

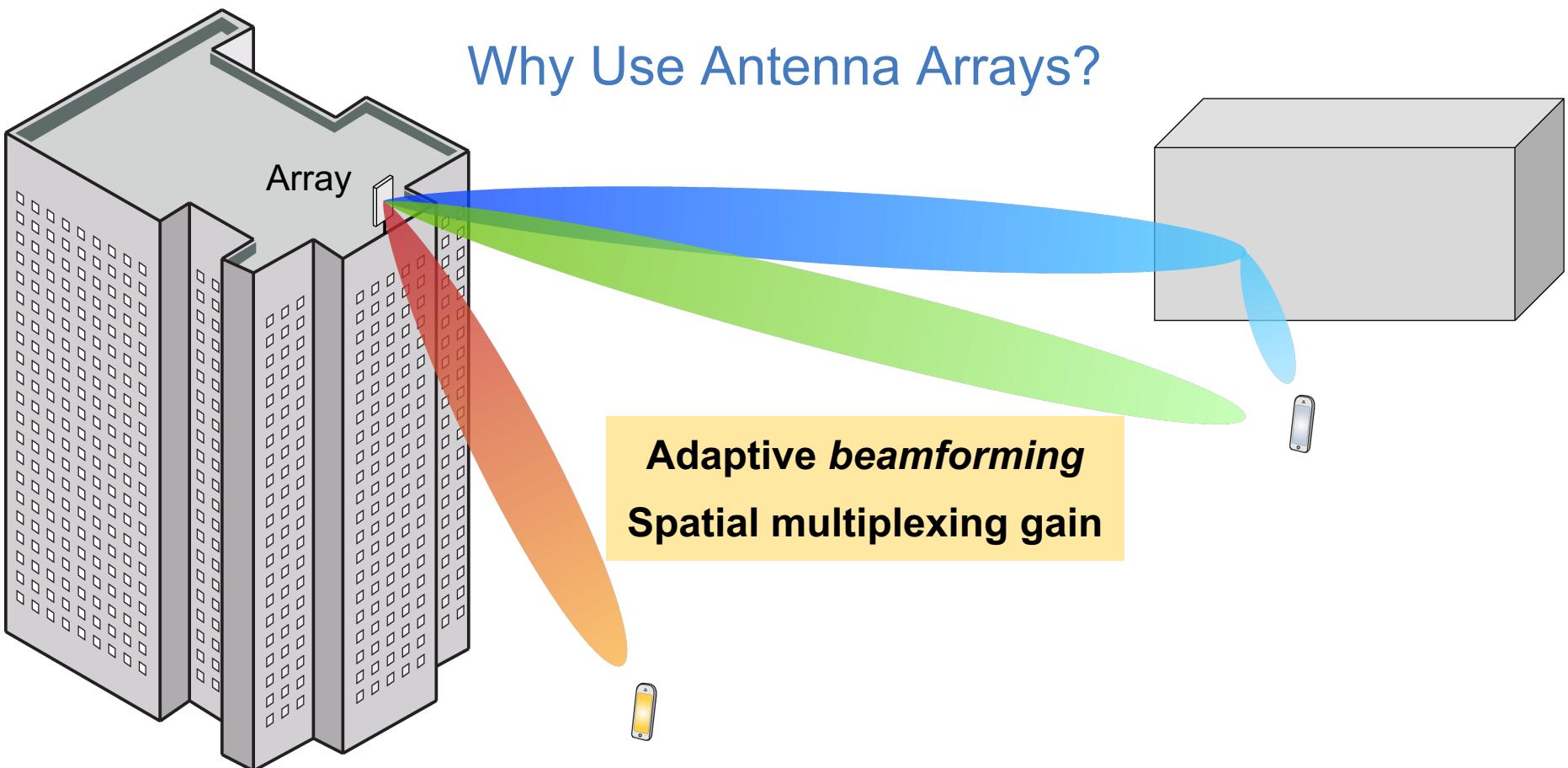
Physically Large Antenna Arrays: When the Near-Field Becomes Far-Reaching

Emil Björnson

Professor, KTH Royal Institute of Technology, Sweden

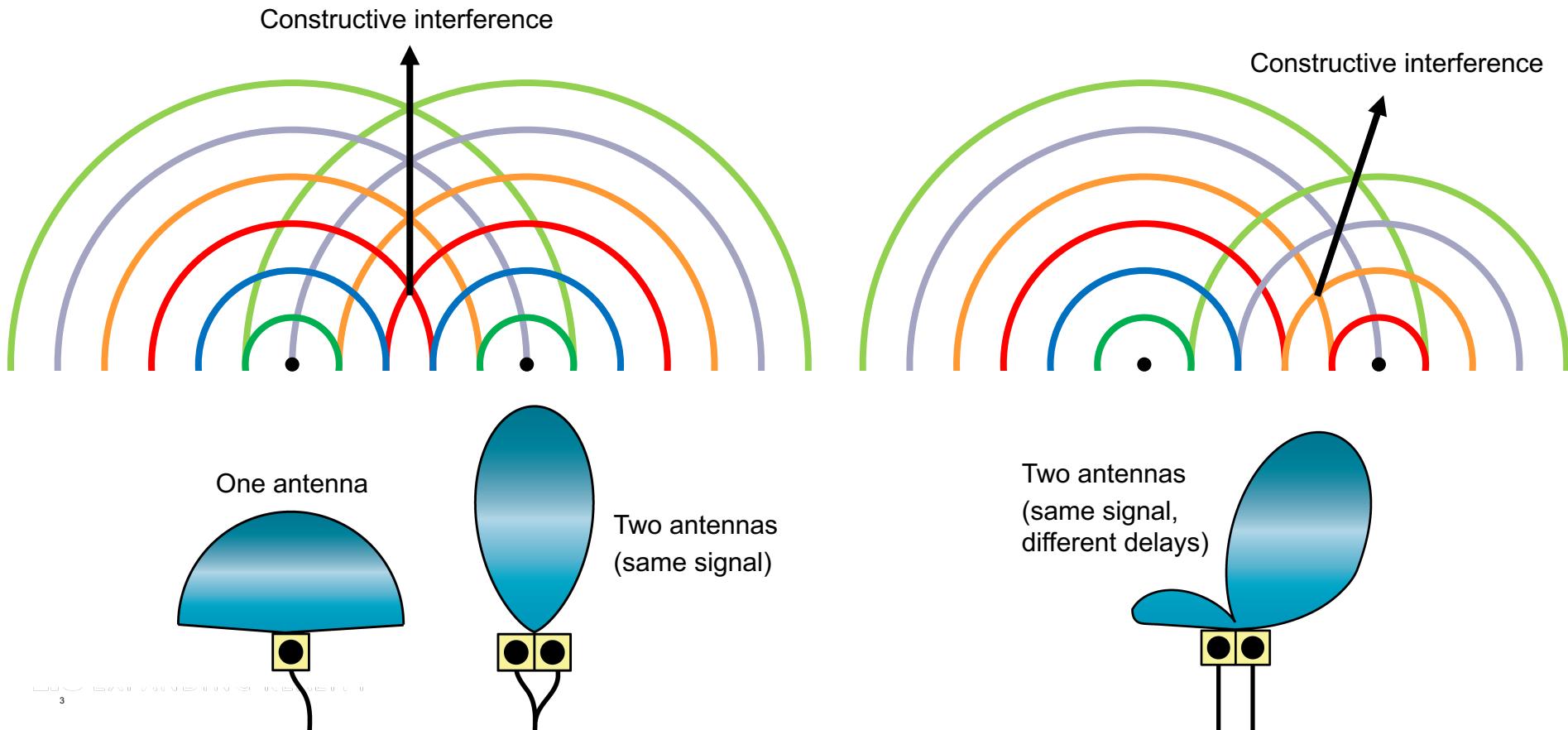
Associate professor, Linköping University, Sweden

Why Use Antenna Arrays?



MIMO = Multiple input multiple output

Adaptive Beamforming in a Nutshell



What is Massive MIMO?

Definition of *massive*

1 : forming or consisting of a large mass:

a : BULKY

// *massive* furniture

b : WEIGHTY, HEAVY

// *massive* walls

// *massive* volume

2 a : large, solid, or heavy in structure

// *massive* jaw

b : large in scope or degree

// *the feeling of frustration, of being ineffectual, is massive*

— David Halberstam



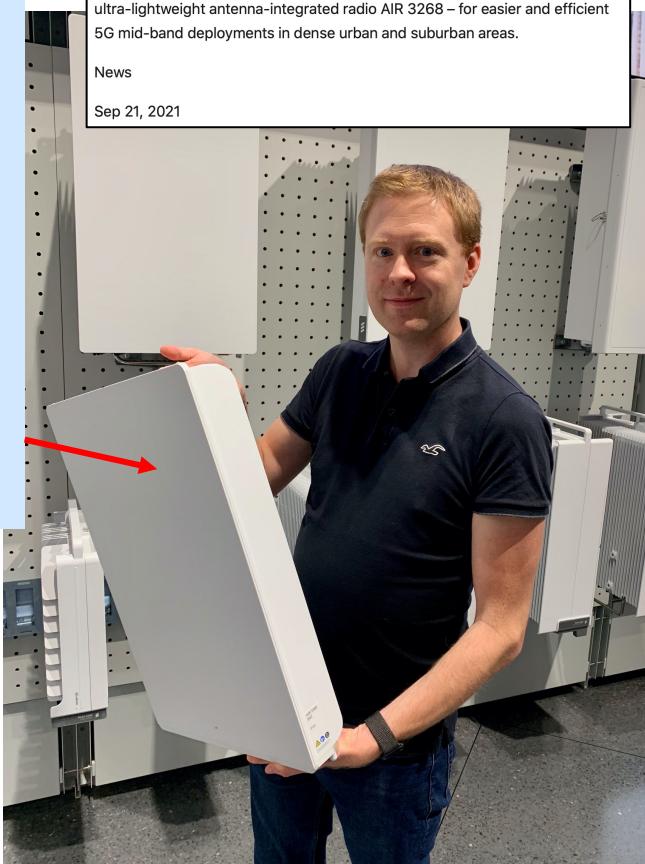
AIR 3268

32 antennas
(128 elements)

12 kg

200 W over
200 MHz in 3.5
GHz band

Integrated
circuitry



Ericsson adds a 12 kg radio to Massive MIMO portfolio for easier 5G mid-band deployment

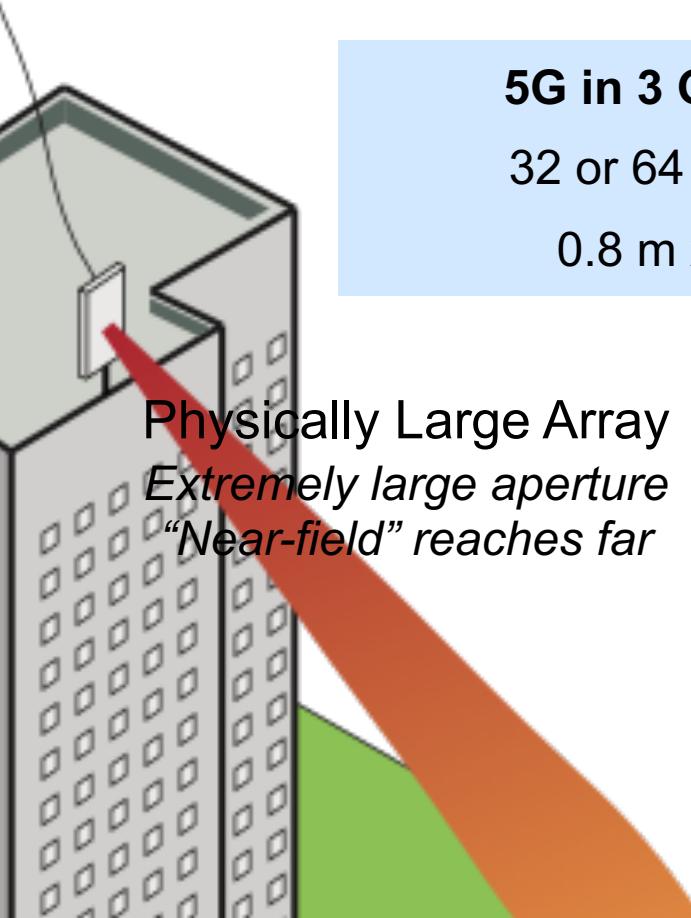
Available in English [Français](#) [Español](#) [日本語](#) [繁體中文](#) [Indonesia](#)

Ericsson today unveils the latest addition to its Massive MIMO portfolio – the ultra-lightweight antenna-integrated radio AIR 3268 – for easier and efficient 5G mid-band deployments in dense urban and suburban areas.

News

Sep 21, 2021

Massive MIMO versus Physically Large Arrays



5G in 3 GHz band

32 or 64 antennas

0.8 m x 0.5 m

Massive MIMO array

Many antennas

Looks small

Physically Large Array

Extremely large aperture

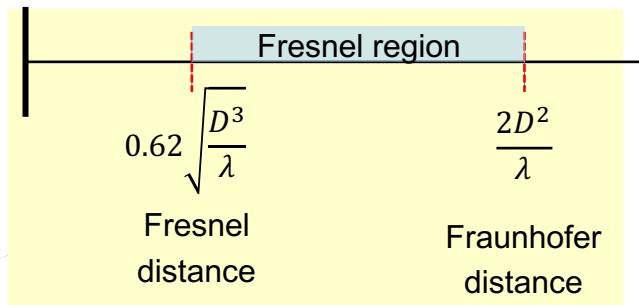
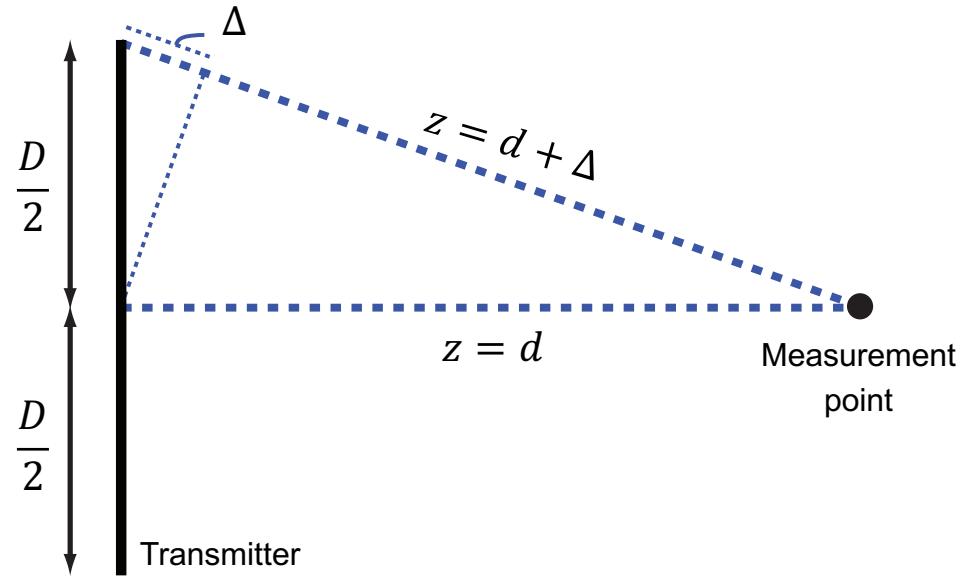
“Near-field” reaches far

**What are the benefits of
this evolution?**

PRELIMINARIES:

NEAR-FIELD OF A *PASSIVE* ANTENNA

Near-Field and Far-Field Regions (Electromagnetic Definition)



Electric field intensity proportional to

$$\int_{\text{Transmitter}} \frac{e^{-j\frac{2\pi}{\lambda}z}}{z} \left(1 + \underbrace{\frac{j}{2\pi z/\lambda} - \frac{1}{(2\pi z/\lambda)^2}}_{\text{Reactive near-field effect}} \right) dz$$

(Negligible for $d > 0.62\sqrt{D^3/\lambda}$)

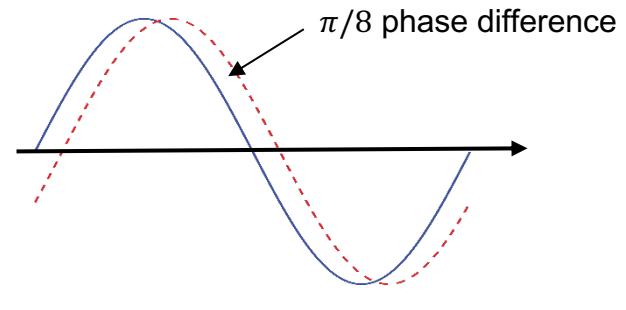
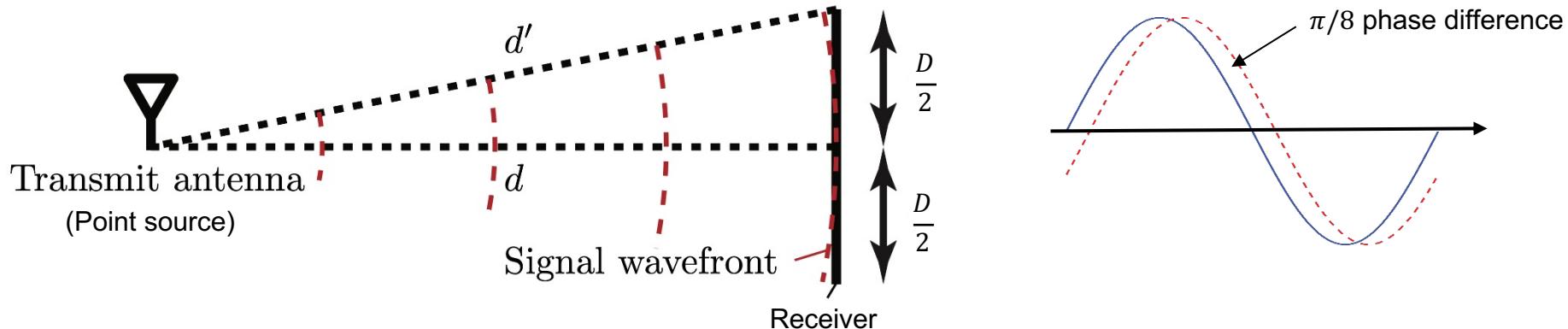
Radiative near-field

Path difference: $\Delta = \sqrt{d^2 + \left(\frac{D}{2}\right)^2} - d \approx \frac{D^2}{8d}$

A graph showing the electric field intensity in the radiative near-field. A solid blue curve represents the field amplitude, which starts at zero at the transmitter, reaches a peak, and then decays. A dashed red curve represents the phase, which starts at zero at the transmitter and increases linearly. An arrow points to the phase curve with the label " $\pi/8$ phase difference".

$$\frac{\pi}{8} \geq \frac{2\pi}{\lambda} \frac{D^2}{8d} \rightarrow d \geq \frac{2D^2}{\lambda}$$

Near-Field and Far-Field Regions (Communication Perspective)



Amplitude difference d/d'

Less than $\cos(\pi/8) \approx 0.92$ if $d \geq 1.2D$

Phase-shift difference less than $\pi/8$ if $d \geq d_F = \frac{2D^2}{\lambda}$

Example: 3 GHz, $D = \lambda = 0.1$ m

$$1.2D = 0.12 \text{ m}$$

$$d_F = \frac{2D^2}{\lambda} = 2\lambda = 0.2 \text{ m}$$

Fraunhofer distance

Locally plane wave
over receiver aperture

NEAR-FIELD OF PHYSICALLY LARGE ARRAYS

Fraunhofer's Array Distance: d_{FA}

The distance after which the angular field distribution is essentially independent of the distance

Transmit antenna



Practical operating regime

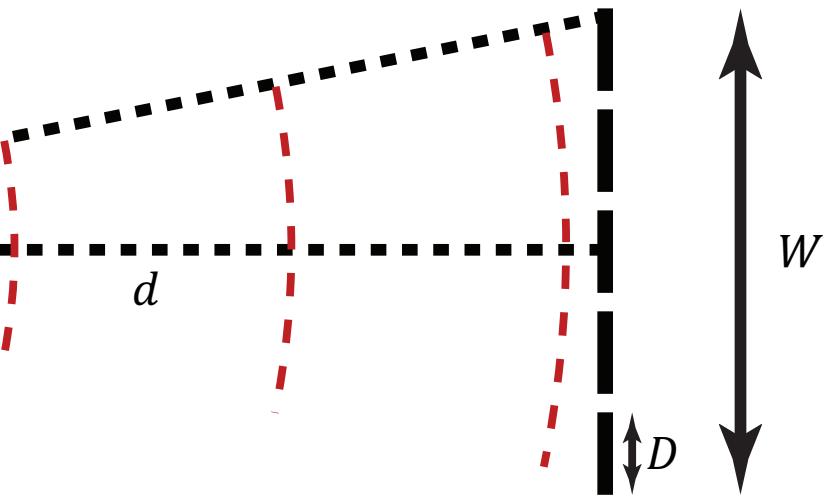
Far-field of antenna: $d \gg 2D^2/\lambda$

Near-field of array if $d \leq 2W^2/\lambda$

Example: 3 GHz, $\lambda = 0.1$ m

$$W = 10\lambda = 1 \text{ m}: d_{FA} = \frac{2W^2}{\lambda} = 20 \text{ m}$$

$$W = 100\lambda = 10 \text{ m}: d_{FA} = \frac{2W^2}{\lambda} = 2 \text{ km}$$



Channel Modeling for Radiative Near-Field Phenomena

Near-field phenomena

1. Different distances to antennas
2. Different effective areas
3. Different polarization losses

Far-field channel gain

$$\beta_z = \frac{L^2}{4\pi z^2}$$

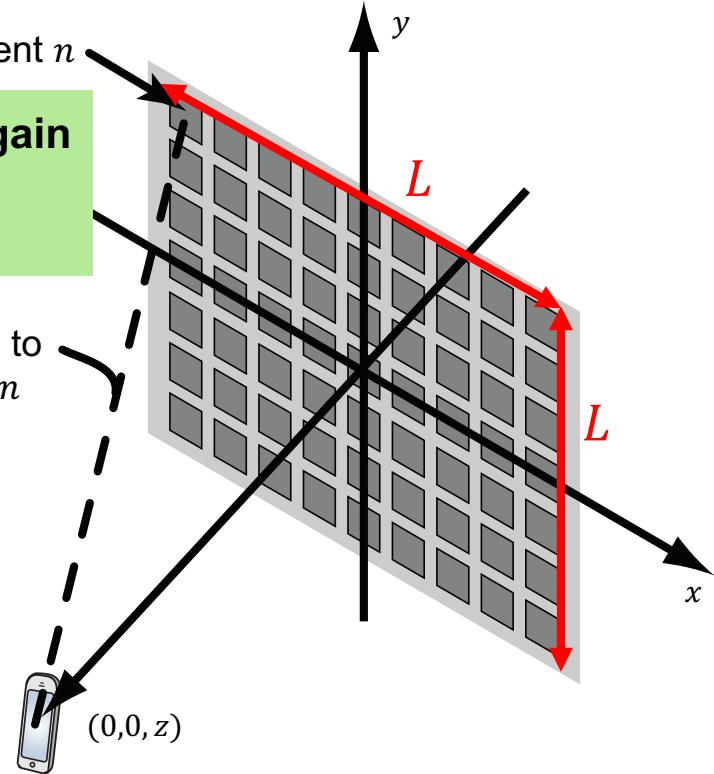
Free-space channel gain (distance z , width/length L):

$$\gamma_z = \frac{\beta_z}{3(\beta_z\pi + 1)\sqrt{2\beta_z\pi + 1}} + \frac{2}{3\pi} \tan^{-1}\left(\frac{\beta_z\pi}{\sqrt{2\beta_z\pi + 1}}\right)$$

Limits: $\gamma_z \approx \beta_z$ for $z \gg L$

$$\gamma_z \rightarrow \frac{1}{3} \text{ as } L \rightarrow \infty$$

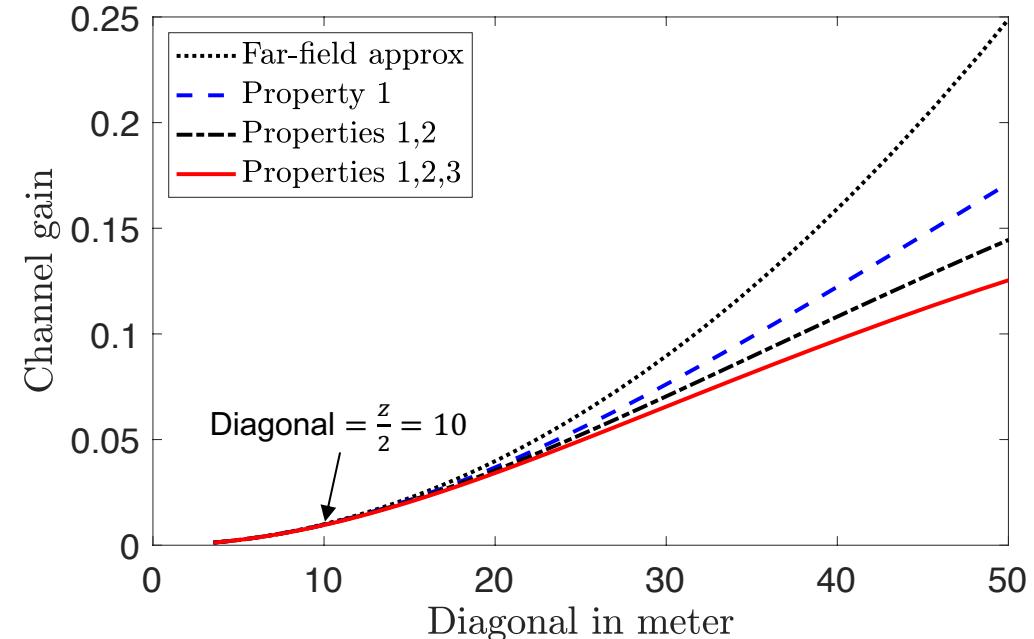
Element n



Reference: E. Björnson, L. Sanguinetti, "Power Scaling Laws and Near-Field Behaviors of Massive MIMO and Intelligent Reflecting Surfaces," 2020

When Are These Phenomena Appearing?

Distance to square array: $z = 20$ m



Near-field phenomena

1. Different distances to antennas
2. Different effective areas
3. Different polarization losses

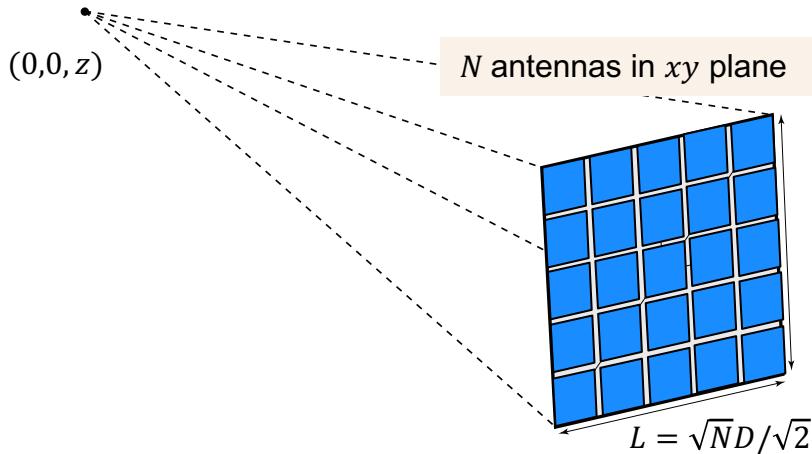
Reference: E. Björnson, L. Sanguinetti, "Power Scaling Laws and Near-Field Behaviors of Massive MIMO and Intelligent Reflecting Surfaces," 2020

Björnson distance d_B

2 × Largest dimension = Distance to array

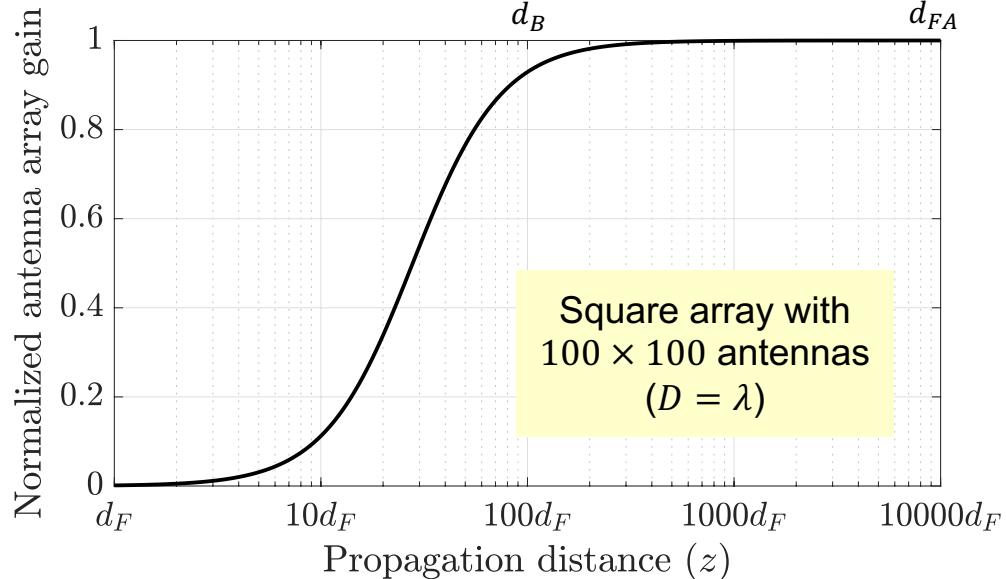
Isotropic
transmitter

Array Gain in Radiative Near-Field



Normalized antenna array gain:

$$G_{\text{array}} = \frac{\text{Actual channel gain}}{\text{Far-field channel gain}}$$



Fraunhofer array
distance:

$$d_{FA} = 2ND^2/\lambda = Nd_F$$

No evident impact
on the gain

Björnson distance

$$d_B = 2D\sqrt{N}$$

Where at least 95%
of gain is achieved

Beamforming in Fresnel Region ($d_B \leq z \leq d_{FA}$)

Fraunhofer array distance is still relevant:

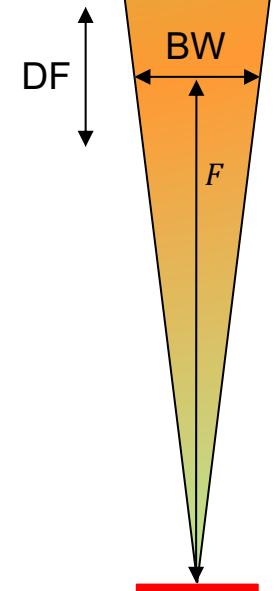
- For $z \geq d_{FA}$: Use array response vector
- For $z \leq d_{FA}$: Use spherical curvature for beamforming

$$\text{Angular 3 dB beam-width (BW) is } \text{BW}_{3\text{dB}}^{\text{ang}} \approx \frac{0.886\lambda}{D\sqrt{N/2}}$$

Theorem: Depth-of-focus (DF), where the array gain is at most 3 dB lower than the maximum, for focusing at $z = F$ is

$$z \in \left[\frac{d_{FA}F}{d_{FA} + 10F}, \frac{d_{FA}F}{d_{FA} - 10F} \right]$$

Finite-depth beamforming closer than $d_{FA}/10$



Reference: E. Björnson, Ö. T. Demir, L. Sanguinetti, "A Primer on Near-Field Beamforming for Arrays and Reconfigurable Intelligent Surfaces," Asilomar 2021, arXiv:2110.06661

Far-field



Far-Field vs. Near-Field Beamforming

$$\frac{d_{FA}}{10}$$

Impact of wavelength

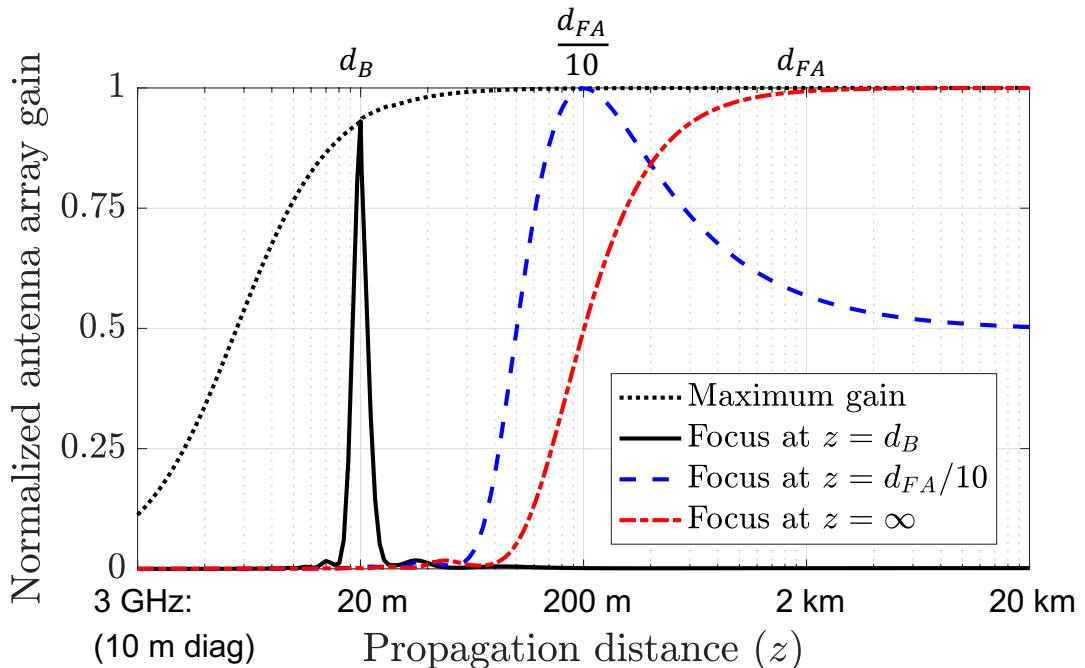
Fixed array size: $d_{FA} \propto \lambda^{-1}$

Fixed number of elements: $d_{FA} \propto \lambda$



Radiative
near-field

Near-Field Finite-Depth Beamforming



Fraunhofer array distance d_{FA}

Nothing special happens here

Finite-depth beamforming

closer than $d_{FA}/10$

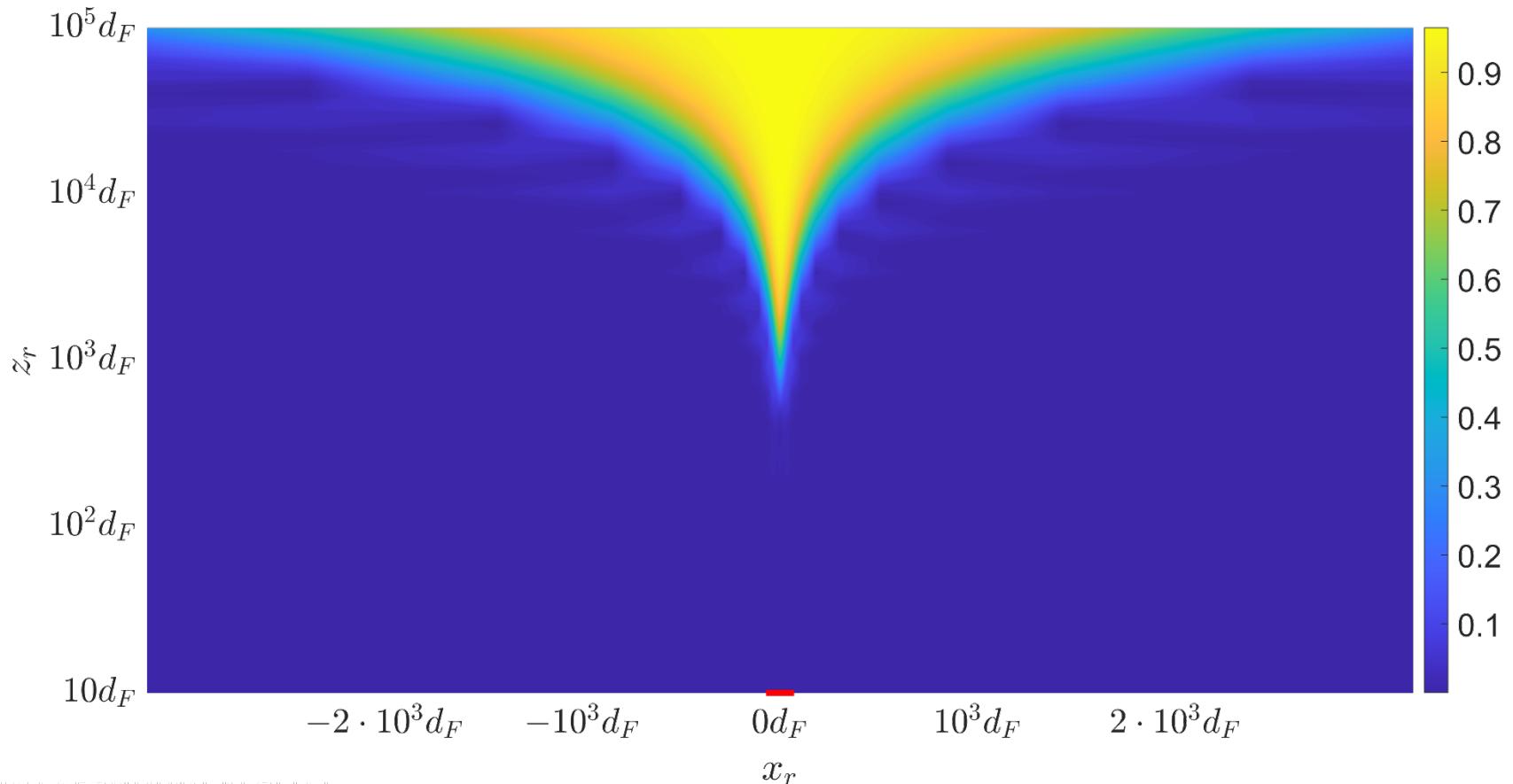
Björnson distance:

Where the array gain tapers off

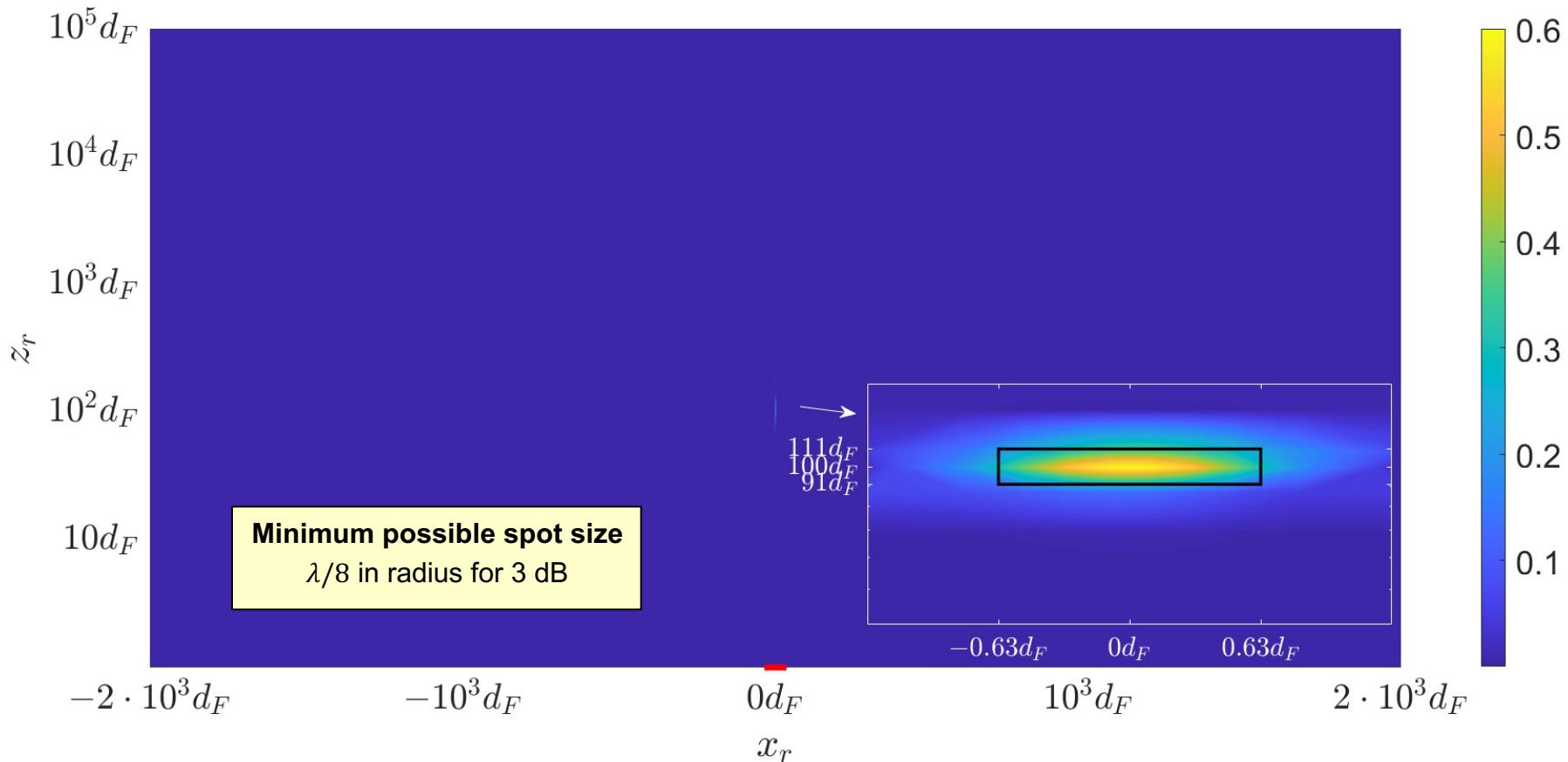
Setup: Square array with 100×100 antennas ($D = \lambda$)

Reference: E. Björnson, Ö. T. Demir, L. Sanguinetti, “A Primer on Near-Field Beamforming for Arrays and Reconfigurable Intelligent Surfaces,” Asilomar 2021, arXiv:2110.06661

Conventional Far-Field Beamforming



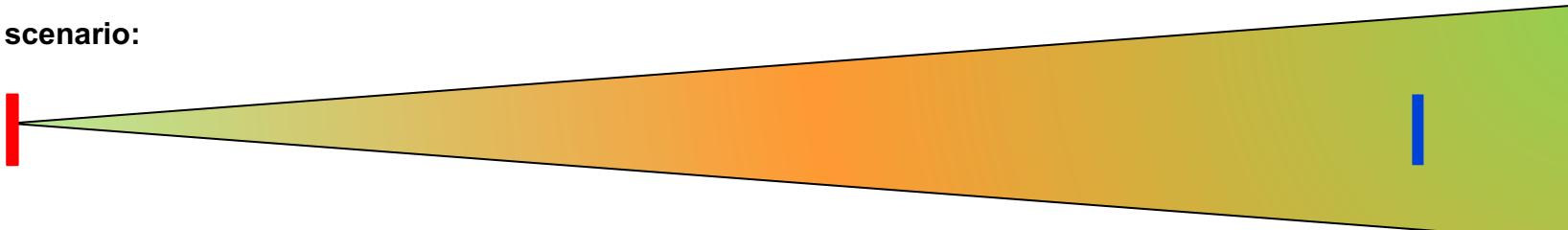
Near-Field Beamforming



NEAR-FIELD MULTIPLEXING

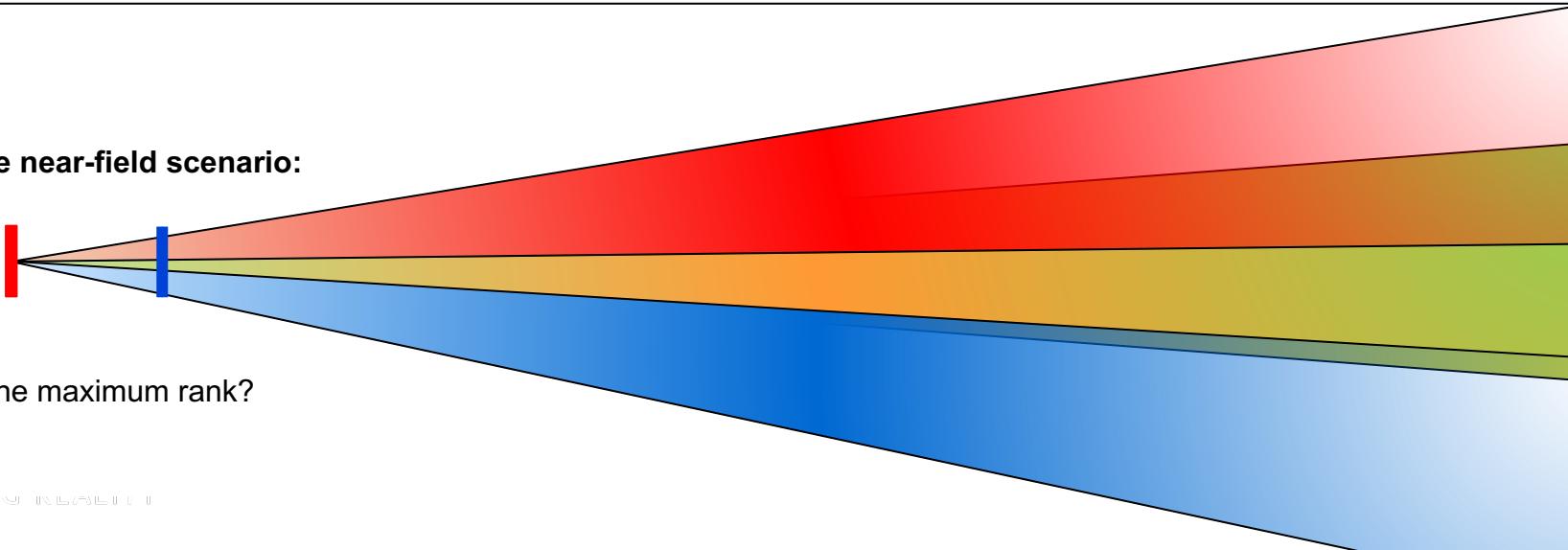
Rank of Channel in Line-of-Sight

Far-field scenario:



Maximum rank: 2 (One per polarization)

Radiative near-field scenario:



What is the maximum rank?

Fundamental Limit: Spatial Degrees-of-Freedom

Theorem

A large antenna array can resolve π signals for each segment of area λ^2

Planar array (length L , height H):

$$\eta \approx \pi \frac{LH}{\lambda^2}$$

We are very far from the limits!

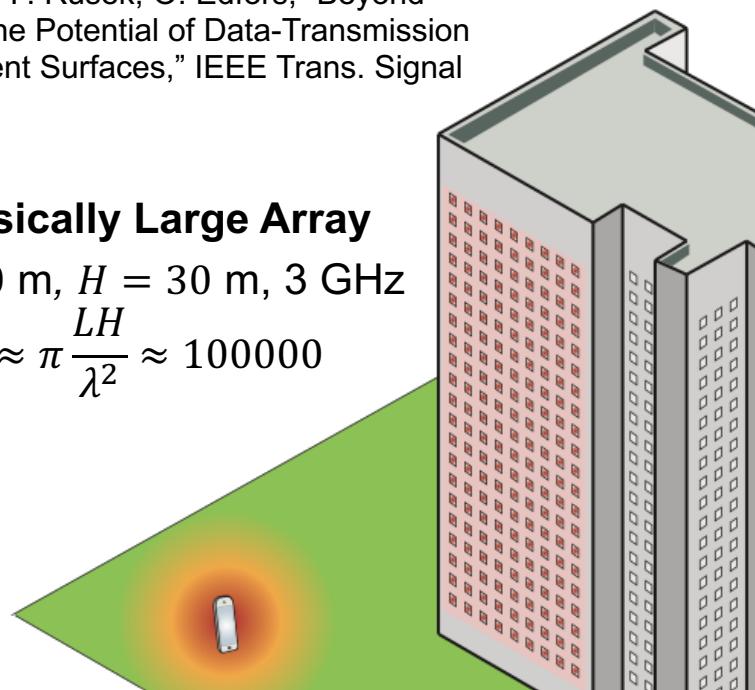
The richness of the channels is key

Reference: S. Hu, F. Rusek, O. Edfors, "Beyond Massive-MIMO: The Potential of Data-Transmission with Large Intelligent Surfaces," IEEE Trans. Signal Process., 2018.

Physically Large Array

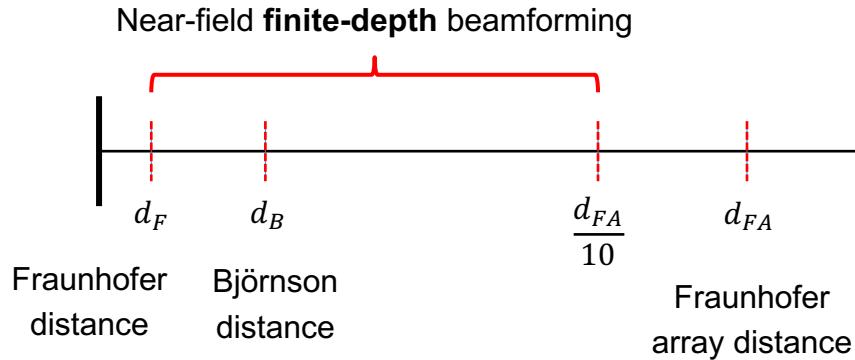
$$L = 10 \text{ m}, H = 30 \text{ m}, 3 \text{ GHz}$$

$$\eta \approx \pi \frac{LH}{\lambda^2} \approx 100000$$



Conclusions

- Physically large arrays:
 - User in far-field of elements
but maybe in near-field of the array



- Beamforming with finite depth and small width
 - Useful for spatial multiplexing
 - Maybe we don't need more bandwidth?

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