

# Reconfigurable Intelligent Surfaces for Wideband Communications

*Challenges and Possible Solutions*

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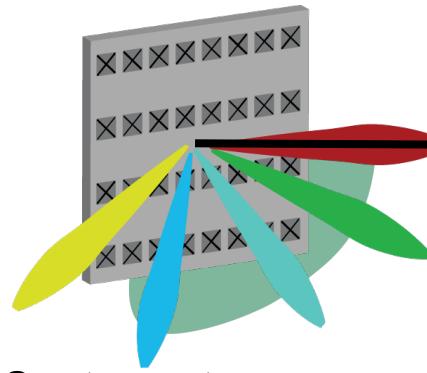
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# INTRODUCTION

# Evolution of Wireless Infrastructure

5G: Adaptive multi-user beamforming

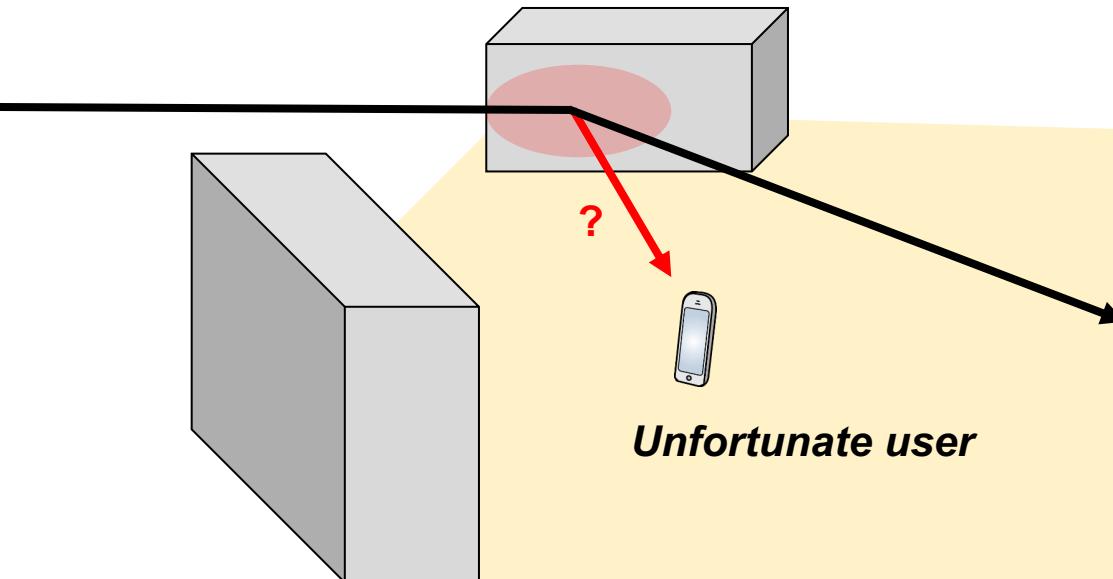


**Sector antenna**

*Fixed beam*

**Fortunate user**

6G: Control objects in environment?

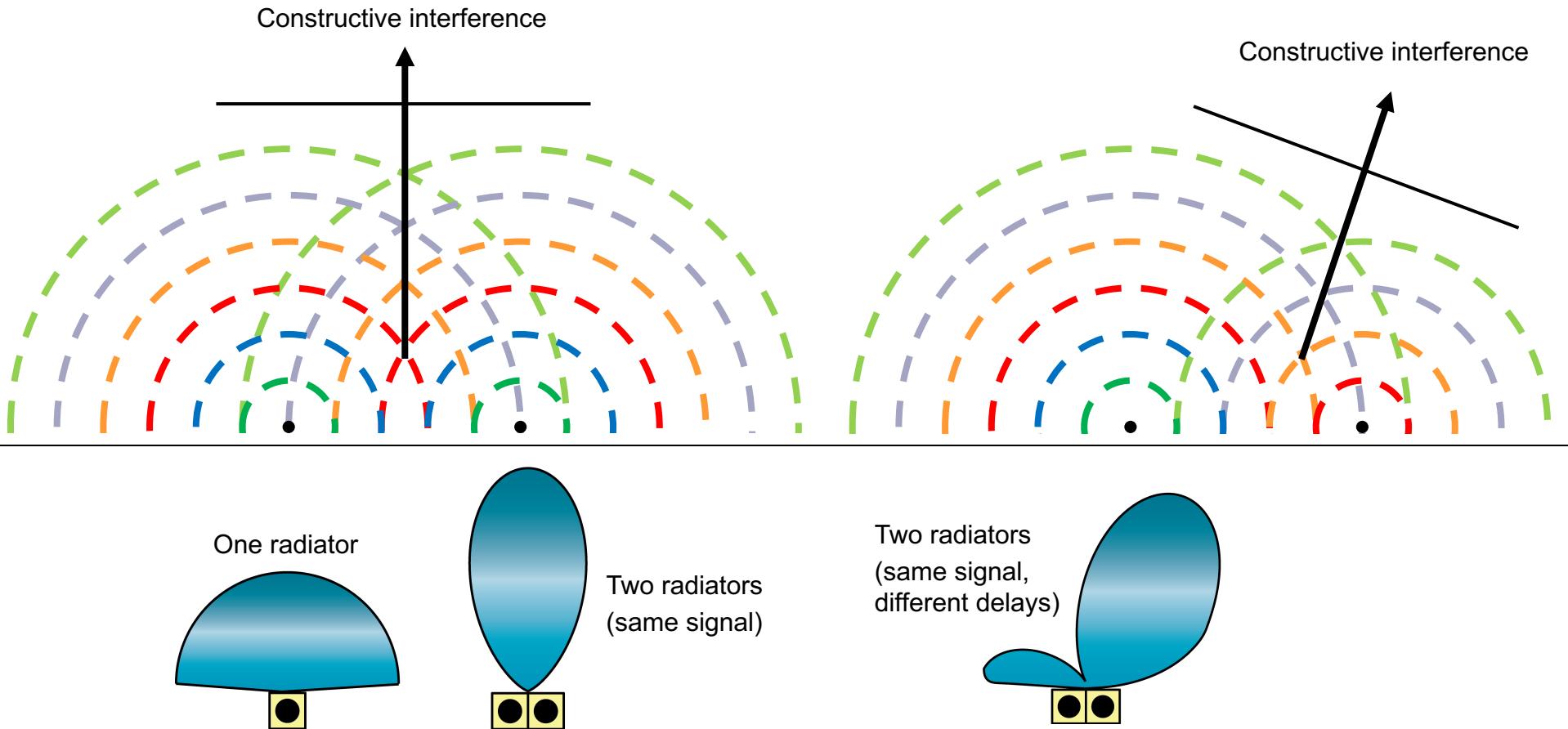


**Reconfigurable metasurfaces**

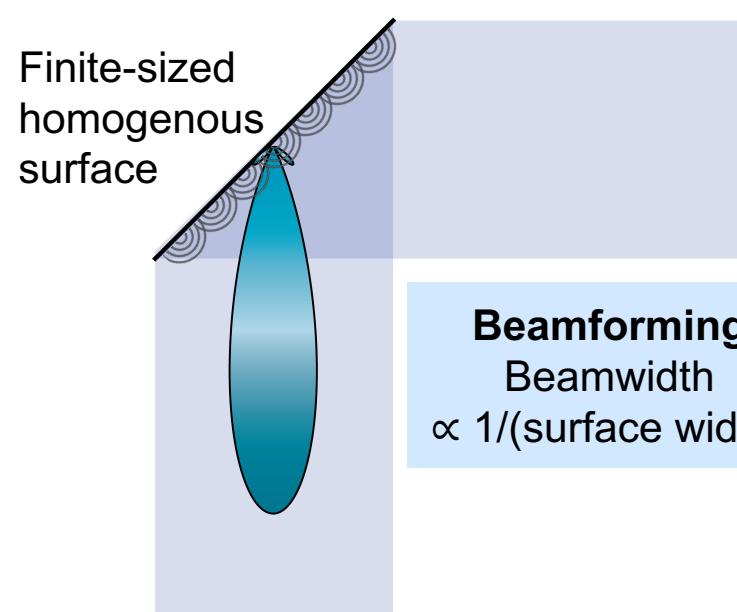
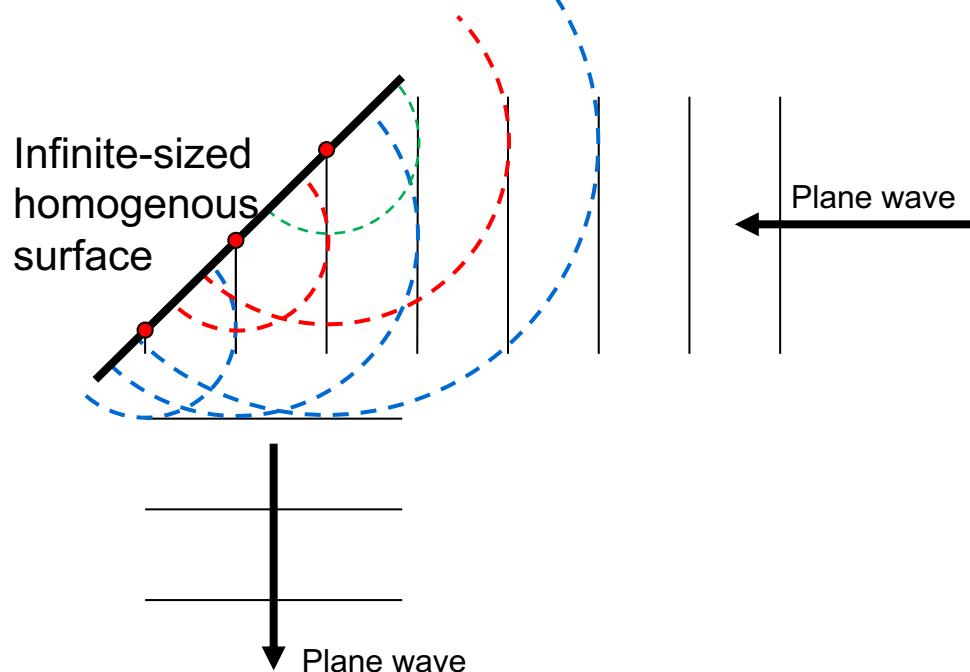
Reconfigurable intelligent surface (RIS)

Intelligent reflecting surface (IRS)

# Beamforming: Directivity by Constructive Interference



# Interpreting Reflection via the Huygens-Fresnel Principle



**Beamforming**  
Beamwidth  
 $\propto 1/(\text{surface width})$

Every point "scatters" a spherical wave:  
Constructive interference determines direction

# Beamforming With RIS

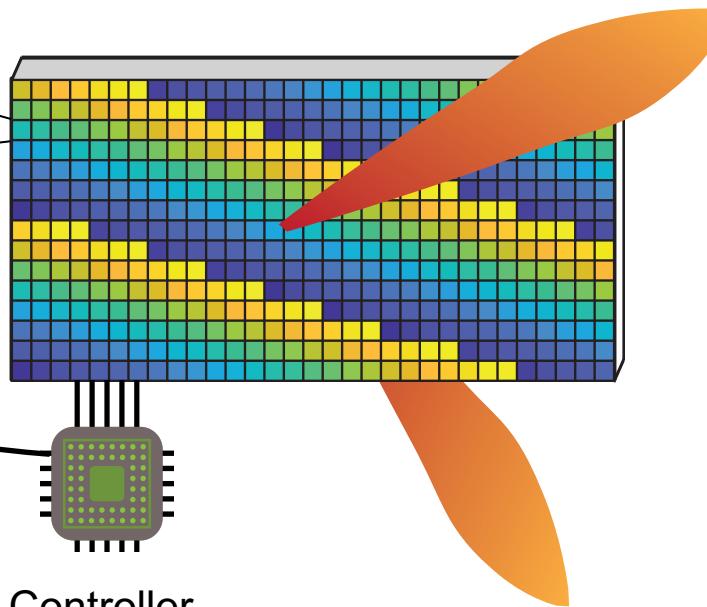


User 2

One element

Passive patch

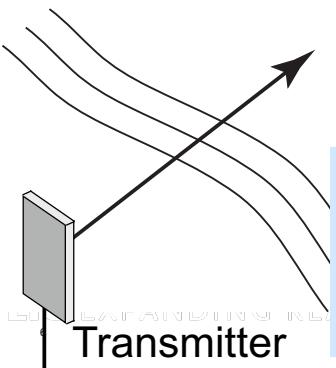
Switch  
(e.g., diod)



Controller

Phase-shifting pattern

Sub- $\lambda$ -sized elements, varying impedance  
Scattering with varying amplitude and phase

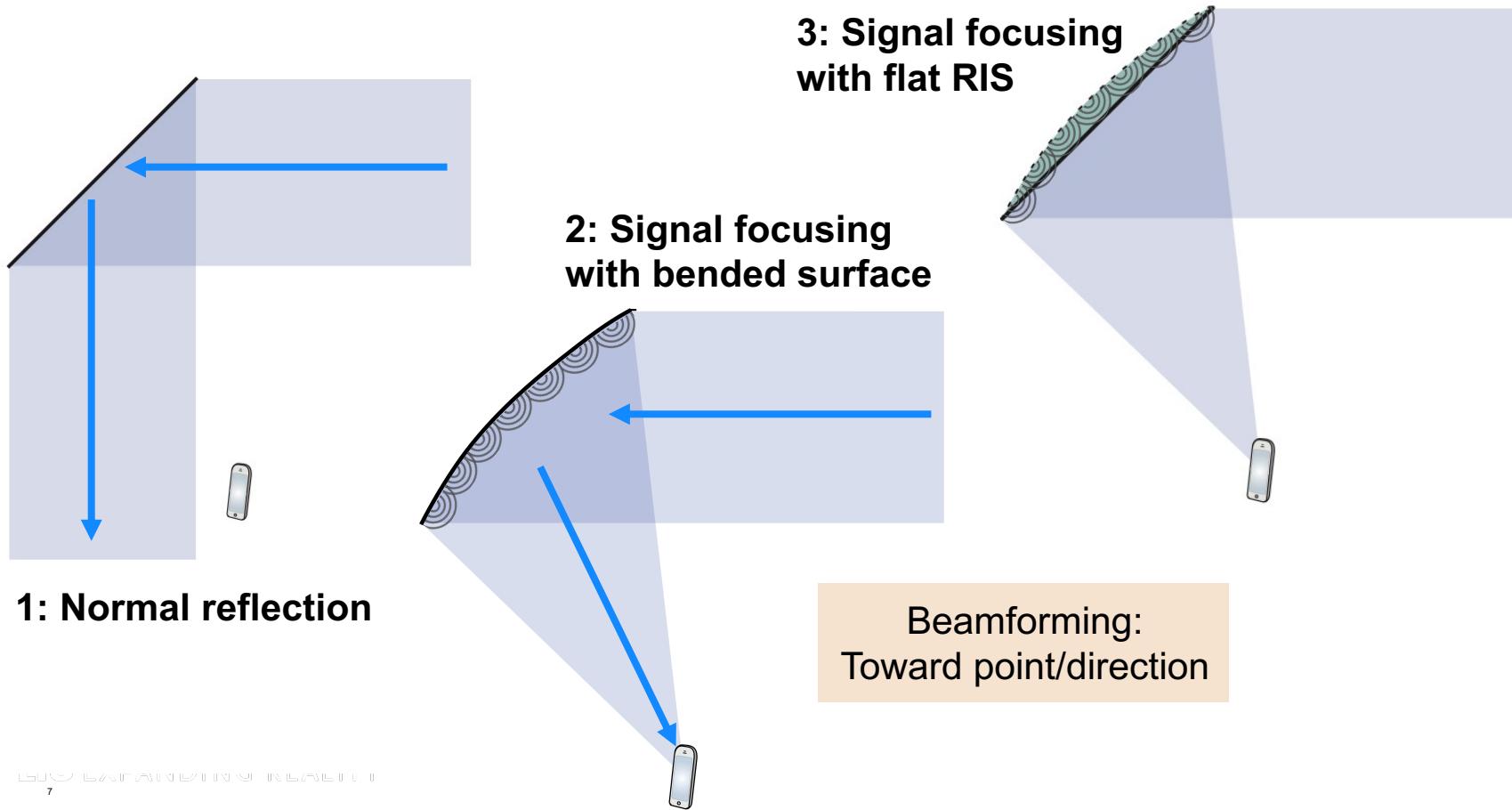


Transmitter



User 1

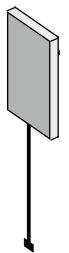
# Geometrical Interpretation at the Global Level



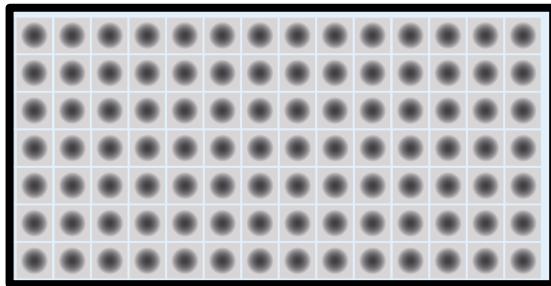
# **SYSTEM MODELING**

# Narrowband System Modelling: $N$ RIS elements

Transmitter



$$\mathbf{h} = \begin{bmatrix} h_1 \\ \vdots \\ h_N \end{bmatrix}$$



Receiver



$$\mathbf{g} = \begin{bmatrix} g_1 \\ \vdots \\ g_N \end{bmatrix}$$

Phase-shifts:  $\boldsymbol{\omega}_\theta = \begin{bmatrix} e^{-j\theta_1} \\ \vdots \\ e^{-j\theta_N} \end{bmatrix}$

**Configure RIS to maximize  $|\rho|^2$ :**

Opposite phases in  $\boldsymbol{\omega}_\theta$  and  $\mathbf{h} \odot \mathbf{g}$ :

$$\theta_n = \arg(h_n g_n)$$

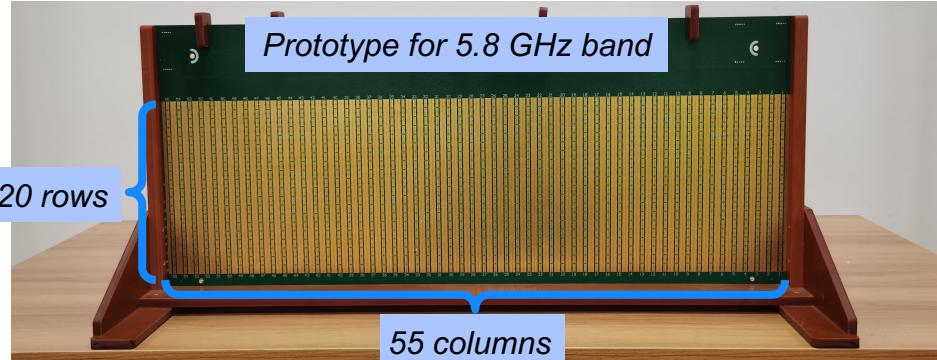
$$|\rho|^2 = \left| \sum_{n=1}^N h_n e^{-j\theta_n} g_n \right|^2 \leq \left| \sum_{n=1}^N |h_n| |g_n| \right|^2 = \|\mathbf{h} \odot \mathbf{g}\|_1^2$$

**End-to-end-channel**

$$\rho = \sum_{n=1}^N h_n e^{-j\theta_n} g_n = (\mathbf{h} \odot \mathbf{g})^T \boldsymbol{\omega}_\theta$$

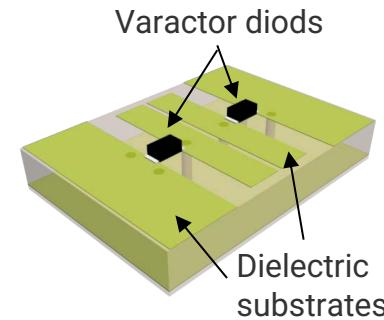
Element-wise product

# How Will an RIS Element Filter the Signal?



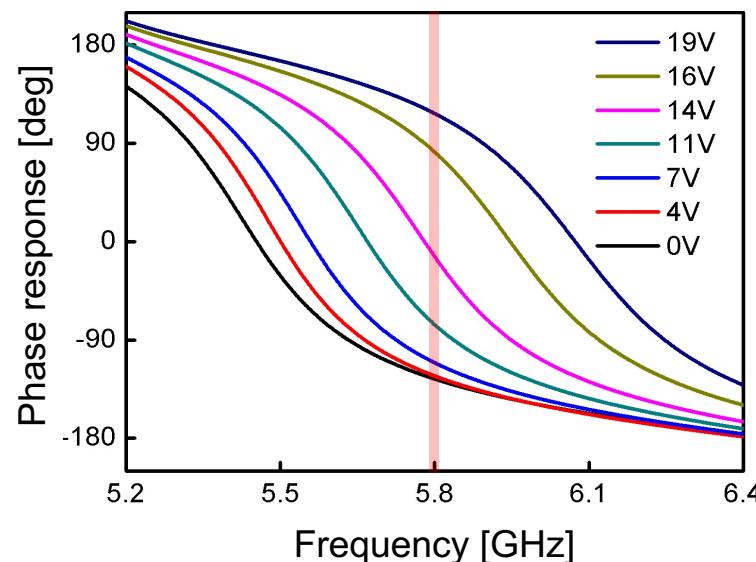
**Example:** Patch with bias voltage  $V$   
Reflection coefficient:

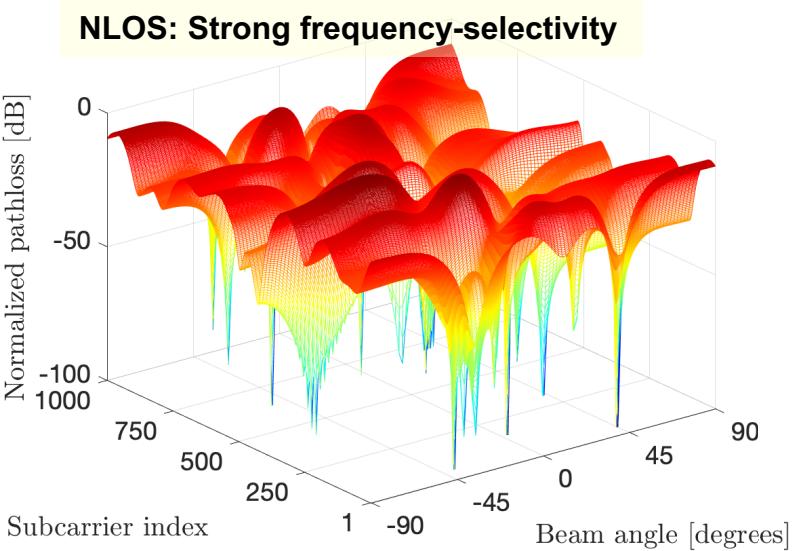
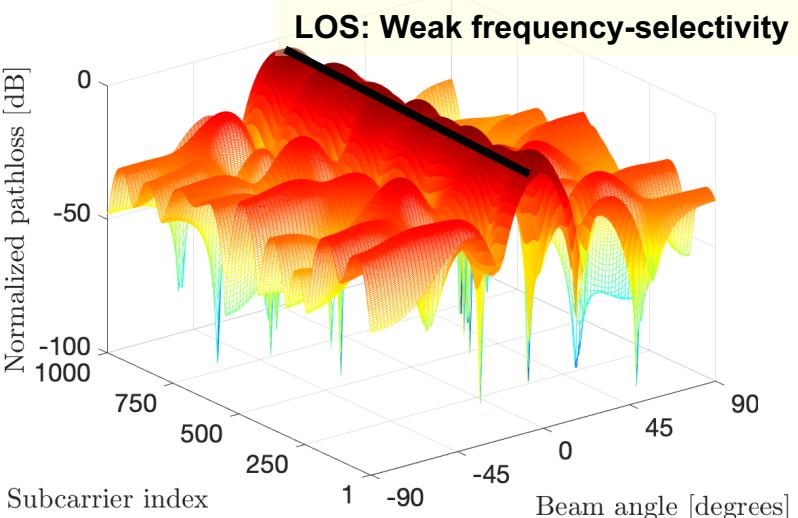
$$\frac{Z_n(V) - Z_0}{Z_n(V) + Z_0}$$



**Reference:** X. Pei, H. Yin, L. Tan, L. Cao, Z. Li, K. Wang, K. Zhang, E. Björnson, "RIS-Aided Wireless Communications: Prototyping, Adaptive Beamforming, and Indoor/Outdoor Field Trials," arXiv:2103.00534

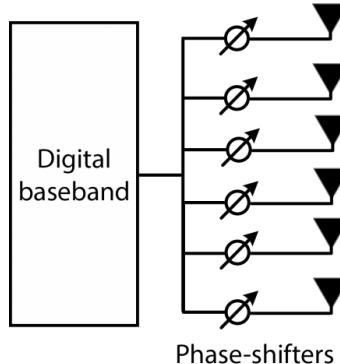
Roughly constant for 20 MHz  
What about the channels?



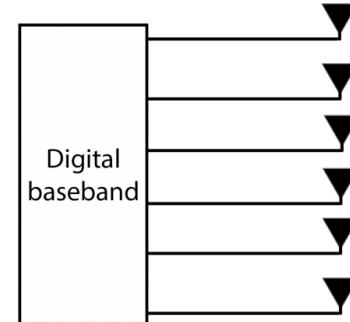


# Frequency selective fading

Analog beamforming



Digital beamforming



**One angular beam**

Same on all subcarriers

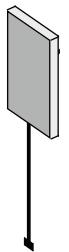
*Same principle as for RIS*

**Sum of many beams**

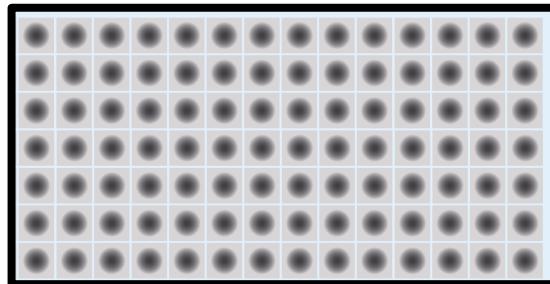
Different over subcarriers

# OFDM System Modeling

Transmitter



Distinct paths:  
 $\mathbf{h}_1, \dots, \mathbf{h}_{L_h}$



Phase-shifts:  $\boldsymbol{\omega}_\theta = \begin{bmatrix} e^{-j\theta_1} \\ \vdots \\ e^{-j\theta_N} \end{bmatrix}$

Receiver



Distinct paths:  
 $\mathbf{g}_1, \dots, \mathbf{g}_{L_g}$

**K subcarrier channels:**

$$\boldsymbol{\rho} = \begin{bmatrix} \rho_1 \\ \vdots \\ \rho_K \end{bmatrix} = \mathbf{F} \mathbf{V}^T \boldsymbol{\omega}_\theta$$

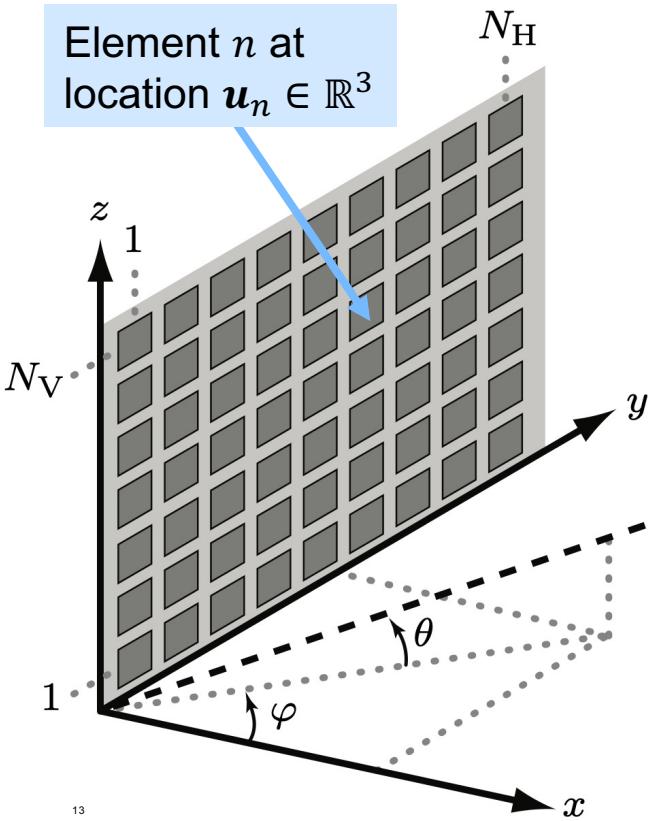
DFT matrix ( $K \times M$ )

Time-domain end-to-end channel matrix ( $M$  taps)

$$\mathbf{V} = \sum_{l=1}^{L_h} \sum_{i=1}^{L_g} (\mathbf{h}_l \odot \mathbf{g}_i) \left[ \delta[\tau_{h,l} + \tau_{g,i}] \dots \delta[\tau_{h,l} + \tau_{g,i} - (M-1)] \right]$$

↑      ↑  
Integer delays

# Channel Modeling Using Array Response Vector



**Far-field:** Incoming/outgoing plane wave determined by

- Azimuth angle  $\varphi \in [-\pi, \pi]$
- Elevation angle  $\theta \in \left[-\frac{\pi}{2}, \frac{\pi}{2}\right]$

**Channels  $h_l$  and  $g_i$  for cluster in direction  $(\varphi, \theta)$**   
constant  $\cdot \mathbf{a}(\varphi, \theta)$

Array response vector:

$$\mathbf{a}(\varphi, \theta) = \begin{bmatrix} e^{jk(\varphi, \theta)^T \mathbf{u}_1} \\ \vdots \\ e^{jk(\varphi, \theta)^T \mathbf{u}_N} \end{bmatrix}$$

$$\mathbf{k}(\varphi, \theta) = \frac{2\pi}{\lambda} \begin{bmatrix} \cos(\theta) \cos(\varphi) \\ \cos(\theta) \sin(\varphi) \\ \sin(\theta) \end{bmatrix}$$

# RIS Optimization for OFDM system

Sum rate expression:

$$R = \frac{B}{K + M - 1} \sum_{k=1}^K \log_2 \left( 1 + \frac{P}{\sigma^2} |\rho_k|^2 \right) = \frac{B}{K + M - 1} \sum_{k=1}^K \log_2 \left( 1 + \frac{P}{\sigma^2} |\mathbf{f}_k^H \mathbf{V}^T \boldsymbol{\omega}_\theta|^2 \right)$$

↑  
Row  $k$  of  $\mathbf{F}$

**Upper bound:** Optimize  $\boldsymbol{\omega}_\theta$  for each subcarrier:

$$|\mathbf{f}_k^H \mathbf{V}^T \boldsymbol{\omega}_\theta|^2 \leq \|\mathbf{f}_k^H \mathbf{V}^T\|_1^2$$

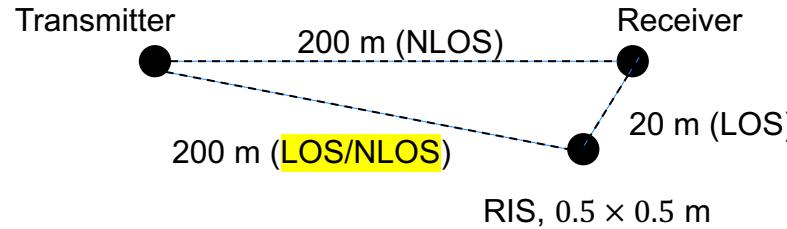
**Heuristic solution:** Strongest-path maximization (STM),

$$m = \arg \max_i \|\mathbf{v}_i\|_1^2 \text{ where } \mathbf{V} = [\mathbf{v}_1 \dots \mathbf{v}_M]$$

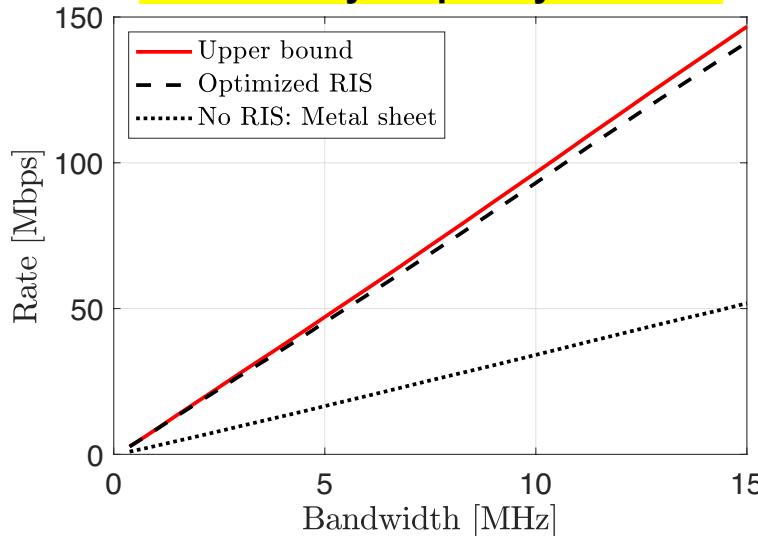
$$\boldsymbol{\omega}_\theta = \arg \max_{\boldsymbol{\omega}} |\mathbf{v}_m^T \boldsymbol{\omega}|^2$$

# RIS in Frequency Selective Channels

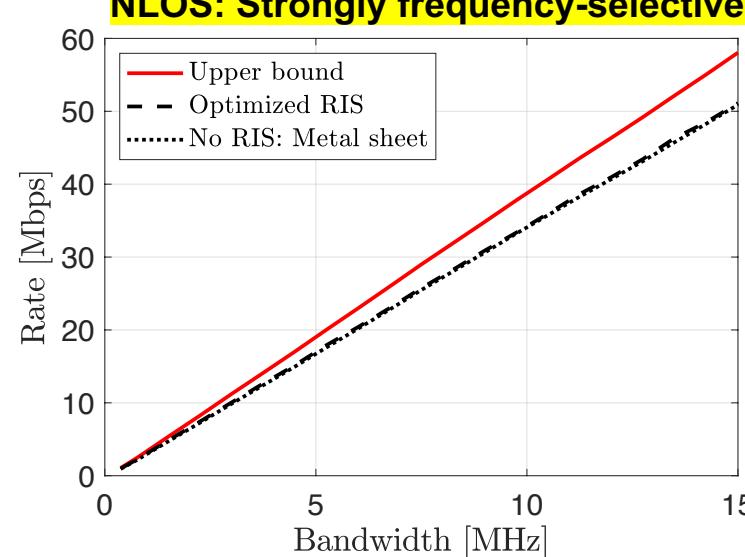
RIS only effective with LOS  
to transmitter and receiver!



**LOS: Weakly frequency-selective**

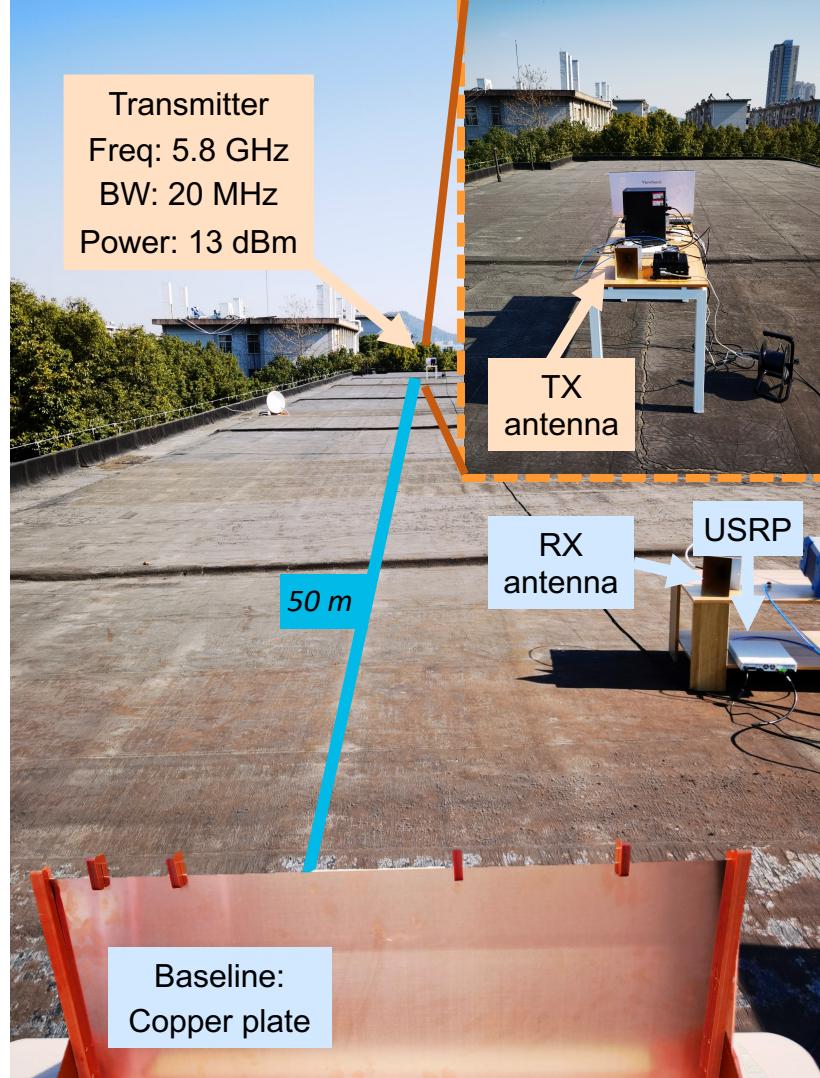
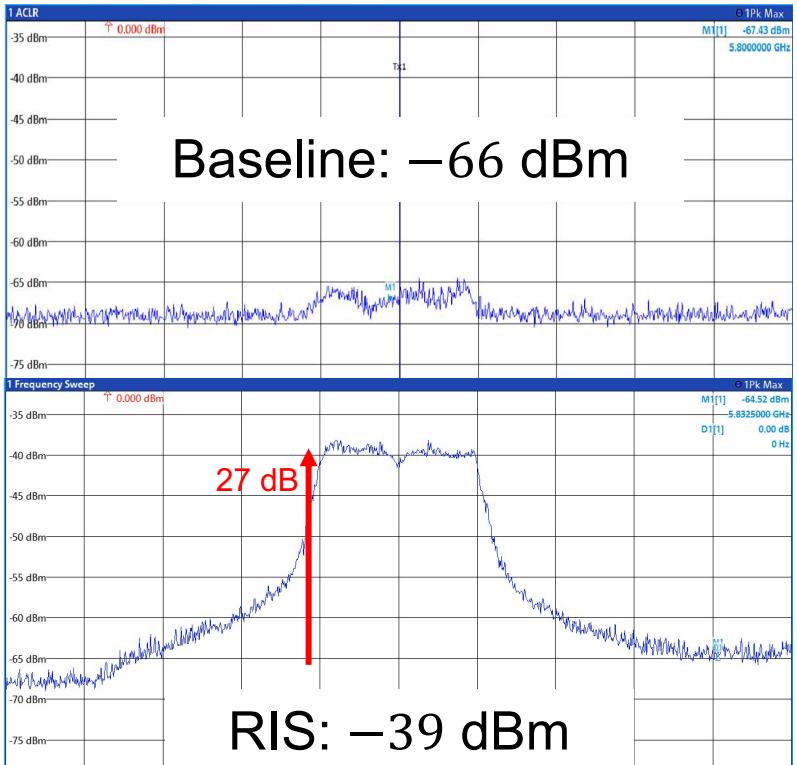


**NLOS: Strongly frequency-selective**



**Reference:** E. Björnson, et al., “Reconfigurable Intelligent Surfaces: A Signal Processing Perspective With Wireless Applications”, Available on arXiv:2102.00742.

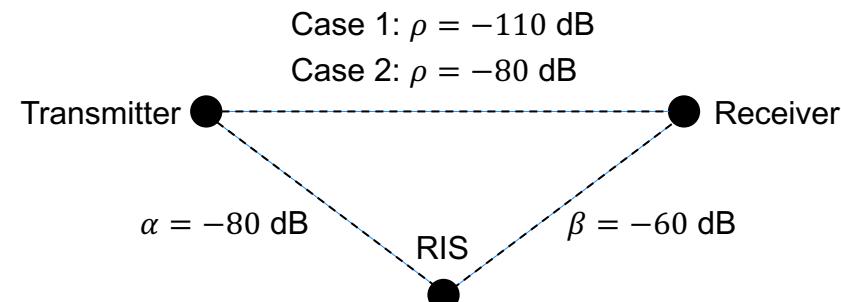
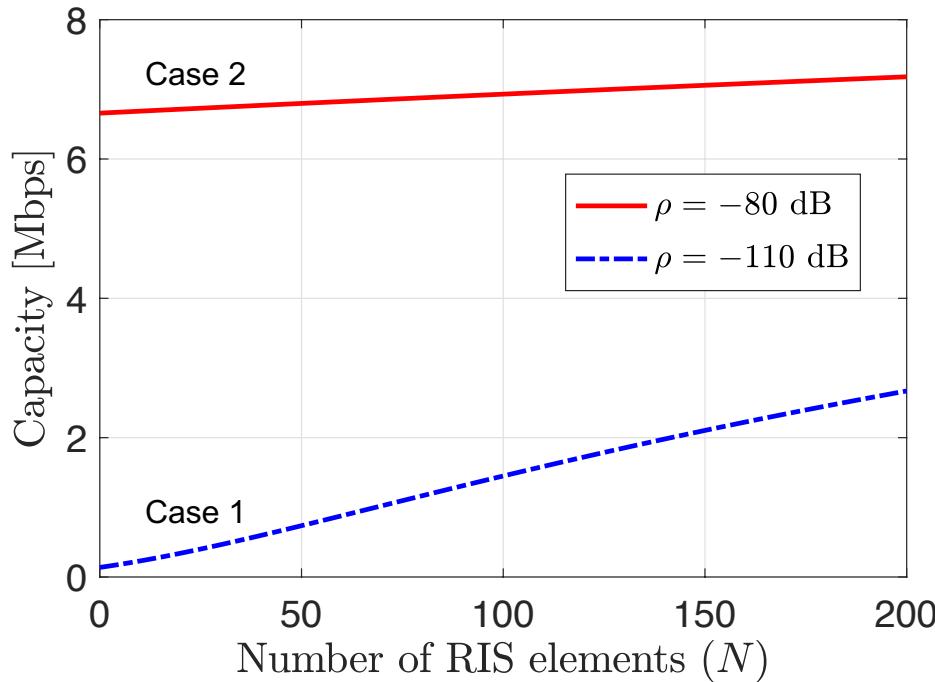
# Experimental Validation



**Reference:** X. Pei, H. Yin, L. Tan, L. Cao, Z. Li, K. Wang, K. Zhang, E. Björnson, "RIS-Aided Wireless Communications: Prototyping, Adaptive Beamforming, and Indoor/Outdoor Field Trials," arXiv:2103.00534

# IDENTIFYING THE BEST USE CASE

# Importance of the Direct Path



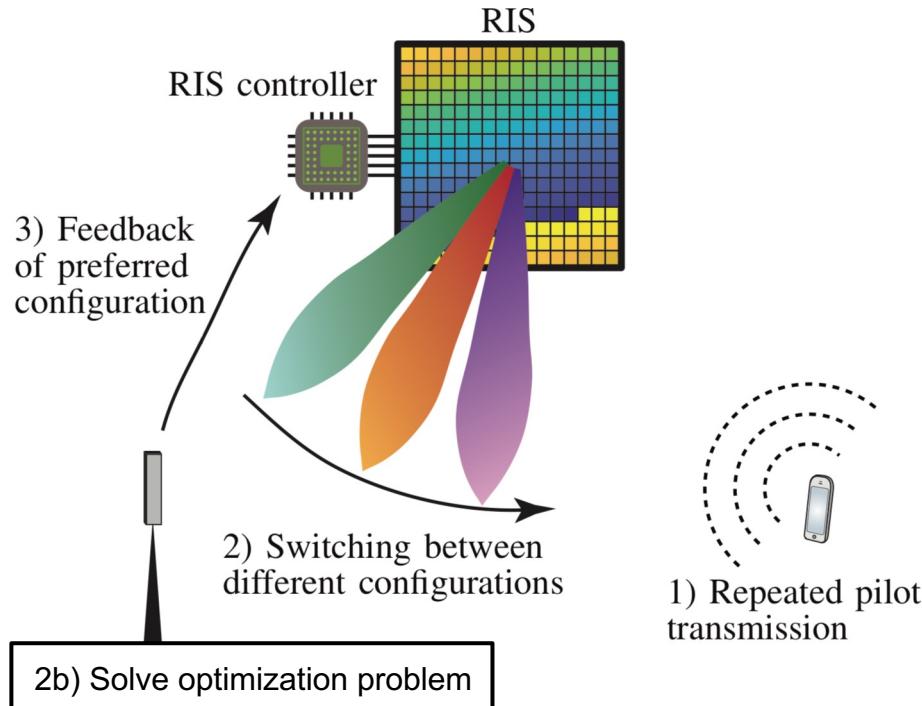
**Bandwidth:**  $B = 1 \text{ MHz}$

**Transmit SNR:** 100 dB

**RIS only effective if direct path is weak**

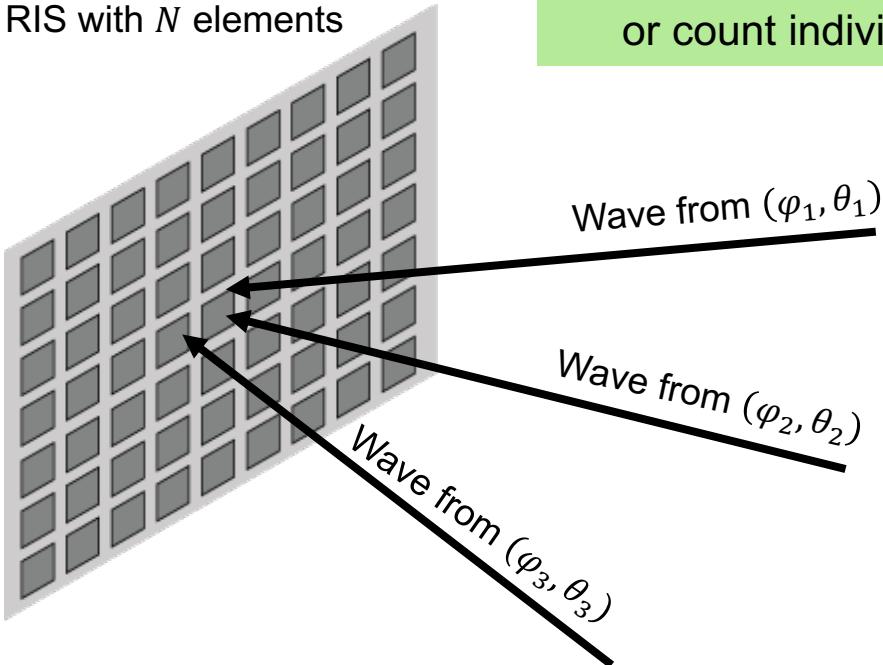
# How Difficult is Channel Estimation?

The RIS is blind!



# How Many Parameters to Estimate?

RIS with  $N$  elements



$L_h$  channel vectors

$$\mathbf{h}_l = c_l \mathbf{a}(\varphi_l, \theta_l)$$

$NM$  complex parameters in  $\mathcal{V}$

or count individual parameters

Parameters to estimate:

Pure line-of-sight

4

Rich scattering

$NM$

# Summary

	NLOS to/from RIS	LOS to/from RIS
Strong direct path	RIS not effective in OFDM Difficult to estimate	RIS adds little to direct path
Weak direct path		Large gains in OFDM Few parameters to estimate

# OPEN PROBLEMS

# Much Deeper Research is Needed!

- Efficient RIS configuration for OFDM

With amplitude variations and limited resolution

*“Optimizing a Binary Intelligent Reflecting Surface for OFDM Communications under Mutual Coupling” arXiv:2106.04280*

- Effective estimation algorithm

Utilize spatial correlation

*“Rayleigh Fading Modeling and Channel Hardening for Reconfigurable Intelligent Surfaces” arXiv:2009.04723*

- Further system modeling

- What if we the RIS is also frequency selective?

# Conclusion: OFDM Works in One Particular Use Case

