START pin is taken high for at least 2 t_{CLK} s or when the START opcode command is sent. As seen in Figure 38, the \overline{DRDY} output goes high when conversions are started and goes low when data are ready. Conversions continue indefinitely until the START pin is taken low or the STOP opcode command is transmitted. When the START pin is pulled low or the stop command is issued, the conversion in progress is allowed to complete. Figure 39 and Table 14 show the required timing of \overline{DRDY} to the START pin and the START/STOP opcode commands when controlling conversions in this mode. To keep the converter running continuously, the START pin can be permanently tied high. Note that when switching from pulse mode to continuous mode, the START signal is pulsed or a STOP command must be issued followed by a START command. This conversion mode is ideal for applications that require a fixed-continuous stream of conversions results.



(1) START and STOP opcode commands take effect on the seventh SCLK falling edge.

Figure 38. Continuous Conversion Mode



(1) START and STOP commands take effect on the seventh SCLK falling edge at the end of the opcode transmission.

Figure 39. START to \overline{DRDY} Timing

Table 14. Timing Characteristics for Figure 39 (1)

| SYMBOL | DESCRIPTION | MIN | UNIT |
|------------|--|-----|-----------|
| t_{SDSU} | START pin low or STOP opcode to \overline{DRDY} setup time to halt further conversions | 16 | t_{CLK} |
| t_{DSHD} | START pin low or STOP opcode to complete current conversion | 16 | t_{CLK} |

(1) START and STOP commands take effect on the seventh SCLK falling edge at the end of the opcode transmission.

Setting the Device for Basic Data Capture

The following section outlines the procedure to configure the device in a basic state and capture data. This procedure is intended to put the device in a data sheet condition to check if the device is working properly in the user's system. It is recommended that this procedure be followed initially to get familiar with the device settings. Once this procedure has been verified, the device can be configured as needed. For details on the timings for commands, refer to the appropriate sections in the data sheet. Also, some sample programming codes are added for the ECG-specific functions.

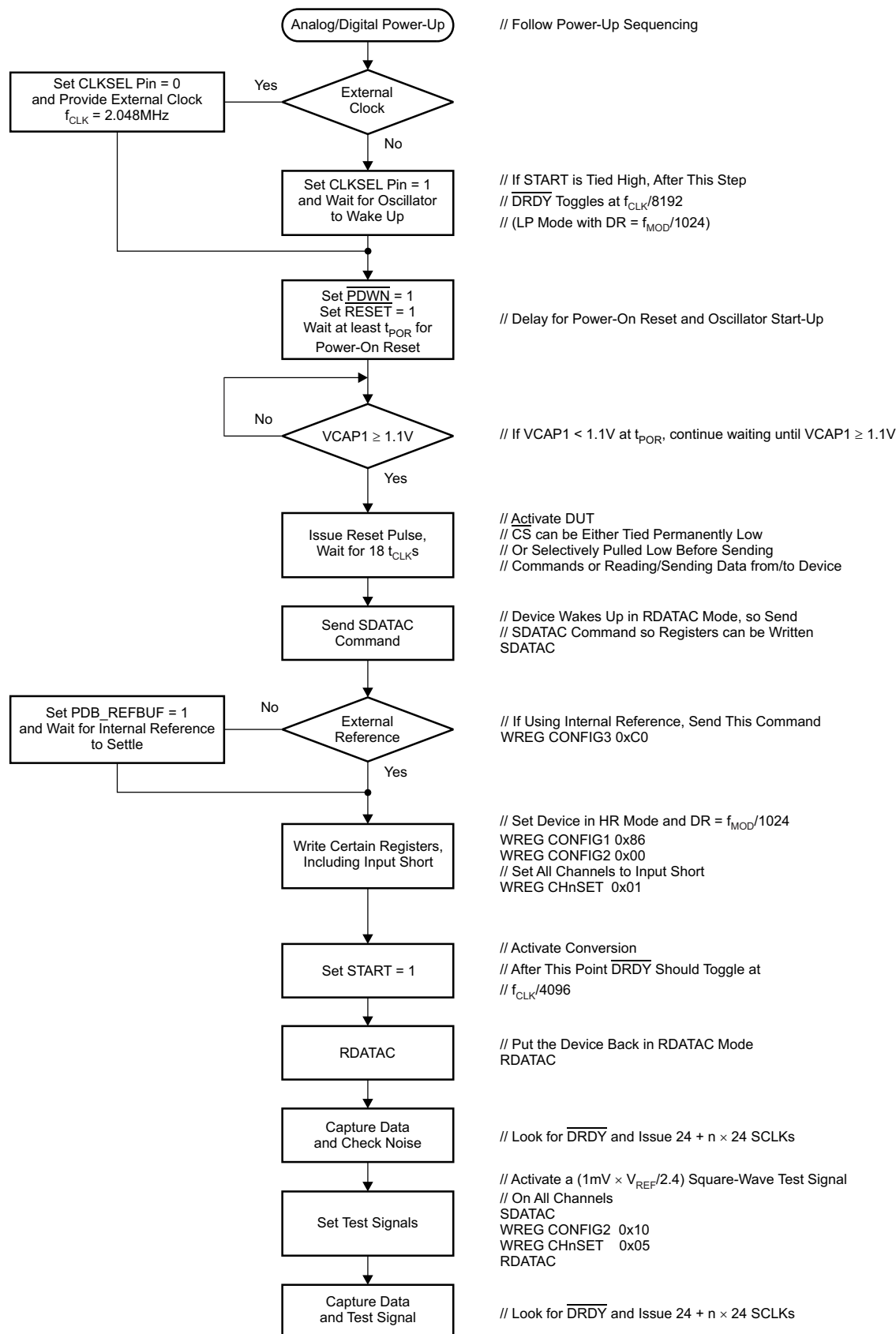


Figure 40. Initial Flow at Power-Up

Lead-Off

Sample code to set dc lead-off with pull-up/pull-down resistors on all channels

```
WREG LOFF 0x13 // Comparator threshold at 95% and 5%, pull-up/pull-down resistor // DC lead-off
```

```
WREG CONFIG4 0x02 // Turn-on dc lead-off comparators
```

```
WREG LOFF_SENSP 0xFF // Turn on the P-side of all channels for lead-off sensing
```

```
WREG LOFF_SENSN 0xFF // Turn on the N-side of all channels for lead-off sensing
```

Observe the status bits of the output data stream to monitor lead-off status.

Right Leg Drive

Sample code to choose RLD as an average of the first three channels.

```
WREG RLD_SENSP 0x07 // Select channel 1—3 P-side for RLD sensing
```

```
WREG RLD_SENSN 0x07 // Select channel 1—3 N-side for RLD sensing
```

```
WREG CONFIG3 b'x1xx 1100 // Turn on RLD amplifier, set internal RLDREF voltage
```

Sample code to route the RLD_OUT signal through channel 4 N-side and measure RLD with channel 5. Make sure the external side to the chip RLDOUT is connected to RLDIN.

```
WREG CONFIG3 b'xxx1 1100 // Turn on RLD amplifier, set internal RLDREF voltage, set RLD measurement bit
```

```
WREG CH4SET b'1xxx 0111 // Route RLDIN to channel 4 N-side
```

```
WREG CH5SET b'1xxx 0010 // Route RLDIN to be measured at channel 5 w.r.t RLDREF
```

PACE Detection

Sample code to select channel 5 and 6 outputs for PACE

```
WREG PACE b'0001 0101 // Power-up PACE amplifier and select channel 5 and 6 for PACE out
```

DIGITAL FUNCTIONALITY

GPIO

The ADS129x have a total of four general-purpose digital I/O (GPIO) pins available in the normal mode of operation. The digital I/O pins are individually configurable as either inputs or as outputs through the GPIOC bits register. The GPIOD bits in the GPIO register control the level of the pins. When reading the GPIOD bits, the data returned are the logic level of the pins, whether they are programmed as inputs or outputs. When the GPIO pin is configured as an input, a write to the corresponding GPIOD bit has no effect. When configured as an output, a write to the GPIOD bit sets the output value.

If configured as inputs, these pins must be driven (do not float). The GPIO pins are set as inputs after power-on or after a reset. [Figure 41](#) shows the GPIO port structure. The pins should be shorted to DGND if not used.

GPIO1 can be used as the PACEIN signal; GPIO2 is multiplexed with RESP_BLK signal; GPIO3 is multiplexed with the RESP signal; and GPIO4 is multiplexed with the RESP_PH signal.

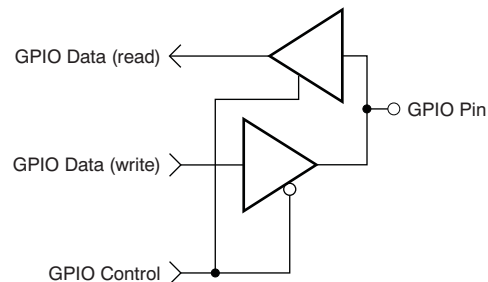


Figure 41. GPIO Port Pin

Power-Down ($\overline{\text{PWDN}}$)

When $\overline{\text{PWDN}}$ is pulled low, all on-chip circuitry is powered down. To exit power-down mode, take the $\overline{\text{PWDN}}$ pin high. Upon exiting from power-down mode, the internal oscillator and the reference require time to wake up. It is recommended that during power-down the external clock is shut down to save power.

Reset ($\overline{\text{RESET}}$)

There are two methods to reset the ADS129x: pull the $\overline{\text{RESET}}$ pin low, or send the RESET opcode command. When using the $\overline{\text{RESET}}$ pin, take it low to force a reset. Make sure to follow the minimum pulse width timing specifications before taking the $\overline{\text{RESET}}$ pin back high. The RESET command takes effect on the eighth SCLK falling edge of the opcode command. On reset it takes $18 t_{\text{CLK}}$ cycles to complete initialization of the configuration registers to the default states and start the conversion cycle. Note that an internal reset is automatically issued to the digital filter whenever registers CONFIG1 and RESP are set to new values with a WREG command.

Digital Decimation Filter

The digital filter receives the modulator output and decimates the data stream. By adjusting the amount of filtering, tradeoffs can be made between resolution and data rate: filter more for higher resolution, filter less for higher data rates. Higher data rates are typically used in ECG applications for implement software PACE detection and ac lead-off detection.

The digital filter on each channel consists of a third-order sinc filter. The decimation ratio on the sinc filters can be adjusted by the DR bits in the CONFIG1 register (see the [Register Map](#) section for details). This setting is a global setting that affects all channels and, therefore, in a device all channels operate at the same data rate.

Sinc Filter Stage ($\text{sinc}x/x$)

The sinc filter is a variable decimation rate, third-order, low-pass filter. Data are supplied to this section of the filter from the modulator at the rate of f_{MOD} . The sinc filter attenuates the high-frequency noise of the modulator, then decimates the data stream into parallel data. The decimation rate affects the overall data rate of the converter.

[Equation 2](#) shows the scaled Z-domain transfer function of the sinc filter.

$$|H(z)| = \left| \frac{1 - Z^{-N}}{1 - Z^{-1}} \right|^3 \quad (2)$$

The frequency domain transfer function of the sinc filter is shown in Equation 3.

$$|H(f)| = \left| \frac{\sin\left(\frac{N\pi f}{f_{MOD}}\right)}{N \times \sin\left(\frac{\pi f}{f_{MOD}}\right)} \right|^3$$

where:

N = decimation ratio (3)

The sinc filter has notches (or zeroes) that occur at the output data rate and multiples thereof. At these frequencies, the filter has infinite attenuation. Figure 42 shows the frequency response of the sinc filter and Figure 43 shows the roll-off of the sinc filter. With a step change at input, the filter takes $3 \times t_{DR}$ to settle. After a rising edge of the START signal, the filter takes t_{SETTLE} time to give the first data output. The settling time of the filters at various data rates are discussed in the *START* subsection of the *SPI Interface* section. Figure 44 and Figure 45 show the filter transfer function until $f_{MOD}/2$ and $f_{MOD}/16$, respectively, at different data rates. Figure 46 shows the transfer function extended until $4 \times f_{MOD}$. It can be seen that the passband of the ADS129x repeats itself at every f_{MOD} . The input R-C anti-aliasing filters in the system should be chosen such that any interference in frequencies around multiples of f_{MOD} are attenuated sufficiently.

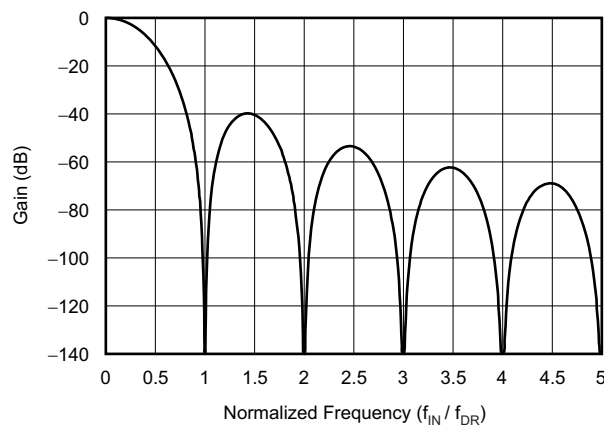


Figure 42. Sinc Filter Frequency Response

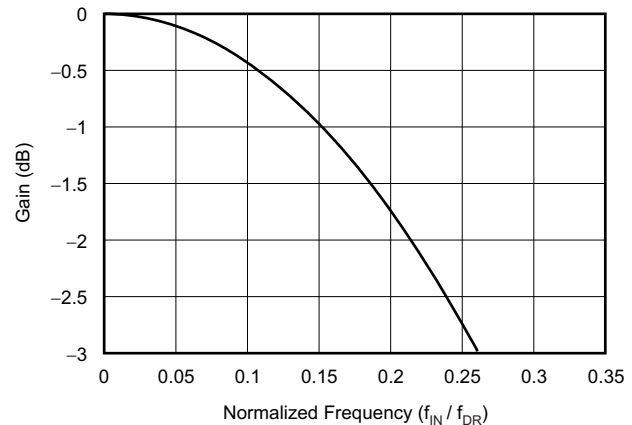


Figure 43. Sinc Filter Roll-Off

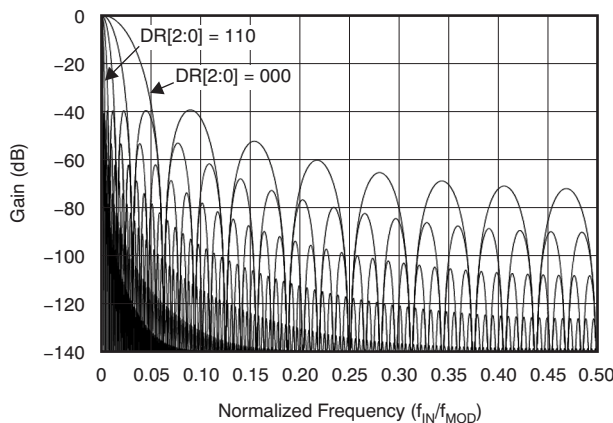


Figure 44. Transfer Function of On-Chip Decimation Filters Until $f_{MOD}/2$

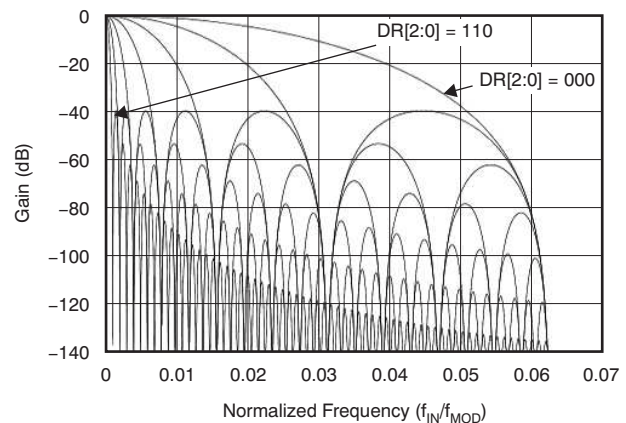
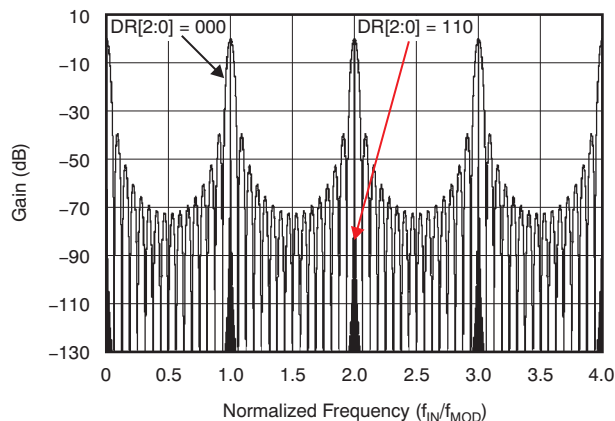


Figure 45. Transfer Function of On-Chip Decimation Filters Until $f_{MOD}/16$



**Figure 46. Transfer Function of On-Chip Decimation Filters
Until $4f_{MOD}$ for DR[2:0] = 000 and DR[2:0] = 110**

Clock

The ADS129x provide two different methods for device clocking: internal and external. Internal clocking is ideally suited for low-power, battery-powered systems. The internal oscillator is trimmed for accuracy at room temperature. Over the specified temperature range the accuracy varies; see the [Electrical Characteristics](#). Clock selection is controlled by the CLKSEL pin and the CLK_EN register bit.

The CLKSEL pin selects either the internal or external clock. The CLK_EN bit in the CONFIG1 register enables and disables the oscillator clock to be output in the CLK pin. A truth table for these two pins is shown in [Table 15](#). The CLK_EN bit is useful when multiple devices are used in a daisy-chain configuration. It is recommended that during power-down the external clock be shut down to save power.

Table 15. CLKSEL Pin and CLK_EN Bit

| CLKSEL PIN | CONFIG1.CLK_EN BIT | CLOCK SOURCE | CLK PIN STATUS |
|------------|--------------------|---------------------------|-----------------------------------|
| 0 | X | External clock | Input: external clock |
| 1 | 0 | Internal clock oscillator | 3-state |
| 1 | 1 | Internal clock oscillator | Output: internal clock oscillator |

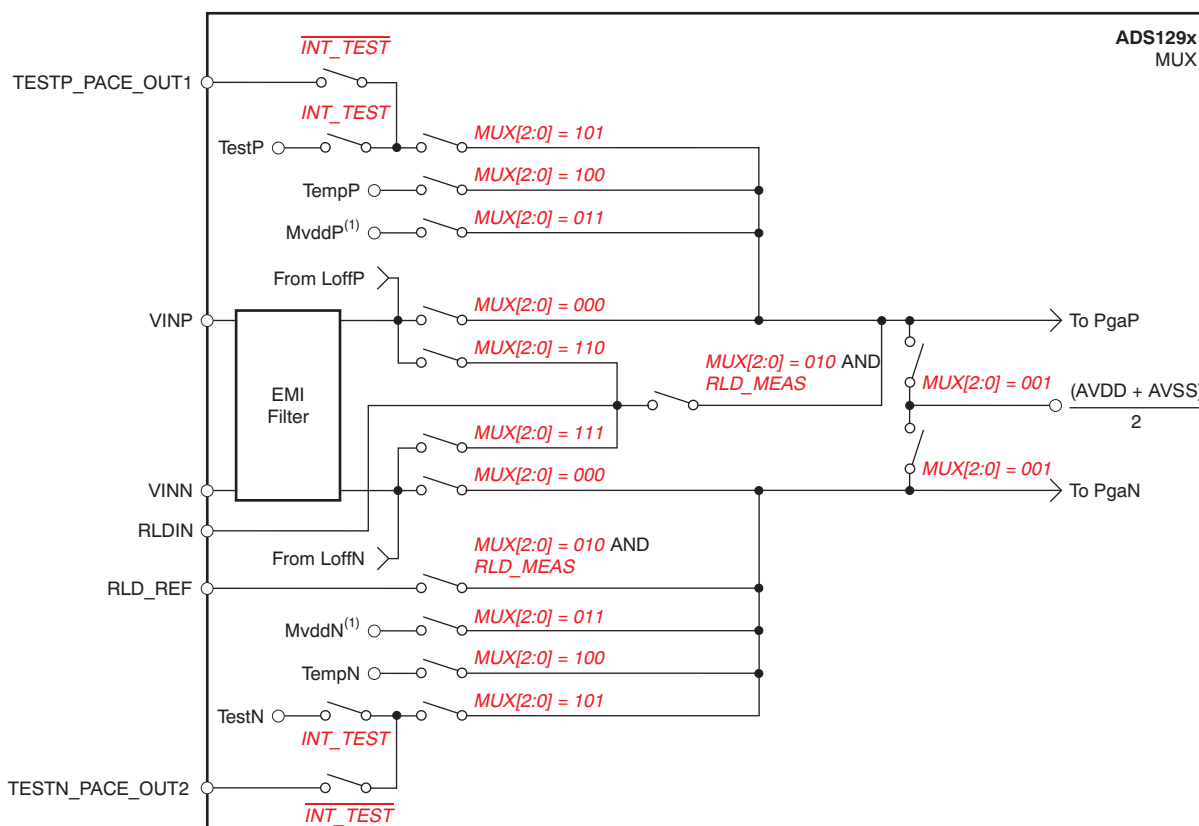
ANALOG FUNCTIONALITY

EMI Filter

An RC filter at the input acts as an EMI filter on all of the channels. The –3dB filter bandwidth is approximately 3MHz.

Input Multiplexer

The ADS129x input multiplexers are very flexible and provide many configurable signal switching options. Figure 47 shows the multiplexer on a single channel of the device. Note that the device has eight such blocks, one for each channel. TEST_PACE_OUT1, TEST_PACE_OUT2, and RLD_IN are common to all eight blocks. VINP and VINN are separate for each of the eight blocks. This flexibility allows for significant device and sub-system diagnostics, calibration, and configuration. Selection of switch settings for each channel is made by writing the appropriate values to the CHnSET[2:0] register (see the [CHnSET: Individual Channel Settings](#) section for details) and by writing the RLD_MEAS bit in the CONFIG3 register (see the [CONFIG3: Configuration Register 3](#) subsection of the [Register Map](#) section for details). More details of the ECG-specific features of the multiplexer are presented in the [Input Multiplexer](#) subsection of the [ECG-Specific Functions](#) section.



(1) MVDD monitor voltage supply depends on channel number; see the [Supply Measurements \(MVDDP, MVDDN\)](#) section.

Figure 47. Input Multiplexer Block for One Channel

Device Noise Measurements

Setting CHnSET[2:0] = 001 sets the common-mode voltage of $(AVDD - AVSS)/2$ to both inputs of the channel. This setting can be used to test the inherent noise of the device in the user system.

Test Signals (TestP and TestN)

Setting CHnSET[2:0] = 101 provides internally-generated test signals for use in sub-system verification at power-up. This functionality allows the entire signal chain to be tested out. Although the test signals are similar to the CAL signals described in the IEC60601-2-51 specification, this feature is not intended for use in compliance testing.

Control of the test signals is accomplished through register settings (see the [CONFIG2: Configuration Register 2](#) subsection in the [Register Map](#) section for details). TEST_AMP controls the signal amplitude and TEST_FREQ controls switching at the required frequency.

The test signals are multiplexed and transmitted out of the device at the TESTP_PACE_OUT1 and TESTN_PACE_OUT2 pins. A bit register (CONFIG2.INT_TEST = 0) deactivates the internal test signals so that the test signal can be driven externally. This feature allows the calibration of multiple devices with the same signal. The test signal feature cannot be used in conjunction with the external hardware PACE feature (see the [External Hardware Approach](#) subsection of the [ECG-Specific Functions](#) section for details).

Auxiliary Differential Input (TESTP_PACE_OUT1, TESTN_PACE_OUT2)

When hardware PACE detect is not used, the TESTP_PACE_OUT1 and TESTN_PACE_OUT2 signals can be used as a multiplexed differential input channel. These inputs can be multiplexed to any of the eight channels. The performance of the differential input signal fed through these pins is identical to the normal channel performance.

Temperature Sensor (TempP, TempN)

The ADS129x contain an on-chip temperature sensor. This sensor uses two internal diodes with one diode having a current density 16x that of the other, as shown in [Figure 48](#). The difference in current densities of the diodes yields a difference in voltage that is proportional to absolute temperature.

As a result of the low thermal resistance of the package to the printed circuit board (PCB), the internal device temperature tracks the PCB temperature closely. Note that self-heating of the ADS129x causes a higher reading than the temperature of the surrounding PCB.

The scale factor of [Equation 4](#) converts the temperature reading to °C. Before using this equation, the temperature reading code must first be scaled to μV.

$$\text{Temperature (}^{\circ}\text{C)} = \left[\frac{\text{Temperature Reading (}\mu\text{V)} - 145,300 \mu\text{V}}{490 \mu\text{V}/^{\circ}\text{C}} \right] + 25^{\circ}\text{C} \quad (4)$$

Temperature Sensor Monitor

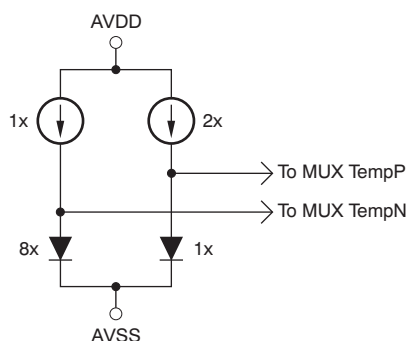


Figure 48. Measurement of the Temperature Sensor in the Input

Supply Measurements (MVDDP, MVDDN)

Setting CHnSET[2:0] = 011 sets the channel inputs to different supply voltages of the device. For channels 1, 2, 5, 6, 7, and 8, (MVDDP – MVDDN) is [0.5 × (AVDD – AVSS)]; for channel 3 and for channel 4, (MVDDP – MVDDN) is DVDD/4. Note that to avoid saturating the PGA while measuring power supplies, the gain must be set to '1'. For example, if AVDD = 2.5V and AVSS = –2.5V, then the measurement result would be 2.5V.

Lead-Off Excitation Signals (LoffP, LoffN)

The lead-off excitation signals are fed into the multiplexer before the switches. The comparators that detect the lead-off condition are also connected to the multiplexer block before the switches. For a detailed description of the lead-off block, refer to the [Lead-Off Detection](#) subsection in the [ECG-Specific Functions](#) section.

Auxiliary Single-Ended Input

The RLD_IN pin is primarily used for routing the right leg drive signal to any of the electrodes in case the right leg drive electrode falls off. However, the RLD_IN pin can be used as a multiple single-ended input channel. The signal at the RLD_IN pin can be measured with respect to the voltage at the RLD_REF pin using any of the eight channels. This measurement is done by setting the channel multiplexer setting to '010' and the RLD_MEAS bit of the CONFIG3 register to '1'.

Analog Input

The analog input to the ADS129x is fully differential. Assuming $PGA = 1$, the differential input ($INP - INN$) can span between $-V_{REF}$ to $+V_{REF}$. Note that the absolute range for INP and INN must be between $AVSS - 0.3V$ and $AVDD + 0.3V$. Refer to [Table 13](#) for an explanation of the correlation between the analog input and the digital codes. There are two general methods of driving the analog input of the ADS129x: single-ended or differential, as shown in [Figure 49](#) and [Figure 50](#). Note that INP and INN are 180° out-of-phase in the differential input method. When the input is single-ended, the INN input is held at the common-mode voltage (CM), preferably at mid-supply. The INP input swings around the same common-mode voltage and the peak-to-peak amplitude swings from $CM - V_{REF}$ to $CM + V_{REF}$. When the input is differential, the common-mode is given by $(INP + INN)/2$. Both the INP and INN inputs swing from $CM + 1/2 V_{REF}$ to $CM - 1/2 V_{REF}$. For optimal performance, it is recommended that the ADS129x devices be used in a differential configuration.

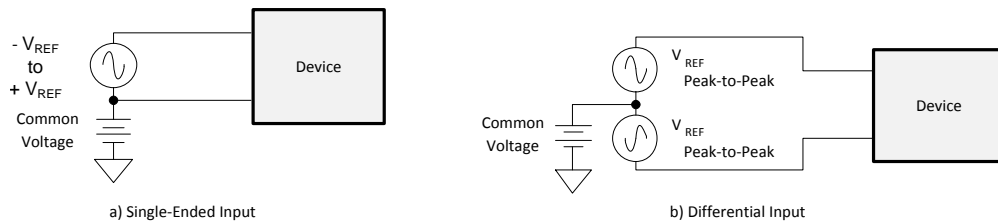


Figure 49. Methods of Driving the ADS1298: Single-Ended or Differential

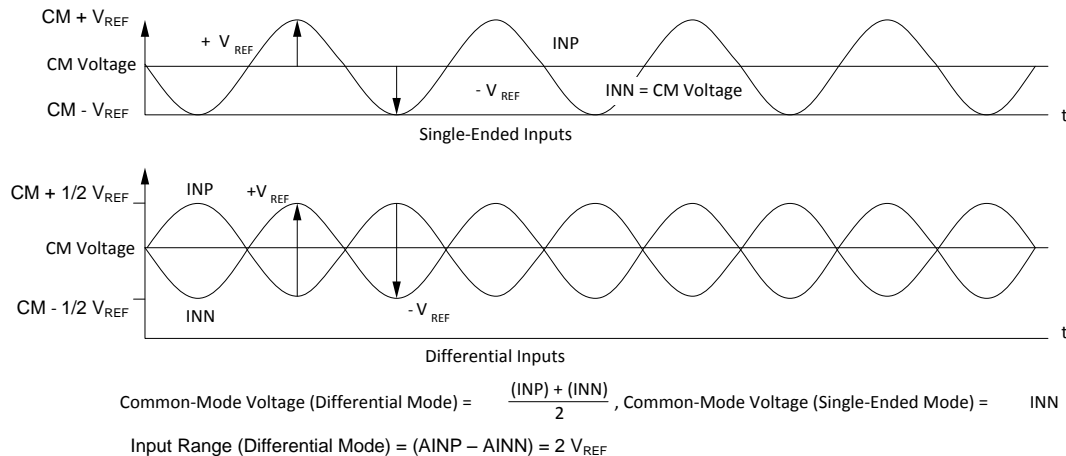


Figure 50. Using the ADS1298 in the Single-Ended and Differential Input Modes

PGA Settings and Input Range

The PGA is a differential input/differential output amplifier, as shown in [Figure 51](#). It has seven gain settings (1, 2, 3, 4, 6, 8, and 12) that can be set by writing to the CHnSET register (see the [CHnSET: Individual Channel Settings](#) subsection of the [Register Map](#) section for details). The ADS129x have CMOS inputs and hence have negligible current noise. [Table 16](#) shows the typical values of bandwidths for various gain settings. Note that [Table 16](#) shows the small-signal bandwidth.

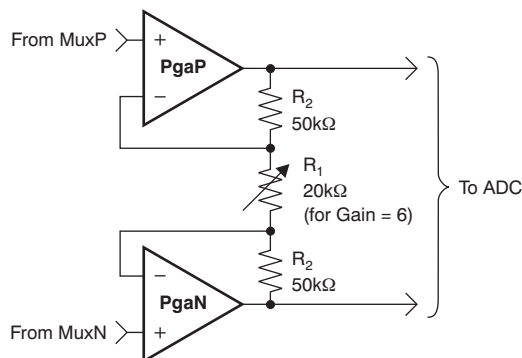


Figure 51. PGA Implementation

Table 16. PGA Gain versus Small-Signal Bandwidth

| GAIN | NOMINAL BANDWIDTH AT ROOM TEMPERATURE (kHz) |
|------|---|
| 1 | 237 |
| 2 | 146 |
| 3 | 127 |
| 4 | 96 |
| 6 | 64 |
| 8 | 48 |
| 12 | 32 |

The resistor string of the PGA that implements the gain has 120kΩ of resistance for a gain of 6. This resistance provides a current path across the outputs of the PGA in the presence of a differential input signal. This current is in addition to the quiescent current specified for the device in the presence of a differential signal at the input.

Input Common-Mode Range

The usable input common-mode range of the front end depends on various parameters, including the maximum differential input signal, supply voltage, PGA gain, etc. This range is described in [Equation 5](#):

$$AVDD - 0.2 - \left[\frac{\text{Gain } V_{\text{MAX_DIFF}}}{2} \right] > CM > AVSS + 0.2 + \left[\frac{\text{Gain } V_{\text{MAX_DIFF}}}{2} \right]$$

where:

$V_{\text{MAX_DIFF}}$ = maximum differential signal at the input of the PGA

CM = common-mode range

(5)

For example:

If $V_{\text{DD}} = 3\text{V}$, gain = 6, and $V_{\text{MAX_DIFF}} = 350\text{mV}$

Then $1.25\text{V} < CM < 1.75\text{V}$

Input Differential Dynamic Range

The differential (INP – INN) signal range depends on the analog supply and reference used in the system. This range is shown in [Equation 6](#).

$$\text{Max (INP – INN)} < \frac{V_{\text{REF}}}{\text{Gain}} ; \quad \text{Full-Scale Range} = \frac{\pm V_{\text{REF}}}{\text{Gain}} = \frac{2 V_{\text{REF}}}{\text{Gain}} \quad (6)$$

The 3V supply, with a reference of 2.4V and a gain of 6 for ECGs, is optimized for power with a differential input signal of approximately 300mV. For higher dynamic range, a 5V supply with a reference of 4V (set by the VREF_4V bit of the CONFIG3 register) can be used to increase the differential dynamic range.

ADC $\Delta\Sigma$ Modulator

Each channel of the ADS129x has a 24-bit $\Delta\Sigma$ ADC. This converter uses a second-order modulator optimized for low-power applications. The modulator samples the input signal at the rate of $f_{\text{MOD}} = f_{\text{CLK}}/4$ for High-Resolution mode and $f_{\text{MOD}} = f_{\text{CLK}}/8$ for Low-Power mode. As in the case of any $\Delta\Sigma$ modulator, the noise of the ADS129x is shaped until $f_{\text{MOD}}/2$, as shown in Figure 52. The on-chip digital decimation filters explained in the next section can be used to filter out the noise at higher frequencies. These on-chip decimation filters also provide antialias filtering. This feature of the $\Delta\Sigma$ converters drastically reduces the complexity of the analog antialiasing filters that are typically needed with Nyquist ADCs.

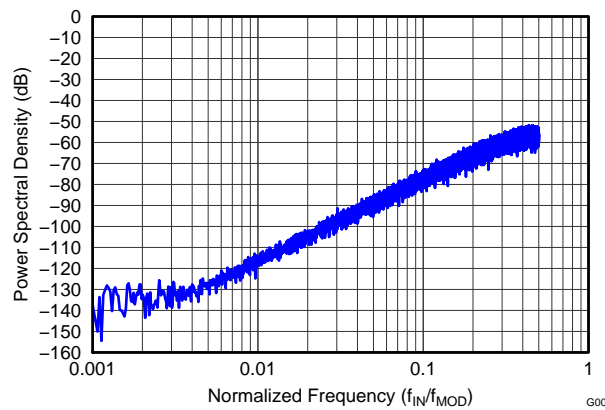
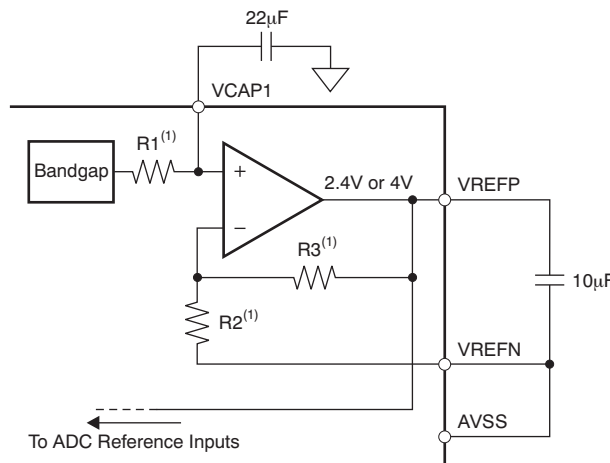


Figure 52. Modulator Noise Spectrum Up To $0.5 \times f_{\text{MOD}}$

Reference

Figure 53 shows a simplified block diagram of the internal reference of the ADS129x. The reference voltage is generated with respect to AVSS. When using the internal voltage reference, connect VREFN to AVSS.



(1) For $V_{\text{REF}} = 2.4\text{V}$: $R_1 = 12.5\text{k}\Omega$, $R_2 = 25\text{k}\Omega$, and $R_3 = 25\text{k}\Omega$. For $V_{\text{REF}} = 4\text{V}$: $R_1 = 10.5\text{k}\Omega$, $R_2 = 15\text{k}\Omega$, and $R_3 = 35\text{k}\Omega$.

Figure 53. Internal Reference

The external band-limiting capacitors determine the amount of reference noise contribution. For high-end ECG systems, the capacitor values should be chosen such that the bandwidth is limited to less than 10Hz, so that the reference noise does not dominate the system noise. When using a 3V analog supply, the internal reference must be set to 2.4V. In case of a 5V analog supply, the internal reference can be set to 4V by setting the VREF_4V bit in the CONFIG2 register.

Alternatively, the internal reference buffer can be powered down and VREFP can be applied externally. [Figure 54](#) shows a typical external reference drive circuitry. Power-down is controlled by the PD_REFBUF bit in the CONFIG3 register. By default the device wakes up in external reference mode.

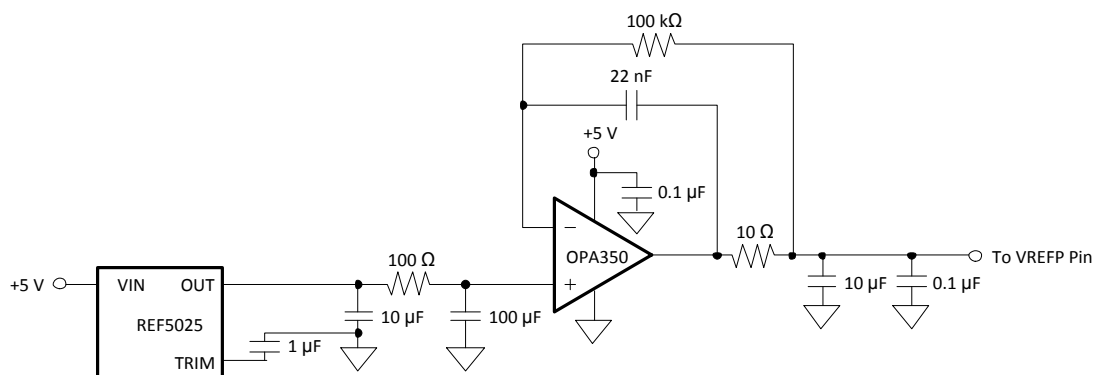


Figure 54. External Reference Driver

MULTIPLE DEVICE CONFIGURATION

The ADS129x are designed to provide configuration flexibility when multiple devices are used in a system. The serial interface typically requires four signals: DIN, DOUT, SCLK, and CS. With one additional chip select signal per device, multiple devices can be connected together. The number of signals needed to interface n devices is $3 + n$.

The right leg drive amplifiers can be daisy-chained as explained in the [RLD Configuration with Multiple Devices](#) subsection of the [ECG-Specific Functions](#) section. To use the internal oscillator in a daisy-chain configuration, one of the devices must be set as the master for the clock source with the internal oscillator enabled (CLKSEL pin = 1) and the internal oscillator clock brought out of the device by setting the CLK_EN register bit to '1'. This master device clock is used as the external clock source for the other devices.

When using multiple devices, the devices can be synchronized with the START signal. The delay from START to the $\overline{\text{DRDY}}$ signal is fixed for a fixed data rate (see the [START](#) subsection of the [SPI Interface](#) section for more details on the settling times). [Figure 55](#) shows the behavior of two devices when synchronized with the START signal as an example.

There are two ways to connect multiple devices with a optimal number of interface pins: cascade mode and daisy-chain mode.

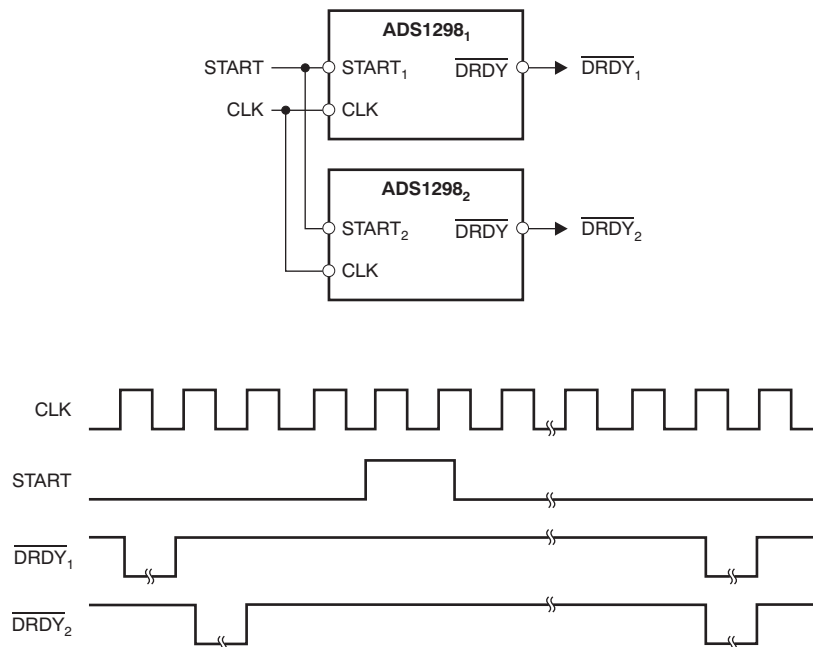


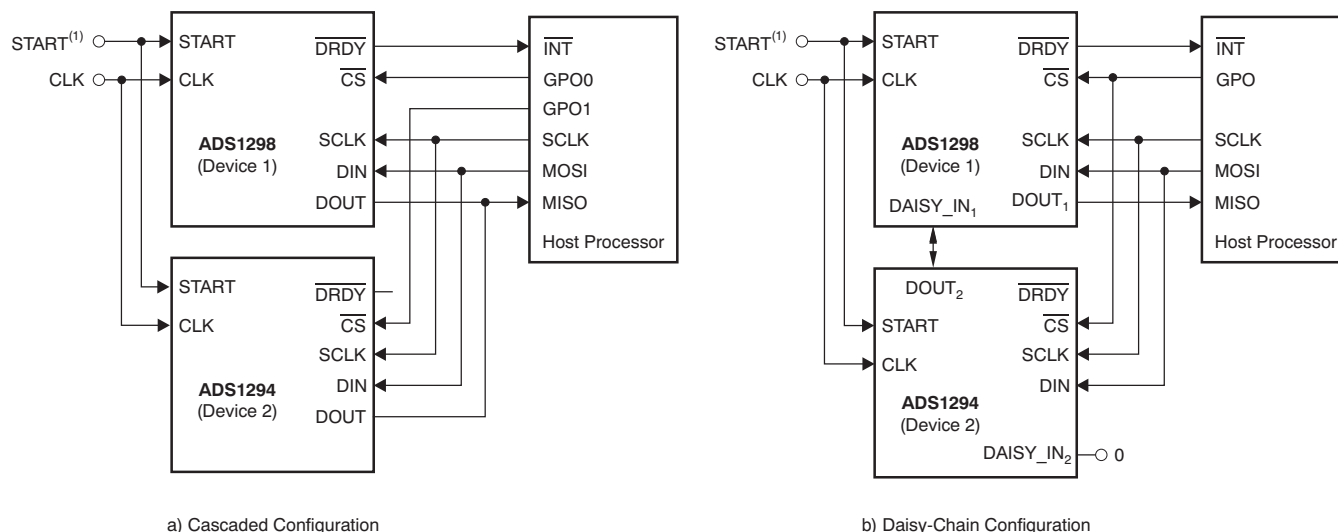
Figure 55. Synchronizing Multiple Converters

Cascaded Mode

[Figure 56a](#) shows a configuration with two devices cascaded together. One of the devices is an ADS1298 (eight channels) and the other is an ADS1294 (four channels). Together, they create a system with 12 channels. DOUT, SCLK, and DIN are shared. Each device has its own chip select. When a device is not selected by the corresponding $\overline{\text{CS}}$ being driven to logic 1, the DOUT of this device is high-impedance. This structure allows the other device to take control of the DOUT bus. This configuration method is suitable for the majority of applications.

Daisy-Chain Mode

Daisy-chain mode is enabled by setting the $\overline{\text{DAISY_EN}}$ bit in the CONFIG1 register. Figure 56b shows the daisy-chain configuration. In this mode SCLK, DIN, and $\overline{\text{CS}}$ are shared across multiple devices. The DOUT of one device is hooked up to the DAISY_IN of the other device, thereby creating a chain. One extra SCLK must be issued between each data set. Also, when using daisy-chain mode, the multiple readback feature is not available. Short the DAISY_IN pin to digital ground if not used. Figure 2 describes the required timing for the ADS1298 shown in Figure 57. Data from the ADS1298 appear first on DOUT, followed by a *don't care* bit, and finally by the status and data words from the ADS1294.



(1) To reduce pin count, set the START pin low and use the START serial command to synchronize and start conversions.

Figure 56. Multiple Device Configurations

Important reminders when using daisy-chain mode:

1. One extra SCLK must be issued between each data set (See Figure 57).
2. All devices are configured to the same register values since $\overline{\text{CS}}$ is shared.
3. Device register readback (RREG) is only valid for Device 0 in the daisy chain. Only data can be read from Device 1 to Device N, where N is the last device in the chain.

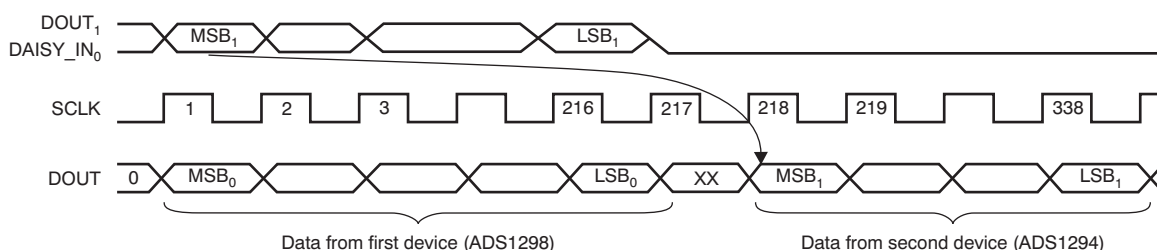


Figure 57. Daisy-Chain Timing for Figure 56b

In a case where all devices in the chain operate in the same register setting, DIN can be shared as well and thereby reduce the SPI communication signals to four, regardless of the number of devices. However, because the individual devices cannot be programmed, the RLD driver cannot be shared among the multiple devices. Furthermore, an external clock must be used.

Note that from [Figure 2](#), the SCLK rising edge shifts data out of the ADS129x on DOUT. The SCLK rising edge is also used to latch data into the device DAISY_IN pin down the chain. This architecture allows for a faster SCLK rate speed, but it also makes the interface sensitive to board level signal delays. The more devices in the chain, the more challenging it could become to adhere to setup and hold times. A star pattern connection of SCLK to all devices, minimizing length of DOUT, and other PCB layout techniques help. Placing delay circuits such as buffers between DOUT and DAISY_IN are ways to mitigate this challenge. One other option is to insert a *D* flip-flop between DOUT and DAISY_IN clocked on an inverted SCLK. Note also that daisy-chain mode requires some software overhead to recombine data bits spread across byte boundaries.

The maximum number of devices that can be daisy-chained depends on the data rate at which the device is being operated. The maximum number of devices can be estimated with [Equation 7](#).

$$N_{\text{DEVICES}} = \frac{f_{\text{SCLK}}}{f_{\text{DR}} (N_{\text{BITS}})(N_{\text{CHANNELS}}) + 24}$$

where:

N_{BITS} = device resolution (depends on data rate), and

N_{CHANNELS} = number of channels in the device (4, 6, or 8). (7)

For example, when the ADS1298 (eight-channel, 24-bit version) is operated at a 2kSPS data rate with a 4MHz f_{SCLK} , 10 devices can be daisy-chained.

INPUT MULTIPLEXER (REROUTING THE RIGHT LEG DRIVE SIGNAL)

Figure 58. Example of RLDOUT Signal Configured to be Routed to IN8N

Also, the RLDOUT signal can be routed to a channel (that is not used for the calculation of RLD) for measurement. [Figure 59](#) shows the register settings to route the RLDIN signal to channel 8. The measurement is done with respect to the voltage on the RLDREF pin. If RLDREF is chosen to be internal, it would be at $(AVDD + AVSS)/2$. This feature is useful for debugging purposes during product development.

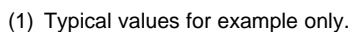


Figure 59. RLDOUT Signal Configured to be Read Back by Channel 8

WILSON CENTRAL TERMINAL (WCT) AND CHEST LEADS

In the standard 12-lead ECG, WCT voltage is defined as the average of Right Arm (RA), Left Arm (LA), and Left Leg (LL) electrodes. This voltage is used as the reference voltage for the measurement of the chest leads. The ADS129x has three integrated low-noise amplifiers that generate the WCT voltage. Figure 60 shows the block diagram of the implementation.

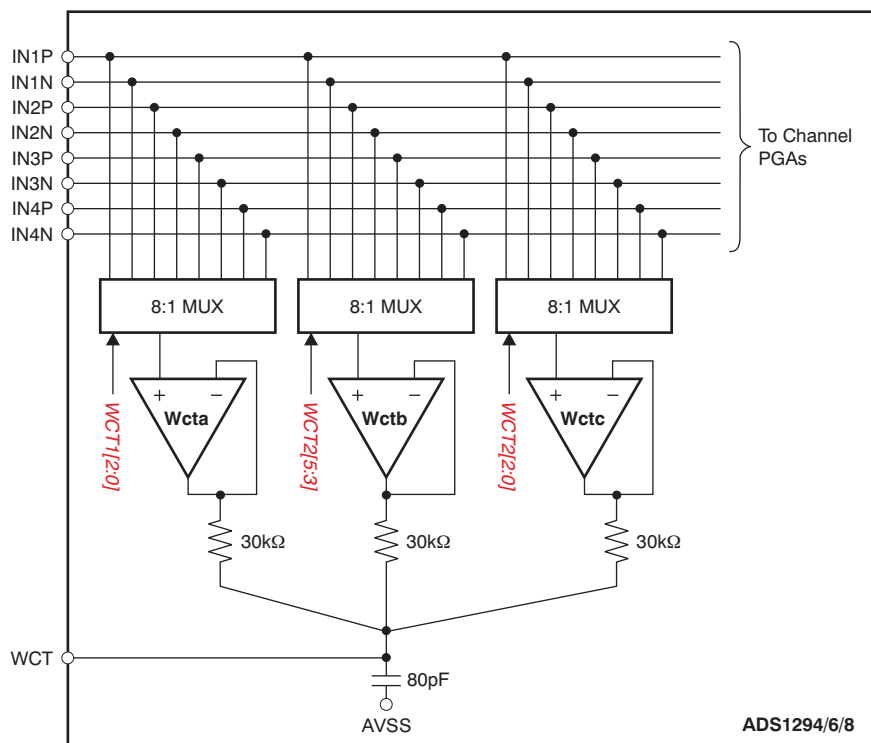


Figure 60. WCT Voltage

The devices provide flexibility to choose any one of the eight signals (IN1P to IN4N) to be routed to each of the amplifiers to generate the average. Having this flexibility allows the RA, LA, and LL electrodes to be connected to any input of the first four channels depending on the lead configuration.

Each of the three amplifiers in the WCT circuitry can be powered down individually with register settings. By powering up two amplifiers, the average of any two electrodes can be generated at the WCT pin. Powering up one amplifier provides the buffered electrode voltage at the WCT pin. Note that the WCT amplifiers have limited drive strength and thus should be buffered if used to drive a low-impedance load.

See Table 6 for performance when using any 1, 2, or 3 of the WCT buffers.

As can be seen in Table 6, the overall noise reduces when more than one WCT amplifier is powered up. This noise reduction is a result of the fact that noise is averaged by the passive summing network at the output of the amplifiers. Powering down individual buffers gives negligible power savings because a significant portion of the circuitry is shared between the three amplifiers. The bandwidth of the WCT node is limited by the RC network. The internal summing network consists of three 30kΩ resistors and a 80pF capacitor. It is recommended that an external 100pF capacitor be added for optimal performance. The effective bandwidth depends on the number of amplifiers that are powered up, as shown in Table 6.

The WCT node should only be used to drive very high input impedances (typically greater than 500MΩ). Typical application would be to connect this WCT signal to the negative inputs of a ADS129x to be used as a reference signal for the chest leads.

As mentioned previously in this section, all three WCT amplifiers can be connected to one of eight analog input pins. The inputs of the amplifiers are chopped and the chopping frequency varies with the data rates of the ADS129x. The chop frequency for the three highest data rates scale 1:1. For example, at 32kSPS data rate, the chop frequency is 32kHz in HR mode with WCT_CHOP = 0. The chopping frequency of the four lower data rates is fixed to 4kHz. When WCT_CHOP = 1, the chop frequency is fixed to highest data rate (that is, $f_{MOD}/16$) frequency, as shown in Table 17. The chop frequency shows itself at the output of the WCT amplifiers as a small square wave riding on dc. The amplitude of the square wave is the offset of the amplifier and is typically 5mV_{PP}. This artifact as a result of chopping is out-of-band and thus does not interfere with ECG-related measurements. As a result of the chopping function, the input current leakage on the pins with WCT amplifiers connected sees increased leakage currents at higher data rates and as the input common voltage swings closer to 0V (AVSS), as described in Figure 61.

Note that if the output of a channel connected to the WCT amplifier (for example, the V lead channels) is connected to one of the PACE amplifiers for external PACE detection, the artifact of chopping appears at the PACE amplifier output.

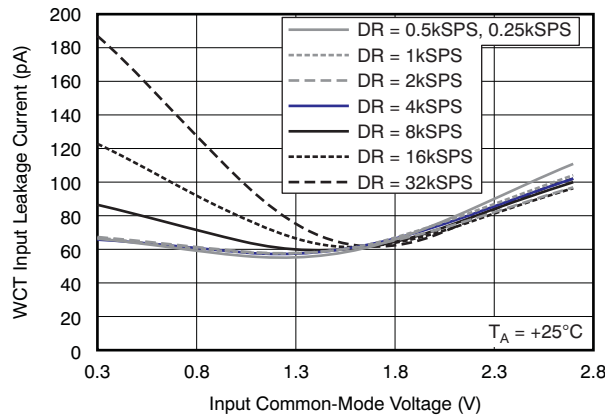


Figure 61. WCT Input Leakage Current versus Input Voltage (WCT_CHOP = 0)

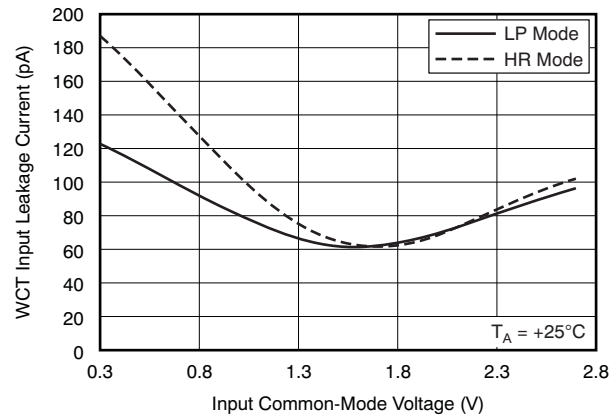


Figure 62. WCT Input Leakage Current versus Input Voltage (WCT_CHOP = 1)

Table 17. WCT Amplifiers Chop Frequency

| CONFIG1.DR[2:0] BIT | CONFIG2.WCT_CHOP = 0 | CONFIG2.WCT_CHOP = 1 |
|---------------------|----------------------|----------------------|
| 000 | $f_{MOD}/16$ | $f_{MOD}/16$ |
| 001 | $f_{MOD}/32$ | $f_{MOD}/16$ |
| 010 | $f_{MOD}/64$ | $f_{MOD}/16$ |
| 011 | $f_{MOD}/128$ | $f_{MOD}/16$ |
| 100 | $f_{MOD}/128$ | $f_{MOD}/16$ |
| 101 | $f_{MOD}/128$ | $f_{MOD}/16$ |
| 110 | $f_{MOD}/128$ | $f_{MOD}/16$ |

Augmented Leads

In the typical implementation of the 12-lead ECG with eight channels, the augmented leads are calculated digitally. In certain applications, it may be required that all leads be derived in analog rather than digital. The ADS1298/8R provides the option to generate the augmented leads by routing appropriate averages to channels 5 to 7. The same three amplifiers that are used to generate the WCT signal are used to generate the Goldberger Central Terminal signals as well. [Figure 63](#) shows an example of generating the augmented leads in analog domain. Note that in this implementation it takes more than eight channels to generate the standard 12 leads. Also, this feature is not available in the ADS1296/6R and ADS1294/4R.

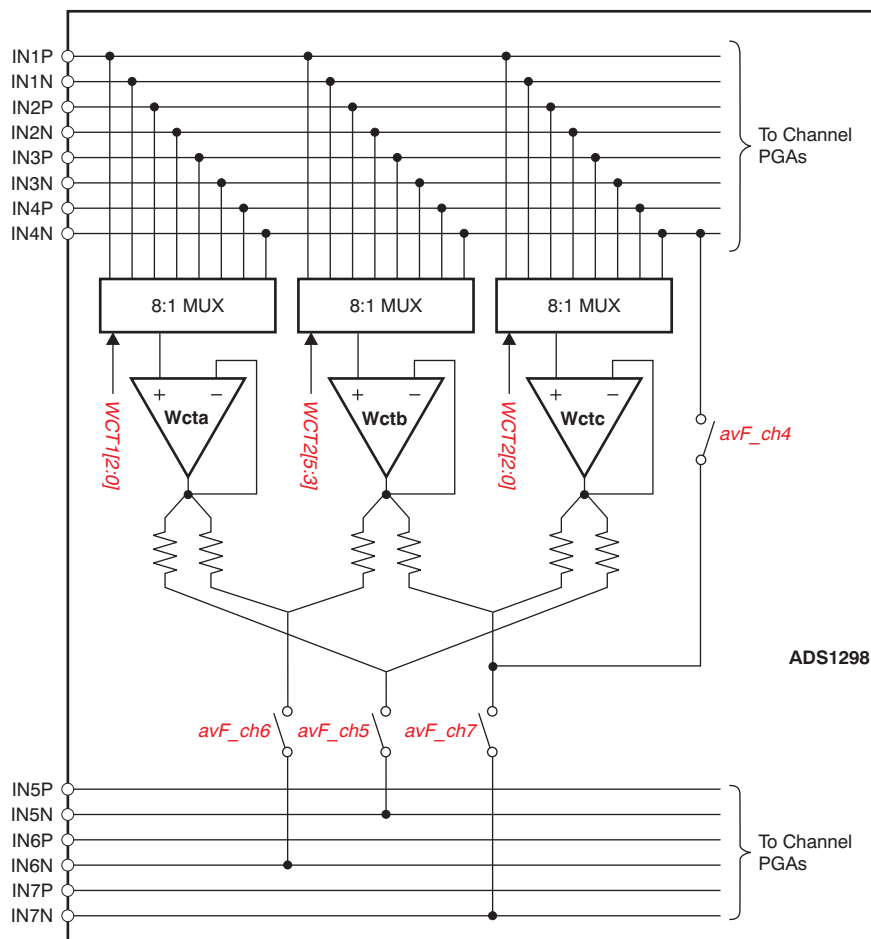


Figure 63. Analog Domain Augmented Leads

Right Leg Drive with the WCT Point

In certain applications, the out-of-phase version of the WCT is used as the right leg drive reference. The ADS1298 provides the option to have a buffered version of the WCT terminal at the RLD_OUT pin. This signal can be inverted in phase using an external amplifier and used as the right leg drive. Refer to the [Right Leg Drive \(RLD DC Bias Circuit\)](#) section for more details.

LEAD-OFF DETECTION

Patient electrode impedances are known to decay over time. It is necessary to continuously monitor these electrode connections to verify a suitable connection is present. The ADS129x lead-off detection functional block provides significant flexibility to the user to choose from various lead-off detection strategies. Though called lead-off detection, this is in fact an *electrode-off* detection.

The basic principle is to inject an excitation signal and measure the response to determine if the electrode is off. As shown in the lead-off detection functional block diagram in Figure 66, this circuit provides two different methods of determining the state of the patient electrode. The methods differ in the frequency content of the excitation signal. Lead-off can be selectively done on a per channel basis using the LOFF_SENSP and LOFF_SENSN registers. Also, the internal excitation circuitry can be disabled and just the sensing circuitry can be enabled.

DC Lead-Off

In this approach, the lead-off excitation is with a dc signal. The dc excitation signal can be chosen from either a pull-up/pull-down resistor or a current source/sink, shown in Figure 64. The selection is done by setting the VLEAD_OFF_EN bit in the LOFF register. One side of the channel is pulled to supply and the other side is pulled to ground. The pull-up resistor and pull-down resistor can be swapped (as shown in Figure 65) by setting the bits in the LOFF_FLIP register. In case of current source/sink, the magnitude of the current can be set by using the ILEAD_OFF[1:0] bits in the LOFF register. The current source/sink gives larger input impedance compared to the 10MΩ pull-up/pull-down resistor.

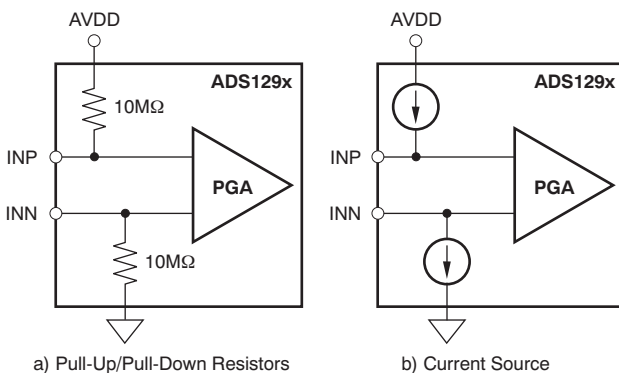


Figure 64. DC Lead-Off Excitation Options

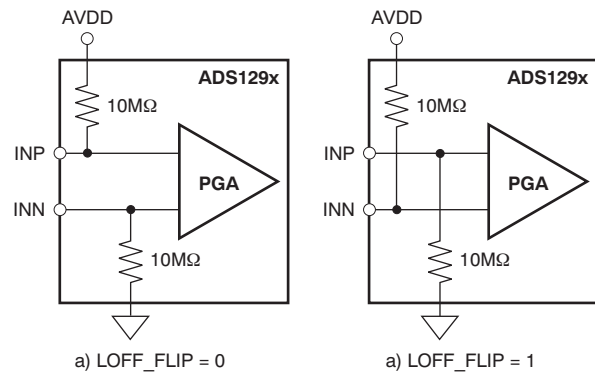


Figure 65. LOFF_FLIP Usage

Sensing of the response can be done either by looking at the digital output code from the device or by monitoring the input voltages with an on-chip comparator. If either of the electrodes is off, the pull-up resistors and/or the pull-down resistors saturate the channel. By looking at the output code it can be determined that either the P-side or the N-side is off. To pinpoint which one is off, the comparators must be used. The input voltage is also monitored using a comparator and a 4-bit DAC whose levels are set by the COMP_TH[2:0] bits in the LOFF register. The output of the comparators are stored in the LOFF_STATP and LOFF_STATN registers. These two registers are available as a part of the output data stream. (See the [Data Output Protocol \(DOUT\)](#) subsection of the [SPI Interface](#) section.) If dc lead-off is not used, the lead-off comparators can be powered down by setting the PD_LOFF_COMP bit in the CONFIG4 register.

An example procedure to turn on dc lead-off is given in the [Lead-Off](#) subsection of the [Guide to Get Up and Running](#) section.

This method uses an out-of-band ac signal for excitation. The ac signal is generated by alternatively providing pull-up resistors and pull-down resistors at the input with a fixed frequency. The ac signal is passed through an anti-aliasing filter to avoid aliasing. The frequency can be chosen by the FLEAD_OFF[1:0] bits in the LOFF register. The excitation frequency is a function of the output data rate and is $f_{DR}/4$. This out-of-band excitation signal is passed through the channel and measured at the output.

The schematic diagram illustrates the EMI Filter and Anti-Aliasing Filter circuitry. It shows the signal path from the Patient Electrode Contact Model through protection resistors and an EMI Filter to the Anti-Aliasing Filter. The filter includes a network of resistors (3.3MΩ, 7MΩ) and capacitors (12pF) connected to AVDD and AVSS. The output of the filter is connected to the PGA inputs (V_INP, V_INN) and the 4-Bit DAC. Control signals like FLEAD_OFF[0:1], LOFF_SENSP, and VLEAD_OFF_EN are shown.

68

RLD LEAD-OFF

The ADS129x provide two modes for determining whether the RLD is correctly connected:

- RLD lead-off detection during normal operation
- RLD lead-off detection during power-up

The following sections provide details of the two modes of operation.

RLD Lead-Off Detection During Normal Operation

During normal operation, the ADS129x RLD lead-off at power-up function cannot be used because it is necessary to power off the RLD amplifier.

RLD Lead-Off Detection At Power-Up

This feature is included in the ADS129x for use in determining whether the right leg electrode is suitably connected. At power-up, the ADS129x provide two measurement procedures to determine the RLD electrode connection status using either a current or a voltage pull-down resistor, as shown in Figure 67. The reference level of the comparator is set to determine the acceptable RLD impedance threshold.

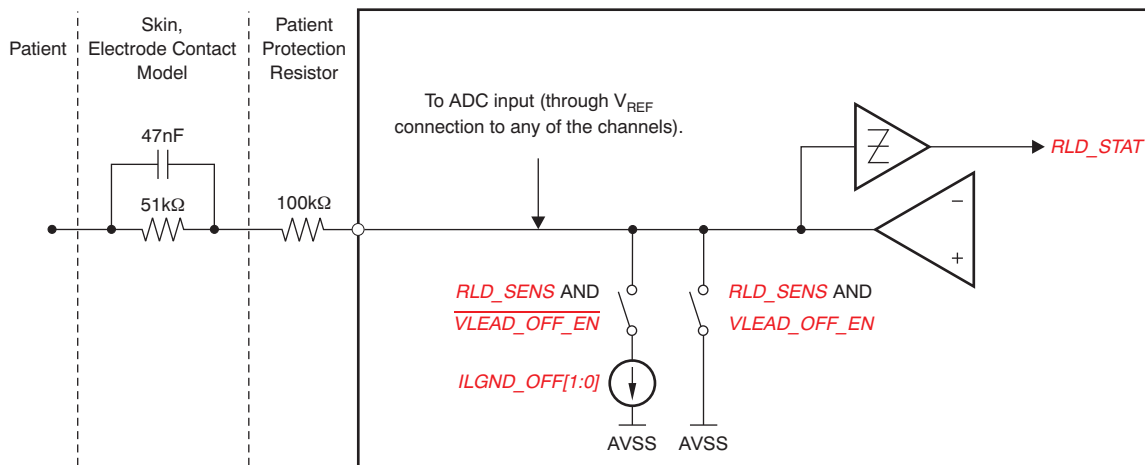


Figure 67. RLD Lead-Off Detection at Power-Up

When the RLD amplifier is powered on, the current source has no function. Only the comparator can be used to sense the voltage at the output of the RLD amplifier. The comparator thresholds are set by the same LOFF[7:5] bits used to set the thresholds for other negative inputs.

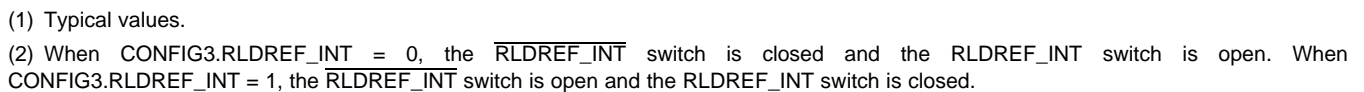
RIGHT LEG DRIVE (RLD DC BIAS CIRCUIT)

The right leg drive (RLD) circuitry is used as a means to counter the common-mode interference in a ECG system as a result of power lines and other sources, including fluorescent lights. The RLD circuit senses the common-mode of a selected set of electrodes and creates a negative feedback loop by driving the body with an inverted common-mode signal. The negative feedback loop restricts the common-mode movement to a narrow range, depending on the loop gain. Stabilizing the entire loop is specific to the individual user system based on the various poles in the loop. The ADS129x integrates the muxes to select the channel and an operational amplifier. All the amplifier terminals are available at the pins, allowing the user to choose the components for the feedback loop. The circuit shown in [Figure 68](#) shows the overall functional connectivity for the RLD bias circuit.

The reference voltage for the right leg drive can be chosen to be internally generated $(AVDD + AVSS)/2$ or it can be provided externally with a resistive divider. The selection of an internal versus external reference voltage for the RLD loop is defined by writing the appropriate value to the RLDREF_INT bit in the CONFIG3 register.

If the RLD function is not used, the amplifier can be powered down using the PD_RLD bit (see the [CONFIG3: Configuration Register 3](#) subsection of the [Register Map](#) section for details). This bit is also used in daisy-chain mode to power-down all but one of the RLD amplifiers.

The functionality of the RLDIN pin is explained in the [Input Multiplexer](#) section. An example procedure to use the RLD amplifier is shown in the [Right Leg Drive](#) subsection of the [Guide to Get Up and Running](#) section.



Product Folder Links: [ADS1294](#) [ADS1294R](#) [ADS1296](#) [ADS1296R](#) [ADS1298](#) [ADS1298R](#)

WCT as RLD

In certain applications, the right leg drive is derived as the average of RA, LA, and LL. This level is the same as the WCT voltage. The WCT amplifier has limited drive strength and thus should be used only to drive very high impedances directly. The ADS129x provide an option to internally buffer the WCT signal by setting the WCT_TO_RLD bit in the CONFIG4 register. The RLD_OUT and RLD_INV pins should be shorted external to the device. Note that before the RLD_OUT signal is connected to the RLD electrode, an external amplifier should be used to invert the phase of the signal for negative feedback.

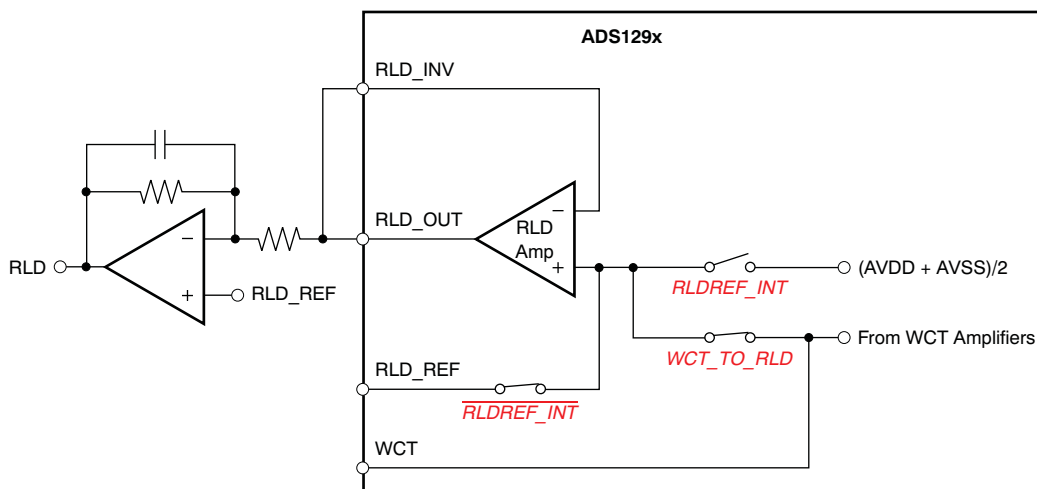


Figure 69. Using the WCT as the Right Leg Drive

RLD Configuration with Multiple Devices

Figure 70 shows multiple devices connected to an RLD.

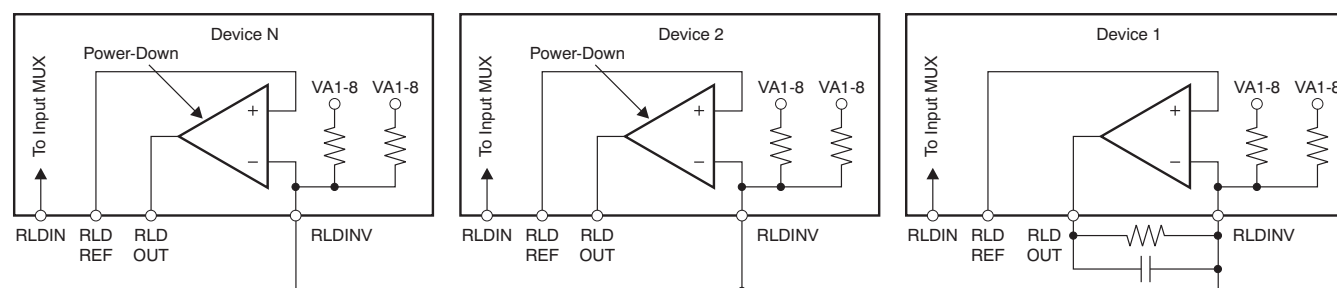


Figure 70. RLD Connection for Multiple Devices

PACE DETECT

The ADS129x provide flexibility for PACE detection either in software or by external hardware. The software approach is made possible by providing sampling rates up to 32kSPS. The external hardware approach is made possible by bringing out the output of the PGA at two pins: TESTP_PACE_OUT1 and TESTN_PACE_OUT2. Note that if the WCT amplifier is connected to the signal path, the user sees switching noise as a result of chopping; see the [Wilson Central Terminal \(WCT\)](#) section for details.

Software Approach

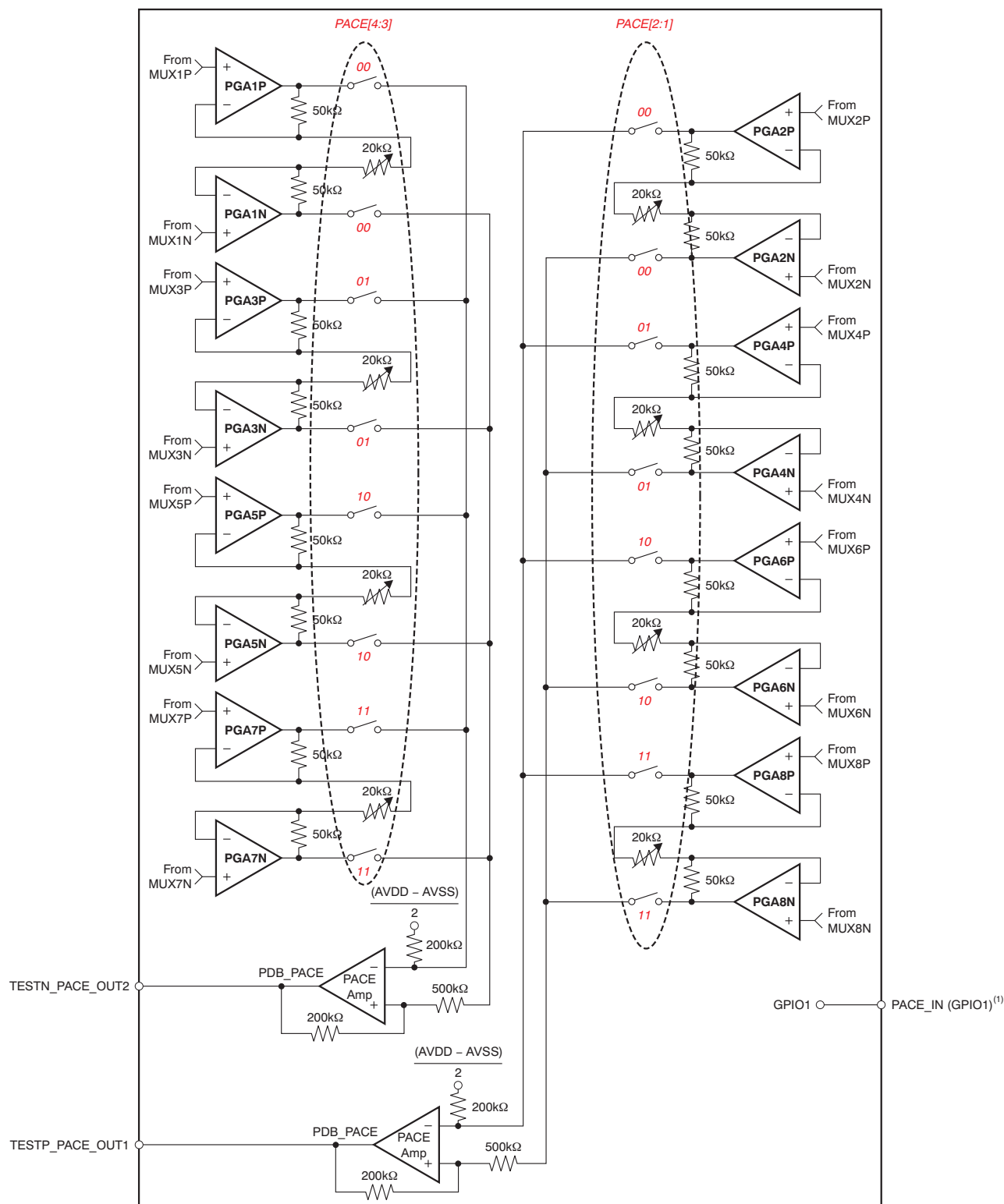
To use the software approach, the device must be operated at 8kSPS or more to be able to capture the fastest pulse. Afterwards, digital signal processing can be used to identify the presence of the pacemaker pulse. The software approach gives the utmost flexibility to the user to be able to program the PACE detect threshold on the fly using software. This becomes increasingly important as pacemakers evolve over time. Two parameters must be considered while measuring fast PACE pulses:

1. The PGA bandwidth shown in [Table 16](#).
2. For a step change in input, the digital decimation filter takes $3 \times t_{DR}$ to settle. The PGA bandwidth determines the gain setting that can be used and the settling time determines the data rate that the device must be operated at.

External Hardware Approach

One of the drawbacks of using the software approach is that all channels on a single device must operate at higher data rates. For systems where it is of concern, the ADS129x provide the option of bringing out the output of the PGA. External hardware circuitry can be used to detect the presence of the pulse. The output of the PACE detection logic can then be fed into the device through one of the GPIO pins. The GPIO data are transmitted through the SPI port and loaded $2t_{CLKS}$ before \overline{DRDY} goes low. Two of the eight channels can be selected using register bits in the PACE register, one from the odd-numbered channels and the other from the even-numbered channels. During the differential to single-ended conversion, there is an attenuation of 0.4. Therefore, the total gain in the PACE path is equal to $(0.4 \times PGA_GAIN)$. The PACE out signals are multiplexed with the TESTP and TESTN signals through the TESTP_PACE_OUT1 and TESTN_PACE_OUT2 pins respectively. The channel selection is done by setting bits[4:1] of the PACE register. If the PACE circuitry is not used, the PACE amplifiers can be turned off using the $\overline{PD_PACE}$ bit in the PACE register.

Note that if the output of a channel connected to the WCT amplifier (for example, the V lead channels) is connected to one of the PACE amplifiers for external PACE detection, the artifact of chopping appears at the PACE amplifier output. Refer to the [Wilson Central Terminal \(WCT\)](#) section for more details.



(1) GPIO1 can be used as the PACE_IN signal.

Figure 71. Hardware PACE Detection Option

RESPIRATION

The ADS1294R/6R/8R provide three modes for respiration impedance measurement: external respiration, internal respiration using on-chip modulation signals, and internal respiration using user-generated modulation signals, as shown in [Table 18](#).

Table 18. Respiration Control

| RESP.RESP_CTRL[1] | RESP.RESP_CTRL[0] | MODE AVAILABLE | DESCRIPTION |
|-------------------|-------------------|---|--|
| 0 | 0 | ADS1294, ADS1296, ADS1298, ADS1294R, ADS1296R, ADS1298R | Respiration disabled |
| 0 | 1 | ADS1294, ADS1296, ADS1298, ADS1294R, ADS1296R, ADS1298R | Generates modulation and demodulation signals for external respiration circuitry. RESP CLK signals on GPIO2, GPIO3, and GPIO4. |
| 1 | 0 | ADS1294R, ADS1296R, ADS1298R | Respiration measurement using internally-generated RESP_MOD signals. |
| 1 | 1 | ADS1294R, ADS1296R, ADS1298R ⁽¹⁾ | Respiration measurement using user-generated modulation and blocking signal. |

For more information on respiration impedance measurement, refer to application note [Respiration Rate Measurement Using Impedance Pneumography \(SBAA181\)](#).

(1) RESP_CTRL[1:0] = 11 is not recommend if CLKSEL = 1 (internal master clock).

External Respiration Circuitry Option (RESP_CTRL = 01b)

In this mode, GPIO2, GPIO3, and GPIO4 are automatically configured as outputs. The phase relationship between the signals is shown in [Figure 72](#). GPIO2 is the exclusive-OR of GPIO3 and GPIO4, as shown in [Figure 73](#). GPIO3 is the modulation signal, and GPIO4 is the de-modulation signal. While in this mode, the general-purpose pin function of GPIO2, GPIO3, and GPIO4 is not available. The modulation frequency can be 64kHz or 32kHz using RESP_FREQ[2:0] bits in the CONFIG4 register. The remaining bit options of RESP_FREQ[2:0] generate square waves on GPIO3 and GPIO4. The exclusive-OR out on GPIO2 is only available on 64kHz and 32kHz modes. The phase of GPIO4, relative to GPIO3, is set by RESP_PH[2:0] bits in the RESP register.

The mode can be used when the user implements custom respiration impedance circuitry external to the ADS129x.

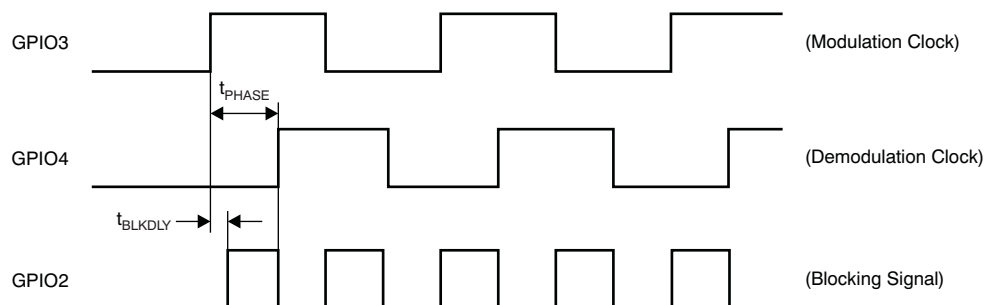


Figure 72. External Respiration (RESP_CTRL = 01b) Timing Diagram

Table 19. Timing Characteristics for [Figure 72](#) ⁽¹⁾

| PARAMETER | DESCRIPTION | 2.7V ≤ DVDD ≤ 3.6V | | | 1.65V ≤ DVDD ≤ 2V | | | UNIT |
|---------------------|--|--------------------|-----|-------|-------------------|-----|-------|---------|
| | | MIN | TYP | MAX | MIN | TYP | MAX | |
| t _{PHASE} | Respiration phase delay, set by RESP.RESP_PH[2:0] bits | 22.5 | | 157.5 | 22.5 | | 157.5 | Degrees |
| t _{BLKDLY} | Modulation clock rising edge to XOR signal | | 1 | | | 5 | | ns |

(1) Specifications apply from –40°C to +85°C.

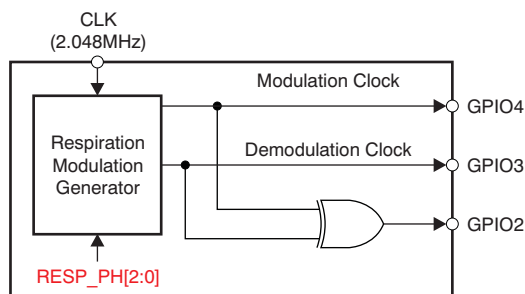


Figure 73. External Respiration (RESP_CTRL = 01b) Block Diagram

Internal Respiration Circuitry with Internal Clock (RESP_CTRL = 10b, ADS1294R/6R/8R Only)

Figure 74 shows a block diagram of the internal respiration circuitry. The internal modulation and demodulator circuitry can be selectively used. The modulation block is controlled by the RESP_MOD_EN bit and the demodulation block is controlled by the RESP_DEMOD_EN bit. The modulation signal is a square wave of magnitude $V_{REFP} - AV_{SS}$. In this mode, the output of the modulation circuitry is available at the RESP_MODP and RESP_MODN terminals of the device. This availability allows custom filtering to be added to the square wave modulation signal. In this mode, GPIO2, GPIO3, and GPIO4 can be used for other purposes. The modulation frequency can be 64kHz or 32kHz, as set by the RESP_FREQ[2:0] bits in the CONFIG4 register. The phase of the internal demodulation signal is set by the RESP_PH[2:0] bits in the RESP register.

The ADS1294R/6R/8R channel 1 with respiration enabled mode cannot be used to acquire ECG signals. If the RA and LA leads are intended to measure respiration and ECG signals, the two leads can be wired into channel 1 for respiration and channel 2 for ECG signals.

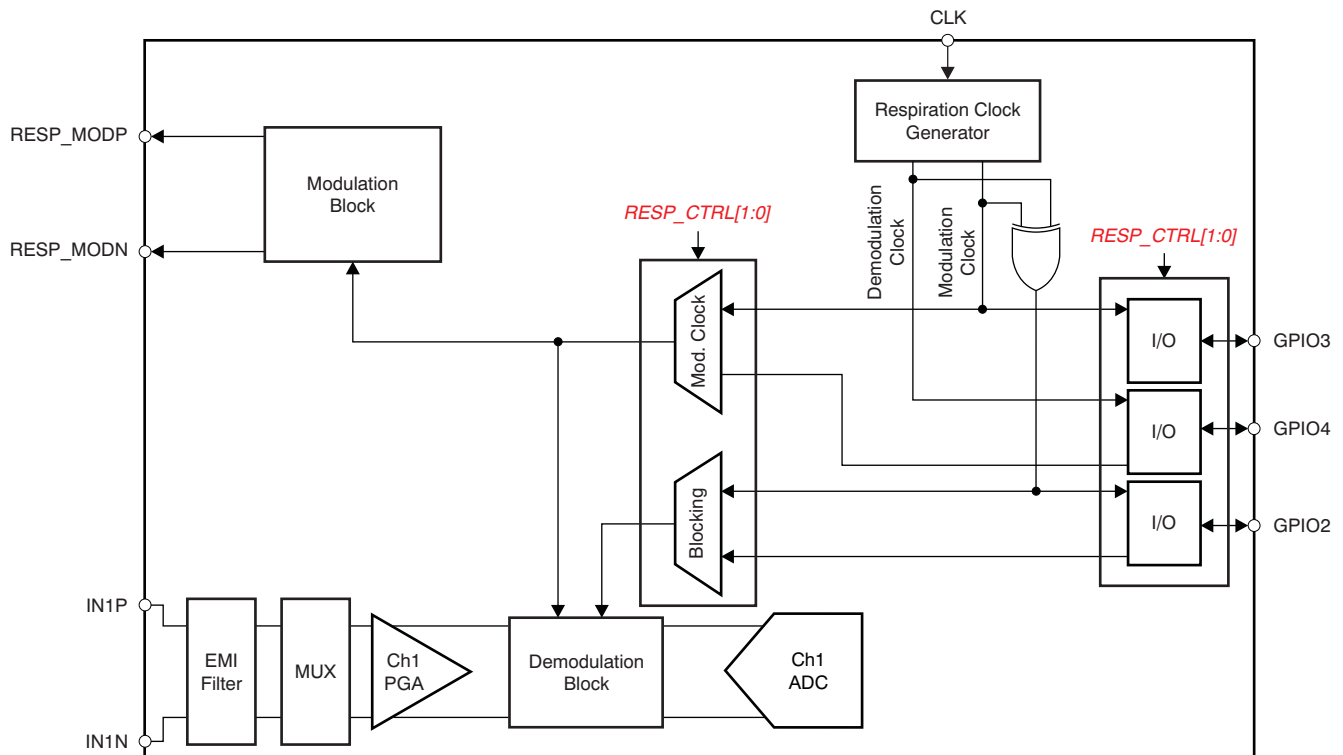


Figure 74. Internal Respiration Block Diagram

Internal Respiration Circuitry with User-Generated Signals (RESP_CTRL = 11b, ADS1294R/6R/8R Only)

In this mode GPIO2, GPIO3, and GPIO4 are automatically configured as inputs. GPIO2, GPIO3, and GPIO4 cannot be used for other purposes. The signals must be provided as described in Figure 75. The internal master clock is not recommended in this mode.

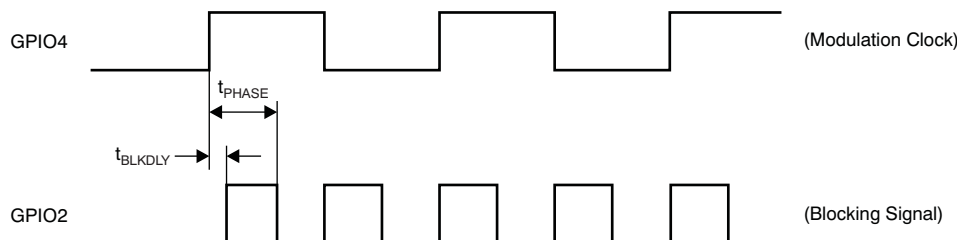


Figure 75. Internal Respiration (RESP_CTRL = 11b) Timing Diagram

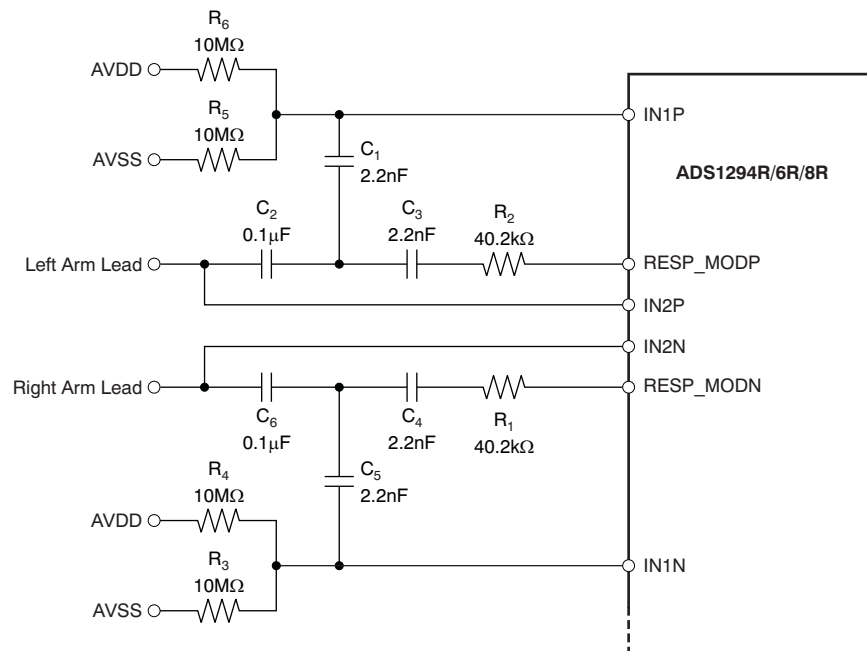
Table 20. Timing Characteristics for Figure 75 ⁽¹⁾

| PARAMETER | DESCRIPTION | 1.65V ≤ DVDD ≤ 3.6V | | | UNIT |
|---------------------|--|---------------------|-----|-------|---------|
| | | MIN | TYP | MAX | |
| t _{PHASE} | Respiration phase delay | 0 | | 157.5 | Degrees |
| t _{BLKDLY} | Modulation clock rising edge to XOR signal | | 0 | 5 | ns |

(1) Specifications apply from –40°C to +85°C.

ADS129xR Application

The ADS1294R, ADS1296R, and ADS1298R channel 1 with respiration enabled mode cannot be used to acquire ECG signals. If the RA and LA leads are intended to measure respiration and ECG signals, the two leads can be wired into channel 1 for respiration and channel 2 for ECG signals, as shown in Figure 76.



NOTE: Patient and input protection circuitry not shown.

Figure 76. Typical Respiration Circuitry

Figure 77 shows a respiration test circuit. Figure 78 and Figure 79 plot noise on channel 1 for the ADS1294R/6R/8R as baseline impedance, gain, and phase are swept. The x-axis is the baseline impedance, normalized to a 29μA modulation current (as shown in Equation 8).

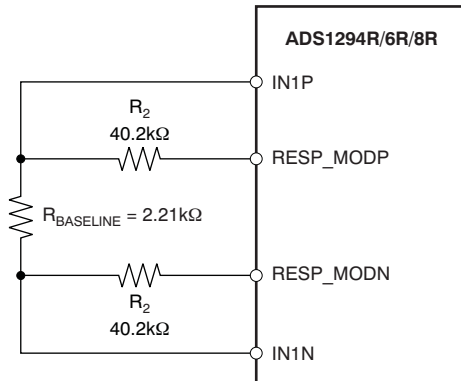


Figure 77. Respiration Noise Test Circuit

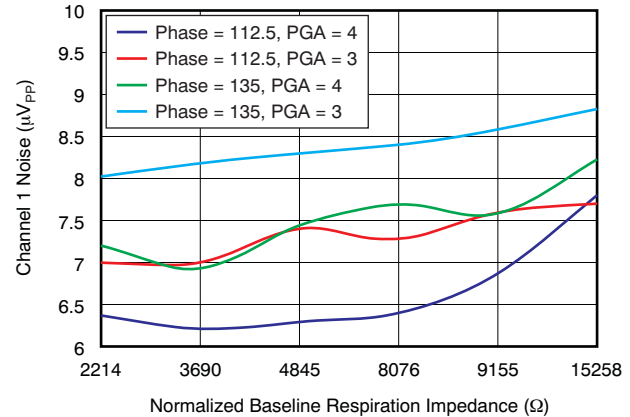


Figure 78. Channel 1 Noise versus Impedance for 32kHz Modulation Clock and Phase (BW = 150Hz, Respiration Modulation Clock = 32kHz)

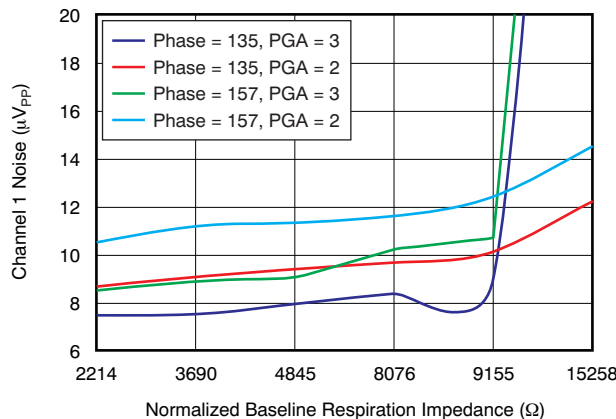


Figure 79. Channel 1 Noise versus Impedance for 64kHz Modulation Clock and Phase (BW = 150Hz, Respiration Modulation Clock = 64kHz)

$$R_{\text{NORMALIZED}} = \frac{R_{\text{ACTUAL}} \times I_{\text{ACTUAL}}}{29\mu\text{A}}$$

where:

R_{ACTUAL} is the baseline body impedance,

I_{ACTUAL} is the modulation current, as calculated by $(V_{\text{REFP}} - AV_{\text{SS}})$ divided by the impedance of the modulation circuit. (8)

For example, if modulation frequency = 32kHz, $R_{\text{ACTUAL}} = 3\text{k}\Omega$, $I_{\text{ACTUAL}} = 50\mu\text{A}$, and $R_{\text{NORMALIZED}} = (3\text{k}\Omega \times 50\mu\text{A})/29\mu\text{A} = 5.1\text{k}\Omega$.

Referring to Figure 78 and Figure 79, it can be noted that gain = 4 and phase = 112.5° yield the best performance at 6.4μV_{PP}. Low-pass filtering this signal with a high-order 2Hz cutoff can reduce the noise to less than 600nV_{PP}. The impedance resolution is 600nV_{PP}/29μA = 20mΩ.

When the modulation frequency is 32kHz, gains of 3 and 4 and phase of 112.5° and 135° are recommended.

When the modulation frequency is 64kHz, gains of 2 and 3 and phase of 135° and 157° are recommended for best performance.

PCB LAYOUT

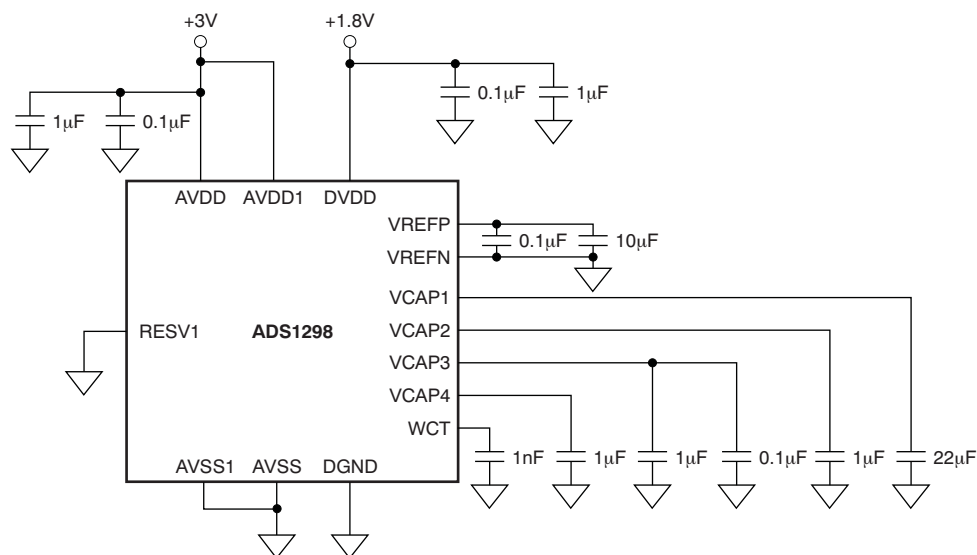
POWER SUPPLIES AND GROUNDING

The ADS129x have three supplies: AVDD, AVDD1, and DVDD. Both AVDD and AVDD1 should be as quiet as possible. AVDD1 provides the supply to the charge pump block and has transients at f_{CLK} . Therefore, it is recommended that AVDD1 and AVSS1 be star-connected to AVDD and AVSS. It is important to eliminate noise from AVDD and AVDD1 that is non-synchronous with the ADS129x operation. Each supply of the ADS129x should be bypassed with $1\mu\text{F}$ and $0.1\mu\text{F}$ solid ceramic capacitors. It is recommended that placement of the digital circuits (DSP, microcontrollers, FPGAs, etc.) in the system is done such that the return currents on those devices do not cross the analog return path of the ADS129x. The ADS129x can be powered from unipolar or bipolar supplies.

Capacitors used for decoupling can be of surface-mount, low-cost, low-profile, multi-layer ceramic type. In most cases, the VCAP1 capacitor can also be a multi-layer ceramic, but in systems where the board is subjected to high- or low-frequency vibration, it is recommend to install a non-ferroelectric capacitor such as a tantalum or class 1 capacitor (C0G or NPO). EIA class 2 and class 3 dielectrics such as (X7R, X5R, X8R, etc.) are ferroelectric. The piezoelectric property of these capacitors can appear as electrical noise coming from the capacitor. When using internal reference, noise on the VCAP1 node results in performance degradation.

Connecting the Device to Unipolar (+3V/+1.8V) Supplies

Figure 80 illustrates the ADS129x connected to a unipolar supply. In this example, analog supply (AVDD) is referenced to analog ground (AVSS) and digital supplies (DVDD) are referenced to digital ground (DGND).

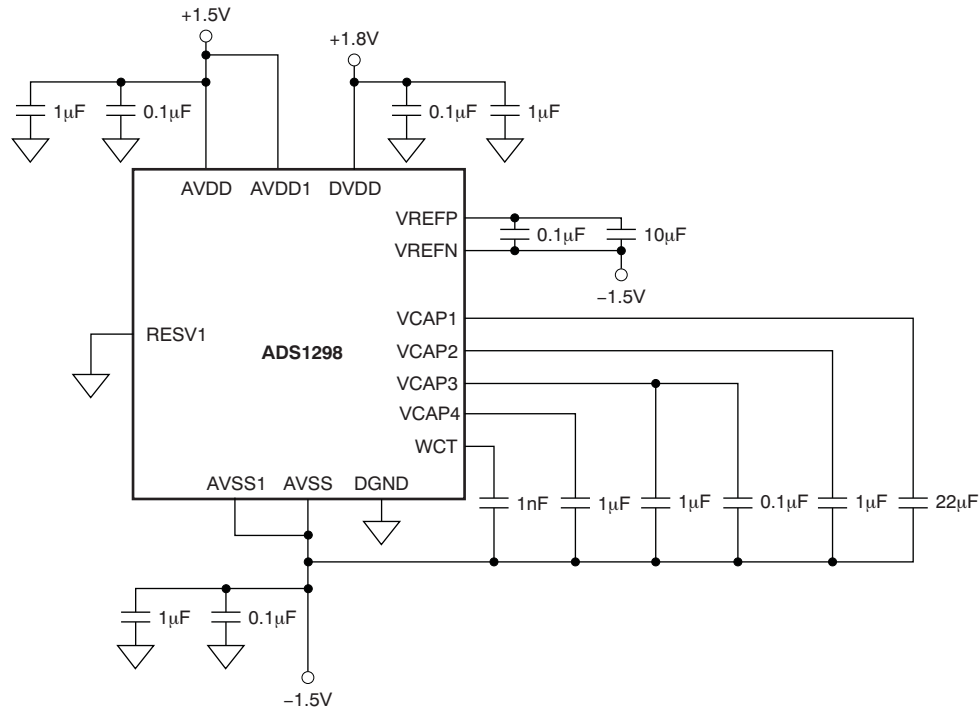


NOTE: Place the capacitors for supply, reference, WCT, and VCAP1 to VCAP4 as close to the package as possible.

Figure 80. Single-Supply Operation

Connecting the Device to Bipolar ($\pm 1.5\text{V}/1.8\text{V}$) Supplies

Figure 81 illustrates the ADS129x connected to a bipolar supply. In this example, the analog supplies connect to the device analog supply (AVDD). This supply is referenced to the device analog return (AVSS), and the digital supply (DVDD) is referenced to the device digital ground return (DGND).



NOTE: Place the capacitors for supply, reference, WCT, and VCAP1 to VCAP4 as close to the package as possible.

Figure 81. Bipolar Supply Operation

SHIELDING ANALOG SIGNAL PATHS

As with any precision circuit, careful PCB layout ensures the best performance. It is essential to make short, direct interconnections and avoid stray wiring capacitance—particularly at the analog input pins and AVSS. These analog input pins are high-impedance and extremely sensitive to extraneous noise. The AVSS pin should be treated as a sensitive analog signal and connected directly to the supply ground with proper shielding. Leakage currents between the PCB traces can exceed the input bias current of the ADS129x if shielding is not implemented. Digital signals should be kept as far as possible from the analog input signals on the PCB.

ANALOG INPUT STRUCTURE

The analog input of the ADS129x is as shown in Figure 82.

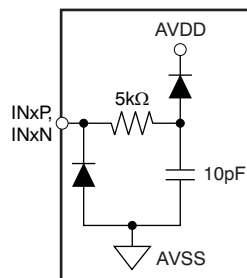


Figure 82. Analog Input Protection Circuit

REVISION HISTORY

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

| Changes from Revision I (January 2012) to Revision J | Page |
|--|------|
| • Changed NC pin discription in <i>Pin Assignments</i> table | 13 |
| • Changed NC pin discription in <i>Pin Assignments</i> table | 13 |
| • Added graph of INTERNAL V_{REF} DRIFT vs TEMPERATURE | 19 |
| • Changed units in TEST_AMP bit description in CONFIG2 register | 25 |
| • Changed order of subsections in the <i>Theory of Operation</i> section | 36 |
| • Added <i>SCLK Clocking Methods</i> section | 36 |
| • Changed <i>Power-Up Sequencing</i> section text to clarify start-up timing | 42 |
| • Changed Figure 31 | 42 |
| • Changed power-up reset wait time in Table 11 | 42 |
| • Added <i>Data Acquisition</i> section | 43 |
| • Updated Figure 32 | 43 |
| • Added Figure 33 | 44 |
| • Added discussion of SCLK/ \overline{DRDY} bus behavior to <i>Data Ready (\overline{DRDY})</i> section | 45 |
| • Added Figure 35 | 45 |
| • Added <i>STATUS Word</i> section and Figure 36 to discuss the STATUS word | 45 |
| • Added <i>Readback Length</i> section | 46 |
| • Changed Figure 40 to clarify Initial Flow at Power-Up | 48 |
| • Changed single-ended input description to correct input range values | 55 |
| • Changed Figure 49 to show correct input range for single-ended inputs | 55 |
| • Changed Figure 50 to show correct input range for single-ended inputs | 55 |
| • Deleted text regarding large scale signal | 56 |
| • Changed Figure 54 to provide a more stable external reference driver circuit | 58 |

Changes from Revision H (October 2011) to Revision I
Page

| | |
|--|----|
| • Added eighth Features bullet (list of standards supported) | 1 |
| • Deleted duplicate <i>Digital input voltage</i> and <i>Digital output voltage</i> rows from Absolute Maximum Ratings table | 2 |
| • Changed parameter name of Channel Performance, <i>Common-mode rejection ratio</i> and <i>Power-supply rejection ratio</i> parameters in Electrical Characteristics table | 4 |
| • Updated BGA pin out | 10 |
| • Updated Figure 24 | 21 |
| • Added power-down recommendation to bit 7 description of <i>CHnSET: Individual Channel Settings</i> section | 28 |
| • Changed description of bit 5 in <i>RESP: Respiration Control Register</i> section | 32 |
| • Corrected name of bit 6 in <i>WCT2: Wilson Central Terminal Control Register</i> section | 35 |
| • Changed description of START pin in <i>START</i> section | 43 |
| • Changed description of <i>Data Ready (DRDY)</i> section | 44 |
| • Changed conversion description in <i>Single-Shot Mode</i> section | 46 |
| • Changed conversion description in <i>Continuous Mode</i> section | 47 |
| • Changed Unit column in Table 10 | 47 |
| • Updated description of <i>Analog Input</i> section | 55 |
| • Updated Figure 52 | 57 |
| • Updated Figure 58 | 62 |
| • Updated Figure 59 | 63 |

| Changes from Revision G (February 2011) to Revision H | Page |
|---|------|
| • Deleted <i>Non-Magnetic BGA Package</i> Features bullet | 1 |
| • Added (ADS1298) to <i>High-Resolution mode</i> and <i>Low-Power mode</i> subsection headers of Supply Current section in Electrical Characteristics table | 6 |
| • Changed 3V Power Dissipation, <i>Quiescent channel power</i> test conditions in Electrical Characteristics table | 6 |
| • Changed 5V Power Dissipation, <i>Quiescent channel power</i> test conditions in Electrical Characteristics table | 7 |
| • Changed footnote 1 of BGA Pin Assignments table | 10 |
| • Added footnote 1 cross-reference to <i>RLDIN</i> , <i>TESTP_PACE_OUT1</i> , and <i>TESTP_PACE_OUT</i> in BGA Pin Assignments table | 11 |
| • Changed footnote 1 of PAG Pin Assignments table | 13 |
| • Added footnote 1 cross-reference to <i>TESTP_PACE_OUT1</i> , <i>TESTP_PACE_OUT2</i> , and <i>RLDIN</i> in PAG Pin Assignments table | 13 |
| • Changed description of <i>AVSS</i> and <i>AVDD</i> in PAG Pin Assignments table | 14 |
| • Changed title of Figure 20 | 18 |
| • Changed bit name for bits 5, 6, and 7 in ID register of Table 9 | 22 |
| • Changed bit name for bits 5, 6, and 7 in <i>ID: ID Control Register</i> section | 23 |
| • Changed description of <i>STANDBY: Enter STANDBY Mode</i> section | 38 |
| • Changed title of Table 13 | 46 |
| • Updated Equation 3 | 51 |
| • Updated Figure 56 | 60 |
| • Updated Figure 67 | 69 |
| • Added new paragraph to <i>Respiration</i> section | 75 |
| • Added footnote to Figure 76 | 78 |
| • Changed description of solid ceramic capacitor in <i>Power Supplies and Grounding</i> section | 80 |
| • Changed description of <i>Connecting the Device to Bipolar ($\pm 1.5V/1.8V$) Supplies</i> section | 81 |

PACKAGING INFORMATION

| Orderable Device | Status (1) | Package Type | Package Drawing | Pins | Package Qty | Eco Plan (2) | Lead/Ball Finish (6) | MSL Peak Temp (3) | Op Temp (°C) | Device Marking (4/5) | Samples |
|------------------|---------------|--------------|--------------------|------|----------------|----------------------------|-------------------------|----------------------|--------------|-------------------------|-------------------------|
| ADS1294CZXGR | ACTIVE | NFBGA | ZXG | 64 | 1000 | Green (RoHS & no Sb/Br) | SNAGCU | Level-3-260C-168 HR | 0 to 70 | ADS1294 | Samples |
| ADS1294CZXGT | ACTIVE | NFBGA | ZXG | 64 | 250 | Green (RoHS & no Sb/Br) | SNAGCU | Level-3-260C-168 HR | 0 to 70 | ADS1294 | Samples |
| ADS1294IPAG | ACTIVE | TQFP | PAG | 64 | 160 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR | -40 to 85 | ADS1294 | Samples |
| ADS1294IPAGR | ACTIVE | TQFP | PAG | 64 | 1500 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR | -40 to 85 | ADS1294 | Samples |
| ADS1294RIZXGR | ACTIVE | NFBGA | ZXG | 64 | 1000 | Green (RoHS & no Sb/Br) | SNAGCU | Level-3-260C-168 HR | -40 to 85 | ADS1294R | Samples |
| ADS1294RIZXGT | ACTIVE | NFBGA | ZXG | 64 | 250 | Green (RoHS & no Sb/Br) | SNAGCU | Level-3-260C-168 HR | -40 to 85 | ADS1294R | Samples |
| ADS1296CZXGR | ACTIVE | NFBGA | ZXG | 64 | 1000 | Green (RoHS & no Sb/Br) | SNAGCU | Level-3-260C-168 HR | 0 to 70 | ADS1296 | Samples |
| ADS1296CZXGT | ACTIVE | NFBGA | ZXG | 64 | 250 | Green (RoHS & no Sb/Br) | SNAGCU | Level-3-260C-168 HR | 0 to 70 | ADS1296 | Samples |
| ADS1296IPAG | ACTIVE | TQFP | PAG | 64 | 160 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR | -40 to 85 | ADS1296 | Samples |
| ADS1296IPAGR | ACTIVE | TQFP | PAG | 64 | 1500 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR | -40 to 85 | ADS1296 | Samples |
| ADS1296RIZXGR | ACTIVE | NFBGA | ZXG | 64 | 1000 | Green (RoHS & no Sb/Br) | SNAGCU | Level-3-260C-168 HR | -40 to 85 | ADS1296R | Samples |
| ADS1296RIZXGT | ACTIVE | NFBGA | ZXG | 64 | 250 | Green (RoHS & no Sb/Br) | SNAGCU | Level-3-260C-168 HR | -40 to 85 | ADS1296R | Samples |
| ADS1298CZXGR | ACTIVE | NFBGA | ZXG | 64 | 1000 | Green (RoHS & no Sb/Br) | SNAGCU | Level-3-260C-168 HR | 0 to 70 | ADS1298 | Samples |
| ADS1298CZXGT | ACTIVE | NFBGA | ZXG | 64 | 250 | Green (RoHS & no Sb/Br) | SNAGCU | Level-3-260C-168 HR | 0 to 70 | ADS1298 | Samples |
| ADS1298IPAG | ACTIVE | TQFP | PAG | 64 | 160 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR | -40 to 85 | ADS1298 | Samples |
| ADS1298IPAGR | ACTIVE | TQFP | PAG | 64 | 1500 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-3-260C-168 HR | -40 to 85 | ADS1298 | Samples |
| ADS1298RIZXGR | ACTIVE | NFBGA | ZXG | 64 | 1000 | Green (RoHS & no Sb/Br) | SNAGCU | Level-3-260C-168 HR | -40 to 85 | ADS1298R | Samples |

| Orderable Device | Status (1) | Package Type | Package Drawing | Pins | Package Qty | Eco Plan (2) | Lead/Ball Finish (6) | MSL Peak Temp (3) | Op Temp (°C) | Device Marking (4/5) | Samples |
|------------------|---------------|--------------|--------------------|------|----------------|----------------------------|-------------------------|----------------------|--------------|-------------------------|-------------------------|
| ADS1298RIZXGT | ACTIVE | NFBGA | ZXG | 64 | 250 | Green (RoHS & no Sb/Br) | SNAGCU | Level-3-260C-168 HR | -40 to 85 | ADS1298R | Samples |

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION


*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
|---------------|--------------|-----------------|------|------|--------------------|--------------------|---------|---------|---------|---------|--------|---------------|
| ADS1294CZXGR | NFBGA | ZXG | 64 | 1000 | 330.0 | 16.4 | 8.3 | 8.3 | 2.25 | 12.0 | 16.0 | Q1 |
| ADS1294CZXGT | NFBGA | ZXG | 64 | 250 | 330.0 | 16.4 | 8.3 | 8.3 | 2.25 | 12.0 | 16.0 | Q1 |
| ADS1294IPAGR | TQFP | PAG | 64 | 1500 | 330.0 | 24.4 | 13.0 | 13.0 | 1.5 | 16.0 | 24.0 | Q2 |
| ADS1294RIZXGR | NFBGA | ZXG | 64 | 1000 | 330.0 | 16.4 | 8.3 | 8.3 | 2.25 | 12.0 | 16.0 | Q1 |
| ADS1294RIZXGT | NFBGA | ZXG | 64 | 250 | 330.0 | 16.4 | 8.3 | 8.3 | 2.25 | 12.0 | 16.0 | Q1 |
| ADS1296CZXGR | NFBGA | ZXG | 64 | 1000 | 330.0 | 16.4 | 8.3 | 8.3 | 2.25 | 12.0 | 16.0 | Q1 |
| ADS1296CZXGT | NFBGA | ZXG | 64 | 250 | 330.0 | 16.4 | 8.3 | 8.3 | 2.25 | 12.0 | 16.0 | Q1 |
| ADS1296IPAGR | TQFP | PAG | 64 | 1500 | 330.0 | 24.4 | 13.0 | 13.0 | 1.5 | 16.0 | 24.0 | Q2 |
| ADS1296RIZXGR | NFBGA | ZXG | 64 | 1000 | 330.0 | 16.4 | 8.3 | 8.3 | 2.25 | 12.0 | 16.0 | Q1 |
| ADS1296RIZXGT | NFBGA | ZXG | 64 | 250 | 330.0 | 16.4 | 8.3 | 8.3 | 2.25 | 12.0 | 16.0 | Q1 |
| ADS1298CZXGR | NFBGA | ZXG | 64 | 1000 | 330.0 | 16.4 | 8.3 | 8.3 | 2.25 | 12.0 | 16.0 | Q1 |
| ADS1298CZXGT | NFBGA | ZXG | 64 | 250 | 330.0 | 16.4 | 8.3 | 8.3 | 2.25 | 12.0 | 16.0 | Q1 |
| ADS1298IPAGR | TQFP | PAG | 64 | 1500 | 330.0 | 24.4 | 13.0 | 13.0 | 1.5 | 16.0 | 24.0 | Q2 |
| ADS1298RIZXGR | NFBGA | ZXG | 64 | 1000 | 330.0 | 16.4 | 8.3 | 8.3 | 2.25 | 12.0 | 16.0 | Q1 |
| ADS1298RIZXGT | NFBGA | ZXG | 64 | 250 | 330.0 | 16.4 | 8.3 | 8.3 | 2.25 | 12.0 | 16.0 | Q1 |

TAPE AND REEL BOX DIMENSIONS

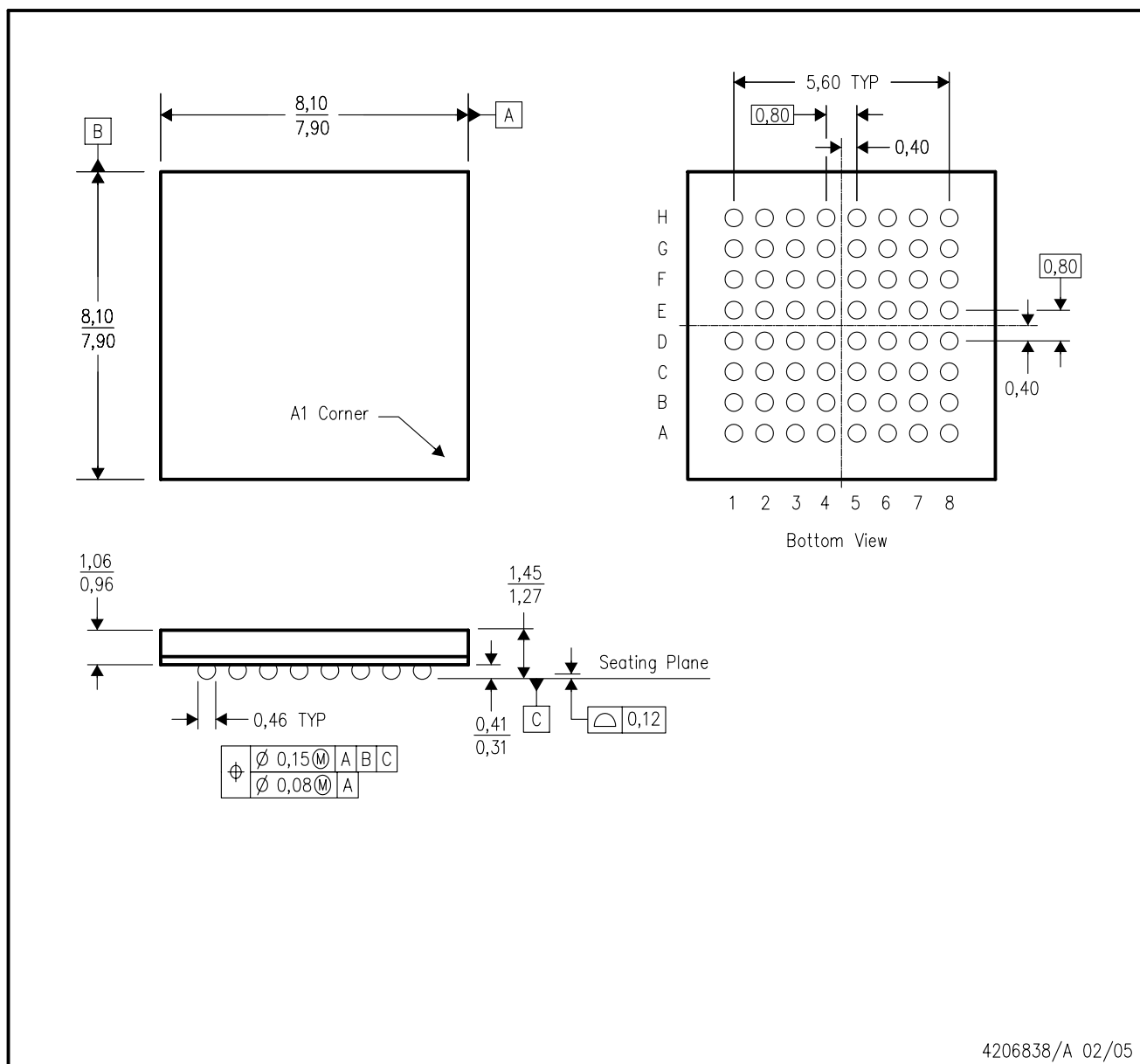


*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Length (mm) | Width (mm) | Height (mm) |
|---------------|--------------|-----------------|------|------|-------------|------------|-------------|
| ADS1294CZXGR | NFBGA | ZXG | 64 | 1000 | 336.6 | 336.6 | 28.6 |
| ADS1294CZXGT | NFBGA | ZXG | 64 | 250 | 336.6 | 336.6 | 28.6 |
| ADS1294IPAGR | TQFP | PAG | 64 | 1500 | 367.0 | 367.0 | 45.0 |
| ADS1294RIZXGR | NFBGA | ZXG | 64 | 1000 | 336.6 | 336.6 | 28.6 |
| ADS1294RIZXGT | NFBGA | ZXG | 64 | 250 | 336.6 | 336.6 | 28.6 |
| ADS1296CZXGR | NFBGA | ZXG | 64 | 1000 | 336.6 | 336.6 | 28.6 |
| ADS1296CZXGT | NFBGA | ZXG | 64 | 250 | 336.6 | 336.6 | 28.6 |
| ADS1296IPAGR | TQFP | PAG | 64 | 1500 | 367.0 | 367.0 | 45.0 |
| ADS1296RIZXGR | NFBGA | ZXG | 64 | 1000 | 336.6 | 336.6 | 28.6 |
| ADS1296RIZXGT | NFBGA | ZXG | 64 | 250 | 336.6 | 336.6 | 28.6 |
| ADS1298CZXGR | NFBGA | ZXG | 64 | 1000 | 336.6 | 336.6 | 28.6 |
| ADS1298CZXGT | NFBGA | ZXG | 64 | 250 | 336.6 | 336.6 | 28.6 |
| ADS1298IPAGR | TQFP | PAG | 64 | 1500 | 367.0 | 367.0 | 45.0 |
| ADS1298RIZXGR | NFBGA | ZXG | 64 | 1000 | 336.6 | 336.6 | 28.6 |
| ADS1298RIZXGT | NFBGA | ZXG | 64 | 250 | 336.6 | 336.6 | 28.6 |

ZXG (S-PBGA-N64)

PLASTIC BALL GRID ARRAY

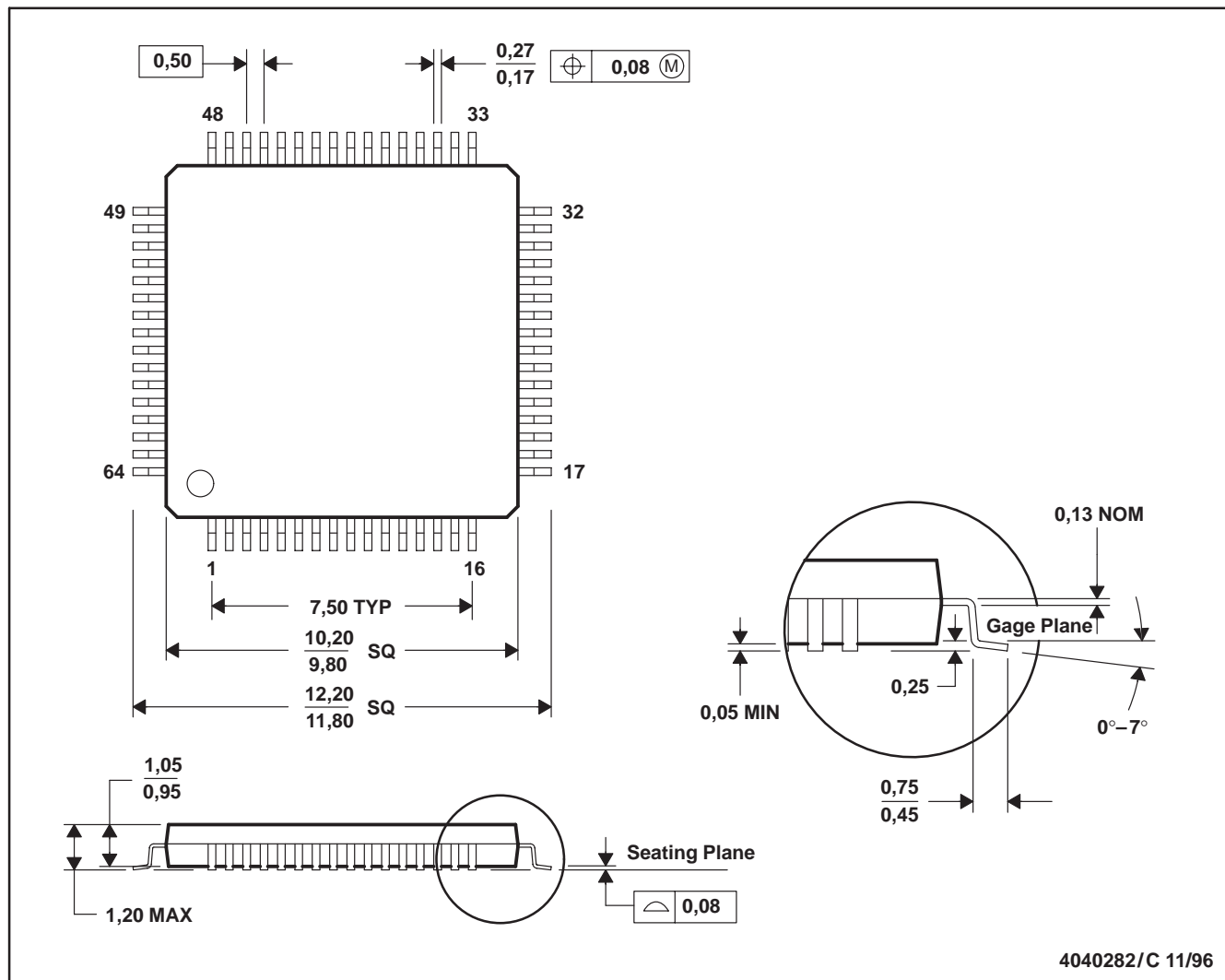


NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. This package is lead-free.

PAG (S-PQFP-G64)

PLASTIC QUAD FLATPACK



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